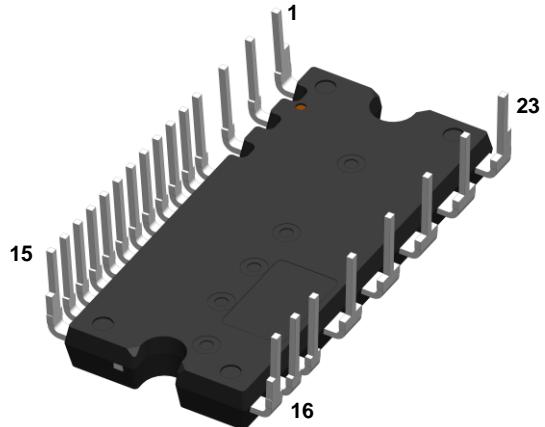


### External View



Size: 33.4 x 15 x 3.6 mm



Green

### Features

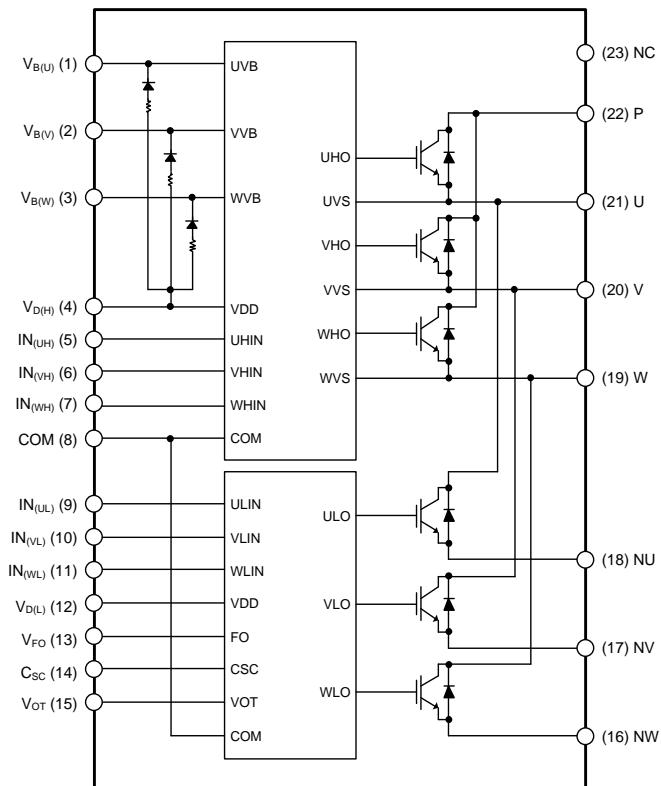
- UL Recognized
- 600V-10A (Trench Shielded Planar Gate IGBT)
- 3 phase Inverter module including HVIC drivers
- Built-in bootstrap diodes with integrated current-limiting resistor
- Control supply under-voltage lockout protection (UVLO)
- Over-temperature (OT) protection and temperature monitoring ( $V_{OT}$ ) – pin open
- Temperature monitoring ( $V_{OT}$ ) – 10k $\Omega$  resistor connection
- Short-circuit current protection ( $C_{SC}$ )
- Controllable fault out signal ( $V_{CF}$ ) corresponding to SC, UV and OT fault
- Wide input interface (3-18V), Schmitt trigger receiver circuit (Active High)
- Isolation ratings of 2000Vrms/min

### Applications

- AC 100-240Vrms class low power motor drives
- Washing Machines, Compressors, Fan Motors, Refrigerators and Dishwashers

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### Internal Equivalent Circuit / Pin Configuration



## Ordering Information

Part Number	Temperature Range	Package	Pin Length Description
AIM5D10L060Q4	-40°C to 150°C	IPM-5	Normal
AIM5D10L060Q4S	-40°C to 150°C	IPM-5A	Short



AOS Green Products use reduced levels of Halogens, and are also RoHS compliant. Please visit [www-aosmd.com/media/AOSGreenPolicy.pdf](http://www-aosmd.com/media/AOSGreenPolicy.pdf) for additional information.

## Pin Description

Pin Number	Pin Name	Pin Function
1	V <sub>B(U)</sub>	High-Side Bias Voltage for U-Phase IGBT Driving
2	V <sub>B(V)</sub>	High-Side Bias Voltage for V-Phase IGBT Driving
3	V <sub>B(W)</sub>	High-Side Bias Voltage for W-Phase IGBT Driving
4	V <sub>D(H)</sub>	High-Side Common Bias Voltage for IC and IGBTs Driving
5	IN <sub>(UH)</sub>	Signal Input for High-Side U-Phase
6	IN <sub>(VH)</sub>	Signal Input for High-Side V-Phase
7	IN <sub>(WH)</sub>	Signal Input for High-Side W-Phase
8	COM	Common Supply Ground
9	IN <sub>(UL)</sub>	Signal Input for Low-Side U-Phase
10	IN <sub>(VL)</sub>	Signal Input for Low-Side V-Phase
11	IN <sub>(WL)</sub>	Signal Input for Low-Side W-Phase
12	V <sub>D(L)</sub>	Low-Side Common Bias Voltage for IC and IGBTs Driving
13	V <sub>FO</sub>	Fault Output
14	C <sub>SC</sub>	Capacitor (Low-Pass Filter) for Short-circuit Current Detection Input
15	V <sub>OT</sub>	Voltage Output of LVIC Temperature
16	NW	Negative DC-Link Input for W-Phase
17	NV	Negative DC-Link Input for V-Phase
18	NU	Negative DC-Link Input for U-Phase
19	W	Output for W-Phase
20	V	Output for V-Phase
21	U	Output for U-Phase
22	P	Positive DC-Link Input
23	NC	No Connection

## Absolute Maximum Ratings

$T_J = 25^\circ\text{C}$ , unless otherwise specified.

Symbol	Parameter	Conditions	Ratings	Units
<b>Inverter</b>				
$V_{PN}$	Supply Voltage	Applied between P - NU,NV,NW	450	V
$V_{PN(\text{surge})}$	Supply Voltage (surge)	Applied between P - NU,NV,NW	500	V
$V_{CES}$	Collector-emitter Voltage		600	V
$I_C$	Output Phase Current	$T_C=25^\circ\text{C}, T_J<150^\circ\text{C}$	10	A
		$T_C=100^\circ\text{C}, T_J<150^\circ\text{C}$	5	A
$\pm I_{PK}$	Output Peak Phase Current	$T_C=25^\circ\text{C}$ , less than 1ms pulse width	20	A
$t_{sc}$	Short Circuit Withstand Time	$V_{PN}\leq 400\text{V}, T_J=150^\circ\text{C}, V_D=15\text{V}$	5	$\mu\text{s}$
$P_C$	Collector Dissipation	$T_C=25^\circ\text{C}$ , per chip	23	W
$T_J$	Operating Junction Temperature		-40 to 150	$^\circ\text{C}$
<b>Control (Protection)</b>				
$V_D$	Control Supply Voltage	Applied between $V_{D(H)}$ -COM, $V_{D(L)}$ -COM	25	V
$V_{DB}$	High-Side Control Bias Voltage	Applied between $V_{B(U)}$ -U, $V_{B(V)}$ -V, $V_{B(W)}$ -W	25	V
$V_{IN}$	Input Voltage	Applied between $IN_{(UH)}$ , $IN_{(VH)}$ , $IN_{(WH)}$ , $IN_{(UL)}$ , $IN_{(VL)}$ , $IN_{(WL)}$ - COM	$V_D\pm 0.5$	V
$V_{FO}$	Fault Output Supply Voltage	Applied between $V_{FO}$ - COM	$5\pm 0.5$	V
$I_{FO}$	Fault Output Current	Sink current at $V_{FO}$ terminal	1	mA
$V_{SC}$	Current Sensing Input Voltage	Applied between $C_{SC}$ - COM	$5\pm 0.5$	V
$V_{OT}$	Temperature Output	Applied between $V_{OT}$ - COM	$5\pm 0.5$	V
<b>Total System</b>				
$V_{PN(\text{PROT})}$	Self Protection Supply Voltage Limit (Short-circuit protection capability)	$V_D=13.5\text{-}16.5\text{V}$ , Inverter part $T_J=150^\circ\text{C}$ , Non-repetitive, less than $2\mu\text{s}$	400	V
$T_c$	Module Case Operation Temperature	Measurement point of $T_c$ is provided in Figure 1	-30 to 125	$^\circ\text{C}$
$T_{STG}$	Storage Temperature		-40 to 150	$^\circ\text{C}$
$V_{Iso}$	Isolation Voltage	60Hz, sinusoidal, AC 1min, between connected all pins and heat sink plate	2000	$V_{rms}$

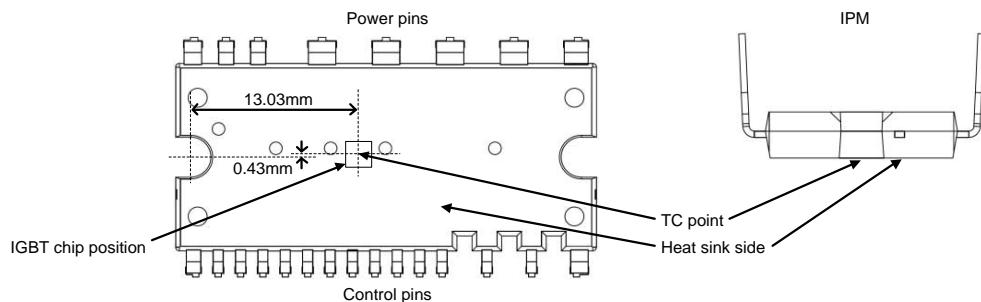


Figure 1.  $T_c$  Measurement Point

## Thermal Resistance

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
$R_{th(j-c)Q}$	Junction to Case Thermal Resistance <sup>(1)</sup>	Inverter IGBT (per 1/6 module)	-	-	5.4	K/W
$R_{th(j-c)F}$		Inverter FWD (per 1/6 module)	-	-	6.9	K/W

### Note:

- For the measurement point of case temperature ( $T_c$ ), please refer to Figure 1.

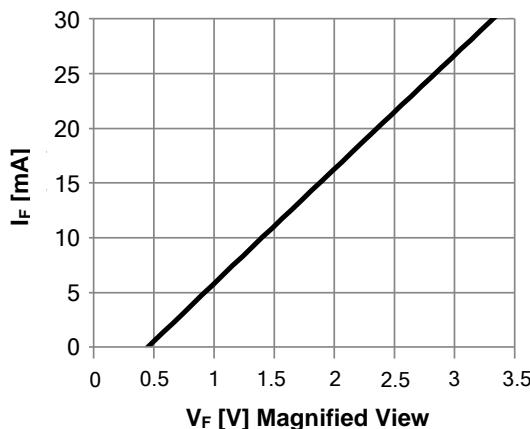
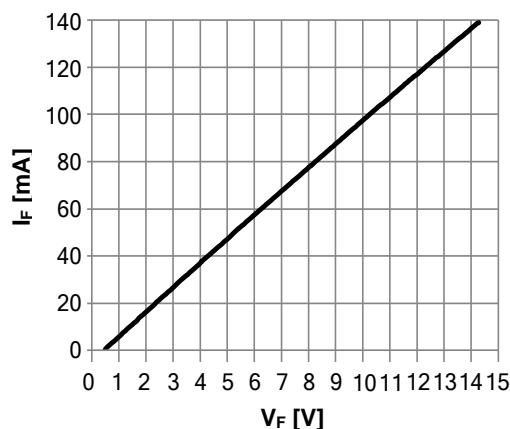
## Electrical Characteristics

$T_J = 25^\circ\text{C}$ , unless otherwise specified.

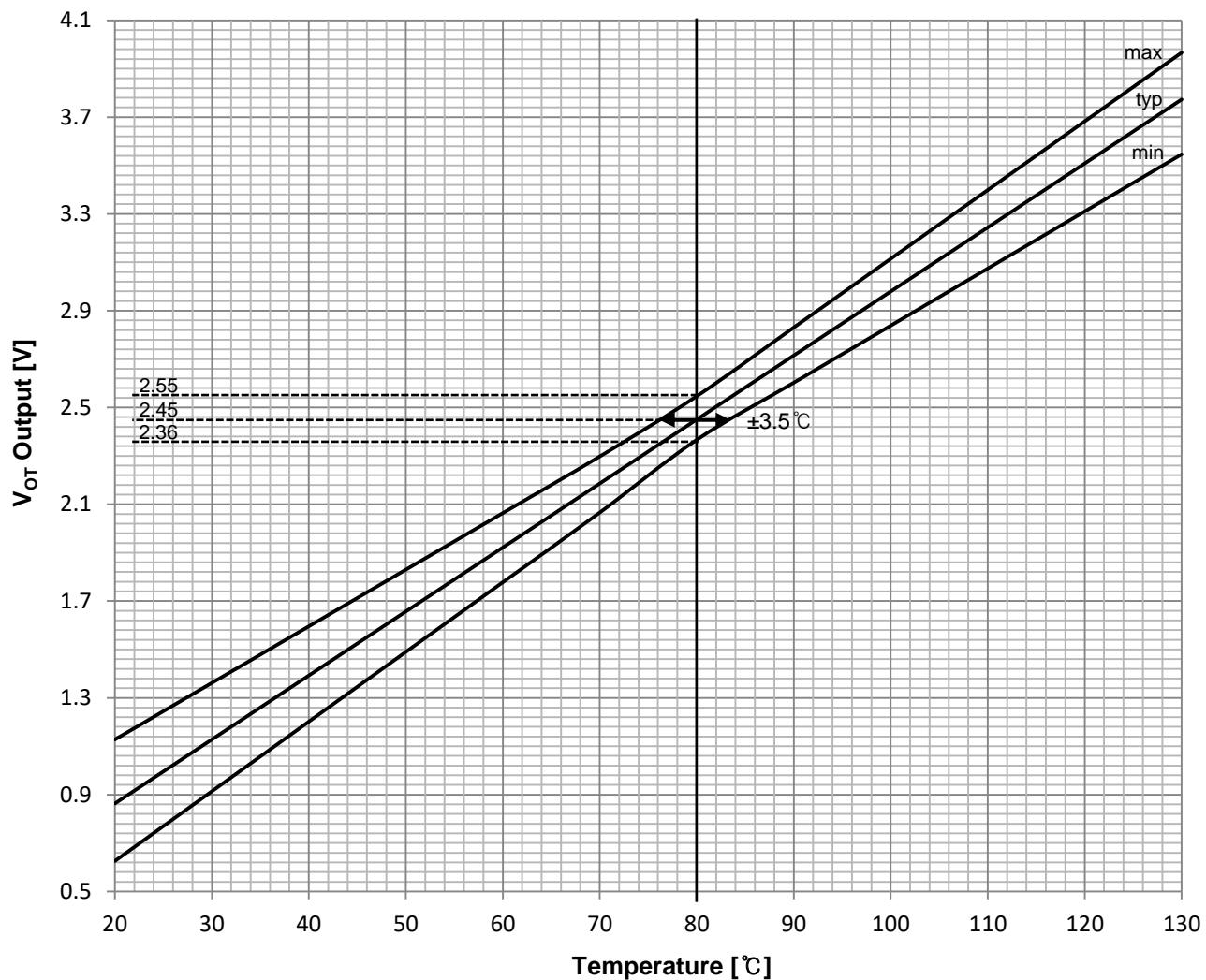
Symbol	Parameter	Conditions		Min.	Typ.	Max.	Units
<b>Inverter</b>							
$V_{CE(\text{SAT})}$	Collector-Emitter Saturation Voltage	$V_D=V_{DB}=15\text{V}$ , $V_{IN}=5\text{V}$	$I_C=5\text{A}$ , $T_J=25^\circ\text{C}$	-	1.60	2.00	V
			$I_C=5\text{A}$ , $T_J=125^\circ\text{C}$	-	1.90	-	V
$V_F$	FWD Forward Voltage	$V_{IN}=0$	$I_F=5\text{A}$ , $T_J=25^\circ\text{C}$	-	1.35	1.80	V
$t_{ON}$	Switching Times	$V_{PN}=300\text{V}$ , $V_D=V_{DB}=15\text{V}$ $I_C=5\text{A}$ , $T_J=25^\circ\text{C}$ , $V_{IN}=0\text{V} \leftrightarrow 5\text{V}$ Inductive load (high-side)	-	0.50	0.95	1.50	$\mu\text{s}$
$t_{C(ON)}$			-	-	0.40	0.70	$\mu\text{s}$
$t_{OFF}$			-	-	1.00	1.50	$\mu\text{s}$
$t_{C(OFF)}$			-	-	0.10	0.40	$\mu\text{s}$
$t_{rr}$			-	-	0.20	-	$\mu\text{s}$
$I_{CES}$	Collector-Emitter Leakage Current	$V_{CE}=V_{CES}$	$T_J=25^\circ\text{C}$	-	-	1	mA
			$T_J=125^\circ\text{C}$	-	-	10	mA
<b>Control (Protection)</b>							
$I_{QDH}$	Quiescent $V_D$ Supply Current	$V_{D(H)}=15\text{V}$ , $V_{IN(UH, VH, WH)}=0\text{V}$	$V_{D(H)}$ - COM	-	-	0.1	mA
$I_{QDL}$		$V_{D(L)}=15\text{V}$ , $V_{IN(UL, VL, WL)}=0\text{V}$	$V_{D(L)}$ - COM	-	-	2.1	mA
$I_{QDB}$	Quiescent $V_{DB}$ Supply Current	$V_{DB}=15\text{V}$ , $V_{IN(UH, VH, WH)}=0\text{V}$	$V_{B(U)}-U$ , $V_{B(V)}-V$ , $V_{B(W)}-W$	-	-	0.3	mA
$V_{SC(\text{ref})}$	Short-Circuit Trip Level	$V_D=15\text{V}^{(2)}$		0.45	0.48	0.51	V
$UV_{DT}$	Supply Circuit Under-Voltage Protection	Trip Level		10.3	11.4	12.5	V
$UV_{DR}$		Reset Level		10.8	11.9	13.0	V
$UV_{DBT}$		Trip Level		8.5	9.5	10.5	V
$UV_{DBR}$		Reset Level		9.5	10.5	11.5	V
$V_{OT}$	Temperature Output	Pull-down $R=10\text{k}\Omega^{(3)}$	$LVIC$ Temperature= $80^\circ\text{C}$	2.36	2.45	2.55	V
			$LVIC$ Temperature= $25^\circ\text{C}$	0.77	1.00	1.25	V
$OT_T$	Over-Temperature Protection <sup>(4)</sup>	$V_D=15\text{V}$ , Detect $LVIC$ Temperature	Trip Level	110	130	150	$^\circ\text{C}$
$OT_{HYS}$			Hysteresis of Trip Reset	-	30	-	$^\circ\text{C}$
$V_{FOH}$	Fault Output Voltage	$V_{SC}=0\text{V}$ , $V_{FO}$ Circuit: $10\text{k}\Omega$ to $5\text{V}$ pull-up	4.9	-	-	-	V
$V_{FOL}$		$V_{SC}=1\text{V}$ , $V_{FO}$ Circuit: $10\text{k}\Omega$ to $5\text{V}$ pull-up	-	-	-	0.5	V
$t_{FO}$	Fault Output Pulse Width <sup>(5)</sup>			20	-	-	$\mu\text{s}$
$I_{IN}$	Input Current	$V_{IN}=5\text{V}$		-	1.0	-	mA
$V_{th(on)}$	ON Threshold Voltage	Applied between $IN_{(UH)}$ , $IN_{(VH)}$ , $IN_{(WH)}$ , $IN_{(UL)}$ , $IN_{(VL)}$ , $IN_{(WL)}$ – COM			2.3	2.6	V
$V_{th(off)}$	OFF Threshold Voltage		0.8	1.2	-	-	V
$V_{th(hys)}$	ON/OFF Threshold Hysteresis Voltage		-	1.1	-	-	V
$V_{F(BSD)}$	Bootstrap Diode Forward Voltage	$I_F=10\text{mA}$ Including Voltage Drop by Limiting Resistor <sup>(6)</sup>		1.0	1.5	2.0	V
$R_{BSD}$	Built-in Limiting Resistance	Included in Bootstrap Diode		80	100	120	$\Omega$

### Notes:

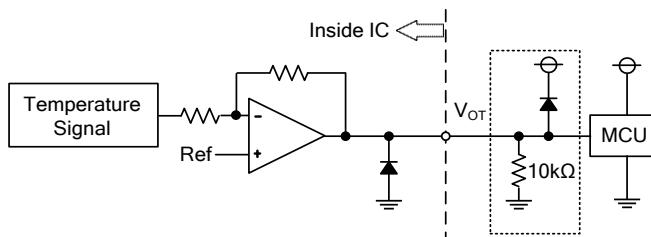
2. Short-circuit protection works only for low sides.
3. The IPM does not shutdown IGBTs and output fault signal automatically when temperature rises excessively. When temperature exceeds the protective level that the user defined, the controller (MCU) should stop the IPM. Temperature of LVIC vs.  $V_{OT}$  output characteristics is described in Figure 3.
4. When the LVIC temperature exceeds OT Trip temperature level ( $OT_T$ ), OT protection is triggered and fault outputs.
5. Fault signal ( $F_O$ ) outputs when SC, UV or OT protection is triggered.  $F_O$  pulse width is different for each protection mode. At SC failure,  $F_O$  pulse width is a fixed width (minimum  $20\mu\text{s}$ ), but at UV or OT failure,  $F_O$  outputs continuously until recovering from UV or OT state. (But minimum  $F_O$  pulse width is  $20\mu\text{s}$ ).
6. The characteristics of bootstrap diodes are described in Figure 2.



**Figure 2. Built-in Bootstrap Diode  $V_F$ - $I_F$  Characteristic (@ $T_A=25^\circ\text{C}$ )**



**Figure 3. Temperature of LVIC vs.  $V_{OT}$  Output Characteristics**


**Figure 4. V<sub>OT</sub> Output Circuit**

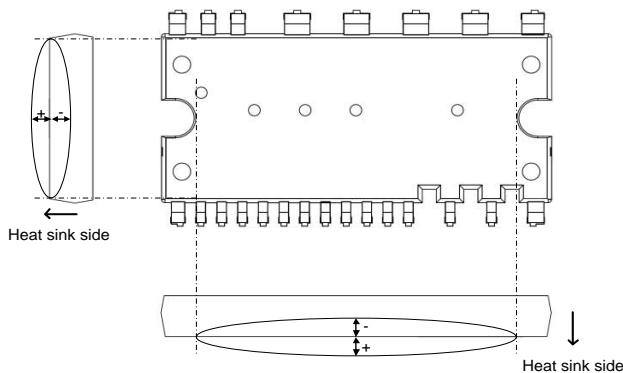
- (1) Connect 10kΩ to V<sub>OT</sub> pin if temperature monitoring function is utilized; otherwise if the V<sub>OT</sub> pin is left unconnected, the internal over-temperature shutdown function is used instead.
- (2) In the case of using V<sub>OT</sub> with low voltage controller like 3.3V MCU, V<sub>OT</sub> output might exceed control supply voltage 3.3V when temperature rises excessively. If system uses low voltage controller, it is recommended to insert a clamp diode between control supply of the controller and V<sub>OT</sub> output for preventing over voltage destruction.

## Mechanical Characteristics and Ratings

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
Mounting torque	Mounting Screw: M3 <sup>(7)</sup>		0.59	0.69	0.78	N m
Weight			-	5.25	-	g
Flatness	Refer to Figure 5		-50	-	100	μm

**Note:**

7. Plain washers (ISO 7089-7094) are recommended.


**Figure 5. Flatness Measurement Positions**

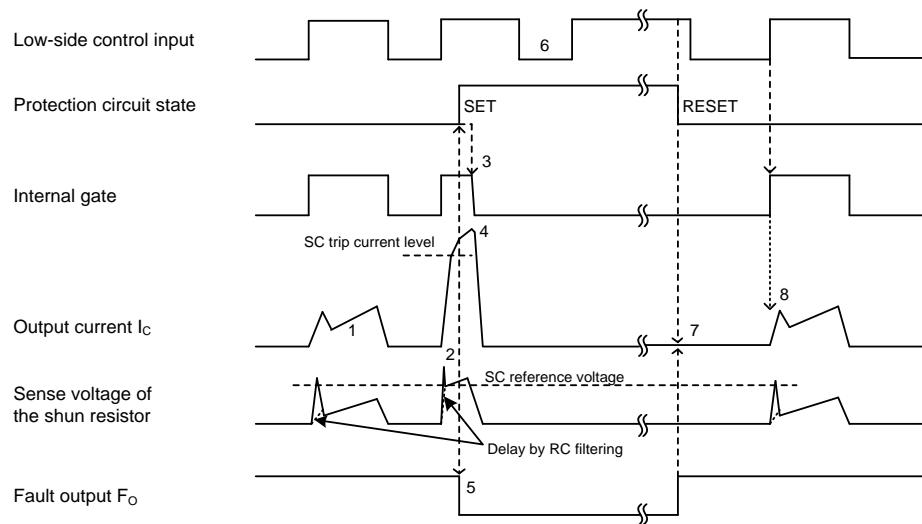
## Recommended Operation Conditions

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Units
V <sub>PN</sub>	Supply Voltage	Applied between P-NU, NV, NW	0	300	400	V
V <sub>D</sub>	Control Supply Voltage	Applied between V <sub>D(H)</sub> - COM, V <sub>D(L)</sub> - COM	13.5	15.0	16.5	V
V <sub>DB</sub>	High-Side Bias Voltage	Applied between V <sub>B(U)</sub> -U, V <sub>B(V)</sub> -V, V <sub>B(W)</sub> -W	13.5	15.0	18.5	V
dV <sub>D</sub> /dt, dV <sub>DB</sub> /dt	Control Supply Variation		-1	-	1	V/μs
t <sub>dead</sub>	Arm Shoot-Through Blocking Time	For each input signal	1.0	-	-	μs
f <sub>PWM</sub>	PWM Input Frequency	-40°C < T <sub>J</sub> < 150°C	-	-	20	kHz
PW <sub>IN(ON)</sub>	Minimum Input Pulse Width <sup>(8)</sup>		0.5	-	-	μs
PW <sub>IN(OFF)</sub>			0.5	-	-	μs
COM	COM Variation	Between COM - NU, NV, NW (including surge)	-5.0	-	5.0	V

**Note:**

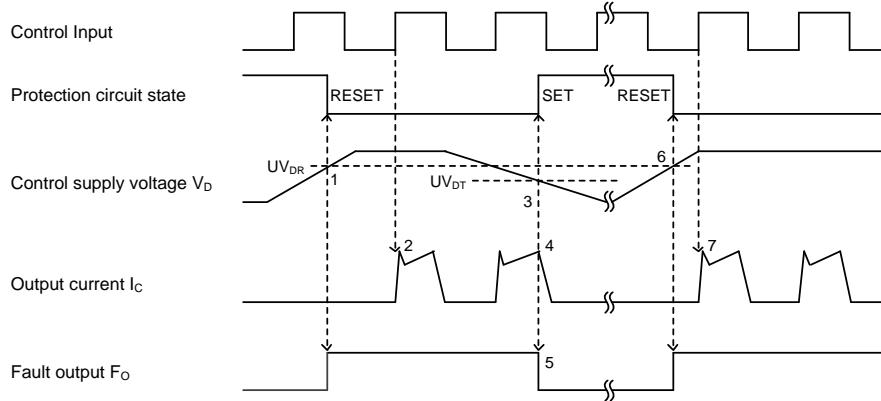
8. IPM may not respond if the input pulse width is less than PW<sub>IN(ON)</sub>, PW<sub>IN(OFF)</sub>.

## Time Charts of the IPM Protective Function



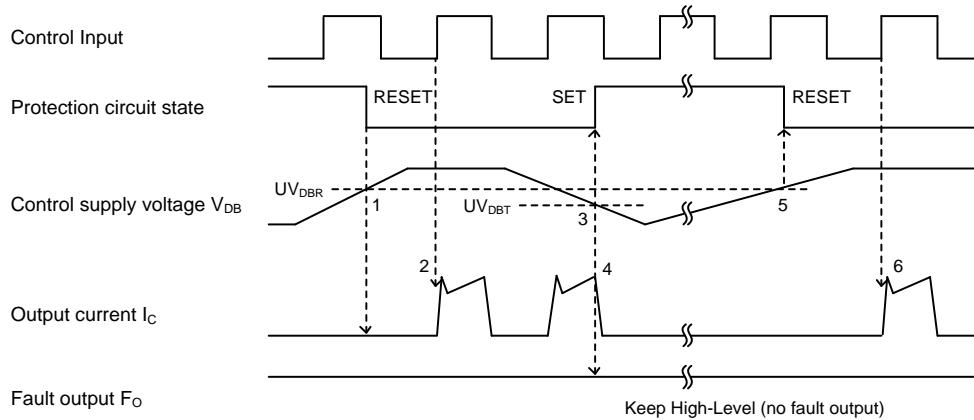
**Figure 6. Short-Circuit Protection  
(Low-side Operation Only with the External Shunt Resistor and RC Filter)**

- (1) Normal operation: IGBT turns on and outputs current.
- (2) Short-circuit current detection (SC triggered).
- (3) All low-side IGBTs' gates are hard interrupted.
- (4) All low-side IGBTs turn OFF.
- (5)  $F_O$  output time ( $t_{FO}$ )=minimum 20μs.
- (6) Input = "L" : IGBT OFF.
- (7) Fault output finishes, but output current will not turn on until next ON signal (L→H).
- (8) Normal operation: IGBT turns on and outputs current.



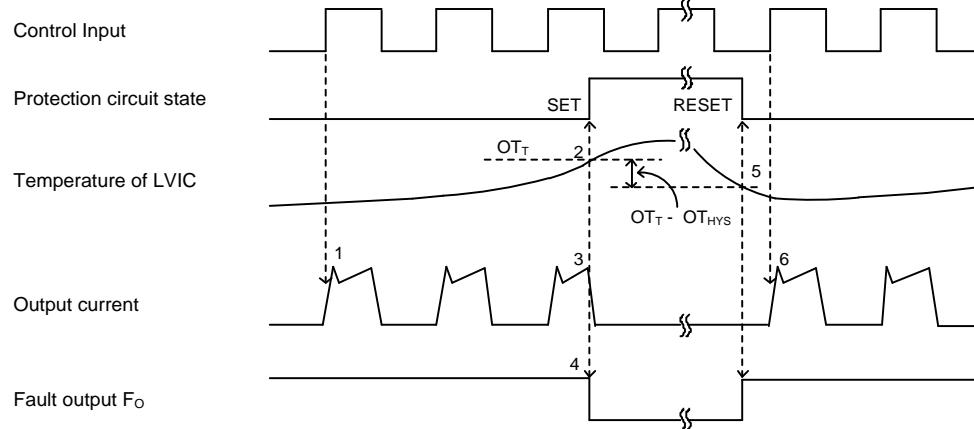
**Figure 7. Under-Voltage Protection (Low-side,  $UV_D$ )**

- (1) Control supply voltage  $V_D$  exceeds under voltage reset level ( $UV_{DR}$ ), but IGBT turns on by next ON signal (L→H).
- (2) Normal operation: IGBT turns on and outputs current.
- (3)  $V_D$  level drops to under voltage trip level ( $UV_{DT}$ ).
- (4) All low-side IGBTs turn OFF regardless of control input condition.
- (5)  $F_O$  output time ( $t_{FO}$ )=minimum 20μs, and  $F_O$  stays low as long as  $V_D$  is below  $UV_{DR}$ .
- (6)  $V_D$  level reaches  $UV_{DR}$ .
- (7) Normal operation: IGBT turns on and outputs current.



**Figure 8. Under-Voltage Protection (High-side, UV<sub>DB</sub>)**

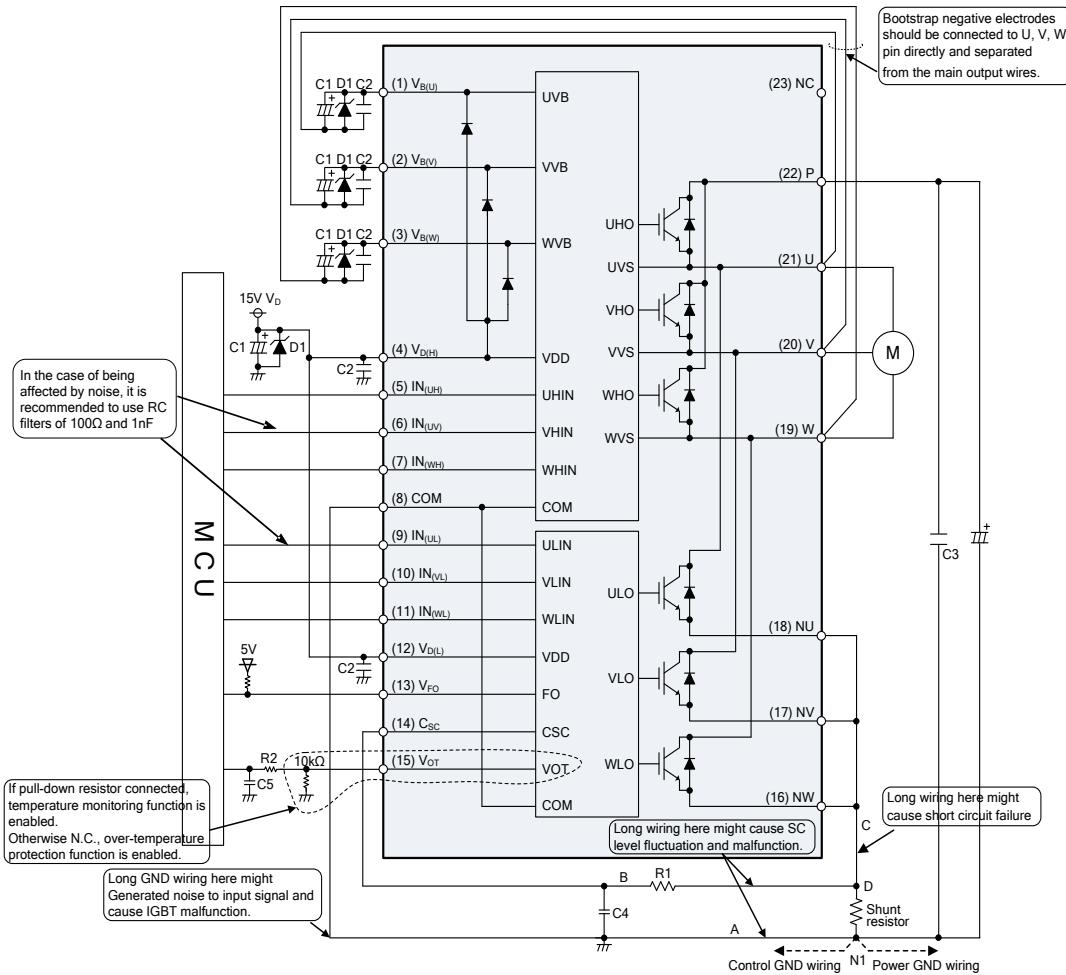
- (1) Control supply voltage  $V_{DB}$  rises. After the voltage reaches under voltage reset level UV<sub>DBR</sub>, IGBT turns on by next ON signal (L→H).
- (2) Normal operation: IGBT turns on and outputs current.
- (3)  $V_{DB}$  level drops to under voltage trip level (UV<sub>DBT</sub>).
- (4) All high-side IGBTs turn OFF regardless of control input condition, but there is no  $F_O$  signal output.
- (5)  $V_{DB}$  level reaches UV<sub>DBR</sub>.
- (6) Normal operation: IGBT turns on and outputs current.



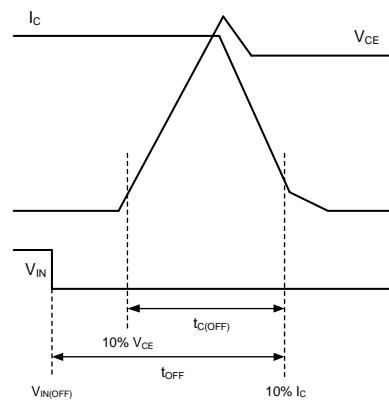
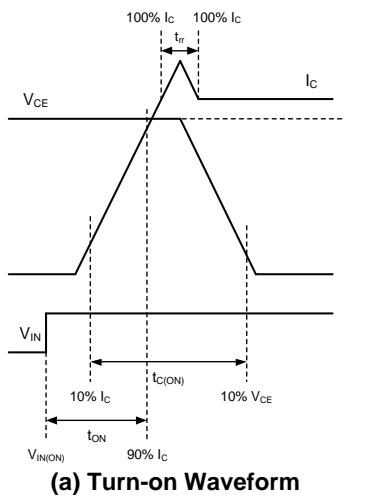
**Figure 9. Over-Temperature Protection (Low-side, Detecting LVIC Temperature)**

- (1) Normal operation: IGBT turns on and outputs current.
- (2) LVIC temperature exceeds over-temperature trip level ( $OT_T$ ).
- (3) All low-side IGBTs turn OFF regardless of control input condition.
- (4)  $F_O$  output time ( $t_{FO}$ )=minimum 20μs, and  $F_O$  stays low as long as LVIC temperature is over  $OT_T$ .
- (5) LVIC temperature drops to over-temperature reset level ( $OT_T - OT_{HYS}$ ).
- (6) Normal operation: IGBT turns on by the next ON signal (L→H).

## Example of Application Circuit

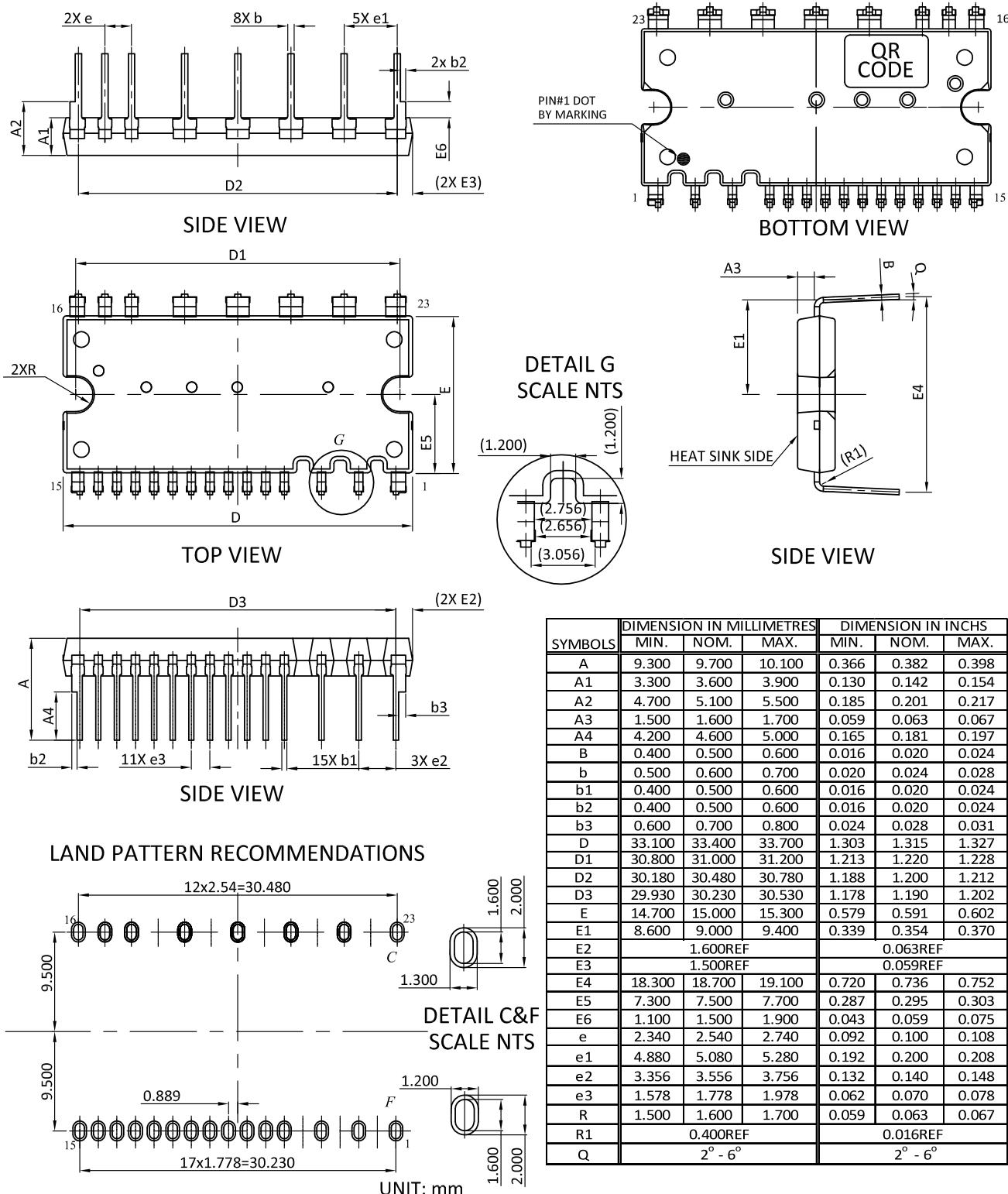


- (1) If the control GND is connected with the power GND by common broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect the control GND and power GND at a single point (N1), near the terminal of the shunt resistor.
- (2) A zener diode D1 (24V/1W) is recommended between each pair of control supply pins to prevent surge destruction.
- (3) Prevention of surge destruction can further be improved by placing the bus capacitor as close to pins P and N1 as possible. Generally a  $0.1\text{-}0.22\mu\text{F}$  snubber capacitor C3 between the P-N1 terminals is recommended.
- (4) Selection of the  $R1 \cdot C4$  filter components for short-circuit protection is recommended to have tight tolerance, and is temperature-compensated type. The  $R1 \cdot C4$  time constant should be set such that SC current is shut down within  $2\mu\text{s}$ ; (typically  $1.5\text{-}2\mu\text{s}$ ). R1 and C4 should be placed as close as possible to the CSC pin. SC interrupting time may vary with layout patterns and components selections, therefore thorough evaluation in the system is necessary.
- (5) Tight tolerance, and temperature-compensated components are also recommended when selecting the  $R2 \cdot C5$  filter for  $V_{OT}$ . The  $R2 \cdot C5$  time constant should be set such that  $V_{OT}$  is immune to noise. Recommended values of R2 and C5 are  $2k\Omega$  and  $10nF$ .
- (6) To prevent malfunction, traces A, B, and C should be as short as possible.
- (7) It is recommended that all capacitors are mounted as close to the IPM as possible. (C1: electrolytic type with good temperature and frequency characteristics. C2: Ceramic type with  $0.1\text{-}2\mu\text{F}$ , good temperature, frequency and DC bias characteristics.)
- (8) Input drives are active-high. There is a minimum  $3.5k\Omega$  pull-down resistor in the input circuit of IC. To prevent malfunction, the layout to each input should be as short as possible. When using RC coupling circuit, make sure the input signal levels meet the required turn-on and turn-off threshold voltages.
- (9)  $V_{FO}$  output is open drain type. It should be pulled up to MCU or control power supply (max=  $5\pm 0.5V$ ), limiting the current ( $I_{FO}$ ) to no more than  $1\text{mA}$ .  $I_{FO}$  is estimated roughly by the formula of control power supply voltage divided by pull-up resistor. For example, if control supply is  $5V$ , a  $10k\Omega$  (over  $5k\Omega$ ) pull-up resistor is recommended.
- (10) Direct drive of the IPM from the MCU is possible without having to use opto-coupler or isolation transformer.
- (11) The IPM may malfunction and erroneous operations may occur if high frequency noise is superimposed to the supply line. To avoid such problems, line ripple voltage is recommended to have  $dV/dt \leq \pm 1V/\mu\text{s}$ , and  $V_{ripple} \leq 2V_{p-p}$ .
- (12) It is not recommended to use the IPM to drive the same load in parallel with another IPM or inverter types.



**Figure 10. Switching Times Definition**

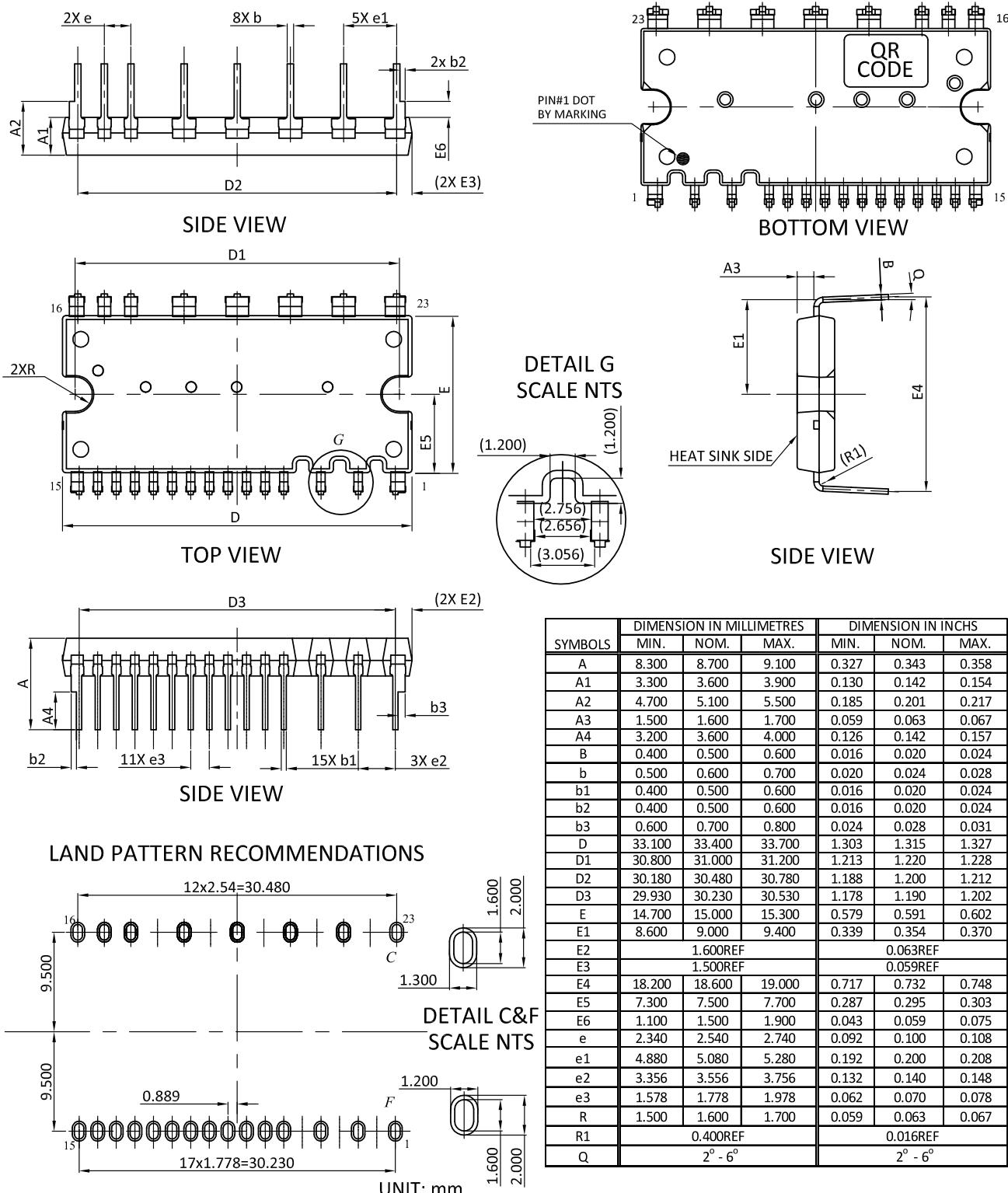
## Package Dimensions, IPM-5



### NOTES

1. PACKAGE BODY SIZES EXCLUDE MOLD FLASH AND GATE BURRS, MOLD FLASH SHOULD BE LESS THAN 6 MIL.
2. TOLERANCE 0.100 MILLIMETERS UNLESS OTHERWISE SPECIFIED.
3. CONTROLLING DIMENSION IS MILLIMETER, CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
4. ( ) IS REFERENCE.

## Package Dimensions, IPM-5A



### NOTES

1. PACKAGE BODY SIZES EXCLUDE MOLD FLASH AND GATE BURRS, MOLD FLASH SHOULD BE LESS THAN 6 MIL.
2. TOLERANCE 0.100 MILLIMETERS UNLESS OTHERWISE SPECIFIED.
3. CONTROLLING DIMENSION IS MILLIMETER, CONVERTED INCH DIMENSIONS ARE NOT NECESSARILY EXACT.
4. ( ) IS REFERENCE.

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## LIFE SUPPORT POLICY

ALPHA & OMEGA SEMICONDUCTOR PRODUCTS ARE NOT AUTHORIZED FOR USE AS CRITICAL COMPONENTS IN LIFE SUPPORT DEVICES OR SYSTEMS.

As used herein:

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body or (b) support or sustain life, and (c) whose failure to perform when properly used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury of the user.

2. A critical component in any component of a life support, device, or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

单击下面可查看定价，库存，交付和生命周期等信息

[>>AOS\(万代\)](#)