

36V, 10MHz

Low-Noise Dual Operational Amplifiers

Features

- Operates on ±2.5V to ±18V Supplies
- Gain Bandwidth Product: 10MHz
- Power Bandwidth: 140kHz
- Slew Rate: 8V/µs
- Offset Voltage: 5mV (Max.)
- Quiescent Current: 2.8mA
- Output Drive Capability: 2kΩ, 10Vrms typ
- Extended Temperature Ranges
 From -40°C to +125°C
- Available in SOP-8/MSOP-8/DIP-8

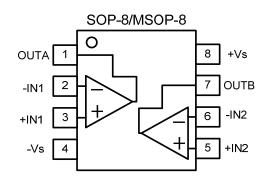
Applications

- Precision Instrumentation
- Professional Audio
- DAC Output Amplifier
- Active Filters
- Low Noise Amplifier Front End

General Description

The COS5532 are high performance, low noise amplifiers operational combining excellent dc and ac characteristics. They feature very low noise, high output-drive unity-gain capability, high and maximum-output-swing bandwidths. low distortion, high slew rate, and output short-circuit protection. These operational amplifiers are compensated internally for unity-gain operation and can operate from ± 2.5 to $\pm 18V$ dual power supplies or from $\pm 5V$ to +36V single supplies.

Pin Configuration



Rev1.0

Copyright@2018 Cosine Nanoelectronics Inc. All rights reserved

The information provided here is believed to be accurate and reliable. Cosine Nanoelectronics assumes no reliability for inaccuracies and omissions. Specifications described and contained here are subjected to change without notice on the purpose of improving the design and performance. All of this information described herein should not be implied or granted for any third party.

www.cosine-ic.com



1. Product Specification

1.1 Absolute Maximum Ratings⁽¹⁾

Parameter	Rating	Units
Power Supply: +Vs to -Vs	36	V
Differential Input Voltage Range	±30	V
Input Voltage (any input)	±15	V
Output Current	50	mA
Storage Temperature Range	-65 to 150	°C
Junction Temperature	150	°C
Operating Temperature Range	-40 to 125	°C
ESD Susceptibility, HBM	2000	V

(1) Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

1.2 Thermal Data

Parameter	Rating	Unit
Package Thermal Resistance	155 (SOP8) 206 (MSOP8) 125 (DIP8)	°C/W

1.3 Recommended Operating Conditions

Parameter	Rating	Unit
DC Supply Voltage	±2.5V ~ ±18V	V
Input common-mode voltage range	-Vs+2 ~ +Vs-2	V
Operating ambient temperature	-40 to +85	°C



1.4 Electrical Characteristics

(+V_S=+15V, -V_S=-15V, T_A=+25°C, R_L=10k Ω to V_S/2, unless otherwise noted)

Parameter	Symbol	Conditions	Min	Тур	Max	Unit
Input Characteristics						
Input Offset Voltage	Vos			0.5	5	mV
Input Offset Voltage Drift	ΔV _{os} /ΔT	-40 to 125°C		2		μV/°C
Input Bias Current	IB			200	800	nA
Input Offset Current	los			50	200	nA
Common-Mode Voltage Range	Vсм			±13		V
Common-Mode Rejection Ratio	CMRR	Rs≪10kΩ	70	100		dB
		$R_{L} \ge 10k\Omega, V_{O} = \pm 10V$	88	110		dB
Open-Loop Voltage Gain	AOL	$R_L \ge 2k\Omega, V_O = \pm 10V$	82	94		dB
Output Characteristics						
Output Voltage Swing	V _{O(PP)}	R _L ≥2kΩ	±12	±13		V
Short-Circuit Current	Isc			60		mA
Power Supply			·			
Operating Voltage Range			±2.5		±18	V
Power Supply Rejection Ratio	PSRR	Rs≪10kΩ	80	110		dB
Quiescent Current / Amplifier	lq			2.8	3.5	mA
Dynamic Performance						
Gain Bandwidth Product	GBWP	C∟=100pF, R∟=2kΩ		10		MHz
Slew Rate	SR	C∟=100pF, R∟=2kΩ, Av=1		8.0		V/µs
Noise Performance			1	1		1
Voltage Noise Density	en	f=1kHz		5.0		nV/√Hz



2.0 Application Notes

Driving Capacitive Loads

Driving large capacitive loads can cause stability problems for voltage feedback op amps. As the load capacitance increases, the feedback loop's phase margin decreases, and the closed loop bandwidth is reduced. This produces gain peaking in the frequency response, with overshoot and ringing in the step response. A unity gain buffer (G = +1) is the most sensitive to capacitive loads, but all gains show the same general behavior.

When driving large capacitive loads with these op amps (e.g., > 100 pF when G = +1), a small series resistor at the output (R_{ISO} in Figure 1) improves the feedback loop's phase margin (stability) by making the output load resistive at higher frequencies. It does not, however, improve the bandwidth.

To select R_{ISO} , check the frequency response peaking (or step response overshoot) on the bench. If the response is reasonable, you do not need R_{ISO} . Otherwise, start R_{ISO} at 1 k Ω and modify its value until the response is reasonable.

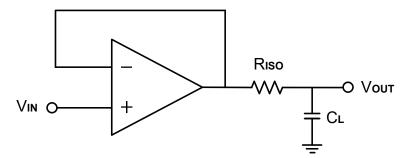


Figure 1. Indirectly Driving Heavy Capacitive Load

An improvement circuit is shown in Figure 2. It provides DC accuracy as well as AC stability. R_F provides the DC accuracy by connecting the inverting signal with the output, C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving phase margin in the overall feedback loop.

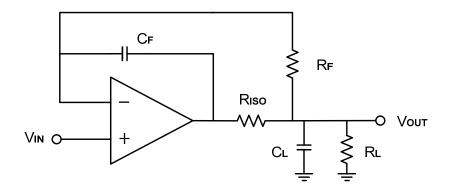


Figure 2. Indirectly Driving Heavy Capacitive Load with DC Accuracy



For noninverting configuration, there are two others ways to increase the phase margin: (a) by increasing the amplifier's gain or (b) by placing a capacitor in parallel with the feedback resistor to counteract the parasitic capacitance associated with inverting node, as shown in Figure 3.

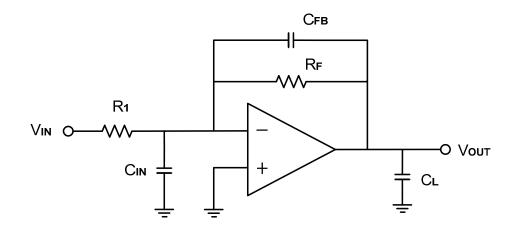


Figure 3. Adding a Feedback Capacitor in the Noninverting Configuration

Power-Supply Bypassing and Layout

The COS5532 operates from a single +5V to +36V supply or dual ±2.5V to ±18V supplies. For single-supply operation, bypass the power supply +Vs with a 0.1μ F ceramic capacitor which should be placed close to the +Vs pin. For dual-supply operation, both the +Vs and the -Vs supplies should be bypassed to ground with separate 0.1μ F ceramic capacitors. 2.2 μ F tantalum capacitor can be added for better performance.

The length of the current path is directly proportional to the magnitude of parasitic inductances and thus the high frequency impedance of the path. High speed currents in an inductive ground return create an unwanted voltage noise. Broad ground plane areas will reduce the parasitic inductance. Thus a ground plane layer is important for high speed circuit design.

Typical Application Circuits

Differential Amplifier

The circuit shown in Figure 4 performs the differential function. If the resistors ratios are equal (R₄ / R₃ = R₂ / R₁), then $V_{OUT} = (V_{IP} - V_{IN}) \times R_2 / R_1 + V_{REF}$.

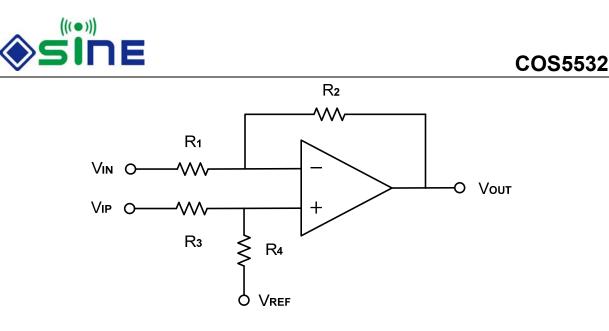


Figure 4. Differential Amplifier

Low Pass Active Filter

When receiving low-level signals, limiting the bandwidth of the incoming signals into the system is often required. The simplest way to establish this limited bandwidth is to place an RC filter at the noninverting terminal of the amplifier. If even more attenuation is needed, a multiple pole filter is required. The Sallen-Key filter can be used for this task, as Figure 5. For best results, the amplifier should have a bandwidth that is 8 to 10 times the filter frequency bandwidth. Failure to follow this guideline can result in reduction of phase margin. The large values of feedback resistors can couple with parasitic capacitance and cause undesired effects such as ringing or oscillation in high-speed amplifiers. Keep resistors value as low as possible and consistent with output loading consideration.

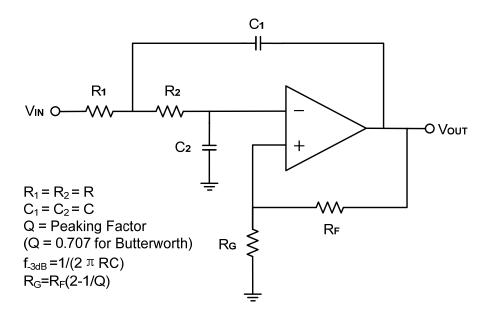
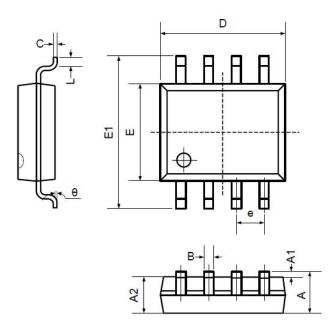


Figure 5. Two-Pole Low-Pass Sallen-Key Active Filter



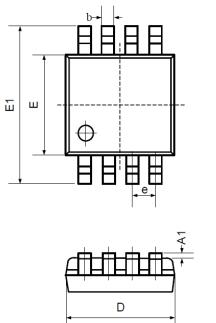
3. Package Information

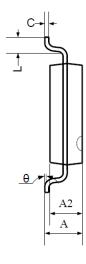
3.1 SOP8 (Package Outline Dimensions)



Symbol	Dimensions In Millimeters		Dimensions In Inches		
	Min	Max	Min	Max	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
В	0.330	0.510	0.013	0.020	
С	0.190	0.250	0.007	0.010	
D	4.780	5.000	0.188	0.197	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.300	0.228	0.248	
е	1.270TYP 0.050TY		DTYP		
L	0.400	1.270	0.016	0.050	
θ	0 °	8°	0°	8°	

3.2 MSOP8 (Package Outline Dimensions)





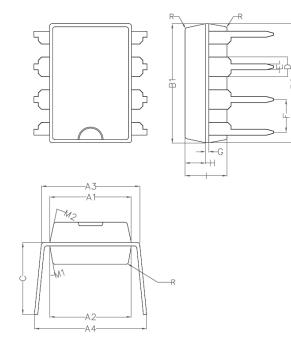
Symbol	Dimensions In Millimeters		Dimer In Ind	nsions ches
	Min	Max	Min	Max
А	0.800	1.200	0.031	0.047
A1	0.000	0.200	0.000	0.008
A2	0.760	0.970	0.030	0.038
b	0.30 TYP		0.012 TYP	
С	0.15 TYP		0.006 TYP	
D	2.900	3.100	0.114	0.122
е	0.65 TYP		0.026	S TYP
E	2.900	3.100	0.114	0.122
E1	4.700	5.100	0.185	0.201
L	0.410	0.650	0.016	0.026
θ	0°	6°	0°	6°

www.cosine-ic.com



COS5532

3.3 DIP8 (Package Outline Dimensions)



Symbol	Min	Non	Max
A1	6.28	6.33	6.38
A2	6.33	6.38	6.43
A3	7.52	7.62	7.72
A4	7.80	8.40	9.00
B1	9.15	9.20	9.25
B2	9.20	9.25	9.30
С		5.57	
D		1.52	
E	0.43	0.45	0.47
F		2.54	
G		0.25	
Н	1.54	1.59	1.64
I	3.22	3.27	3.32
R		0.20	
M1	9°	10°	11°
M2	11°	12°	13°

4. Package and Ordering Information

Model	Channel	Order Number	Package	Package Option	Marking Information
0005530	COS5532SR	SOP-8	Tape and Reel, 3000	COS5532SR	
	COS5532MR	MSOP-8	Tape and Reel, 3000	COS5532MR	
COS5532	2 2	COS5532DR	DIP-8	Tape and Reel, 1500	COS5532DR
		COS5532DT	DIP-8	Tube, 50	COS5532DT

5. Related Parts

Part Number	Description
COS1177/2177/4177	36V high precision Op Amps, 5 to 36V Supply, Vos<50µV
COS1347/2347/4347	350kHz, 15µA, RRIO Op Amps, 1.8 to 5.5V Supply
COS6001/2/4	1.5MHz, 50µA, RRIO Op Amps, 1.8 to 5.5V Supply
COS721/2/4	10MHz, 650µA, RRIO Op Amps, 2.1 to 5.5V Supply
COS1333/2333/4333	0.35MHz, 18µA, RRIO Op Amps, 1.8 to 5.5V Supply, Zero Drift, Vos<20µV
COS8551/2/4	1.5MHz, 55 μ A, RRIO Op Amps, 1.8 to 5.5V Supply, Zero Drift, Vos<10 μ V

www.cosine-ic.com

单击下面可查看定价,库存,交付和生命周期等信息

>>COSINE(科山芯创)