

### Description

The APR349 is a secondary-side MOSFET driver for synchronous rectification, which can effectively reduce the secondary-side rectifier power dissipation and provide a high-performance solution.

The APR349 supports continuous or discontinuous conduction mode (CCM and DCM) and quasi-resonant flyback operation. It utilizes an intelligent control method to realize on-time control, which not only provides a short turn-on and turn-off delay to reduce power loss, but also eliminate the false trigger of resonant ring and keep safe operation without adding any external components or circuitry.

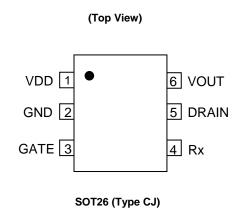
The APR349 can be charged by internal PLR (Pulse Linear Regulator) at low system output voltage, and provide a wide output operation voltage from 3.3V to 24V.

The APR349 is available in SOT26 (Type CJ) package.

### Features

- Support Flyback Synchronous Rectification with CCM/DCM/QR Operation Mode
- Intelligent Control to Minimize Turn-On and Turn-Off Delay
- Smart-Blanking Control to Prevent Shoot-Through
- Supports Both High-Side and Low-Side Application
- Built-in PLR Power Supply for Low System Output Voltage
- Internal UVLO Protection
- Fewest External Components
- Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)
- Halogen and Antimony Free. "Green" Device (Note 3)
- For automotive applications requiring specific change control (i.e. parts qualified to AEC-Q100/101/200, PPAP capable, and manufactured in IATF 16949 certified facilities), please <u>contact us</u> or your local Diodes representative. <u>https://www.diodes.com/quality/product-definitions/</u>

### Pin Assignments

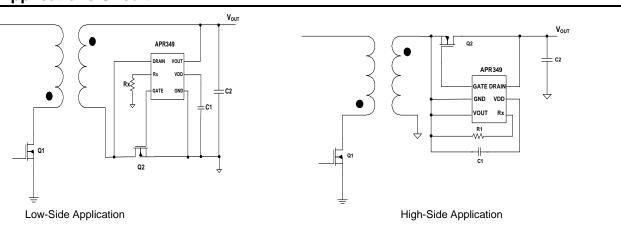


### Applications

- Adaptive Chargers for Cellular Phones, Digital Cameras, and Power Tools
- AC/DC Adapters
- Battery-Powered Systems
- All SMPS with High-Efficiency Requirements

- Notes: 1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
  - 2. See https://www.diodes.com/quality/lead-free/ for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
  - 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.

## **Typical Applications Circuit**

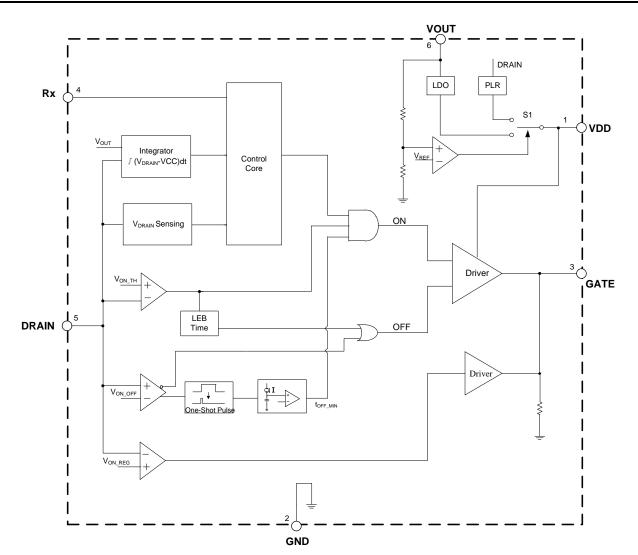




# **Pin Descriptions**

Pin Number	Pin Name	Function	
1	VDD	Supply input. It provides bias voltage for the internal logic circuit and the MOSFET driver.	
2	GND	Ground, also used as FET source sense reference for DRAIN	
3	GATE	Synchronous Rectification MOSFET Gate Drive	
4	Rx	Program a voltage reference with a resistor from Rx to GND, to enable synchronous rectification MOSFET drive signal	
5	DRAIN	Synchronous Rectification MOSFET Drain Voltage Sense Input	
6	VOUT	Internal Linear Regulator Input	

# **Functional Block Diagram**





## Absolute Maximum Ratings (Note 4)

Symbol	Parameter	Rating	Unit
Vout	Supply Voltage	-0.3 to 28	V
Vdrain	Voltage at DRAIN Pin (Note 5)	-0.7 to 120	V
Vgate	Voltage at GATE Pin	-0.3 to 7	V
Vrx	Voltage at Rx Pin	-0.3 to 7	V
PD	Power Dissipation at T <sub>A</sub> = +25°C	0.6	W
TJ	Operating Junction Temperature	+150	°C
Тѕтс	Storage Temperature	-65 to +150	°C
TLEAD	Lead Temperature (Soldering, 10s)	+300	°C
θja	Thermal Resistance (Junction to Ambient) (Note 6)	200	°C/W
	Human Body Model	5000	V
ESD	Charged Device Model	1500	V

4. Stresses greater than those listed under Absolute Maximum Ratings can cause permanent damage to the device. These are stress ratings only, and Notes: functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to Absolute Maximum Ratings for extended periods can affect device reliability.

5. If -0.7V to -2V negative voltage is applied to DRAIN pin, the period of negative pulse is lower than 0.1µs. 6. Test condition: Device mounted on FR-4 substrate PC board, 1mm width,20mm length, 2oz power lead.

# **Recommended Operating Conditions**

Symbol	Parameter	Min	Max	Unit
Vout	Supply Voltage	0	24	V
T <sub>A</sub>	Ambient Temperature	-40	+85	°C



Symbol	Parameter	Condition	Min	Тур	Max	Unit
Output Voltage ( V	/OUT Pin )					
Vout_on	LDO Enable Threshold When VOUT Rising	_	4.1	4.4	4.7	V
Vout_off	LDO Disable Threshold When VOUT Falling	_	4.0	4.3	4.6	V
Internal Supply Se	ection (VDD Pin )					
Vdd_on	V <sub>DD</sub> Startup Voltage	_	3.55	3.75	3.95	V
V <sub>DD_OFF</sub>	V <sub>DD</sub> UVLO Voltage	_	3.35	3.55	3.75	V
	VDD Regulation Voltage (Include	Vvdd_reg_out, Vout = 12V, Vdrain = 12V	5	5.5	6	V
Vdd_reg	Vvdd_reg_out , Vvdd_reg_drain )	$V_{VDD_{REG_{DRAIN}}}, V_{OUT} = 3V,$ $V_{DRAIN} = 12V$	5.3	5.7	6.5	V
ISTARTUP	VDD Startup Current	Vout = 0, Vdd = Vdd_st - 0.1V	_	140	165	μA
Imax_vdd	Maximum Charging Current from LDO to VDD	Vout = 12V	65	90	120	mA
IQ_VDD	Quiescent Current	V <sub>DD</sub> = 5.0V, V <sub>OUT</sub> = 0		215	255	μA
DRAIN Detection	(DRAIN Pin )					
Von_th	Gate Turn-On Threshold Voltage	Voltage at DRAIN Pin	-225	-200	-175	mV
Von_reg	Gate Turn-On Regulation Voltage	Voltage at DRAIN Pin	-50	-40	-30	mV
Von_off	Gate Turn-Off Threshold Voltage	Voltage at DRAIN Pin	-15	-7	0	mV
ICHG_VDD	Maximum Charging Current from DRAIN to VDD	VDRAIN = 12V	55	75	100	mA
Gate Driver (GAT	E Pin )					
Vgate_h	Gate Output High Voltage	$V_{DD} = 5V$	4.9	—	5	V
Vgate_l_pre	Gate Output Low Voltage before Startup	$V_{DD} = 0, V_{DRAIN} = 8V,$ $I_{LOAD} = 100mA$	_	200	300	mV
Vgate_l	Gate Output Low Voltage after Startup	$V_{DD} \ge V_{DD_ON},$ $I_{LOAD} = 100 \text{mA}$		70	140	mV
ISOURCE_MAX	Maximum GATE Sourcing Current	(Note 7)	_	0.6	—	А
ISINK_MAX	Maximum GATE Sinking Current	(Note 7)	_	3.5		А
RGATE_L	GATE Pull-Low Resistance	$I_{G\_LOAD} = 100 \text{mA}, V_{DD} = 5 \text{V}$	_	0.7	1.4	Ω
Control Section						
ton_min	GATE Minimum On Time	_	0.65	0.9	1.15	μs
toff_min	GATE Minimum Off Time	_	0.35	0.5	0.71	μs
t <sub>D_ON</sub>	Turn-On Delay	$C_{LOAD} = 2.2 nF$	_	30	60	ns
tD_OFF	Turn-Off Propagation Delay	_	_	15	45	ns
Kqs	A Fixed Coefficient Correlated with Voltage-Second Area Criteria	_	0.43	0.5	0.57	mA*µs

**Electrical Characteristics** (@  $V_{OUT} = 5V$ ,  $T_A = -40^{\circ}C < T_A < +85^{\circ}C$ , unless otherwise specified.)

Note: 7. Guaranteed by design.



## **Operation Description**

#### VDD Regulator and UVLO Protection

The V<sub>DD</sub> is supplied by the PLR (Pulse Linear Regulator) from the drain voltage, or LDO from the V<sub>DUT</sub> voltage. When the APR349 operates at a low voltage output condition, a Pulse Linear Regulator is integrated in the controller to provide voltage to the V<sub>DD</sub> pin, in which the system output voltage may drop to 2V. Meanwhile, when the system output voltage is higher than 4.75V, the V<sub>DD</sub> bias voltage generating circuit will change from the PLR to the LDO circuit.

When the V<sub>DD</sub> provides bias voltage for the controller, a capacitor (typical 3.3µF) should be connected between the V<sub>DD</sub> pin and GND pin. The APR349 also has UVLO protection. When V<sub>DD</sub> drops below V<sub>DD\_UVLO</sub>, the IC will stop providing gate drive pulse.

#### GATE Turn-On Logic

The APR349 determines the synchronous rectification MOSFET turn-on moment by monitoring the MOSFET drain-to-source voltage. When the drain voltage is lower than the turn-on threshold voltage  $V_{ON_TH}$ , the IC outputs a positive drive voltage after a turn-on delay time ( $t_{D_ON}$ ). The MOSFET will turn on and the current will transfer from the body diode into the MOSFET's channel.

Moreover, the device also sets an internally fixed, volt-sec product threshold to prevent false turn-on within ringing voltage cycle. Regarding the sensed voltage of the DRAIN pin, the volt-second product above the V<sub>CC</sub> voltage at the primary switch on-time is much higher than the volt-second product of ringing voltage above V<sub>CC</sub>. Therefore, the APR349 can turn on the synchronous rectifier depending on the detected volt-second product of drain voltage above V<sub>CC</sub>. The APR349 has an internal parameter Kqs, which multiplies Rx resistor value to generate volt-second product threshold. Thus, changing the Rx resistor can adjust the volt-second product threshold.

Area2 = Rx \* Kqs

In general, the Area1 and Area3 values depend on system design. Area2 should be set in the middle of Area1 and Area3 to ensure suitable design margins.

Area3 < Rx \* Kqs < Area1

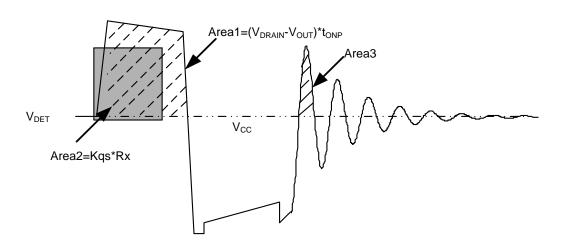


Figure 1. Volt-Second Product Function



### **Operation Description** (continued)

#### GATE Voltage Regulation/Auto-Tracking

The minimum on-time ( $t_{ON\_MIN}$ ) blanking period aims to prevent an accidental turn-off due to interference of ringing noise voltage. When the controlled MOSFET gate turns on, some ringing noise will be generated. Utilizing the minimum on-time timer to blank the  $V_{THOFF}$  comparator, this keeps the SR MOSFET on for at least the minimum on-time. The gate driver can be pulled down quickly to zero when  $V_{DRAIN}$  ringing voltage goes over 2V during this period.

Once the synchronous rectification gate outputs a high-level voltage, the secondary-side current goes through the synchronous rectification MOSFET. The voltage drop on the MOSFET is calculated by  $R_{DS(ON)} \times$  secondary-side current. After minimum turn-on time  $t_{ON\_MIN}$ , the IC continuously monitors  $V_{DRAIN}$  by the DRAIN pin and generates a pull-down current from the MOSFET gate until drain voltage equals -40mV. As shown in Figure 2, the MOSFET drain-to-source voltage would remain around -40mV ( $V_{ON\_REG}$ ) with the secondary-side current decreasing.

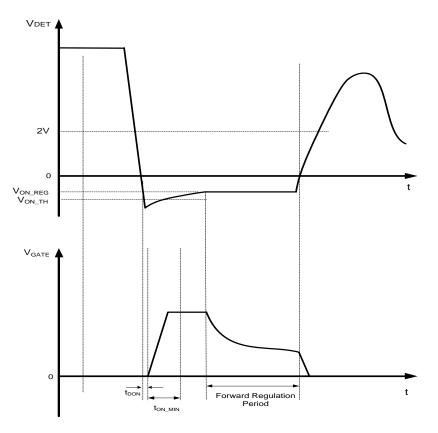


Figure 2. Forward Regulation Operation Principle



### **Operation Description** (continued)

#### Smart-Blanking Control

Because of the resonance oscillation of the primary inductance and equivalent output capacitance of the switching device, the voltage on the MOSFET drain pin has an under damped voltage ring in DCM operation mode. In some special occasions, such as output transition periods or light-load conditions, the SR MOSFET would turn on falsely during the t<sub>OFF</sub> period when the ring of the V<sub>DRAIN</sub> meets the conduction requirement. When the turn–on of the primary MOSFET meets the false turn-on of the SR, the transformer would immediately be shoot through. To avoid such faults, the device has a built-in smart blanking function to prevent continuous false turn-ons. The SR gate driver will blank for a period of 20µs when detecting a false turn-on twice in a period of 4µs in t<sub>OFF</sub> period. Figure 3 shows a smart-blanking control method.

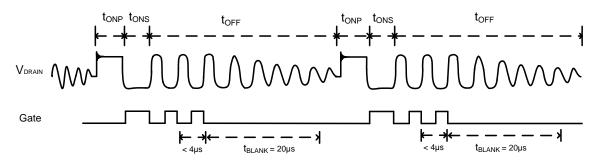


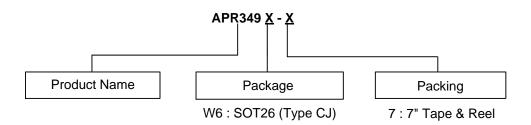
Figure 3. Smart-Blanking Control Method

#### GATE Turn-Off Logic

When V<sub>DRAIN</sub> rises to the turn-off threshold (-7mV), the synchronous rectification MOSFET gate voltage will be pulled down quickly from a low voltage to zero after a very short turn-off delay. For CCM mode, the secondary V<sub>DRAIN</sub> will rise up as the primary MOSFET turns on. Once drain voltage rises to the turn-off threshold, the gate driving signal will immediately shut down.



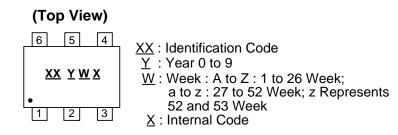
## **Ordering Information**



Package	Temperature Range	Part Number	Identification	7"Tape and Reel		
гаскауе			Code	Quantity	Part Number Suffix	
SOT26 (Type CJ)	-40°C to +85°C	APR349W6-7	D9	3000/Tape and Reel	-7	

# **Marking Information**

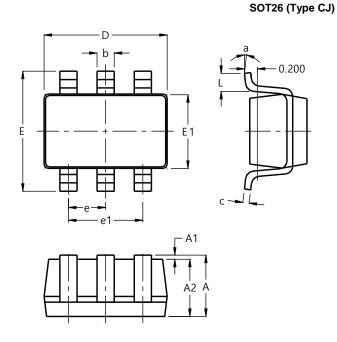
SOT26 (Type CJ)





# **Package Outline Dimensions**

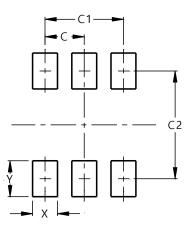
Please see http://www.diodes.com/package-outlines.html for the latest version.



SOT26 (Type CJ)				
Dim	Min	Max	Тур	
Α	1.050	1.250		
A1	0.00	0.10		
A2	1.050	1.150		
b	0.300	0.500		
c	0.100	0.200		
D	2.820	3.020		
Е	2.650	2.950		
E1	1.500	1.700		
е	0.950BSC			
e1	1.800	2.000		
L	0.300	0.600		
а	0°	8°		
All Dimensions in mm				

## **Suggested Pad Layout**

Please see http://www.diodes.com/package-outlines.html for the latest version.



Dimensions	Value (in mm)
С	0.95
C1	1.90
C2	2.40
Х	0.60
Y	1.00

### **Mechanical Data**

- Moisture Sensitivity: MSL Level 1 per JESD22-A113
- Terminals: Finish Matte Tin Plated Leads, Solderable per JESD22-B102 (3)
- Weight: 0.016/0.017 grams (Approximate)

SOT26 (Type CJ)



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