

## Description

The AL5814Q is an 8-terminal adjustable linear LED controller offering excellent temperature stability and output current capability. It works with a wide-input voltage range from 4.5V to 60V. With an external LED driving power device, its internal power dissipation is minimized compared with traditional linear LED drivers. This makes it ideal for medium-to-high current automotive LED lighting applications.

The AL5814Q has internal output drive capability up to 15mA, which enables it to drive external bipolar transistors or MOSFETs. It also provides the capability to drive longer LED chains with low dropout voltage and multiple LED channels.

VSET pin is used to directly set output current feedback level. Using a resistor divider between REF pin and VSET pin, the output current can be set. Additionally, the use of an NTC resistor allows the creation of an accurate and configurable thermal fold-back behavior.

The AL5814Q provides an LED-open detection feature through its VFAULT pin. If VFAULT is brought lower than 2.5V (by any one of the multiple LED channels going open), the AL5814Q's output goes low turning off the external transistors. The device recovers when the open condition is removed. This ability of VFAULT to turn off the external transistors also allows PWM dimming of the LED current by adding PWM control signal on VFAULT pin.

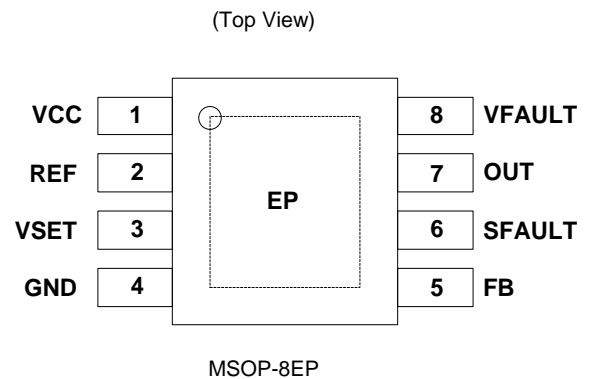
AL5814Q is available in the thermally enhanced MSOP-8EP package and is automotive-compliant, qualified to AEC-Q100 Grade 1, supporting PPAP documentation.

## Features

- Qualified to AEC-Q100 Grade 1
- Wide Input Voltage Range from 4.5V to 60V
- Configurable LED Current Setting
- 5% Reference Voltage Tolerance Across Operating Temperature
- Low Temperature Drift
- 15mA Output Drive Capability for MOSFET or Bipolar Transistor
- LED Open Protection Detected by VFAULT Pin
- PWM Dimming Realized Through VFAULT Pin
- LED Thermal Foldback Configured by VSET
- Overtemperature Protection (OTP)
- Input Undervoltage Lockout
- **Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)**
- **Halogen and Antimony Free. "Green" Device (Note 3)**
- **PPAP Capable (Note 4)**

- Notes:
1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS), 2011/65/EU (RoHS 2) & 2015/863/EU (RoHS 3) compliant.
  2. See <https://www.diodes.com/quality/lead-free/> for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
  3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
  4. Automotive products are AEC-Q100 qualified and are PPAP capable. Automotive, AEC-Q100 and standard products are electrically and thermally the same, except where specified. For more information, please refer to <https://www.diodes.com/quality/>.

## Pin Assignments



## Applications

- Automotive Rear Lamps
- Automotive Interior Lamps
- Automotive Instrumentation Illumination
- Automotive Position Lamps
- Automotive License Plate Illumination

**Typical Application Circuit**

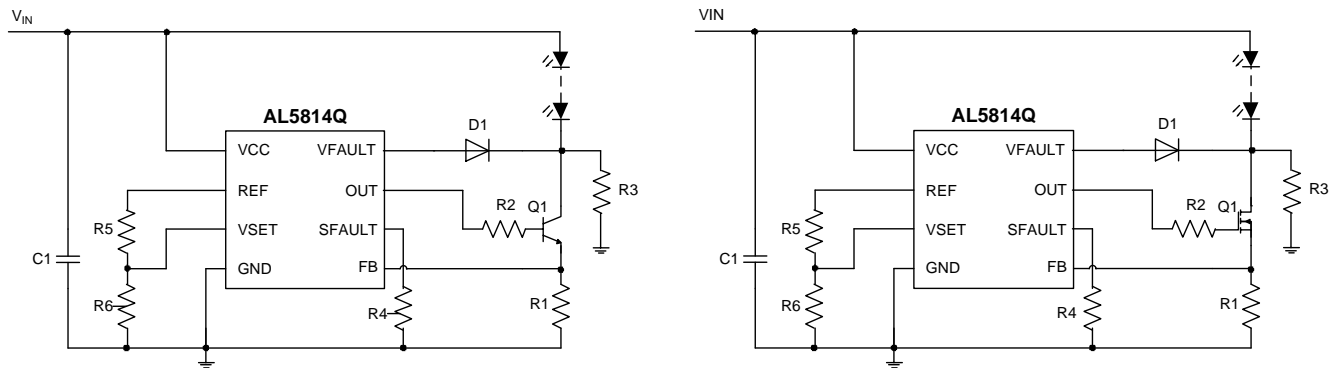


Figure 1

**Pin Descriptions**

Pin Number	Pin Name	Function
1	VCC	Supply input
2	REF	Internal reference voltage. The maximum recommended output current is 50µA and the maximum recommended capacitor connected from REF pin to GND is less than 100pF. A potential divider from REF to VSET can be used to accurately set the output LED current.
3	VSET	LED current setting threshold. The voltage on VSET sets the reference voltage for the FB pin. This pin is pulled down internally if left floating. The capacitance if needed between VSET and GND is recommended to be less than 1nF.
4	GND	Ground
5	FB	Feedback input. When control loop is operating in linear mode, the FB pin will be regulated to the level set by VSET. This pin is pulled up internally if left floating. A capacitor may be added between FB and GND to improve noise rejection. Value needs to be limited to 100pF or less.
6	SFAULT	VCC enable level setting for LED-open protection activation. Used in conjunction with VCC to determine 1. If LED-open is activated and 2. At what VCC voltage it becomes active This function prevents false triggering on power up.
7	OUT	Driving output For 15mA drive current, VCC must be at least 3V greater than V <sub>OUT</sub>
8	VFAULT	Input for LED-open detection. When LED-open detection function is active, if VFAULT is brought lower than approximately 2.5V, the device output will be turned off and will auto-retry driving the output to see if the fault still exists. A PWM dimming function will be realized by adding an open collector/drain signal on this pin.
EP	EP	Exposed Pad (Bottom). Recommended to be connected to a large-area contiguous copper GND plane for effective thermal dissipation. Do not use as the IC's only electrical GND connection.

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**Functional Block Diagram**

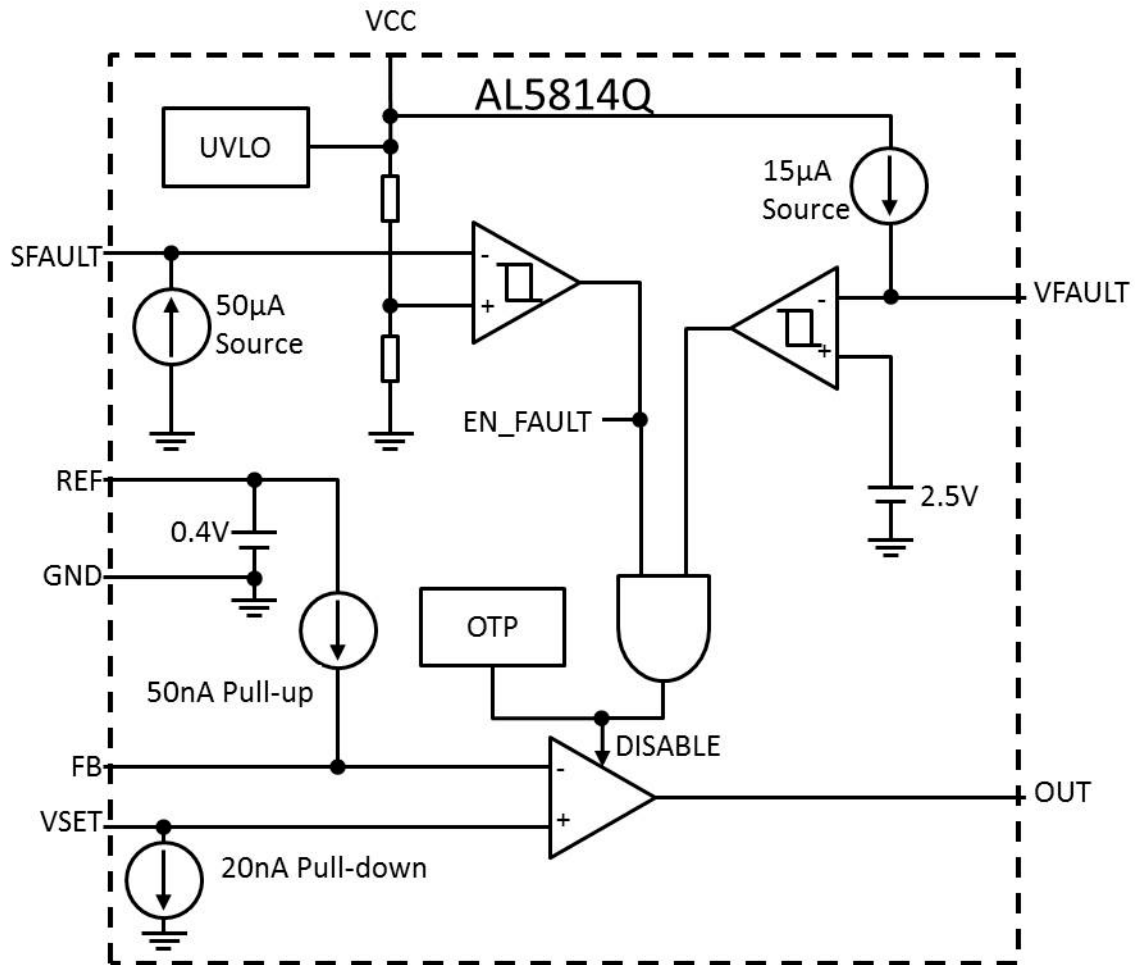


Figure 2

**Absolute Maximum Ratings** (Note 5)

Symbol	Parameter	Rating	Unit
$V_{VCC}$	Supply Voltage Relative to GND	-0.3 to 65	V
$I_{VCC}$	IC Supply Current	18	mA
$V_{VFAULT}$	Input Voltage Relative to GND	-0.3 to 65	V
$V_{REF}, V_{VSET}, V_{SFAULT}, V_{OUT}, V_{FB}$	Input Voltage of REF, VSET, SFAULT, OUT, FB Relative to GND	-0.3 to 6	V
$T_J$	Operating Junction Temperature	-40 to +150	°C
$T_{ST}$	Storage Temperature	-55 to +150	°C

Note: 5. Stresses greater than those listed under *Absolute Maximum Ratings* can cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to *Absolute Maximum Ratings* for extended periods may affect device reliability.

## ESD Ratings

Symbol	Parameter	Rating	Unit	
V <sub>ESD</sub>	Human-Body Model (HMB), Per AEC Q100-002 (Note 6)	±2000	V	
	Charged-Device Model (CDM), per AEC Q100-011	Other Pins		±500
		Corner Pins (1, 4, 5, 8,)		±750

Note: 6. AEC-Q100-002 indicates that HBM stressing shall be accordance with the ANSI/ESDA/JEDEC JS-001 specification

## Recommended Operating Conditions

Symbol	Parameter	Min	Max	Unit
V <sub>VCC</sub>	Supply Voltage Range Relative to GND Pin	4.5	60	V
V <sub>OUT</sub>	OUT Voltage Range (Note 7)	0	3	V
I <sub>OUT</sub>	OUT Pin Current (Note 7)	0	15	mA
V <sub>VSET</sub>	VSET Pin Operating Input Voltage Range	0	0.6	V
T <sub>A</sub>	Operating Ambient Temperature	-40	+125	°C

Note: 7. The maximum source current and drive voltage out of the OUT pin is limited by an internal clamp as well as maximum recommended junction temperature not exceeding 150°C.  
 OUT voltages greater than 3V are supported but at reduced output currents.  
 As VCC reduces below 6V the maximum output voltage supporting output currents above 10mA also reduces; VCC-V<sub>OUT</sub> ≥ 3V.

## Thermal Information (Note 8)

Symbol	Parameter	Rating	Unit
θ <sub>JA</sub>	Junction-To-Ambient Thermal Resistance	56	°C/W
θ <sub>JC</sub>	Junction-To-Case (Top) Thermal Resistance	8	°C/W

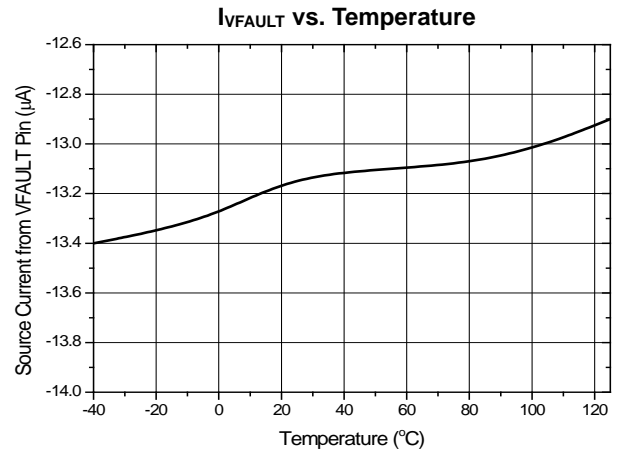
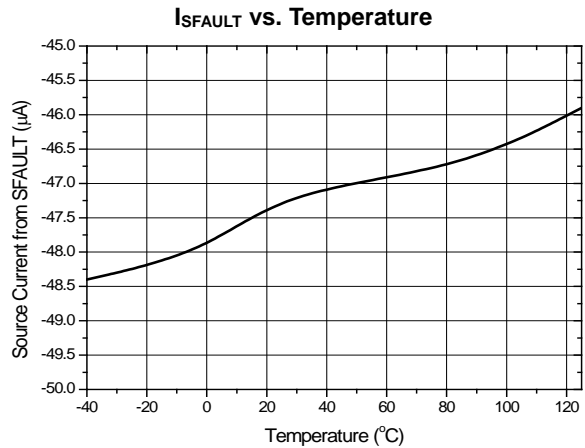
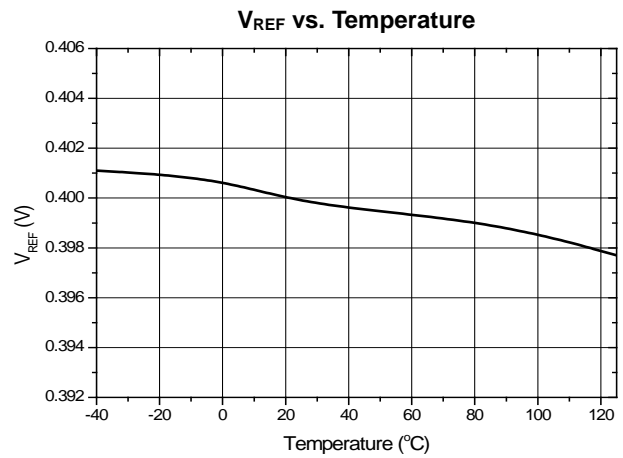
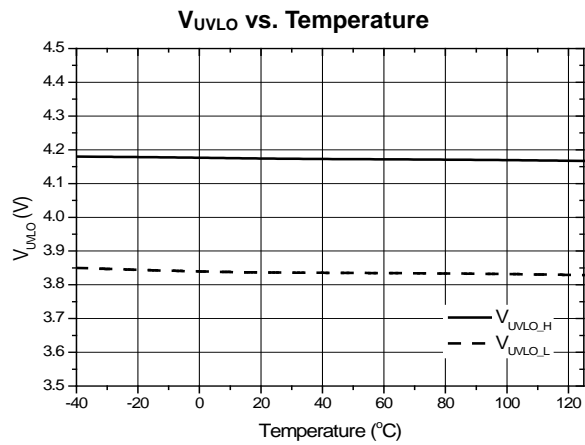
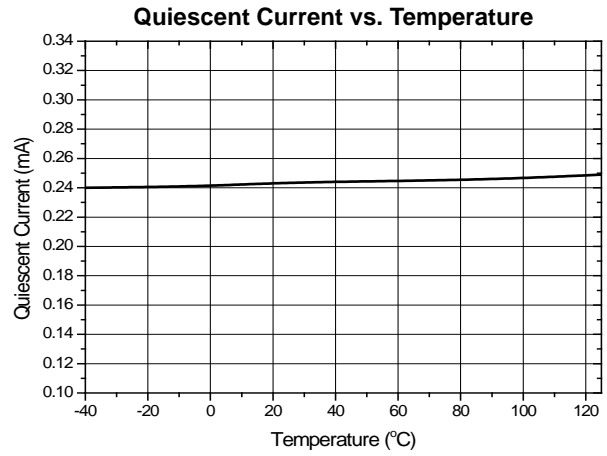
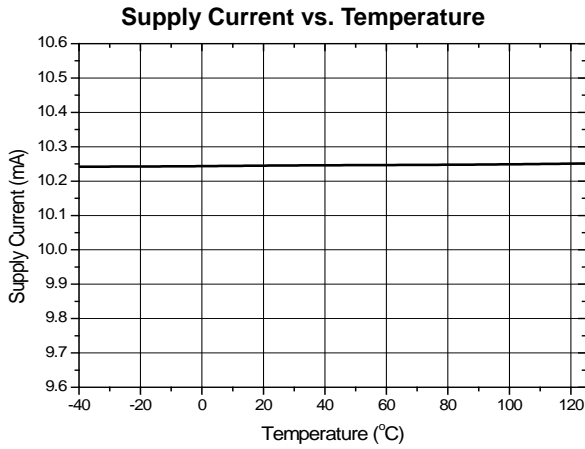
Note: 8. Device mounted on FR-4 PCB (51mm × 51mm 2oz copper); minimum recommended pad layout on top layer and thermal vias to bottom layer ground plane. For better thermal performance, larger copper pad for heat-sink is required.

**Electrical Characteristics** ( $V_{CC} = 12V$ ,  $T_A = -40^{\circ}C$  to  $+125^{\circ}C$ . Typical values are at  $T_A = 25^{\circ}C$ , unless otherwise specified.)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
<b>Power Supply</b>							
$V_{UVLO}$	Undervoltage Lockout Voltage	$V_{IN}$ Rising	—	4.2	4.4	V	
		$V_{IN}$ Falling	3.6	3.85	—		
$I_{CC}$	Supply Current	$V_{CC} = 4.5V$ to $60V$ , $I_{OUT} = 10mA$	—	10.25	11	mA	
$I_Q$	No Load Quiescent Current	$V_{CC} = 4.5V$ to $60V$ , $I_{OUT} = 0$	—	0.25	1	mA	
<b>Reference and Feedback Loop</b>							
$V_{REF}$	Reference Voltage	$V_{CC} = 4.5V$ to $60V$ , $I_{OUT} = 0$ , $I_{REF} = 20\mu A$	0.38	0.4	0.42	V	
$V_{REF\_LINE}$	Reference Voltage Line Regulation	$V_{CC} = 4.5V$ to $19V$	—	0.2	3	mV	
$V_{REF\_LOAD}$	Reference Voltage Load Regulation	$I_{REF} = 0$ to $-50\mu A$	—	1.4	6	mV	
$I_{FB}$	FB Input Bias Current	$V_{FB} = 0.2V$	-200	-125	-70	nA	
$I_{SET}$	SET Input Bias Current	$V_{SET} = 0.6V$	—	25	50	nA	
$V_{OFFSET}$	VSET to $V_{FB}$ Offset	$V_{SET} = 0.2V$ , $V_{FB} = V_{OUT}$	—	—	7	mV	
<b>Fault Detection and Protection</b>							
$I_{SFAULT}$	Source Current from SFAULT Pin	$V_{CC} = 4.5V$ to $60V$	-58	-50	-41	$\mu A$	
$I_{VFAULT}$	Source Current from VFAULT Pin	$V_{CC} = 4.5V$ to $60V$	-21	-15	-9	$\mu A$	
$V_{VFAULT}$	VFAULT Pin Threshold Voltage	$V_{CC} = 4.5V$ to $60V$ , $V_{VFAULT}$ Falling	2.36	2.5	2.64	V	
$V_{VFAULT\_HYS}$	VFAULT Hysteresis	—	—	100	—	mV	
$t_{VF\_OUTR}$	VFAULT to OUT Rising Delay	—	—	2	—	$\mu s$	
$t_{VF\_OUTF}$	VFAULT to OUT Falling Delay	—	—	1	—		
$T_{SHDN}$	Thermal Shutdown	—	—	+160	—	$^{\circ}C$	
$T_{HYS}$	Thermal Shutdown Hysteresis	—	—	+30	—	$^{\circ}C$	
<b>Output Driver Error Amplifier</b>							
$I_{OUTSOURCE}$	Maximum Source Current $V_{SET} - V_{FB} = 10mV$	$V_{OUT} = 1V$ , $V_{CC} = 4.5V$ to $60V$	$T_A = 25^{\circ}C$	—	—	-15	mA
			$-40^{\circ}C \leq T_A \leq 125^{\circ}C$	—	—	-13	
		$V_{OUT} = 2V$ , $V_{CC} = 4.5V$ to $60V$	$T_A = 25^{\circ}C$	—	—	-11	
			$-40^{\circ}C \leq T_A \leq 125^{\circ}C$	—	—	-9	
$V_{OUT} = 3V$ , $V_{CC} = 6.0V$ to $60V$	$T_A = 25^{\circ}C$	—	—	-15			
	$-40^{\circ}C \leq T_A \leq 125^{\circ}C$	—	—	-13			
$I_{OUTSINK}$	Maximum Sink Current	$V_{CC} = 12V$ , $V_{OUT} = 4V$ , $V_{VFAULT} > 2.7V$ , $V_{SET} - V_{FB} = -50mV$	20	—	—	$\mu A$	
		$V_{CC} = 12V$ , $V_{OUT} = 4V$ , VFAULT enabled by SFAULT $V_{SET} - V_{FB} = -50mV$ , $V_{VFAULT} < 2.3V$	0.3	—	—	mA	
$G_m$	Trans-Conductance of Error Amplifier	$V_{SET} - V_{FB} = 5mV$ , Sourcing Current	—	4	—	A/V	
$BW$	Bandwidth	—	—	50	—	kHz	
$V_{OUTMAX}$	Maximum Output Voltage	$V_{CC} \geq 6V$ , $I_{OUT} = -1mA$	4	—	—	V	
$V_{OUTMIN}$	Minimum Output Voltage	$V_{CC} = 12V$ , $I_{OUT} = 0.1mA$ , $V_{FB} = 250mV$	—	—	300	mV	
<b>Thermal Shutdown</b>							
$T_{SHDN}$	Thermal Shutdown	—	—	+160	—	$^{\circ}C$	
$T_{HYS}$	Thermal Shutdown Hysteresis	—	—	+30	—	$^{\circ}C$	

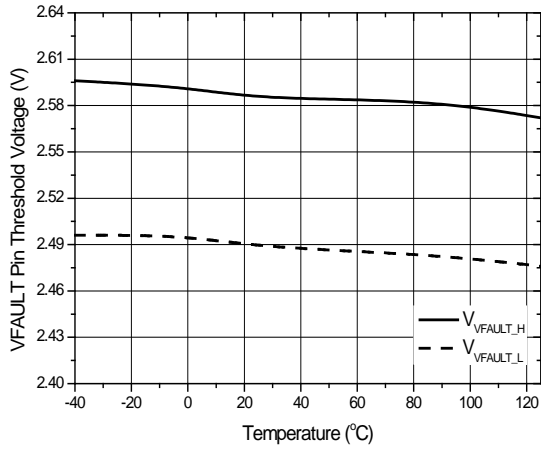
**Typical Performance Characteristics** ( $V_{CC} = 12V$ ,  $T_A = +25^\circ C$ , unless otherwise specified.)

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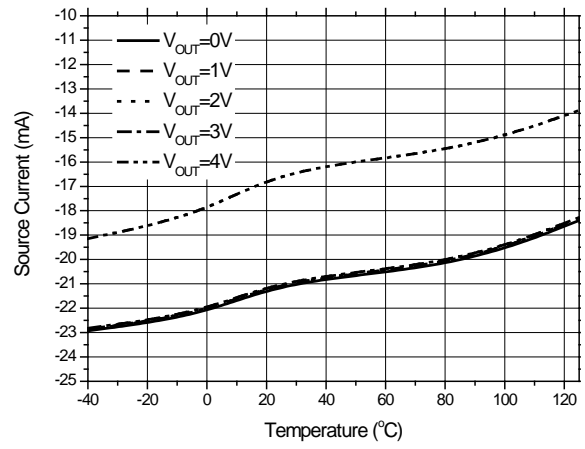


**Typical Performance Characteristics** (cont.) ( $V_{CC} = 12V$ ,  $T_A = +25^\circ C$ , unless otherwise specified.)

**$V_{FAULT}$  vs. Temperature**

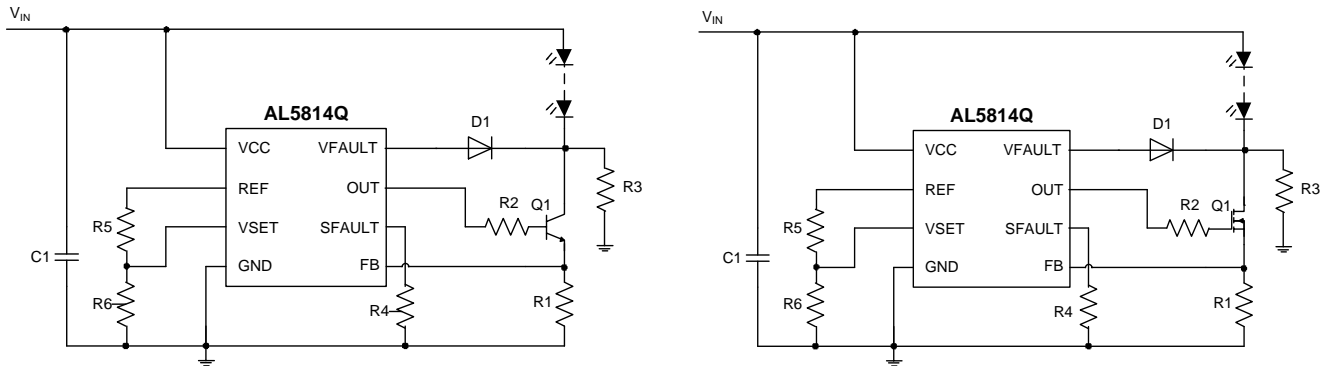


**Source Current vs. Temperature**



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**Application Information**



**Figure 3. Typical Application Circuits**

**Output Drive**

Figure 3 shows the typical output drive configuration. The feedback loop regulates the current through the external LEDs. The voltage across the external sense resistor (R1) is fed to the FB pin for sensing. When the voltage exceeds VSET voltage the OUT goes lower, decreasing the drive to the external transistor.

The output current can be set as:

$$I_{LED} = \frac{V_{VSET}}{R_1}$$

Where  $I_{LED}$  is the desired LED current,  $V_{VSET}$  is determined by R5 and R6 resistor divider, and  $R_1$  is the sense resistor.

The power in the sense resistor is calculated as:

$$Power = V_{VSET} * I_{LED}$$

Where  $V_{VSET} = VSET$  voltage, and  $I_{LED}$  is the desired LED string current. For most cases, a standard 1/4W resistor works. Similarly, the external transistor's power dissipation must also be considered to prevent thermal damage to the transistor, which can further damage the LED controller IC.

**Power Consideration of the NMOS or BJT Device**

The power rating of the transistor (either BJT or NMOS) used in the typical application circuit is important. A correctly mounted transistor used in a typical application can dissipate power of up to 2W. To calculate power dissipation, first calculate the voltage drop across the transistor as follows:

$$V_{DS} = V_{CC} - V_{LED} - V_{VSET}$$

Then calculate the power dissipation requirement:

$$Power = V_{DS} * I_{LED}$$

If power dissipation is higher than the transistor package and layout can dissipate then a higher power dissipation transistor must be selected and/or use a better PCB layout.

**Multiple LED Strings in Parallel**

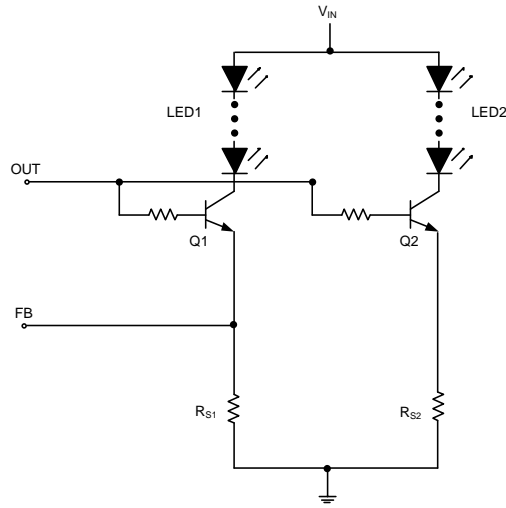
The AL5814Q can drive more than one channel of LED strings. As shown in Figure 4, the sense voltage of two channels' (or more) output current can be implemented by connecting the voltage of one sense resistor to the FB pin.

By utilizing the same type transistors, sense resistors and series base resistors, the currents in all channels match.

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**Application Information** (continued)



**Figure 4. Two LED Strings in Parallel**

The output current can be set as:

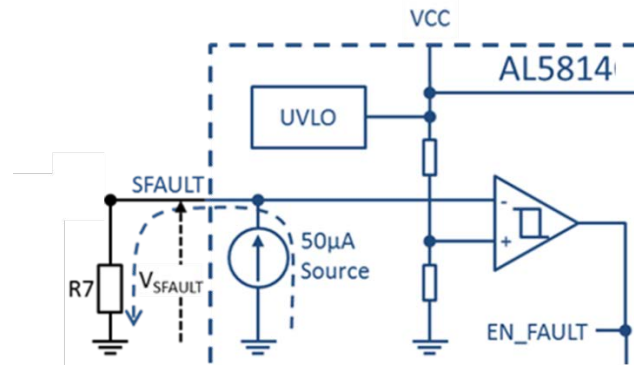
$$I_{LED1} = I_{LED2} = \dots = V_{FB}/R_{S1}$$

Where  $I_{LEDx}$  is the desired LED current,  $V_{FB}$  is  $V_{VSET}$ , and  $R_{S1}$  is the sense resistor. To keep the LED currents as equal as possible, transistors Q1 and Q2 must be matched, and  $R_{S1}$  and  $R_{S2}$  must be matched.

**Open Load Detection**

**SFAULT Setup:**

The AL5814Q can be configured to detect the presence of the LEDs on the external output transistor going open circuit. This function is activated by the SFAULT pin. A 50µA current source from SFAULT creates a voltage ( $V_{SFAULT}$ ) across an external resistor (R7), which is compared to potential divided down  $V_{CC}$  voltage. See Figure 5 (blue components are internal to the AL5814Q).



**Figure 5. SFAULT Configuration**

Once  $V_{CC}$  rises above the set voltage on the VSFAULT pin, the EN\_FAULT becomes high, and LED-open detection becomes functional. This avoids false triggering of open-load protection during startup and powerdown procedure.

Supply voltage where LED-open detection becomes operational is set by the following equation:

$$V_{CC} = 11 \times 50 \times 10^{-6} \times R7$$

**Application Information** (cont.)

The table below shows approximate supply voltages where the LED-open detection becomes active by different resistor values.

V <sub>CC</sub> (V)	R7 (kΩ)
5.01	9.1
5.50	10
6.60	12
8.25	15
9.90	18
11.55	21

If SFAULT is shorted to GND, by default LED-open fault detection is automatically entered. If SFAULT is left open, LED-open detection is inhibited.

**VFAULT**

The AL5814Q detects open conditions on the collector/drain of the external transistor driving the LEDs using the VFAULT pin (see Figure 6). The VFAULT pin has a 14μA current source generated out of this pin.

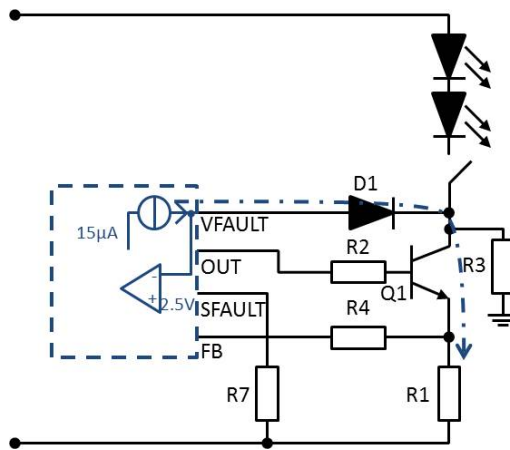


Figure 6. LED-Open Detection Using VFAULT

If the LED string is open, VFAULT pin is pulled down by the power bipolar or MOSFET below its internal 2.5V threshold. This condition triggers an output disable condition causing OUT to go low, turning off the external MOSFET/BJT. A resistor (R3, 100k) is required to keep the VFAULT signal low during a fault condition. A 100kΩ keeps this node low. When the fault condition is fixed, and VFAULT pin rises above 2.63V, the device operates normally.

**Feedback Loop**

The device has internal compensation; therefore, it is not recommended that any components are added to the feedback loop.

**LED Thermal Compensation**

The LED current is set by the reference voltage at the emitter/drain voltage of the LED driving transistor. This reference voltage is determined by the resistor divider between REF and VSET. The LED driver's LED current behavior under different ambient or LED temperatures can be configured by using NTC for R6 shown in Figure 7.

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**Application Information** (cont.)

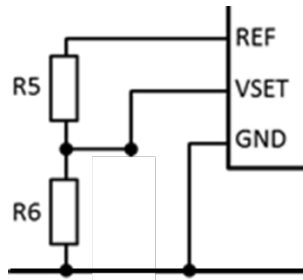


Figure 7

In the example shown in Figure 8, the resistor network is comprised of one NTC and a resistor. When the temperature of the detecting point is rising, the NTC resistance decreases and makes the voltage at VSET decrease. Consequentially, the output current decreases to prevent the system from overheating.

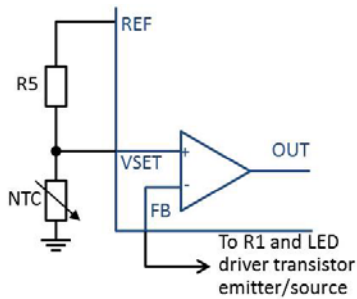
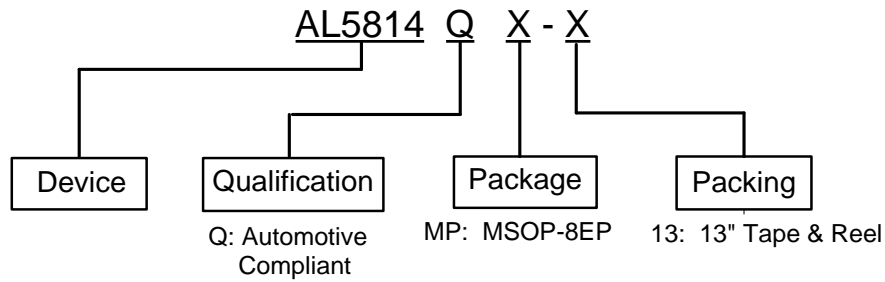


Figure 8. Thermal Fold Back Circuit Basing on NTC

**Thermal Protection**

The AL5814Q has internal overtemperature protection (OTP). When the junction temperature is over +160°C, the IC shuts down. When the junction temperature drops by +30°C, the IC turns back on.

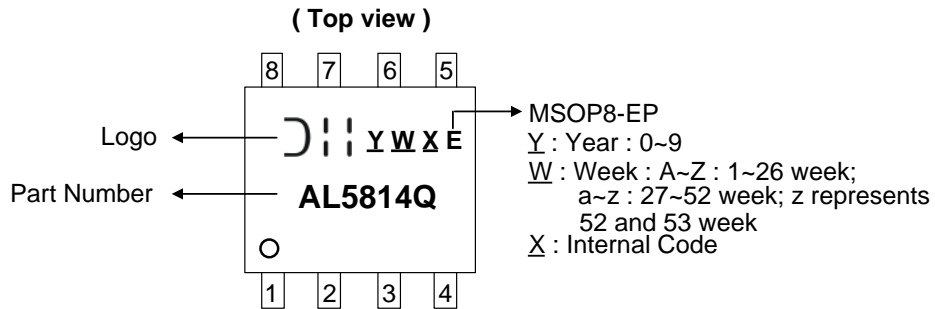
**Ordering Information** (Note 9)



Part Number	Package Code	Packaging	13" Tape and Reel	
			Quantity	Part Number Suffix
AL5814QMP-13	MP	MSOP-8EP	2500	-13

Note: 9. For packaging details, go to our website at <https://www.diodes.com/design/support/packaging/diodes-packaging/>.

**Marking Information**



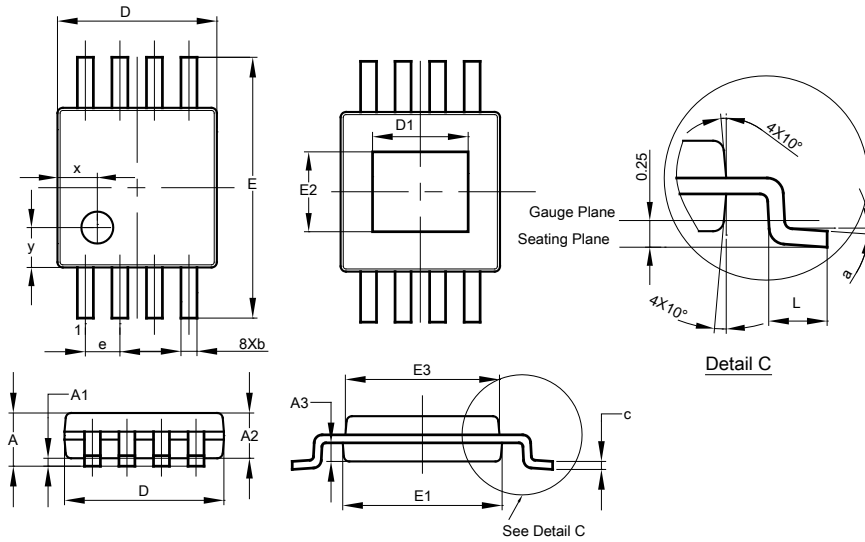
Device	Package
AL5814QMP-13	MSOP-8EP

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**Package Outline Dimensions**

Please see <http://www.diodes.com/package-outlines.html> for the latest version.

**MSOP-8EP**

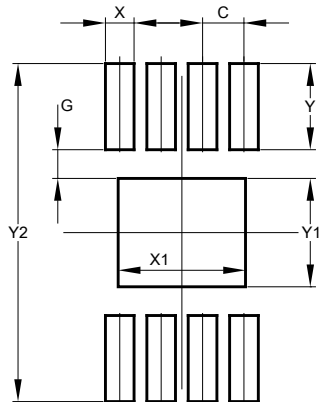


MSOP-8EP			
Dim	Min	Max	Typ
A	-	1.10	-
A1	0.05	0.15	0.10
A2	0.75	0.95	0.86
A3	0.29	0.49	0.39
b	0.22	0.38	0.30
c	0.08	0.23	0.15
D	2.90	3.10	3.00
D1	1.60	2.00	1.80
E	4.70	5.10	4.90
E1	2.90	3.10	3.00
E2	1.30	1.70	1.50
E3	2.85	3.05	2.95
e	-	-	0.65
L	0.40	0.80	0.60
a	0°	8°	4°
x	-	-	0.750
y	-	-	0.750
<b>All Dimensions in mm</b>			

**Suggested Pad Layout**

Please see <http://www.diodes.com/package-outlines.html> for the latest version.

**MSOP8-EP**



Dimensions	Value (in mm)
C	0.650
G	0.450
X	0.450
X1	2.000
Y	1.350
Y1	1.700
Y2	5.300

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