

DIO69309

High-Efficiency 3 A, 24 V Input Synchronous Step Down Converter

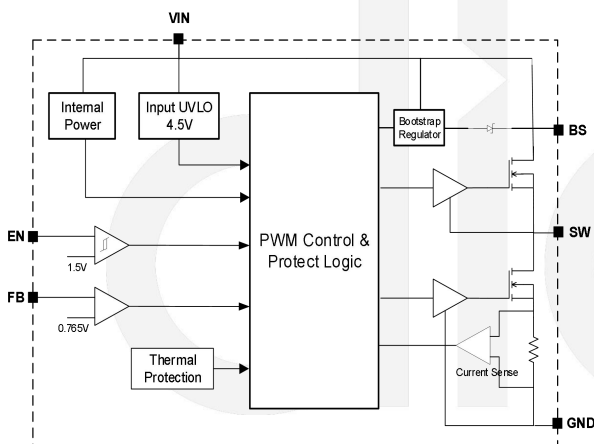
Features

- Low $R_{DS(ON)}$ for internal switches (top/bottom) 80m Ω /40m Ω , 3.0 A
- 4.5-24 V input voltage range
- High-Efficiency Synchronous-Mode
- Internal soft-start limits the inrush current
- Over Current protection
- Output short circuit protection with hiccup mode
- Thermal shutdown
- Available in TSOT23-6 packages

Applications

- Portable Navigation Devices
- Set Top Boxes
- Portable TVs
- LCD TVs

Function Block



Descriptions

The DIO69309 is a high-efficiency, high-frequency synchronous step-down DC-DC regulator ICs capable of delivering up to 3 A output currents. The DIO69309 family operates over a wide input voltage range from 4.5 V to 24 V and integrates the main switch and synchronous switch with very low $R_{DS(ON)}$ to minimize the conduction loss. The pure COT architecture with Pseudo fixed switching frequency operation provides a fast transient response and eases loop stabilization. Protection features include over-current protection and thermal shutdown.

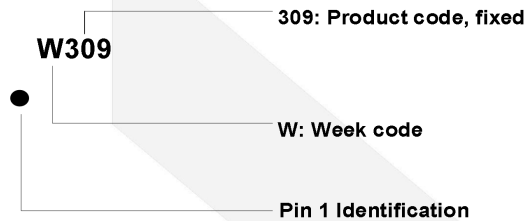
The DIO69309 always operates in continuous conduction mode, which reduces the output ripple voltage in light load compared to discontinuous conduction mode. The DIO69309 is stable with extremely low ESR, and high capacitance.

The DIO69309 requires a minimal number of readily-available, standard, external components and is available in a space-saving TSOT23-6 package.

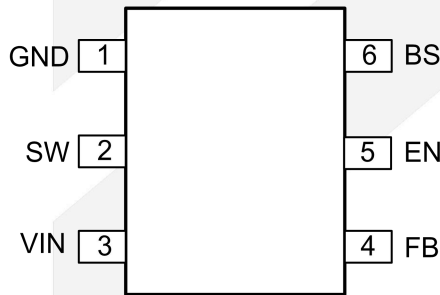
Ordering Information

Order Part Number	Top Marking		T _A	Package	
DIO69309TST6	W309	Green	-40 to 85°C	TSOT23-6	Tape & Reel, 3000

Marking Definition



Pin Assignments



TSOT23-6

Figure 1 Pin Assignment (Top View)

Pin Definitions

Pin Name	Description
GND	Power Ground
SW	Inductor pin. Connect this pin to the switching node of inductor.
VIN	Power Input
FB	Output Feedback Pin. Connect this pin to the center point of the output resistor divider (as shown in Figure 1) to program the output voltage: $V_{OUT}=0.765*(1+R1/R2)$.
EN	Enable control. Pull high to turn on. Do not float.
BS	Bootstrap. Connect a capacitor and a resistor between SW and BS pins to form a floating supply across the high-side switch driver. Recommend to use 0.1μF BS capacitor.



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Absolute Maximum Ratings

Stresses beyond those listed under the Absolute Maximum Rating table may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other condition beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Parameter	Rating	Unit
Supply Voltage (V+ – V-)	28	V
EN, SW Voltage	$V_{IN}+0.3$	V
FB Voltage	6	V
BS Voltage	SW+6	V
Package Thermal Resistance	θ_{JA}	°C/W
	θ_{JC}	
Storage Temperature Range	-65 to 150	°C
Junction Temperature Range	150	°C
Lead Temperature Range	260	°C

Recommend Operating Conditions

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. DIOO does not recommend exceeding them or designing to absolute maximum ratings.

Parameter	Rating	Unit
Supply Voltage	4.5 to 24	V
Junction Temperature Range	-40 to 125	°C
Ambient Temperature Range	-40 to 85	°C



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Electrical Characteristics

$V_{IN} = 12\text{ V}$, $V_{OUT} = 1.2\text{ V}$, $L = 1.5\ \mu\text{H}$, $C_{OUT} = 47\ \mu\text{F}$, $T_j = -40^\circ\text{C}$ to 125°C , $I_{OUT} = 1\text{ A}$ unless otherwise specified.

Symbol	Parameter	Test Conditions	Min	Typ	Max	Unit
V_{IN}	Input Voltage Range		4.5		24	V
I_Q	Quiescent Current	$I_{OUT} = 0$, $V_{FB} = V_{REF} \cdot 105\%$		700	1000	μA
I_{SHDN}	Shutdown Current	EN=0		5	10	μA
V_{REF}	Feedback Reference Voltage	$T_j = -40^\circ\text{C}$ to 125°C	0.745	0.765	0.780	V
		$T_j = 0^\circ\text{C}$ to 125°C	0.750	0.765	0.780	
I_{FB}	FB Input Current	$V_{FB} = 3.3\text{ V}$	-50		50	nA
$R_{DS(ON)}$	Top FET R_{ON}	$T_A = 25^\circ\text{C}$		80		m Ω
$R_{DS(ON)}$	Bottom FET R_{ON}	$T_A = 25^\circ\text{C}$		40		m Ω
$I_{LIM}^{(1)}$	Low side power FET current limit		3.3	3.9	4.5	A
V_{ENH}	EN Rising Threshold		1.5			V
V_{ENL}	EN Falling Threshold				0.4	V
V_{UVLO}	V_{IN} Under-Voltage Unlock Threshold, Rising		3.65	4.05	4.45	V
	Hysteresis V_{IN} voltage		0.3	0.4	0.5	V
f_{SW}	Switching Frequency	$V_{IN} = 12\text{ V}$; $V_{OUT} = 3.3\text{ V}$; Load = 3 A	600	700	800	kHz
	Min ON Time			140		ns
	Min OFF Time			90	140	ns
	Maximum Duty Cycle			90%		
T_{SS}	Soft Start Time			1		ms
T_{SD}	Thermal Shutdown Temperature			148		$^\circ\text{C}$
T_{HYS}	Thermal Shutdown Hysteresis			20		$^\circ\text{C}$

(1) Not production tested.

Specifications subject to change without notice.

Typical Application

DIO69309 4.5 V to 24 V Input, 1.05 V Output Converter

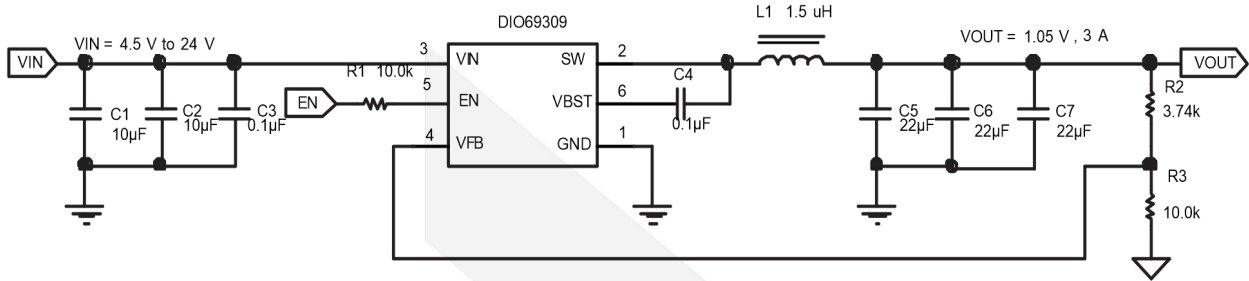
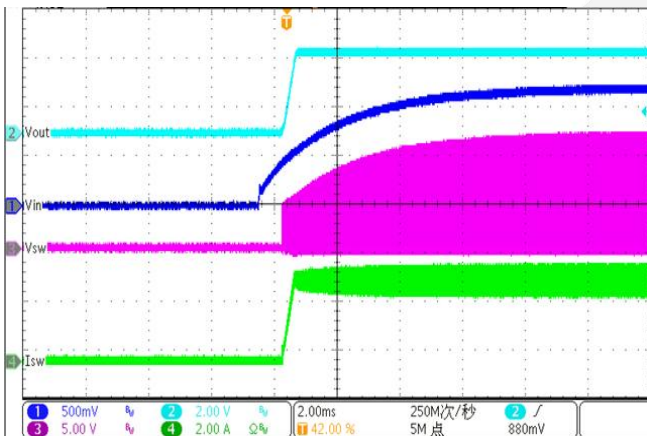


Figure 2. DIO69309 1.05V/3A Reference Design

Typical Performance Characteristics

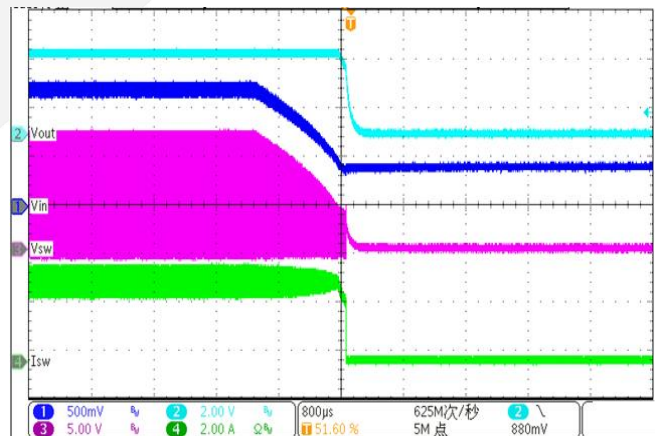
Start up from VIN

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=1 Ω)



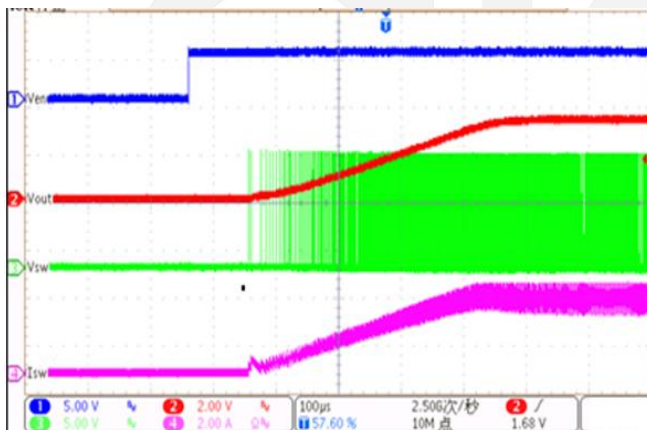
Shut down from VIN

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=1 Ω)



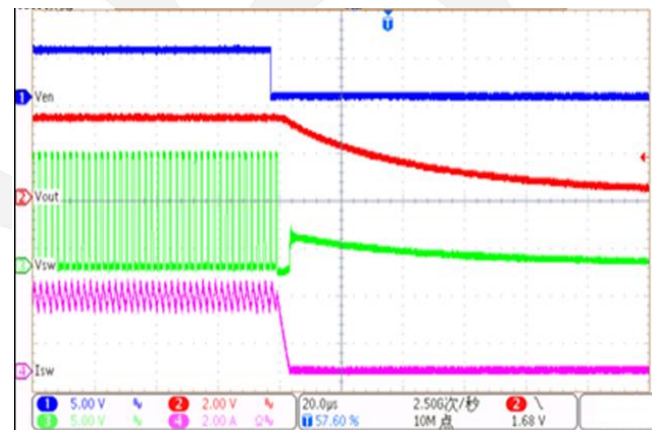
Start up from Enable

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=1 Ω)



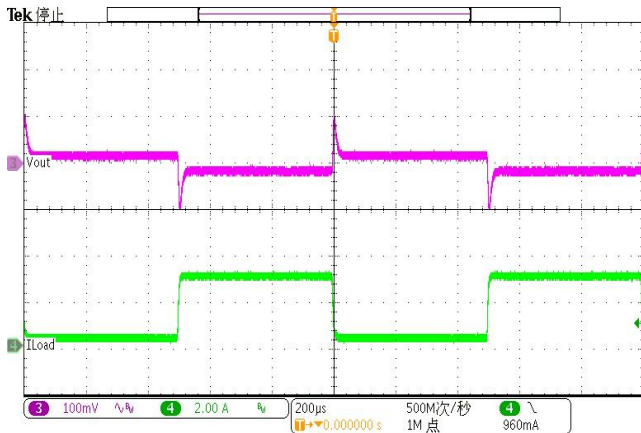
Shut down from Enable

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=1 Ω)



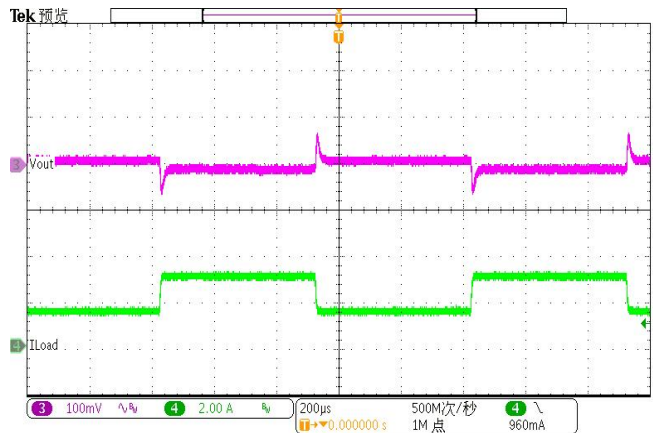
Load transient

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=0.3-3 A)



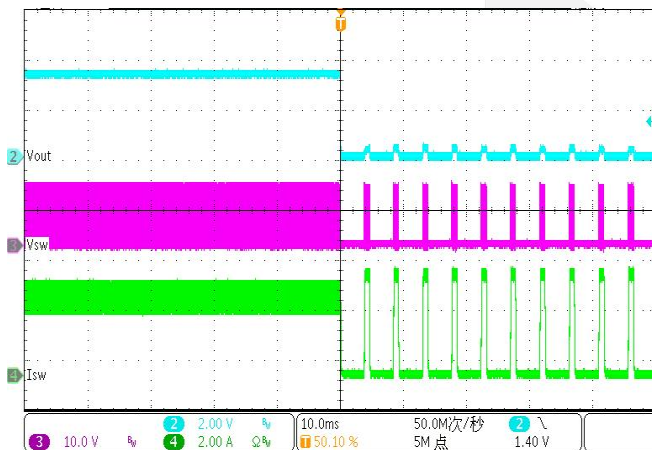
Load transient

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=1.5-3 A)



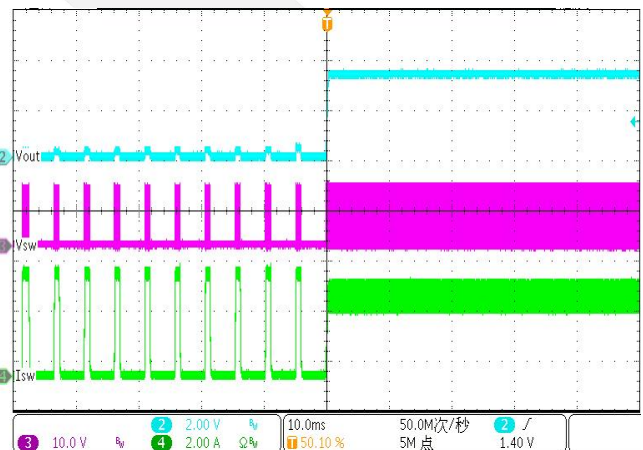
Short Circuit Protection

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=3 A)



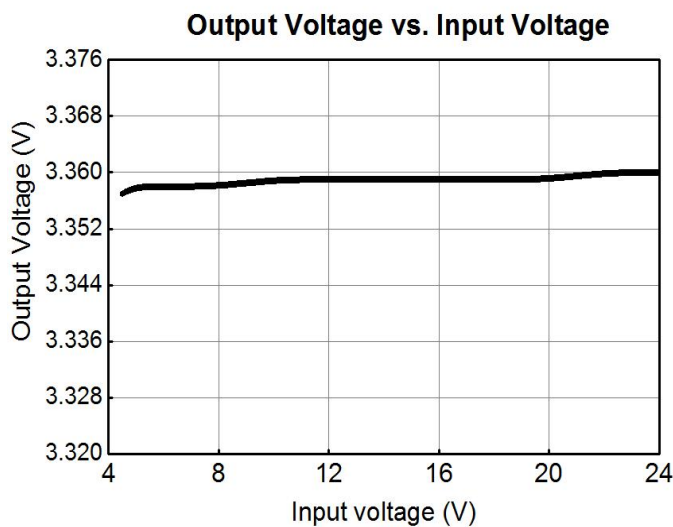
Short Circuit Recovery

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, Load=3 A)



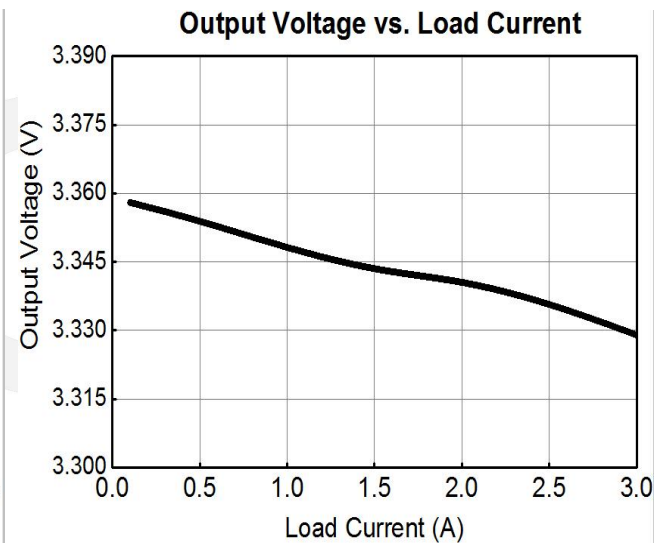
Line Regulation

($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$, $I_{Load}=10\text{ mA}$)

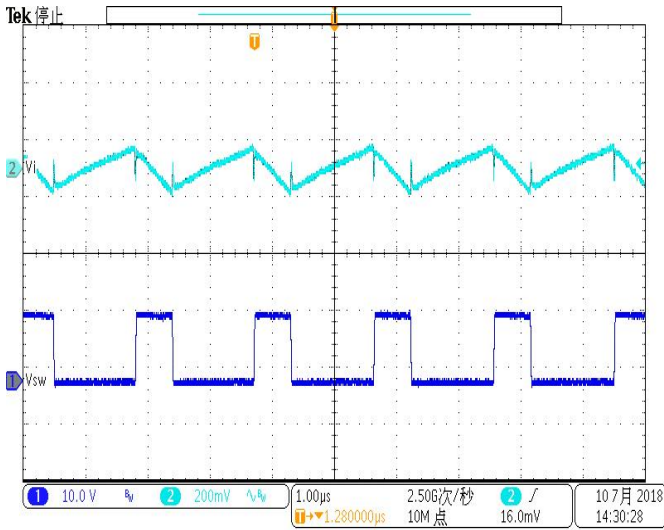


Load Regulation

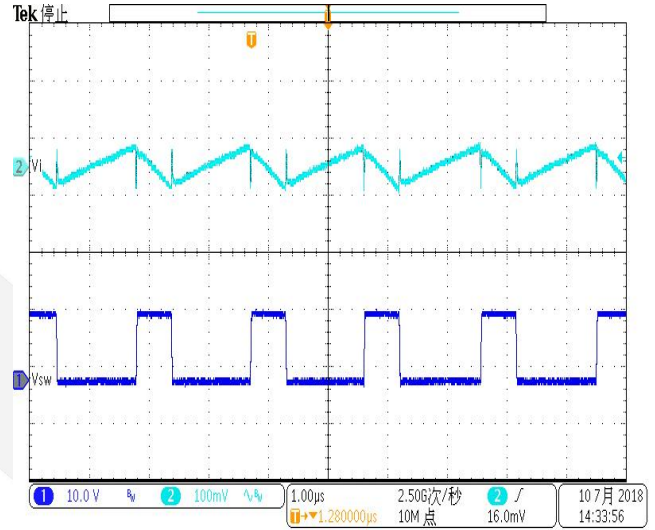
($V_{IN}=12\text{ V}$, $V_{OUT}=3.3\text{ V}$)



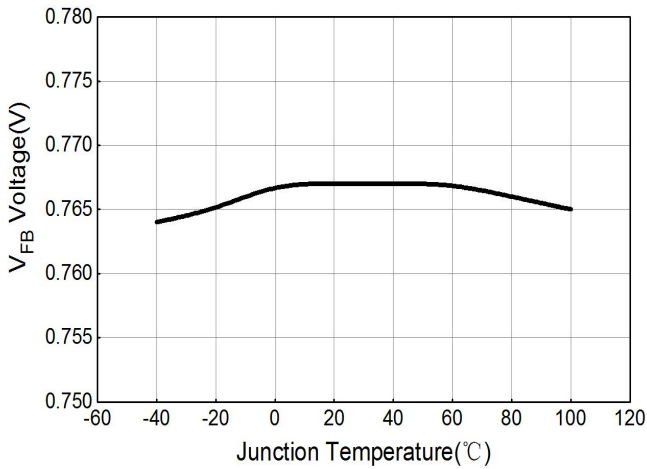
VIN Ripple
($V_{IN}=12\text{ V}$, Load=3 A, $C_{IN}=22\text{ }\mu\text{F}$)



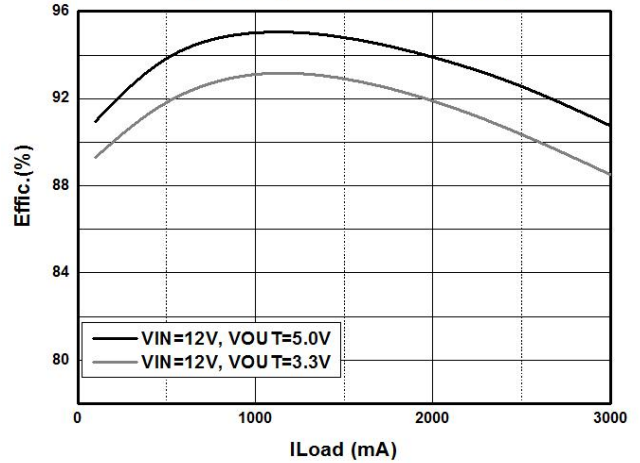
VIN Ripple
($V_{IN}=12\text{ V}$, Load=3 A, $C_{IN}=44\text{ }\mu\text{F}$)



V_{FB} vs. Temperature



Efficiency vs Output Current



Application Information

The DIO69309 is a synchronous buck regulator IC that integrates the pure COT control, and top and bottom switches on the same die to minimize the switching transition loss and conduction loss. With ultra-low $R_{DS(ON)}$ power switches and proprietary pure COT control, this regulator IC can achieve the highest efficiency and the highest switch frequency simultaneously to minimize the external inductor and capacitor size, and thus achieve the minimum solution footprint.

Table 1. Design Parameters

PARAMETER	VALUE
Input voltage range	4.5 V to 24 V
Output voltage	1.05 V
Output current	3 A
Output voltage ripple	20 mVpp

Table 2. Recommended Component Values

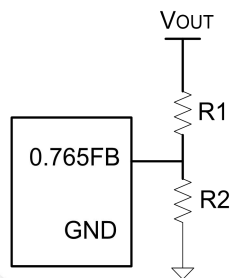
OUTPUT VOLTAGE (V)	R2 (kΩ)	R3 (kΩ)	L1 (μH)			C5 + C6 (μF)
			MIN	TYP	MAX	
1	3.09	10.0	1	1.5	2.2	44 - 66
1.05	3.74	10.0	1	1.5	2.2	44 - 66
1.2	5.76	10.0	1	1.5	2.2	44 - 66
1.5	9.53	10.0	1	1.5	2.2	44 - 66
1.8	13.7	10.0	1	1.5	2.2	44 - 66
2.5	22.6	10.0	1.5	2.2	3.3	44 - 66
3.3	33.2	10.0	1.5	2.2	3.3	44 - 66
5	54.9	10.0	1.5	3.3	4.7	44 - 66

Because of the high integration in the DIO69309 IC, the application circuit based on this regulator IC is rather simple. Only input capacitor C_{IN} , output capacitor C_{OUT} , output inductor L and feedback resistors (R1 and R2) need to be selected for the targeted applications specifications.

Feedback resistor dividers R1 and R2

Choose R1 and R2 to program the proper output voltage. To minimize the power consumption under light loads, it is desirable to choose large resistance values for both R1 and R2. A value of between 10k and 1 M is highly recommended for both resistors. If V_{out} is 3.3V, $R_1=40.2$ k is chosen, then R2 can be calculated to be 12 k.

$$R_2 = \frac{0.765V}{V_{OUT} - 0.765V} R_1$$



Current Protection

The output over-current limit (OCL) is implemented using a cycle-by-cycle valley detect control circuit. The switch current is monitored during the OFF state by measuring the low-side FET drain to source voltage. This voltage is proportional to the switch current. To improve accuracy, the voltage sensing is temperature compensated. During the on time of the high-side FET switch, the switch current increases at a linear rate determined by V_{in} , V_{out} , the on-time and the output inductor value.

During the on time of the low-side FET switch, this current decreases linearly. The average value of the switch current is the load current I_{out} . If the monitored current is above the OCL level, the converter maintains low-side FET on and delays the creation of a new set pulse, even if the voltage feedback loop requires one, until the current level becomes OCL level or lower. In subsequent switching cycles, the on-time is set to a fixed value and the current is monitored in the same manner. If the over current condition exists consecutive switching cycles, the internal OCL threshold is set to a lower level, reducing the available output current. When a switching cycle occurs where the switch current is not above the lower OCL threshold, the counter is reset and the OCL threshold is returned to the higher value.

There are some important considerations for this type of over-current protection. The load current is higher than the over-current threshold by one half of the peak-to-peak inductor ripple current. Also, when the current is being limited, the output voltage tends to fall as the demanded load current may be higher than the current available from the converter. This may cause the output voltage to fall. When the VFB voltage falls below the UVP threshold voltage, the UVP comparator detects it. And then, the device will shut down after the UVP delay time (typically 14 μ s) and re-start after the hiccup time (typically 8ms).

When the over current condition is removed, the output voltage returns to the regulated value.

UVLO Protection

Under voltage lock-out protection (UVLO) monitors the device input voltage. When the voltage is lower than the UVLO threshold voltage, the device is shut off. This protection is non-latching.

Thermal Shutdown

The device monitors the temperature of itself. If the temperature exceeds the threshold value (typically 155°C), the device is shut off. This is a non-latch protection.

Input capacitor C_{IN}

This ripple current through input capacitor is calculated as:

$$I_{CIN_RMS} = I_{OUT} \cdot \sqrt{D(1-D)}$$

This formula has a maximum at $V_{IN}=2V_{OUT}$ condition, where $I_{CIN_RMS}=I_{OUT}/2$. This simple worst-case condition is commonly used for DC/DC design.

To minimize the potential noise problem, place a typical X5R or better grade ceramic capacitor really close to the IN and GND pins. Care should be taken to minimize the loop area formed by C_{IN} , and IN/GND pins. In this case, a 10 μ F low ESR ceramic capacitor is recommended.

Output capacitor C_{OUT}

The output capacitor is selected to handle the output ripple noise requirements. Both steady state ripple and transient requirements must be taken into consideration when selecting this capacitor. For the best performance, it is recommended to use X5R or better grade ceramic capacitor greater than 22 μ F capacitance.

Output inductor L:

There are several considerations in choosing this inductor.

- 1) Choose the inductance to provide the desired ripple current. It is suggested to choose the ripple current to be about 40% of the maximum output current. The inductance is calculated as:

$$L = \frac{V_{OUT}(1 - V_{OUT} / V_{IN,MAX})}{F_{SW} \times I_{OUT,MAX} \times 40\%}$$

where F_{sw} is the switching frequency and $I_{OUT,MAX}$ is the maximum load current. The DIO69309 regulator IC is quite tolerant of different ripple current amplitudes. Consequently, the final choice of inductance can be slightly off the calculation value without significantly impacting the performance.

- 2) The saturation current rating of the inductor must be selected to be greater than the peak inductor current under full load conditions.

$$I_{SAT,MIN} > I_{OUT,MAX} + \frac{V_{OUT}(1 - V_{OUT} / V_{IN,MAX})}{2 \cdot F_{SW} \cdot L}$$

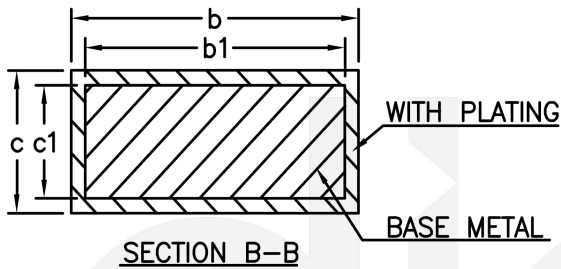
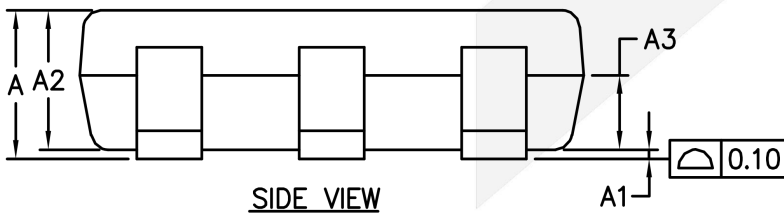
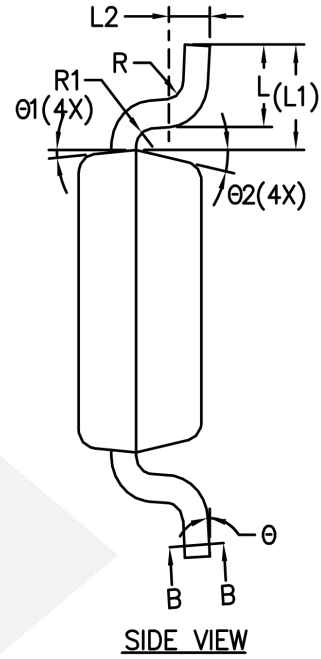
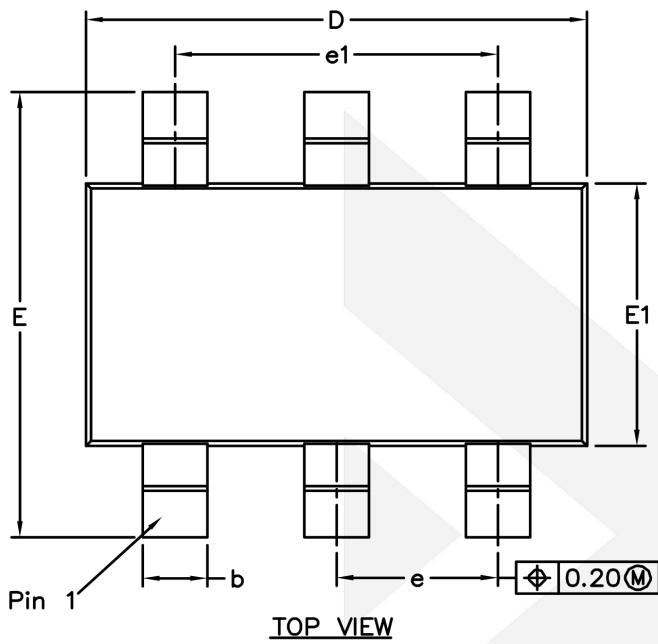
- 3) The DCR of the inductor and the core loss at the switching frequency must be low enough to achieve the desired efficiency requirement. It is desirable to choose an inductor with $DCR < 50m\Omega$ to achieve a good overall efficiency.

Layout Design:

The layout design of the DIO69309 regulator is relatively simple. For the best efficiency and minimum noise problems, we should place the following components close to the IC: C_{IN} , L, R1, and R2.

- 1) It is desirable to maximize the PCB copper area connecting to the GND pin to achieve the best thermal and noise performance. If the board space allows, a ground plane is highly desirable.
- 2) C_{IN} must be close to Pins IN and GND. The loop area formed by C_{IN} and GND must be minimized.
- 3) The PCB copper area associated with the SW pin must be minimized to avoid the potential noise problem.
- 4) The components R1 and R2, and the trace connecting to the FB pin must NOT be adjacent to the SW net on the PCB layout to avoid the noise problem.
- 5) If the system chip interfacing with the EN pin has a high impedance state at shutdown mode and the IN pin is connected directly to a power source such as a Li-Ion battery, it is desirable to add a pull down 1M Ω resistor between the EN and GND pins to prevent the noise from falsely turning on the regulator at shutdown mode.

Physical Dimensions: TSOT23-6



COMMON DIMENSIONS (UNITS OF MEASURE=MILLIMETER)			
Symbol	MIN	NOM	MAX
A	-	-	0.90
A1	0	-	0.15
A2	0.65	0.75	0.85
A3	0.35	0.40	0.45
b	0.36	-	0.50
b1	0.36	0.38	0.45
c	0.14	-	0.20
c1	0.14	0.15	0.16
D	2.85	2.95	3.05
E	2.60	2.80	3.00
E1	1.60	1.65	1.70
e	0.90	0.95	1.00
e1	1.80	1.90	2.00
L	0.30	0.45	0.60
L1	0.575REF		
L2	0.25BSC		
R	-	-	0.25
R1	-	-	0.25
theta	0°	-	8°
theta1	3°	5°	7°
theta2	10°	12°	14°



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CONTACT US

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dioo

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