

#### **Features**

Single-Supply Operation from +2.1V ~ +5.5V

· Rail-to-Rail Input / Output

Gain-Bandwidth Product: 1MHz (Typ)

Low Input Bias Current: 1pA (Typ)

Low Offset Voltage: 3.5mV (Max)

Quiescent Current: 40µA per Amplifier (Typ)
Operating Temperature: -40°C ~ +125°C

· Embedded RF Anti-EMI Filter

#### Small Package:

GS321 Available in SOT23-5 and SC70-5 Packages GS358 Available in SOP-8, MSOP-8, DIP-8 and DFN-8 Packages

GS324 Available in SOP-14 and TSSOP-14 Packages

## **General Description**

The GS321/358/324 family have a high gain-bandwidth product of 1MHz, a slew rate of 0.6V/μs, and a quiescent current of 40μA /amplifier at 5V. The GS321/358/324 family is designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for GS321/358/324 family. They are specified over the extended industrial temperature range (-40°C to +125°C). The operating range is from 2.1V to 5.5V. The GS321 single is available in Green SC70-5 and SOT-23-5 packages. The GS358 Dual is available in Green SOP-8, MSOP-8, DIP-8 and DFN-8 packages. The GS324 Quad is available in Green SOP-14 and TSSOP-14 packages.

## **Applications**

- ASIC Input or Output Amplifier
- Sensor Interface
- · Medical Communication
- Smoke Detectors

- · Audio Output
- · Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Portable Systems

## **Pin Configuration**

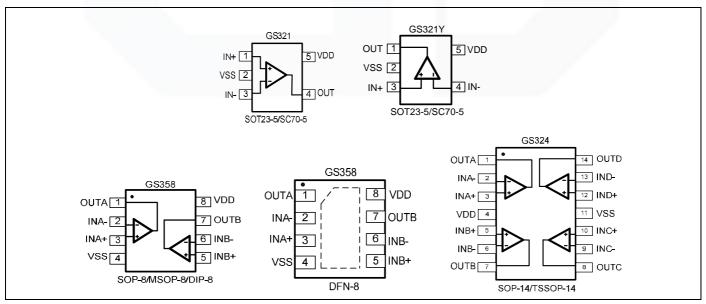


Figure 1. Pin Assignment Diagram





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## **Absolute Maximum Ratings**

Condition	Min	Max			
Power Supply Voltage (V <sub>DD</sub> to Vss)	-0.5V	+7.5V			
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V <sub>DD</sub> +0.5V			
PDB Input Voltage	Vss-0.5V	+7V			
Operating Temperature Range	-40°C	+125°C			
Junction Temperature	+160	D°C			
Storage Temperature Range	-55°C	+150°C			
Lead Temperature (soldering, 10sec)	+260	+260°C			
Package Thermal Resistance (T <sub>A</sub> =+25℃)					
SOP-8, θ <sub>JA</sub>	125°0	C/W			
MSOP-8, θ <sub>JA</sub>	216°0	C/W			
SOT23-5, θ <sub>JA</sub>	190°0	190°C/W			
SC70-5, θ <sub>JA</sub>	333°0	333°C/W			
ESD Susceptibility					
НВМ	6K	V			
MM	300	V			

**Note**: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

# Package/Ordering Information

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION	
		GS321-CR	SC70-5	Tape and Reel,3000	321	
C6334	GS321 Single	GS321-TR	SOT23-5	Tape and Reel,3000	321/GS321	
GS321		GS321Y-CR	SC70-5	Tape and Reel,3000	321Y	
		GS321Y-TR	SOT23-5	Tape and Reel,3000	321Y	
	GS358 Dual		GS358-SR	SOP-8	Tape and Reel,4000	GS358
00050		GS358-MR	MSOP-8	Tape and Reel,3000	GS358	
GS358		GS358-DR	DIP-8	20Tube(1000pcs)	GS358	
		GS358-FR	DFN-8	Tape and Reel,3000	GS358	
C6334	Oued	GS324-TR	TSSOP-14	Tape and Reel,3000	GS324	
G3324	GS324 Quad	GS324-SR	SOP-14	Tape and Reel,2500	GS324	



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### **Electrical Characteristics**

(At Vs = +5V,  $RL = 100k\Omega$  connected to Vs/2, and Vout = Vs/2, unless otherwise noted.)

			GS321/358/324					
PARAMETER	SYMBOL	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE				
			+25℃	+25℃	-40℃ to +85℃	UNITS	MIN/MAX	
INPUT CHARACTERISTICS								
Input Offset Voltage	Vos	V <sub>CM</sub> = V <sub>S</sub> /2	0.4	3.5	5.6	mV	MAX	
Input Bias Current	I <sub>B</sub>		1			pА	TYP	
Input Offset Current	los		1			pА	TYP	
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>S</sub> = 5.5V	-0.1 to +5.6			V	TYP	
Common Made Dejection Datio	CMRR	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to 4V	70	62	62	dB	MINI	
Common-Mode Rejection Ratio	CIVIRR	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to $5.6V$	68	56	55		MIN	
Open Lean Voltage Coin	^	$R_L = 5k\Omega$ , $V_O = +0.1V$ to +4.9V	80	70	70	dB	MINI	
Open-Loop Voltage Gain	A <sub>OL</sub>	$R_L = 10k\Omega$ , $V_O = +0.1V$ to +4.9V	100	94	85		MIN	
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$		2.7			μV/°C	TYP	
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	V <sub>OH</sub>	$R_L = 100k\Omega$	4.997	4.990	4.980	V	MIN	
	V <sub>OL</sub>	$R_L = 100k\Omega$	3	10	20	mV	MAX	
	V <sub>OH</sub>	$R_L = 10k\Omega$	4.992	4.970	4.960	V	MIN	
	V <sub>OL</sub>	$R_L = 10k\Omega$	8	30	40	mV	MAX	
Output Compant	I <sub>SOURCE</sub>	D 400 to V /2	84	60	45	<b>∞</b> Λ Μ	MINI	
Output Current	I <sub>SINK</sub>	$R_L = 10\Omega$ to $V_S/2$	75	60	45	mA	MIN	
POWER SUPPLY								
On arcting Valtage Dange				2.1	2.5	V	MIN	
Operating Voltage Range				5.5	5.5	V	MAX	
Power Supply Rejection Ratio	PSRR	$V_S = +2.5V \text{ to } +5.5V, V_{CM} = +0.5V$	82	60	58	dB	MIN	
Quiescent Current / Amplifier	ΙQ		40			μA	TYP	
DYNAMIC PERFORMANCE (CL	. = 100pF)							
Gain-Bandwidth Product	GBP		1			MHz	TYP	
Slew Rate	SR	G = +1, 2V Output Step	0.6			V/µs	TYP	
Settling Time to 0.1%	t <sub>S</sub>	G = +1, 2V Output Step	5			μs	TYP	
Overload Recovery Time		V <sub>IN</sub> ·Gain = V <sub>S</sub>	2.6			μs	TYP	
NOISE PERFORMANCE								
Voltage Noise Density		f = 1kHz	27			$nV / \sqrt{Hz}$	TYP	
voltage Noise Density	e <sub>n</sub>	f = 10kHz	20			$nV / \sqrt{Hz}$	TYP	



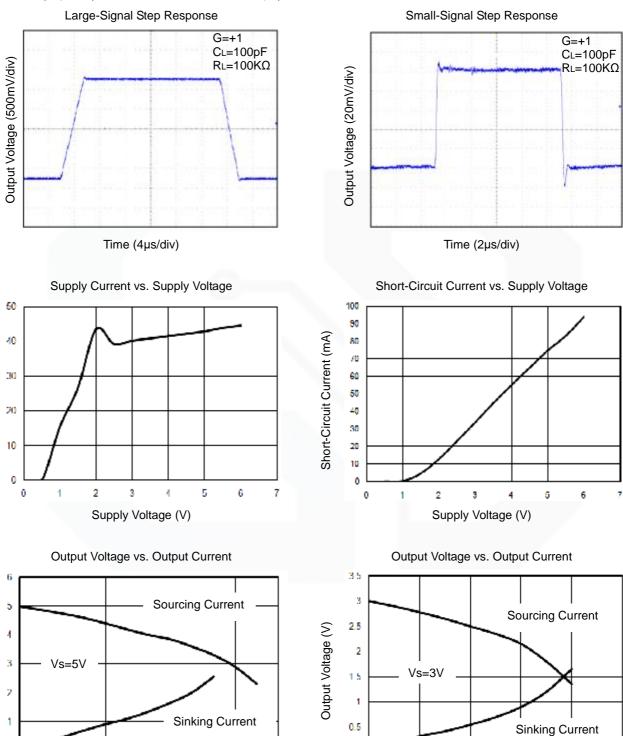


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## **Typical Performance characteristics**

At  $T_A$ =+25°C,  $V_S$ =+5V, and  $R_L$ =100K $\Omega$  connected to  $V_S$ /2, unless otherwise noted.





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20

40

60

Output Current (mA)

100

120

0 L

Supply Current (uA)

Output Voltage (V)



Output Current (mA)

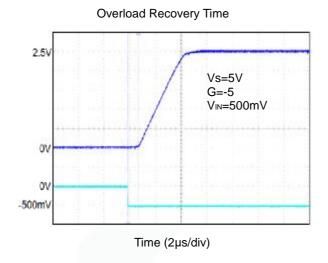
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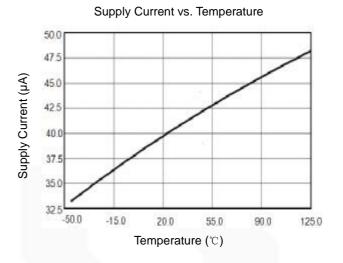
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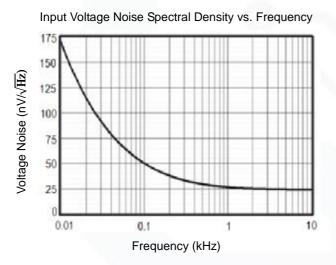


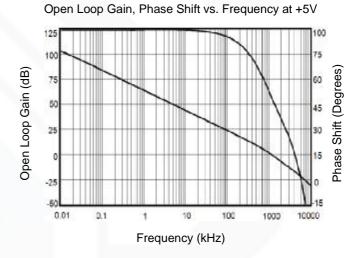
### **Typical Performance characteristics**

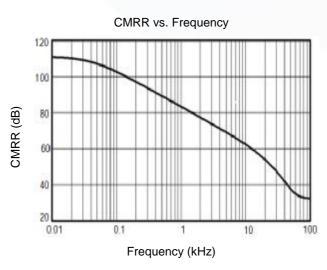
At  $T_A$ =+25°C,  $V_S$ =+5V, and  $R_L$ =100K $\Omega$  connected to  $V_S$ /2, unless otherwise noted.

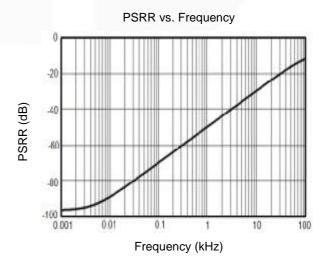












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### **Application Note**

#### Size

GS321/358/324 family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the GS321/358/324 family packages save space on printed circuit boards and enable the design of smaller electronic products.

#### Power Supply Bypassing and Board Layout

GS321/358/324 family series operates from a single 2.1V to 5.5V supply or dual  $\pm 1.05$ V to  $\pm 2.75$ V supplies. For best performance, a  $0.1\mu$ F ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu$ F ceramic capacitors.

#### Low Supply Current

The low supply current (typical 40µA per channel) of GS321/358/324 family will help to maximize battery life. They are ideal for battery powered systems.

#### **Operating Voltage**

GS321/358/324 family operates under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

#### Rail-to-Rail Input

The input common-mode range of GS321/358/324 family extends 100mV beyond the supply rails ( $V_{SS}$ -0.1V to  $V_{DD}$ +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

#### Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of GS321/358/324 family can typically swing to less than 5mV from supply rail in light resistive loads (> $100k\Omega$ ), and 30mV of supply rail in moderate resistive loads ( $10k\Omega$ ).

#### Capacitive Load Tolerance

The GS321/358/324 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create apole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

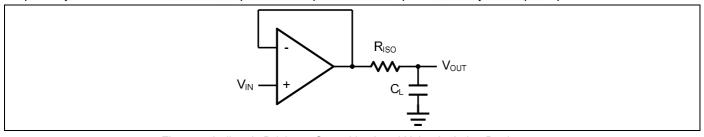


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor





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The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the  $V_{IN}$  to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

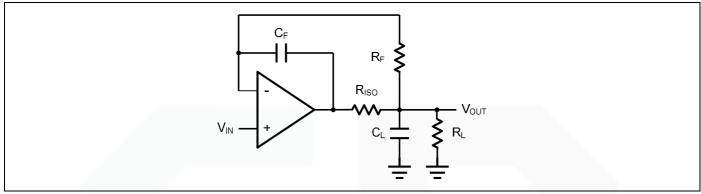


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy





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## **Typical Application Circuits**

### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using GS321/358/324 family.

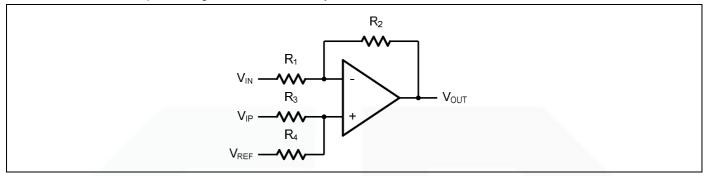


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R<sub>1</sub>=R<sub>3</sub> and R<sub>2</sub>=R<sub>4</sub>), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

### Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

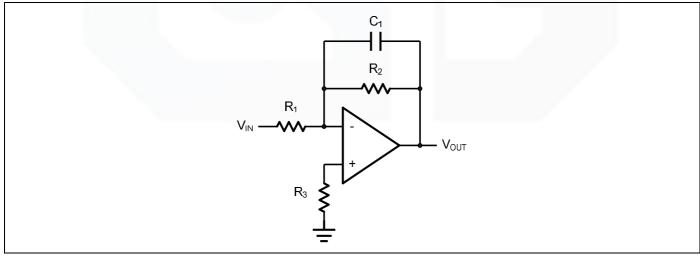


Figure 5. Low Pass Active Filter





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### **Instrumentation Amplifier**

The triple GS321/358/324 family can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R2/R1. The two differential voltage followers assure the high input impedance of the amplifier.

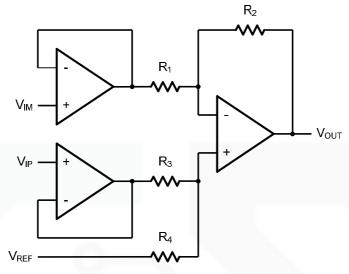


Figure 6. Instrument Amplifier

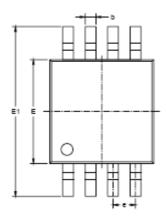


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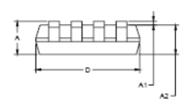


# Package Information

### MSOP-8



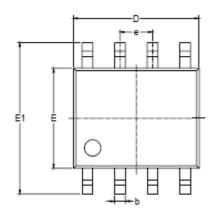


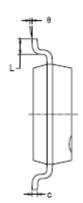


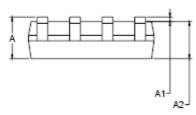
Symbol	Dimen In Milli		Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
A	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.008	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650	BSC	0.026 BSC		
L	0.400	0.800	0.016	0.031	
0	0°	6°	0°	6°	



SOP-8





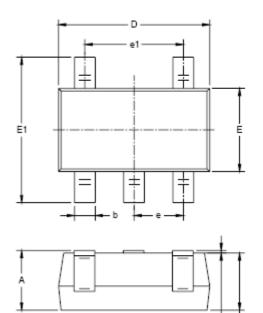


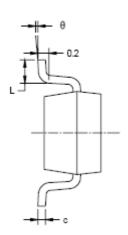
Symbol		nsions imeters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27	1.27 BSC		BSC	
L	0.400	1.270	0.016	0.050	
е	0° 8°		0°	8°	

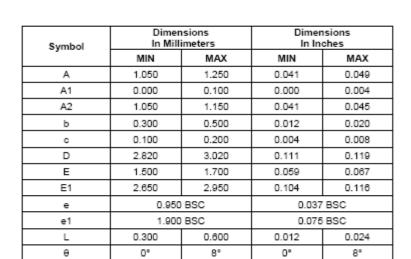
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#### SOT23-5



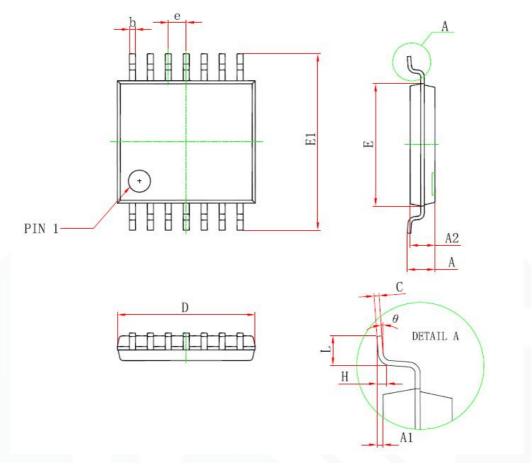




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### TSSOP-14

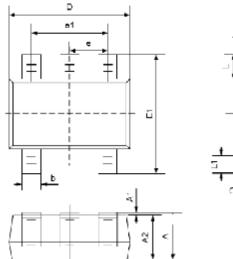


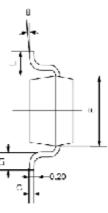
	Dimensions I	n Millimeters	Dimensions In Inche		
Symbol	Min	Max	Min	Max	
D	4. 900	5. 100	0.193	0.201	
E	4.300	4.500	0.169	0.177	
b	0.190	0.300	0.007	0.012	
С	0.090	0.200	0.004	0.008	
E1	6. 250	6.550	0.246	0.258	
A		1.200		0.047	
A2	0.800	1.000	0.031	0.039	
A1	0.050	0.150	0.002	0.006	
e	0.65	(BSC)	0.026	(BSC)	
L	0.500	0.700	0.020	0.028	
Н	0.25(	TYP)	0.01(	TYP)	
θ	1 °	7°	1 °	7°	





#### SC70-5





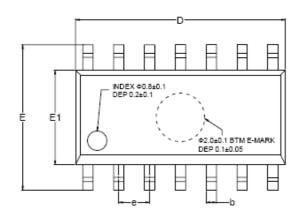
	Dimensions		Dimens	sions	
Symbol	In Millir	meters	In Inches		
	Min Max		Min	Max	
A	0.900	1.100	0.035	0.043	
A1	0.000	0.100	0.000	0.004	
A2	0.900	1.000	0.035	0.039	
b	0.150	0.350	0.006	0.014	
С	0.080	0.150	0.003	0.006	
D	2.000	2.200	0.079	0.087	
E	1.150	1.350	0.045	0.053	
E1	2.150	2.450	0.085	0.096	
e	0.650T	ΥP	0.026T	ΥP	
e1	1.200	1.400	0.047	0.055	
L	0.525REF		0.021REF		
L1	0.260	0.460	0.010 0.01		
Ð	0.	8°	0°	8°	

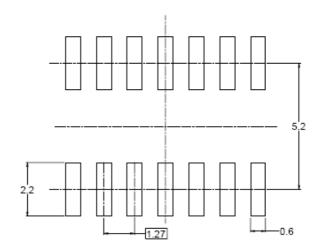




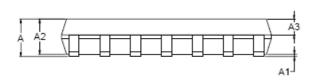


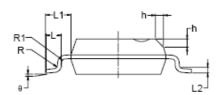
### SOP-14





RECOMMENDED LAND PATTERN (Unit: mm)

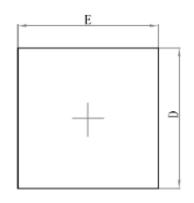


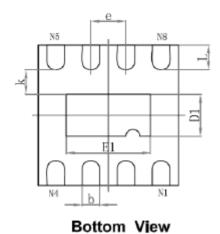


Symbol	Dimen	sions In Mill	imeters	Dimensions In Inches		
Symbol	MIN	MOD	MAX	MIN	MOD	MAX
Α	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.004		0.010
A2	1.25		1.65	0.049		0.065
A3	0.55		0.75	0.022		0.030
b	0.36		0.49	0.014		0.019
D	8.53		8.73	0.336		0.344
E	5.80		6.20	0.228		0.244
E1	3.80		4.00	0.150		0.157
е		1.27 BSC			0.050 BSC	
L	0.45		0.80	0.018		0.032
L1		1.04 REF			0.040 REF	
L2		0.25 BSC			0.01 BSC	
R	0.07			0.003		
R1	0.07			0.003		
h	0.30		0.50	0.012		0.020
θ	0°		8°	0°		8°

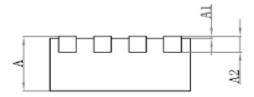


#### DFN-8





Top View



Side View

	Dimensions			Dimensions In Inches		
Symbol	In Milli	meters				
	Min	Nom	Max	Min	Nom	Max
Α	0.80	0.85	0.9	0.031	0.033	0.035
A1	0.00	0.02	0.05	0.000	0.001	0.002
A2	0.153	0.203	0.253	0.006	0.008	0.010
b	0.18	0.24	0.30	0.007	0.009	0.012
D	1.9	2.0	2.1	0.075	0.079	0.083
E	1.9	2.0	2.1	0.075	0.079	0.083
D1	0.5	0.6	0.7	0.020	0.024	0.028
E1	1.1	1.2	1.3	0.043	0.047	0.051
е		0.50			0.20	
k	0.2			0.008		
L	0.25	0.35	0.45	0.010	0.014	0.018

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