

#### **Features**

Single-Supply Operation from +1.8V ~ +6V

• Rail-to-Rail Input / Output

Gain-Bandwidth Product: 1MHz (Typ.)

Low Input Bias Current: 1pA (Typ.)

• Low Offset Voltage: 3.5mV (Max.)

Quiescent Current: 75µA per Amplifier (Typ.)

• Embedded RF Anti-EMI Filter

Operating Temperature: -40°C ~ +125°C

Small Package:

GS6001 Available in SOT23-5 and SC70-5 Packages GS6002 Available in SOP-8 and MSOP-8 Packages GS6004 Available in SOP-14 and TSSOP-14 Packages

### **General Description**

The GS6001 family have a high gain-bandwidth product of 1MHz, a slew rate of  $0.8V/\mu\,s$ , and a quiescent current of 75  $\mu$  A/amplifier at 5V. The GS6001 family is designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for GS6001 family. They are specified over the extended industrial temperature range (-40 °C to +125 °C). The operating range is from 1.8V to 6V. The GS6001 single is available in Green SC70-5 and SOT23-5 packages. The GS6002 dual is available in Green SOP-8 and MSOP-8 packages. The GS6004 Quad is available in Green SOP-14 and TSSOP-14 packages.

## **Applications**

- ASIC Input or Output Amplifier
- Sensor Interface
- Medical Communication
- Smoke Detectors

- Audio Output
- Piezoelectric Transducer Amplifier
- Medical Instrumentation
- Portable Systems

## **Pin Configuration**

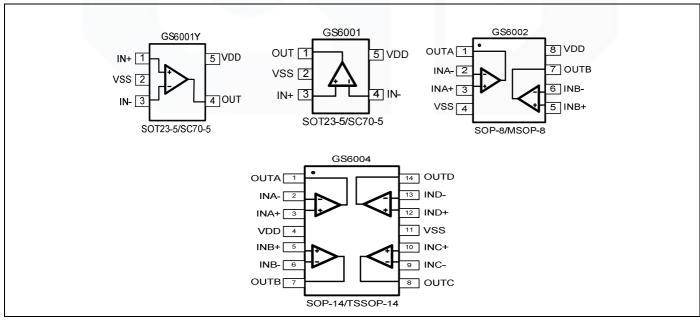


Figure 1. Pin Assignment Diagram





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## **Absolute Maximum Ratings**

Condition	Min	Max
Power Supply Voltage (V <sub>DD</sub> to Vss)	-0.5V	+7.5V
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V <sub>DD</sub> +0.5V
PDB Input Voltage	Vss-0.5V	+7V
Operating Temperature Range	-40°C	+125°C
Junction Temperature	+160	)°C
Storage Temperature Range	-55°C	+150°C
Lead Temperature (soldering, 10sec)	+260	)°C
Package Thermal Resistance (T <sub>A</sub> =+25℃)		
SOP-8, θ <sub>JA</sub>	125°0	C/W
MSOP-8, $\theta_{JA}$	216°0	C/W
SOT23-5, θ <sub>JA</sub>	190°C	C/W
SC70-5, θ <sub>JA</sub>	333°0	C/W
ESD Susceptibility		
НВМ	6K	V
MM	400	)V

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

# **Package/Ordering Information**

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
		GS6001-CR	SC70-5	Tape and Reel,3000	6001
GS6001	000004	GS6001-TR	SOT23-5	Tape and Reel,3000	6001
G36001	Single	GS6001Y-CR	SC70-5	Tape and Reel,3000	6001Y
		GS6001Y-TR	SOT23-5	Tape and Reel,3000	6001Y
000000	Duel	GS6002-SR	SOP-8	Tape and Reel,4000	GS6002
GS6002 Dual	GS6002-MR	MSOP-8	Tape and Reel,3000	GS6002	
GS6004	Overal	GS6004-TR	TSSOP-14	Tape and Reel,3000	GS6004
	Quad	GS6004-SR	SOP-14	Tape and Reel,2500	GS6004



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## **Electrical Characteristics**

(At VS = +5V, RL =  $100k\Omega$  connected to VS/2, and VOUT = VS/2, unless otherwise noted.)

			GS6001/2/4					
PARAMETER	SYMBOL	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE				
			+25℃	+25℃	-40℃ to +85℃	UNITS	MIN/MA	
INPUT CHARACTERISTICS								
Input Offset Voltage	Vos	$V_{CM} = V_S/2$	0.8	3.5	5.6	mV	MAX	
Input Bias Current	I <sub>B</sub>		1			pA	TYP	
Input Offset Current	los		1			pA	TYP	
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>S</sub> = 5.5V	-0.1 to +5.6			٧	TYP	
Common Mode Dejection Datio	CMDD	V <sub>S</sub> = 5.5V, V <sub>CM</sub> = -0.1V to 4V	70	62	62	dB		
Common-Mode Rejection Ratio	CMRR	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to 5.6V	68	56	55		MIN	
Open Lean Valtage Coin	Δ.	$R_L = 5k\Omega$ , $V_O = +0.1V$ to +4.9V	80	70	70	dB		
Open-Loop Voltage Gain	A <sub>OL</sub>	$R_L = 10k\Omega$ , $V_O = +0.1V$ to +4.9V	100	94	85		MIN	
Input Offset Voltage Drift	$\Delta V_{OS}/\Delta_T$	A \	2.7			μV/°C	TYP	
OUTPUT CHARACTERISTICS								
	V <sub>OH</sub>	R <sub>L</sub> = 100kΩ	4.997	4.980	4.970	٧	MIN	
Output Valtage Cuing from Deil	V <sub>OL</sub>	R <sub>L</sub> = 100kΩ	5	20	30	mV	MAX	
Output Voltage Swing from Rail	V <sub>OH</sub>	R <sub>L</sub> = 10kΩ	4.992	4.970	4.960	V	MIN	
	V <sub>OL</sub>	R <sub>L</sub> = 10kΩ	8	30	40	mV	MAX	
Output Current	I <sub>SOURCE</sub>	D = 100 to 1/ /2	84	60	45	A	MIN	
Output Current	I <sub>SINK</sub>	$R_L = 10\Omega$ to $V_S/2$	75	60	45	mA	IVIIIN	
POWER SUPPLY								
Operating Voltage Dange				1.8	1.8	٧	MIN	
Operating Voltage Range				6	6	٧	MAX	
Power Supply Rejection Ratio	PSRR	$V_S = +2.5V \text{ to } +6V, V_{CM} = +0.5V$	82	60	58	dB	MIN	
Quiescent Current / Amplifier	ΙQ		75	110	125	μA	MAX	
DYNAMIC PERFORMANCE (CL	= 100pF)							
Gain-Bandwidth Product	GBP		1			MHz	TYP	
Slew Rate	SR	G = +1, 2V Output Step	0.8			V/µs	TYP	
Settling Time to 0.1%	t <sub>S</sub>	G = +1, 2V Output Step	5.3			μs	TYP	
Overload Recovery Time		V <sub>IN</sub> ·Gain = V <sub>S</sub>	2.6			μs	TYP	
NOISE PERFORMANCE								
Voltage Noise Demaits	_	f = 1kHz	27			$nV/\sqrt{Hz}$	TYP	
Voltage Noise Density	e <sub>n</sub>	f = 10kHz	20			$nV/\sqrt{Hz}$	TYP	





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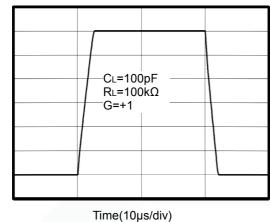


Output Voltage (250mV/div)

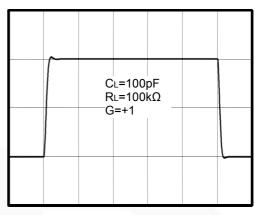
## **Typical Performance characteristics**

At  $T_A$ =+25°C, Vs=5V,  $R_L$ =100K $\Omega$  connected to  $V_S$ /2 and  $V_{OUT}$ =  $V_S$ /2, unless otherwise noted.





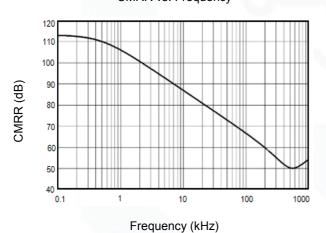
Small Signal Transient Response



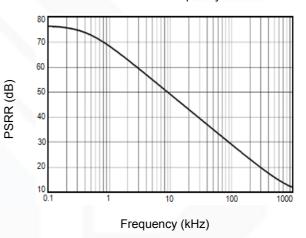
Time(2µs/div)

Output Voltage (50mV/div)

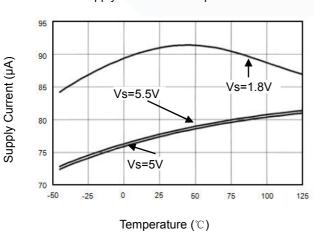
CMRR vs. Frequency



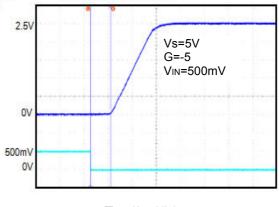
PSRR vs. Frequency



Supply Current vs. Temperature



Overload Recovery Time



Time(2µs/div)



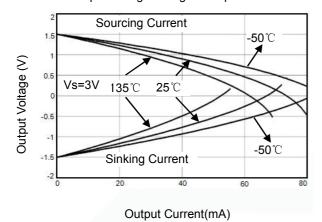
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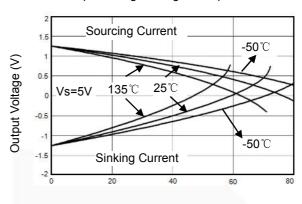
## **Typical Performance characteristics**

At  $T_A$ =+25°C,  $R_L$ =100K $\Omega$  connected to  $V_S$ /2 and  $V_{OUT}$ =  $V_S$ /2, unless otherwise noted.

Output Voltage Swing vs.Output Current

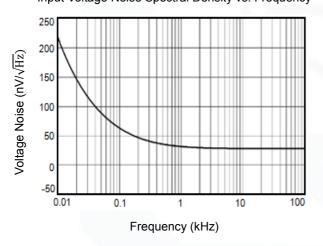


Output Voltage Swing vs.Output Current

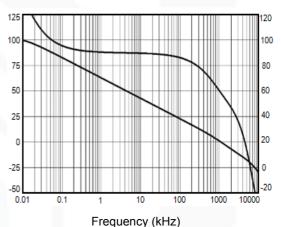


Output Current(mA)

Input Voltage Noise Spectral Density vs. Frequency



Open Loop Gain, Phase Shift vs. Frequency



Phase Shift (Degrees)

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Open Loop Gain (dB)



## **Application Note**

#### Size

GS6001 family series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the GS6001 family packages save space on printed circuit boards and enable the design of smaller electronic products.

#### **Power Supply Bypassing and Board Layout**

GS6001 family series operates from a single 1.8V to 6V supply or dual  $\pm 0.9$ V to  $\pm 3$ V supplies. For best performance, a  $0.1\mu$ F ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu$ F ceramic capacitors.

#### **Low Supply Current**

The low supply current (typical 75µA per channel) of GS6001 family will help to maximize battery life. They are ideal for battery powered systems

#### **Operating Voltage**

GS6001 family operates under wide input supply voltage (1.8V to 6V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime

#### Rail-to-Rail Input

The input common-mode range of GS6001 family extends 100mV beyond the supply rails ( $V_{SS}$ -0.1V to  $V_{DD}$ +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

#### **Rail-to-Rail Output**

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of GS6001 family can typically swing to less than 10mV from supply rail in light resistive loads (>100k $\Omega$ ), and 60mV of supply rail in moderate resistive loads (10k $\Omega$ ).

#### **Capacitive Load Tolerance**

The GS6001 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2 shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

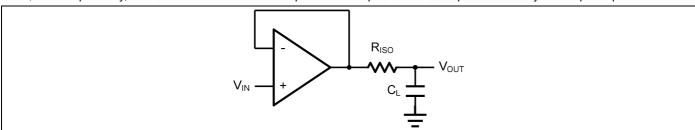


Figure 2 Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R<sub>F</sub> provides the DC accuracy by feed-forward the V<sub>IN</sub> to R<sub>L</sub>. C<sub>F</sub>



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and  $R_{\rm ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

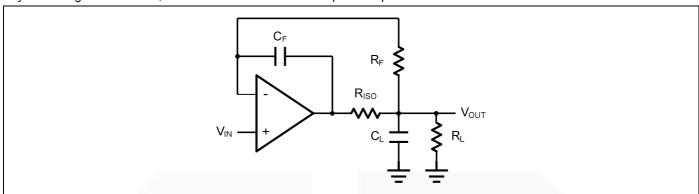


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy







## **Typical Application Circuits**

## **Differential amplifier**

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using GS6001 family.

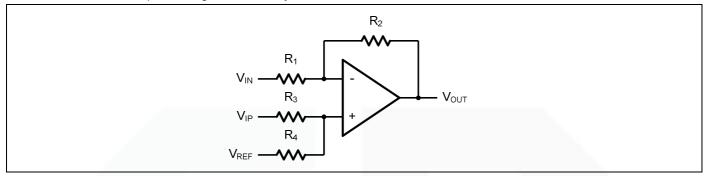


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + \left(\frac{R_1 + R_2}{R_3 + R_4}\right) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R<sub>1</sub>=R<sub>3</sub> and R<sub>2</sub>=R<sub>4</sub>), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

#### **Low Pass Active Filter**

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

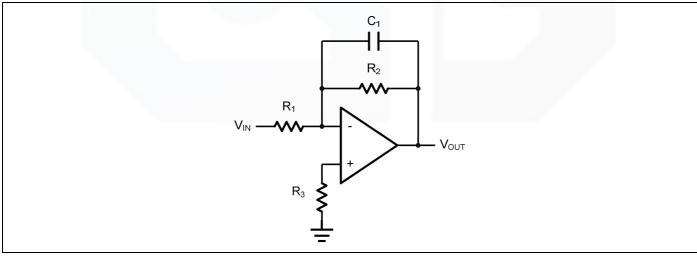


Figure 5. Low Pass Active Filter



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## **Instrumentation Amplifier**

The triple GS6001 family can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

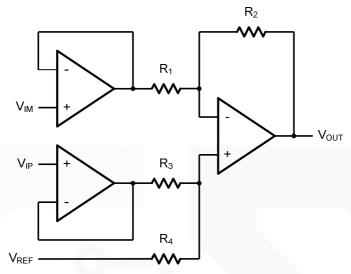


Figure 6. Instrument Amplifier



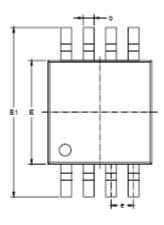
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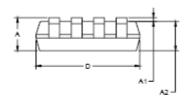


# **Package Information**

## MSOP-8





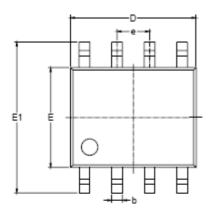


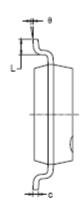
Symbol	Dimer In Milli		Dimensions In Inches		
_	MIN	MAX	MIN	MAX	
A	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.008	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650	0.650 BSC		BSC	
L	0.400	0.800	0.016 0.031		
θ	0°	6°	0° 6°		

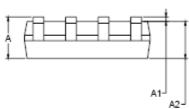




## SOP-8





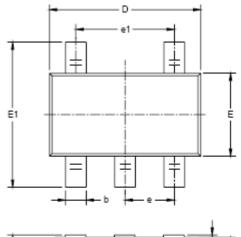


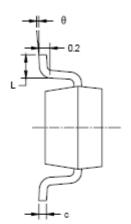
Symbol		nsions meters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27 BSC		0.050	BSC	
L	0.400	1.270	0.016	0.050	
6	0°	8°	0°	8°	

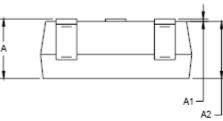




# SOT23-5





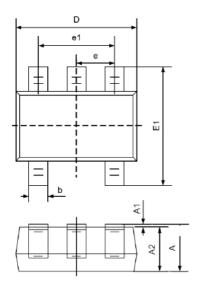


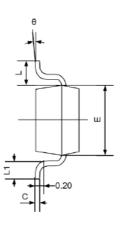
Symbol		nsions imeters	Dimensions In Inches		
,	MIN	MAX	MIN	MAX	
A	1.050	1.250	0.041	0.049	
A1	0.000	0.100	0.000	0.004	
A2	1.050	1.150	0.041	0.045	
b	0.300	0.500	0.012	0.020	
С	0.100	0.200	0.004	0.008	
D	2.820	3.020	0.111	0.119	
E	1.500	1.700	0.059	0.067	
E1	2.650	2.950	0.104	0.116	
e	0.950	BSC	0.037 BSC		
e1	1.900	1.900 BSC		BSC	
L	0.300	0.600	0.012	0.024	
θ	0°	8°	0°	8°	





# SC70-5





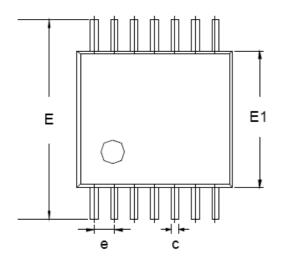
	Dimens	sions	Dimensions		
Symbol	In Milli	meters	In Inches		
	Min	Max	Min	Max	
Α	0.900	1.100	0.035	0.043	
A1	0.000	0.100	0.000	0.004	
A2	0.900	1.000	0.035	0.039	
b	0.150	0.350	0.006	0.014	
С	0.080	0.150	0.003	0.006	
D	2.000	2.200	0.079	0.087	
E	1.150	1.350	0.045	0.053	
E1	2.150	2.450	0.085	0.096	
е	0.650T	ΥP	0.026TYP		
e1	1.200	1.400	0.047	0.055	
L	0.525REF		0.021REF		
L1	0.260	0.460	0.010	0.018	
θ	0°	8°	0°	8°	

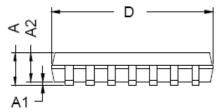


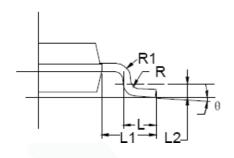




## TSSOP-14







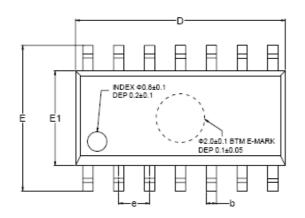
	Dimensions				
Symbol	In Millimeters				
Symbol	MIN	TYP	MAX		
Α	-	-	1.20		
A1	0.05	-	0.15		
A2	0.90	1.00	1.05		
b	0.20	-	0.28		
С	0.10	-	0.19		
D	4.86	4.96	5.06		
E	6.20	6.40	6.60		
E1	4.30	4.40	4.50		
е		0.65 BSC			
L	0.45	0.60	0.75		
L1	1.00 REF				
L2	0.25 BSC				
R	0.09	-	-		
θ	0°	-	8°		

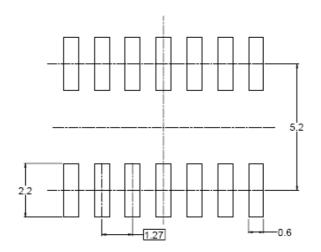


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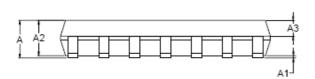


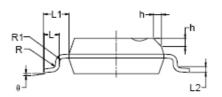
#### **SOP-14**





RECOMMENDED LAND PATTERN (Unit: mm)





S. mahad	Dimens	Dimensions In Millimeters			Dimensions In Inches		
Symbol	MIN	MOD	MAX	MIN	MOD	MAX	
Α	1.35		1.75	0.053		0.069	
A1	0.10		0.25	0.004		0.010	
A2	1.25		1.65	0.049		0.065	
A3	0.55		0.75	0.022		0.030	
b	0.36		0.49	0.014		0.019	
D	8.53		8.73	0.336		0.344	
E	5.80		6.20	0.228		0.244	
E1	3.80		4.00	0.150		0.157	
е	1.27 BSC				0.050 BSC		
L	0.45		0.80	0.018		0.032	
L1	1.04 REF				0.040 REF		
L2	0.25 BSC				0.01 BSC		
R	0.07			0.003			
R1	0.07			0.003			
h	0.30		0.50	0.012		0.020	
θ	0°		8°	0°		8°	

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