



Features

• Single-Supply Operation from +2.1V ~ +5.5V

• Rail-to-Rail Input / Output

• Gain-Bandwidth Product: 3MHz (Typ)

• Low Input Bias Current: 1pA (Typ)

Low Offset Voltage: 3.5mV (Max)

Quiescent Current: 250µA per Amplifier (Typ)

• Operating Temperature: -40°C ~ +125°C

• Small Package:

GS8622 Available in SOP-8 and MSOP-8 Packages GS8624 Available in SOP-14 and TSSOP-14 Packages

General Description

The GS8622/24 have a high gain-bandwidth product of 3MHz, a slew rate of $1.66\text{V}/\mu\text{s}$, and a quiescent current of $250\mu\text{A}$ per amplifier at 5V. The GS8622/24 are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for GS8622/24. They are specified over the extended industrial temperature range (-40°C to +125°C). The operating range is from 2.1V to 5.5V. The GS8622 dual is available in Green SOP-8 and MSOP-8 packages. The GS8624 Quad is available in Green SOP-14 and TSSOP-14 packages.

Applications

- Sensors
- Active Filters
- · Cellular and Cordless Phones
- Laptops and PDAs

- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters

Pin Configuration

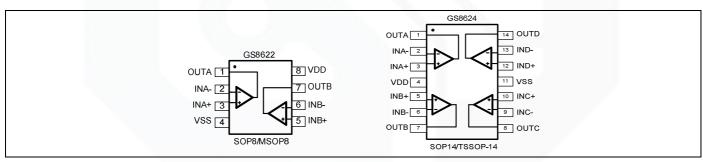


Figure 1. Pin Assignment Diagram



Absolute Maximum Ratings

Condition	Min	Max
Power Supply Voltage (V _{DD} to Vss)	-0.5V	+7.5V
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	V _{DD} +0.5V
PDB Input Voltage	Vss-0.5V	+7V
Operating Temperature Range	-40°C	+125°C
Junction Temperature	+160)°C
Storage Temperature Range	-55°C	+150°C
Lead Temperature (soldering, 10sec)	+260)°C
Package Thermal Resistance (TA=+25℃)		
SOP-8, θ _{JA}	125°	C/W
MSOP-8, θ _{JA}	216°	C/W
ESD Susceptibility		
НВМ	84	(V
MM	40	0V

Note: Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

Package/Ordering Information

MODEL	CHANNEL	ORDER NUMBER	PACKAGE DESCRIPTION	PACKAGE OPTION	MARKING INFORMATION
000000	S8622 Dual	GS8622-SR	SOP-8	Tape and Reel,4000	GS8622
G58622		GS8622-MR	MSOP-8	Tape and Reel,3000	GS8622
CC0624	Quad	GS8624-TR	TSSOP-14	Tape and Reel,3000	GS8624
GS8624 Qua	Quad	GS8624-SR	SOP-14	Tape and Reel,2500	GS8624



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Electrical Characteristics

(At Vs=5V, T_A = +25 °C, V_{CM} = $V_S/2$, R_L = 600 Ω , unless otherwise noted.)

		GS8622/4						
DADAMETER	CONDITIONS	TYP	MIN/MAX OVER TEMPERATURE					
PARAMETER	CONDITIONS	+25℃	+25℃	0℃ to 70℃	-40 ℃ to 85 ℃	-40 ℃ to	UNITS	MIN /
INPUT CHARACTERISTICS		1		I			ı	<u> </u>
Input Offset Voltage (Vos)		0.8	3.5	3.9	4.3	4.6	mV	MAX
Input Bias Current (I _B)		1					pA	TYP
Input Offset Current (Ios)		1					pA	TYP
Input Common Mode Voltage Range (V_{CM})	V _S = 5.5V	-0.1 to					V	TYP
		+5.6						
Common Mode Rejection Ratio (CMRR)	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 4V	82	65	64	64	63	dB	MIN
	$V_S = 5.5V$, $V_{CM} = -0.1V$ to 5.6V	71					dB	MIN
Open-Loop Voltage Gain (A _{OL})	$R_L = 600\Omega, V_O = 0.15V \text{ to } 4.85V$	90	80	76	75	68	dB	MIN
	$R_L = 10k\Omega, V_O = 0.05V \text{ to } 4.95V$	100					dB	MIN
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta_T$)		2.4					μV/°C	TYP
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	R _L = 600Ω	0.1					V	TYP
	$R_L = 10k\Omega$	0.015					V	TYP
Output Current (I _{OUT})		53	49	45	40	35	mA	MIN
Closed-Loop Output Impedance	f = 100kHz, G = 1	10					Ω	TYP
POWER-DOWN DISABLE								
Turn-On Time		4					μs	TYP
Turn-Off Time		1.2					μs	TYP
DISABLE Voltage-Off			0.8				V	MAX
DISABLE Voltage-On			2				V	MIN
POWER SUPPLY								
Operating Voltage Range			2.1	2.1	2.1	2.1	V	MIN
			5.5	5.5	5.5	5.5	V	MAX
Power Supply Rejection Ratio (PSRR)	V_S = +2.5V to +5.5V							
	$V_{CM} = (-V_S) + 0.5V$	91	74	72	72	68	dB	MIN
Quiescent Current/Amplifier (I _Q)	I _{OUT} = 0	250	350	427	450	515	μΑ	MAX
Supply Current when Disabled		90					nA	MAX
(GS8621N Only)								





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Electrical Characteristics

(At Vs=5V, T_A = +25 $^{\circ}\text{C}$, V_{CM} = V_S/2, R_L = 600 $^{\Omega}$, unless otherwise noted.)

		GS8622/4							
PARAMETER	CONDITIONS	TYP	TYP MIN/MAX OVER TEMPERATURE						
	CONDITIONS	+25℃		0°C to	-40℃ to	-40℃to	UNITS	MIN /	
		+25 C	+25℃	70 ℃	85℃	125℃		MAX	
DYNAMIC PERFORMANCE									
Gain-Bandwidth Product (GBP)	$R_L = 10k\Omega, C_L = 100pF$	3					MHz	TYP	
Phase Margin (φ _O)	$R_L = 10k\Omega, C_L = 100pF$	50					Degrees	TYP	
Full Power Bandwidth (BWP)	$<$ 1% distortion, R_L = 600Ω	50					kHz	TYP	
Slew Rate (SR)	$G = +1$, 2V Step, $R_L = 10$ kΩ	1.66					V/µs	TYP	
Settling Time to 0.1% (t _S)	$G = +1$, 2V Step, $R_L = 600Ω$	0.5					μs	TYP	
Overload Recovery Time	V_{IN} ·Gain = VS, R_L = 600 Ω	4.5					μs	TYP	
NOISE PERFORMANCE	NOISE PERFORMANCE								
Voltage Noise Density (e _n)	f = 1kHz	18					nV/\sqrt{Hz}	TYP	
Current Noise Density (in)	f = 1kHz	4.5					fA/\sqrt{Hz}	TYP	





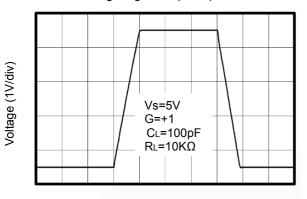
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Typical Performance characteristics

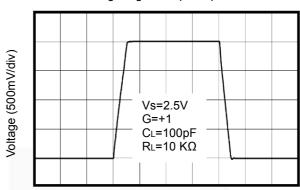
(At Vs=5V, T_A = +25 °C, V_{CM} = Vs/2, R_L = 600 Ω , unless otherwise noted.)

Large-Signal Step Response



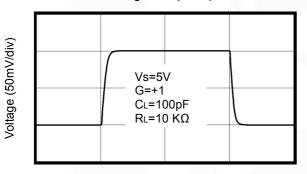
Time (2µs/div)

Large-Signal Step Response



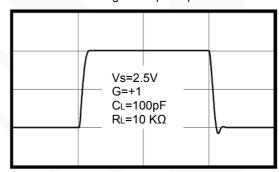
Time (2µs/div)

Small-Signal Step Response



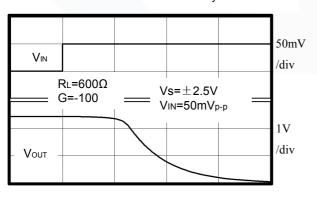
Time (1µs/div)

Small-Signal Step Response



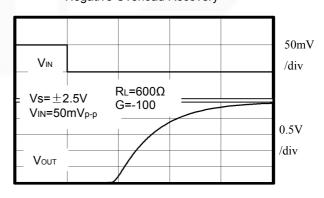
Time (1µs/div)

Positive Overload Recovery



Time (5µs/div)

Negative Overload Recovery



Time (5µs/div)

Voltage (50mV/div)

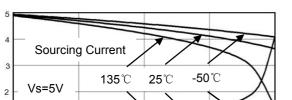


Output Voltage (V)

Typical Performance characteristics

Output Voltage Swing vs.Output Current

(At Vs=5V, T_A = +25 °C, V_{CM} = Vs/2, R_L = 600 Ω , unless otherwise noted.)

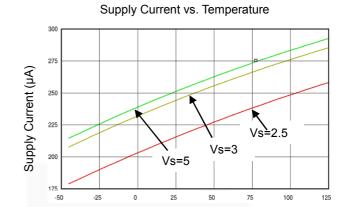


Sinking Current

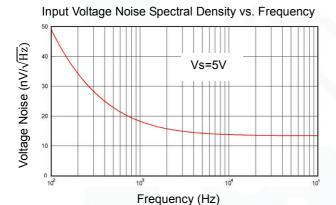
60

Output Current(mA)

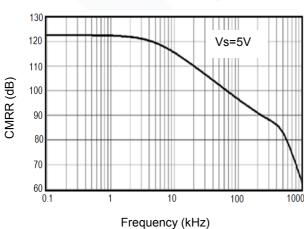
20



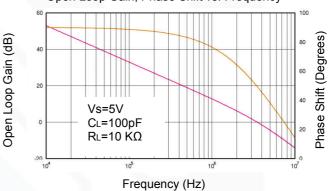
Temperature (°C)



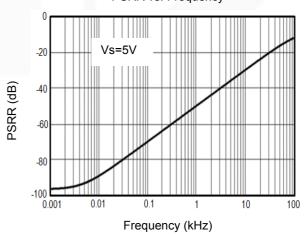
CMRR vs. Frequency



Open Loop Gain, Phase Shift vs. Frequency



PSRR vs. Frequency





Application Note

Size

GS8622/24 series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the GS8622/24 series packages save space on printed circuit boards and enable the design of smaller electronic products.

Power Supply Bypassing and Board Layout

GS8622/24 series operates from a single 2.1V to 5.5V supply or dual ± 1.05 V to ± 2.75 V supplies. For best performance, a 0.1 μ F ceramic capacitor should be placed close to the V_{DD} pin in single supply operation. For dual supply operation, both V_{DD} and V_{SS} supplies should be bypassed to ground with separate 0.1 μ F ceramic capacitors.

Low Supply Current

The low supply current (typical 250µA per channel) of GS8622/24 series will help to maximize battery life. They are ideal for battery powered systems.

Operating Voltage

GS8622/24 series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40 °C to +125 °C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime.

Rail-to-Rail Input

The input common-mode range of GS8622/24 series extends 100mV beyond the supply rails (V_{SS} -0.1V to V_{DD} +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of GS8622/24 series can typically swing to less than 2mV from supply rail in light resistive loads (>100k Ω), and 60mV of supply rail in moderate resistive loads (10k Ω).

Capacitive Load Tolerance

The GS8622/24 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

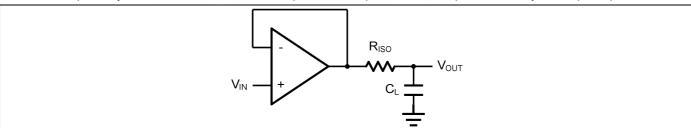


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor





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The bigger the R_{ISO} resistor value, the more stable V_{OUT} will be. However, if there is a resistive load R_L in parallel with the capacitive load, a voltage divider (proportional to R_{ISO}/R_L) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2. R_F provides the DC accuracy by feed-forward the V_{IN} to R_L . C_F and R_{ISO} serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of C_F . This in turn will slow down the pulse response.

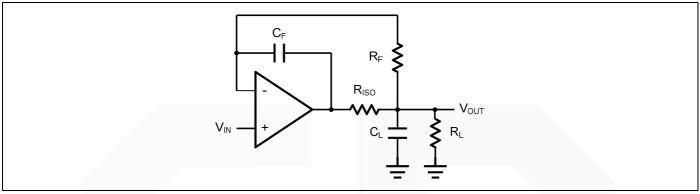


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy





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Typical Application Circuits

Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using GS8622/24.

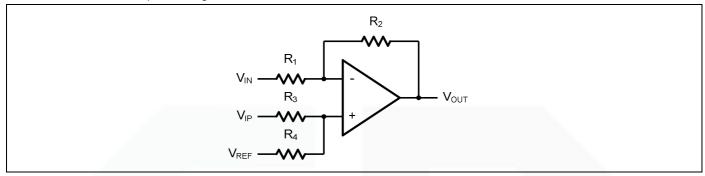


Figure 4. Differential Amplifier

$$V_{\text{OUT}} = (\frac{R_1 + R_2}{R_2 + R_4}) \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + (\frac{R_1 + R_2}{R_2 + R_4}) \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e. R₁=R₃ and R₂=R₄), then

$$V_{\text{OUT}} = \frac{R_2}{R_1} (V_{\text{IP}} - V_{\text{IN}}) + V_{\text{REF}}$$

Low Pass Active Filter

The low pass active filter is shown in Figure 5. The DC gain is defined by $-R_2/R_1$. The filter has a -20dB/decade roll-off after its corner frequency $f_C=1/(2\pi R_3C_1)$.

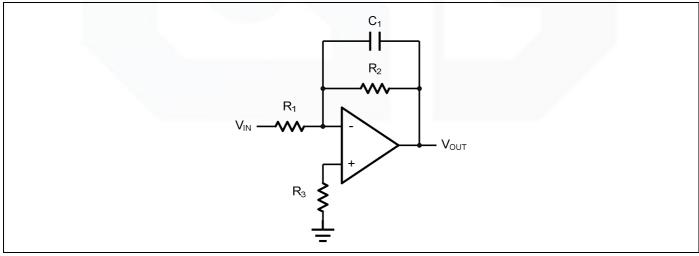


Figure 5. Low Pass Active Filter



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Instrumentation Amplifier

The triple GS8622/24 can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of R_2/R_1 . The two differential voltage followers assure the high input impedance of the amplifier.

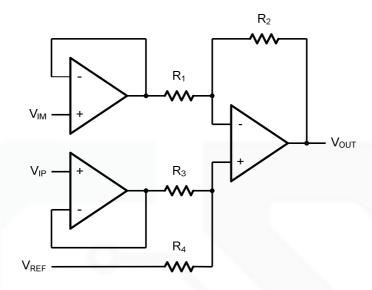


Figure 6. Instrument Amplifier



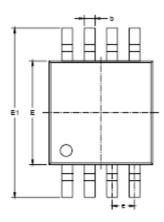


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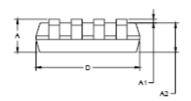


Package Information

MSOP8



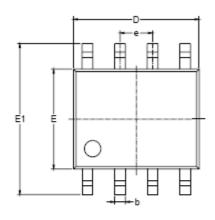


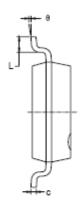


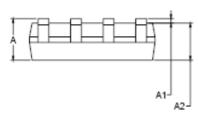
Symbol	Dimer In Milli	nsions meters	Dimensions In Inches		
-	MIN	MAX	MIN	MAX	
Α	0.820	1.100	0.032	0.043	
A1	0.020	0.150	0.001	0.008	
A2	0.750	0.950	0.030	0.037	
b	0.250	0.380	0.010	0.015	
С	0.090	0.230	0.004	0.009	
D	2.900	3.100	0.114	0.122	
E	2.900	3.100	0.114	0.122	
E1	4.750	5.050	0.187	0.199	
e	0.650	BSC	0.026	BSC	
L	0.400	0.800	0.016	0.031	
θ	0°	6°	0°	6°	



SOP8



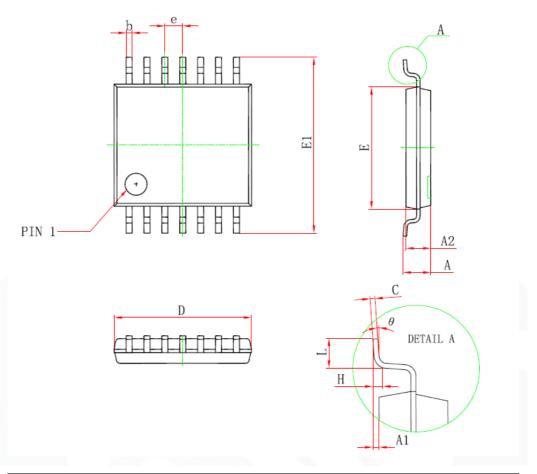




Symbol		nsions imeters	Dimensions In Inches		
-	MIN	MAX	MIN	MAX	
Α	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.350	1.550	0.053	0.061	
b	0.330	0.510	0.013	0.020	
С	0.170	0.250	0.006	0.010	
D	4.700	5.100	0.185	0.200	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
e	1.27 BSC		0.050	BSC	
L	0.400	1.270	0.016	0.050	
е	0°	8°	0°	8°	



TSSOP-14

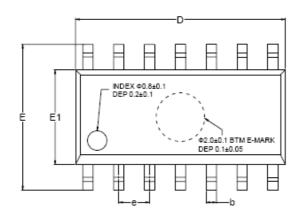


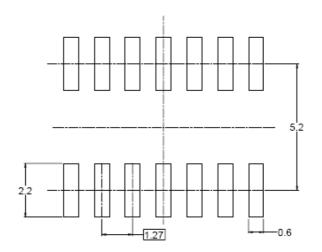
Symbol	Dimensions In	Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
D	4.900	5. 100	0.193	0. 201	
E	4.300	4. 500	0.169	0.177	
b	0.190	0.300	0.007	0.012	
С	0. 090	0.200	0.004	0.008	
E1	6.250	6. 550	0.246	0. 258	
A		1. 200		0.047	
A2	0.800	1.000	0.031	0.039	
A1	0.050	0. 150	0.002	0.006	
e	0.65 (BSC)	0.026(BSC)		
L	0.500	0. 700	0.020	0.028	
H	0.25(TYP)		0.01(TYP)	
θ	1 °	7°	1°	7°	



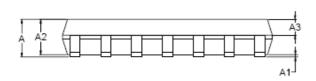


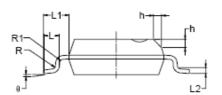
SOP-14





RECOMMENDED LAND PATTERN (Unit: mm)





Dimensions In Millimeters			Dimensions In Inches			
MIN	MOD	MAX	MIN	MOD	MAX	
1.35		1.75	0.053		0.069	
0.10		0.25	0.004		0.010	
1.25		1.65	0.049		0.065	
0.55		0.75	0.022		0.030	
0.36		0.49	0.014		0.019	
8.53		8.73	0.336		0.344	
5.80		6.20	0.228		0.244	
3.80		4.00	0.150		0.157	
1.27 BSC				0.050 BSC		
0.45		0.80	0.018		0.032	
1.04 REF				0.040 REF		
0.25 BSC				0.01 BSC		
0.07			0.003			
0.07			0.003			
0.30		0.50	0.012		0.020	
0°		8°	0°		8°	
	MIN 1.35 0.10 1.25 0.55 0.36 8.53 5.80 3.80 0.45	MIN MOD 1.35 0.10 1.25 0.55 0.36 8.53 5.80 3.80 1.27 BSC 0.45 1.04 REF 0.25 BSC 0.07 0.07 0.30	MIN MOD MAX 1.35 1.75 0.10 0.25 1.25 1.65 0.55 0.75 0.36 0.49 8.53 8.73 5.80 6.20 3.80 4.00 1.27 BSC 0.80 1.04 REF 0.25 BSC 0.07 0.07 0.30 0.50	MIN MOD MAX MIN 1.35 1.75 0.053 0.10 0.25 0.004 1.25 1.65 0.049 0.55 0.75 0.022 0.36 0.49 0.014 8.53 8.73 0.336 5.80 6.20 0.228 3.80 4.00 0.150 1.27 BSC 0.80 0.018 1.04 REF 0.25 BSC 0.003 0.07 0.003 0.003 0.30 0.50 0.012	MIN MOD MAX MIN MOD 1.35 1.75 0.053 0.004 0.10 0.25 0.004 0.049 1.25 1.65 0.049 0.014 0.36 0.49 0.014 0.036 8.53 8.73 0.336 5.80 6.20 0.228 3.80 4.00 0.150 1.27 BSC 0.050 BSC 0.45 0.80 0.018 1.04 REF 0.040 REF 0.25 BSC 0.003 0.07 0.003 0.30 0.50 0.012	

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