

# 3MHz CMOS Rail-to-Rail IO Opamps

#### **Features**

Single-Supply Operation from +2.1V ~ +5.5V

• Rail-to-Rail Input / Output

Gain-Bandwidth Product: 3MHz (Typ)
 Low Input Bias Current: 1pA (Typ)

Low Offset Voltage: 3.5mV (Max)

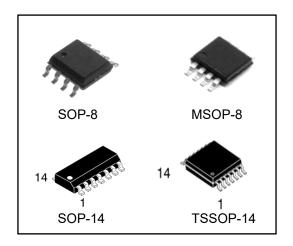
Quiescent Current: 250µA per Amplifier (Typ)

Operating Temperature: -40°C ~ +125°C

Small Package:

LMV552 Available in SOP-8 and MSOP-8 Packages

LMV554 Available in SOP-14 and TSSOP-14 Packages



## **Ordering Information**

DEVICE	Package Type	MARKING	Packing	Packing Qty
LMV552M/TR	SOP-8	LMV552	REEL	2500pcs/Reel
LMV552MM/TR	MSOP-8	LMV552,V552	REEL	3000pcs/Reel
LMV554M/TR	SOP-14	LMV554	REEL	2500pcs/Reel
LMV554MT/TR	TSSOP-14	LMV554	REEL	2500pcs/Reel

# **General Description**

The LMV552/24 have a high gain-bandwidth product of 3MHz, a slew rate of 1.66V/μs, and a quiescent current of 250μA per amplifier at 5V. The LMV552/24 are designed to provide optimal performance in low voltage and low noise systems. They provide rail-to-rail output swing into heavy loads. The input common mode voltage range includes ground, and the maximum input offset voltage is 3.5mV for LMV552/24. They are specified over the extended industrial temperature range (-40°C to+125°C). The operating range is from 2.1V to 5.5V. The LMV552 dual is available in Green SOP-8 and MSOP-8 packages. The LMV554 Quad is available in Green SOP-14 and TSSOP-14 packages.

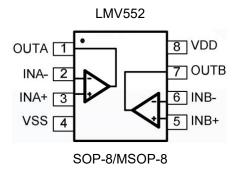
# **Applications**

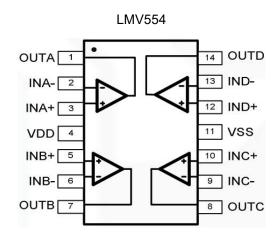
- Sensors
- Active Filters
- Cellular and Cordless Phones
- Laptops and PDAs

- Audio
- Handheld Test Equipment
- Battery-Powered Instrumentation
- A/D Converters



## **Pin Configuration**





SOP-14/TSSOP-14
Figure 1. Pin Assignment Diagram

# **Absolute Maximum Ratings**

Condition	Min	Max			
Power Supply Voltage (VDD to Vss)	-0.5V	+7.5V			
Analog Input Voltage (IN+ or IN-)	Vss-0.5V	VDD+0.5V			
PDB Input Voltage	Vss-0.5V	+7V			
Operating Temperature Range	-40°C	+125°C			
Junction Temperature	+160°C				
Storage Temperature Range	-55°C	+150°C			
Lead Temperature (soldering, 10sec)	+245°C				
Package Thermal Resistance (TA=+25°C)					
SOP-8, θJA	125°0	C/W			
MSOP-8, θJA	216°0	C/W			
ESD Susceptibility					
НВМ	8KV				
MM	400	)V			

**Note:** Stress greater than those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions outside those indicated in the operational sections of this specification are not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.



## **Electrical Characteristics**

(At Vs=5V,  $T_A$  = +25  $^{\circ}\! C$  ,  $V_{CM}$  = Vs/2,  $R_L$  = 600  $\! \Omega$  , unless otherwise noted.)

					LMV552	/4		
PARAMETER	CONDITIONS	TYP		MIN/M	AX OVER		RATURE	
PARAMETER	CONDITIONS	+25℃	+25℃	0℃ to	-40℃ to 85℃	-40 ℃ to125℃	UNITS	MIN / MAX
INPUT CHARACTERISTICS								
Input Offset Voltage (V <sub>OS</sub> )		0.8	3.5	3.9	4.3	4.6	mV	MAX
Input Bias Current (I <sub>B</sub> )		1					рА	TYP
Input Offset Current (I <sub>OS</sub> )		1					pА	TYP
Input Common Mode Voltage Range $(V_{CM})$	V <sub>S</sub> = 5.5V	-0.1 to					V	TYP
		+5.6						
Common Mode Rejection Ratio(CMRR)	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to 4V	82	65	64	64	63	dB	MIN
	$V_S = 5.5V$ , $V_{CM} = -0.1V$ to 5.6V	71					dB	MIN
Open-Loop Voltage Gain (A <sub>OL</sub> )	$R_L = 600\Omega, V_O = 0.15V \text{ to } 4.85V$	90	80	76	75	68	dB	MIN
	$R_L = 10k\Omega, V_O = 0.05V \text{ to } 4.95V$	100					dB	MIN
Input Offset Voltage Drift ( $\Delta V_{OS}/\Delta T$ )		2.4					μV/℃	TYP
OUTPUT CHARACTERISTICS								
Output Voltage Swing from Rail	$R_L = 600\Omega$	0.1					V	TYP
	$R_L = 10k\Omega$	0.015					V	TYP
Output Current (IOUT)		53	49	45	40	35	mA	MIN
Closed-Loop Output Impedance	f = 100kHz, G = 1	10					Ω	TYP
POWER-DOWN DISABLE					1	1		
Turn-On Time		4					μs	TYP
Turn-Off Time		1.2					μs	TYP
DISABLE Voltage-Off			0.8				V	MAX
DISABLE Voltage-On			2				V	MIN
POWER SUPPLY								
Operating Voltage Range			2.1	2.1	2.1	2.1	V	MIN
Power Supply Rejection Ratio (PSRR) Quiescent Current/Amplifier (I <sub>Q</sub> ) Supply Current when Disabled (LMV551N Only)	$V_S = +2.5V \text{ to } +5.5V$ $V_{CM} = (-V_S) + 0.5V$ $I_{OUT} = 0$	91 250 90	5.5 74 350	5.5 72 427	5.5 72 450	5.5 68 515	V dB μa nA	MAX MIN MAX MAX



## **Electrical Characteristics**

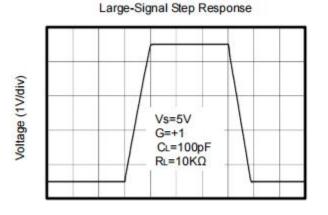
(At Vs=5V,  $T_A$  = +25°C, VCM = V<sub>S</sub>/2,  $R_L$  = 600 $\Omega$ , unless otherwise noted.)

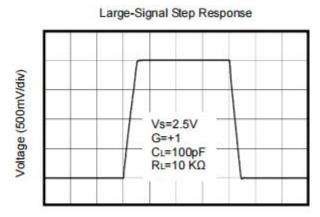
		LMV552/4									
PARAMETER	CONDITIONS	TYP	P MIN/MAX OVER TEMPERATURE								
	CONDITIONS	+25℃	+25℃	0℃ to 70℃	-40℃ to 85℃	-40℃to 125℃	UNITS	MIN / MAX			
DYNAMIC PERFORMANCE											
Gain-Bandwidth Product (GBP)	R <sub>L</sub> = 10kΩ, C <sub>L</sub> = 100pF	3					MHz	TYP			
Phase Margin (φ <sub>0</sub> )	RL = 10kΩ, CL = 100pF	50					Degrees	TYP			
Full Power Bandwidth (BWP)	$<$ 1% distortion, R <sub>L</sub> = 600 $\Omega$	50					kHz	TYP			
Slew Rate (SR)	G = +1, 2V Step, RL = 10kΩ	1.66					V/µs	TYP			
Settling Time to 0.1% (ts)	G = +1, 2V Step, R <sub>L</sub> = 600Ω	0.5					μs	TYP			
Overload Recovery Time	V <sub>IN</sub> ·Gain = VS, R <sub>L</sub> = 600Ω	4.5					μs	TYP			
NOISE PERFORMANCE											
Voltage Noise Density (e <sub>n</sub> )	f = 1kHz	18					$nV\sqrt{Hz}$	TYP			
Current Noise Density (in)	f = 1kHz	4.5					$fA\sqrt{Hz}$	TYP			



# **Typical Performance characteristics**

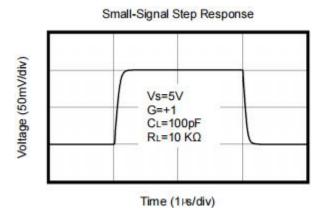
(At Vs=5V, T\_A= +25  $^{\circ}\text{C}$  , V<sub>CM</sub> = V<sub>S</sub>/2, R<sub>L</sub> = 600  $\Omega$ , unless otherwise noted.)

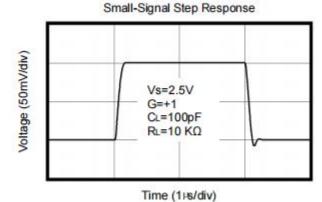












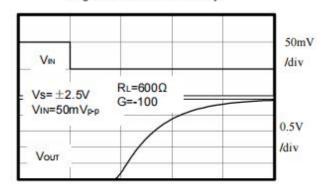


V<sub>IN</sub>

RL=600Ω
G=-100
Vs=±2.5V
V<sub>IN</sub>=50mV<sub>P-P</sub>

1V
/div

Negative Overload Recovery



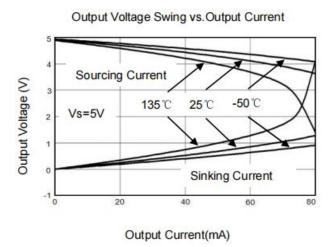
Time (5Hs/div)

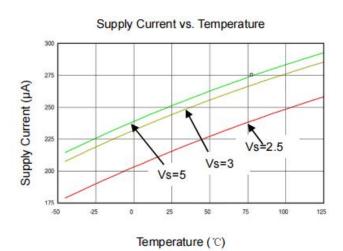
Time (5µs/div)

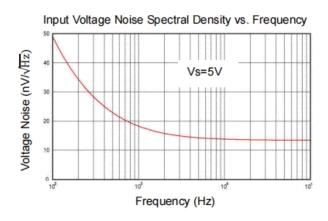


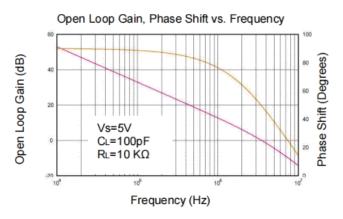
# **Typical Performance characteristics**

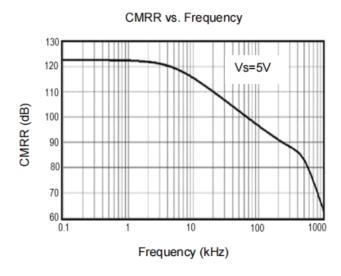
(At Vs=5V,  $T_A$ = +25°C,  $V_{CM}$  = Vs/2, RL = 600 $\Omega$ , unless otherwise noted.)

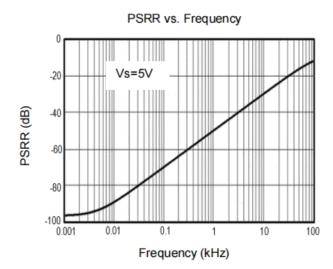














### **Application Note**

#### Size

LMV552/554 series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. The small footprints of the LMV552/554 series packages save space on printed circuit boards and enable the design of smaller electronic products.

#### **Power Supply Bypassing and Board Layout**

LMV552/554 series operates from a single 2.1V to 5.5V supply or dual  $\pm 1.05$ V to  $\pm 2.75$ V supplies. For best performance, a 0.1 $\mu$ F ceramic capacitor should be placed close to the V<sub>DD</sub> pin in single supply operation. For dual supply operation, both V<sub>DD</sub> and V<sub>SS</sub> supplies should be bypassed to ground with separate 0.1 $\mu$ F ceramic capacitors.

#### **Low Supply Current**

The low supply current (typical 250µA per channel) of LMV552/554 series will help to maximize battery life. They are ideal for battery powered systems.

#### **Operating Voltage**

LMV552/554 series operate under wide input supply voltage (2.1V to 5.5V). In addition, all temperature specifications apply from -40 $^{\circ}$ C to +125 $^{\circ}$ C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-lon battery lifetime.

#### Rail-to-Rail Input

The input common-mode range of LMV552/554 series extends 100mV beyond the supply rails ( $V_{SS}$ -0.1V to  $V_{DD}$ +0.1V). This is achieved by using complementary input stage. For normal operation, inputs should be limited to this range.

#### Rail-to-Rail Output

Rail-to-Rail output swing provides maximum possible dynamic range at the output. This is particularly important when operating in low supply voltages. The output voltage of LMV552/554 series can typically swing to less than 2mV from supply rail in light resistive loads (>100k $\Omega$ ), and 60mV of supply rail in moderate resistive loads (10k $\Omega$ ).

#### **Capacitive Load Tolerance**

The LMV552/554 family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain.

Figure 2. shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and, more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

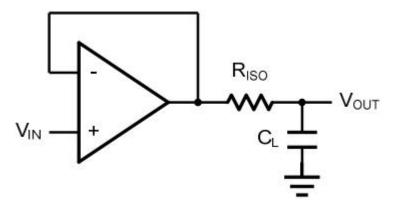


Figure 2. Indirectly Driving a Capacitive Load Using Isolation Resistor



The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load RL in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.

The circuit in Figure 3 is an improvement to the one in Figure 2.  $R_F$  provides the DC accuracy by feed-forward the VIN to  $R_L$ .  $C_F$  and  $R_{ISO}$  serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of  $C_F$ . This in turn will slow down the pulse response.

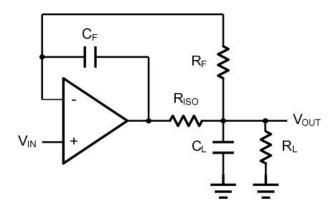


Figure 3. Indirectly Driving a Capacitive Load with DC Accuracy



### **Typical Application Circuits**

#### Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4. shown the differential amplifier using LMV552/554.

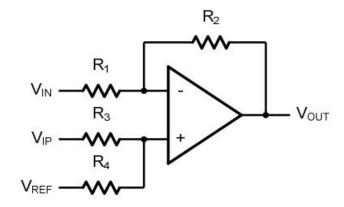


Figure 4. Differential Amplifier

$$V_{OUT} = (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_4}{R_1} V_{IN} - \frac{R2}{R1} V_{IP} + (\frac{R_1 + R_2}{R_3 + R_4}) \frac{R_3}{R_1} V_{REF}$$

If the resistor ratios are equal (i.e.  $R_1=R_3$  and  $R_2=R_4$ ), then

$$V_{OUT} = \frac{R_2}{R_1} (V_{IP} - V_{IN}) + V_{REF}$$

## **Low Pass Active Filter**

The low pass active filter is shown in Figure 5. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/decade roll-off after its corner frequency  $f_C=1/(2\pi R_3C_1)$ .

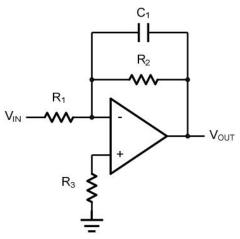


Figure 5. Low Pass Active Filter



# **Instrumentation Amplifier**

The triple LMV552/554 can be used to build a three-op-amp instrumentation amplifier as shown in Figure 6. The amplifier in Figure 6 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

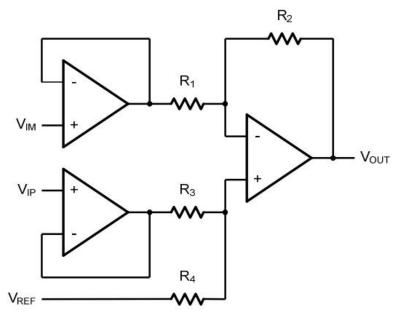
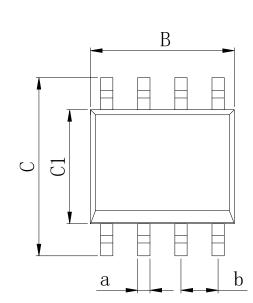


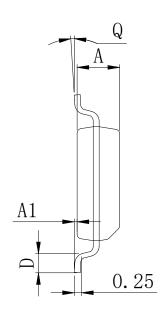
Figure 6. Instrument Amplifier



# **Physical Dimensions**

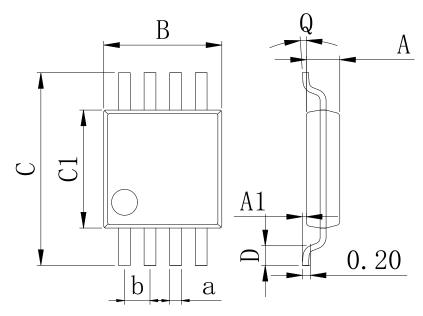
# SOP-8





Dimensions In Millimeters(SOP-8)									
Symbol:	А	A1	В	С	C1	D	Q	а	b
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	1.27 650

MSOP-8

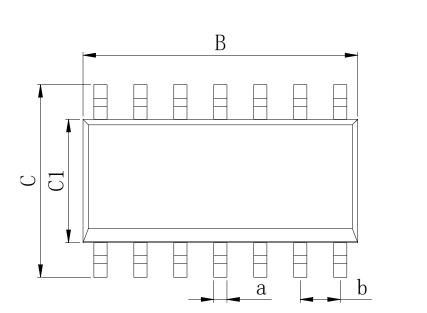


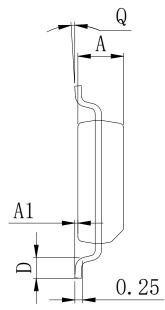
Dimensions In Millimeters(MSOP-8)									
Symbol:	Α	A1	В	С	C1	D	Q	а	b
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65.000
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	0.65 BSC



# **Physical Dimensions**

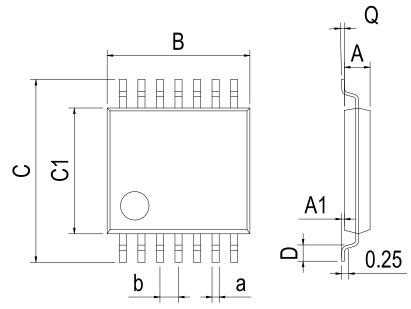
# SOP-14





Dimensions In Millimeters(SOP-14)									
Symbol:	Α	A1	В	С	C1	D	Q	а	b
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	1 07 DCC
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	1.27 BSC

TSSOP-14



Dimensions In Millimeters(TSSOP-14)									
Symbol:	Α	A1	В	С	C1	D	Q	а	b
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0.05.000
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25	0.65 BSC

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# **Revision History**

DATE	REVISION	PAGE
2015-10-6	New	1-14
2023-7-24	Update encapsulation type、Update Lead Temperature	1、2



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