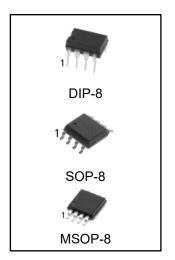


DC-to- DC Converter Control Circuits

FEATURES

- Operation from 3.0V to 40V input
- Low standby current
- Current limiting
- Output switch current up to 1.5A
- Adjustable output voltage
- Operation at frequencies up to 100kHz
- Precision reference (2%)

ORDERING INFORMATION



DEVICE	Package Type	MARKING	Packing	Packing Qty
MC33063N	DIP-8	MC33063	TUBE	2000pcs/box
MC33063M/TR	SOP-8	MC33063	REEL	2500pcs/reel
MC33063MM/TR	MSOP-8	33063	REEL	2500pcs/reel

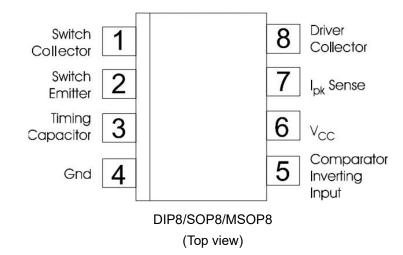
DESCRIPTION

The MC33063 series is a monolithic control circuit containing primary functions required for DC-to-DC converters.

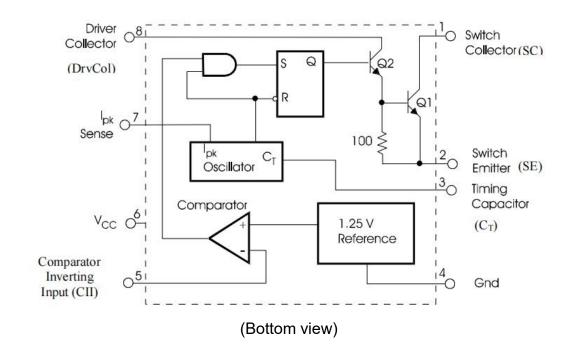
These devices consist of an internal temperature-compensated reference, comparator, controlled duty cycle oscillator withan active current limit circuit, driver and high current output switch. This series was specifically designed to be incorporated in step-down and step-up and voltage-inverting applications with a minimum number of external components.



PIN CONNECTIONS



SCHEMATIC DIAGRAM





ABSOLUTE MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Power supply voltage	Vcc	40	V
Comparator input voltage range	V _{IR}	-0.3 to +40	V
Switch collector voltage	V _{C(Switch)}	40	V
Switch emitter voltage (VPin1=40V)	V _{E(Switch)}	40	V
Switch collector-to-emitter voltage	V _{CE(Switch)}	40	V
Driver collector voltage	V _{C(Driver)}	40	V
Driver collector current (Note 1)	I _{C(Driver)}	100	mA
Switch current	I _{Sw}	1.5	А
Operating junction temperature	TJ	+150	°C
Operating ambient temperature range	T _A	-40 to +85	°C
Storage temperature range	T _{STG}	-65 to + 150	°C
ESD (HBM)		2500	V

ELECTRICAL CHARACTERISTICS (Vcc=5.0V, TA=TLow to THigh, unless otherwise specified.)

Characteristics	Symbol	Min	Тур	Max	Unit
OSCILLATOR					
Frequency (VPin5=0V, C _T =1.0nF, T _A =25°C)	fosc	24	33	42	kHz
Charge current (V _{CC} =5.0V to 40V, T_A =25°C)	Ichg	24	35	42	μA
Discharge current (V _{CC} =5.0V to 40V, T_A =25°C)	Idischg	140	220	260	μA
Discharge-to-charge current ratio (Pin7 to V_{CC} , $T_A=25^{\circ}C$)	ldischg/lchg	5.2	6.5	7.5	-
Current limit sense voltage (Ichg=Idischg, $T_A=25^{\circ}C$)	Vlpk(sense)	250	300	350	mV
OUTPUT SWITCH (Note 2)					
Saturation voltage, Darlington connection	N		1.0	1.3	V
ISw=1.0A, Pins1, 8 connected	V _{CE(sat)}	-			V
Saturation voltage, Darlington connection	V _{CE(sat)}	-	0.45	0.7	V
(ISw=1.0A, R _{Pin8} =82 Ω to V _{CC} , forced β =20)					
DC current gain (ISw=1.0A, V _{CE} =5.0, T _A =25°C)	h _{FE}	50	75	-	-
Collector off-state current (V _{CE} =40V)	I _{C(off)}	-	1.0	100	μA
COMPARATOR					
Threshold voltage		1.225	1.25	1.275	V
Threshold voltage	V _{th}	1.21	-	1.29	v
Threshold voltage line regulation(V_{CC} =3.0V to 40V)	Regline	-	1.4	5.0	mV
Input bias current(Vin=0V)	I _{IB}	-	-20	-400	nA
TOTAL DEVICE					
Supply current					
(V _{CC} =5.0V to 40V, C _T =1.0nF, Pin7=V _{CC} ,	Icc	-	-	4.0	mA
VPin5>Vth, Pin2 =Gnd, remaining pins - open					

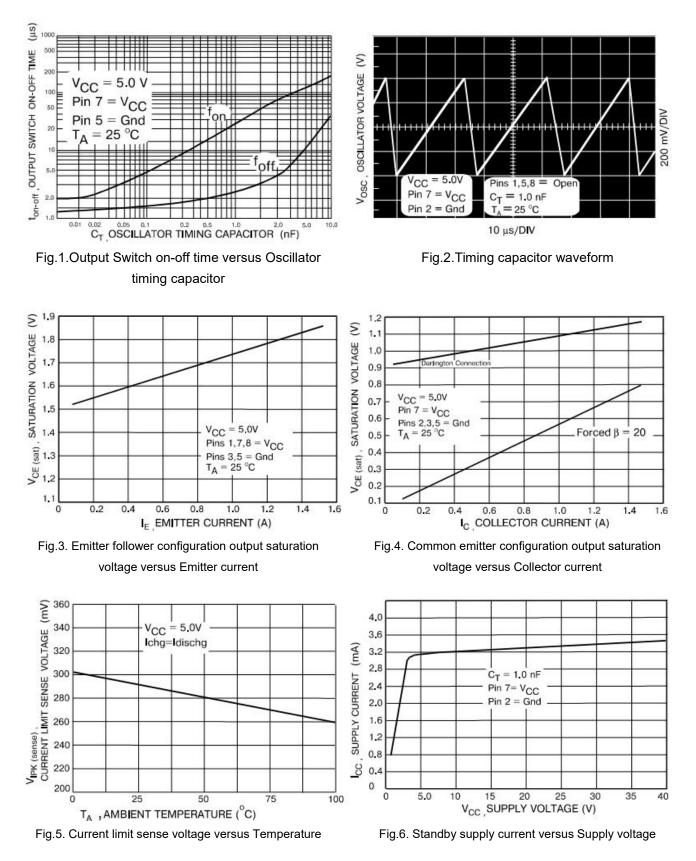
Notes:

1. Maximum package power dissipation limits must be observed.

2. Low duty cycle pulse techniques are used during the test to maintain the junction temperature as close to the ambient temperature as possible.



TYPICAL PERFORMANCE CHARACTERISTICS





APPLICATION INFORMATION

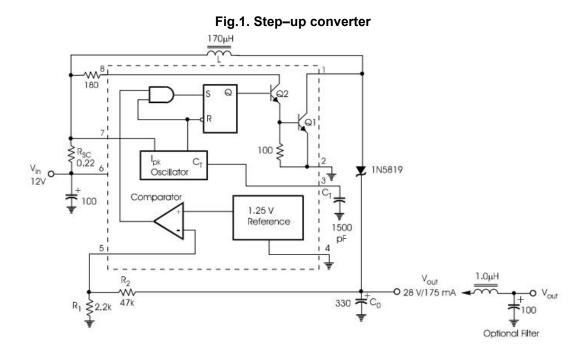


Fig.2. External current boost connections for IC Peak greater than 1.5A2a. External NPN switch2b. External NPN saturated switch

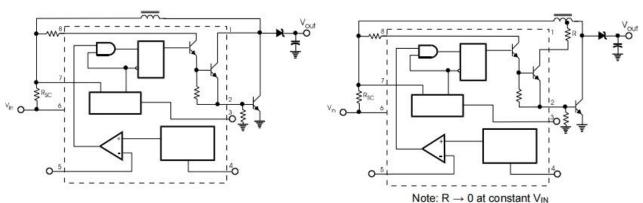




Fig.3. Step-down Converter

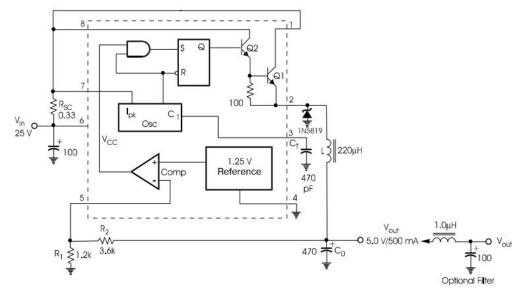
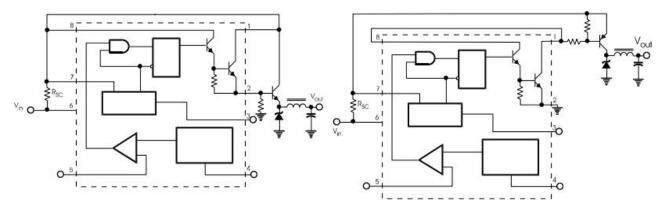


Fig.4. External current boost connections for Ic Peak greater than 1.5A4a. External NPN switch4b. External PNP saturated switch





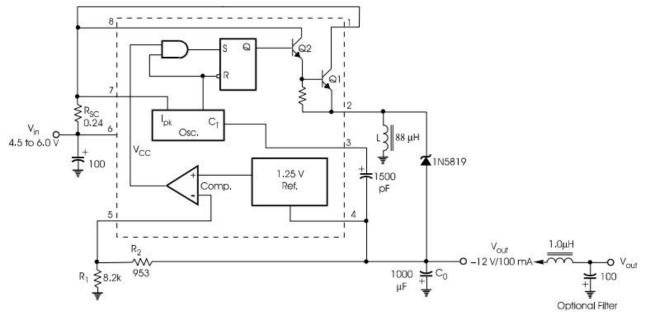
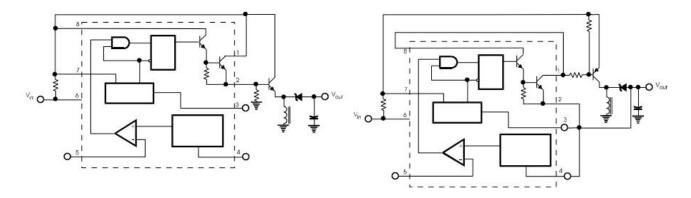


Fig.5. Voltage inverting converter

Fig.6. External current boost connections for Ic Peak greater than 1.5A6a. External NPN switch6b. External PNP saturated switch





DESIGN FORMULA

Calculation	Step-up	Step-down	Voltage-inverting		
ton	Vout + VF – Vin(min) Vin(min) – Vsat	Vout + VF Vin(min) — Vsat — Vout	Vout + VF Vin + Vsat		
(ton + toff)max	1 fmin	1 fmin	1 fmin		
Ст	4.0 x 10⁻⁵ ton	4.0 x 10 ⁻⁵ ton	4.0 x 10 ⁻⁵ ton		
I _{pk(switch)}	$2I_{out (max)} \left(\frac{ton}{toff} + 1\right)$	21 _{out(max)}	$2I_{out (max)} \left(\frac{ton}{toff} + 1\right)$		
Rsc	0.3/Ipk(Switch)	0.3/Ipk(Switch)	0.3/Ipk(Switch)		
L(min)	$\left(\frac{Vin(min)-Vsat}{Ipk (switch)}\right) \times ton(max)$	$\left(\frac{Vin(min)-Vsat-Vout}{Ipk (switch)}\right) \times ton(max)$	$\left(\frac{Vin(min)-Vsat}{Ipk (switch)}\right) \times ton(max)$		
Co	9 loutton Vripple(pp)	lpk(switch)(ton + toff) 8Vripple(pp)	9 loutton Vripple(pp)		

TERMS AND DEFINITIONS

Vsat – Saturation voltage of the output switch.

VF-Forward voltage drop of the output rectifier.

The following power supply characteristics must be chosen:

Vin-Nominal input voltage.

 V_{out} - Desired output voltage, |Vout| = 1.25 $\left(1 + \frac{R^2}{R^1}\right)$

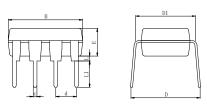
fmin – Minimum desired output switching frequency at the selected values of Vin and Iout.

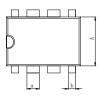
 $V_{ripple(p-p)}$ – Desired peak-to-peak output ripple voltage. In practice, the calculated capacitor value will need to be increased due to its equivalent series resistance and board layout. The ripple voltage should be kept to a low value since it will directly affect the line and load regulation.



Physical Dimensions

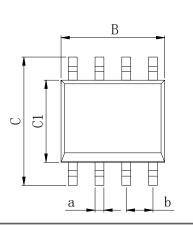
DIP8

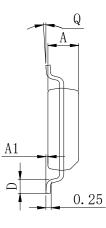




Dimensions In Millimeters(DIP8)											
Symbol:	A	В	D	D1	E	L	L1	а	b	С	d
Min:	6.10	9.00	8.40	7.42	3.10	0.50	3.00	1.50	0.85	0.40	2.54.000
Max:	6.68	9.50	9.00	7.82	3.55	0.70	3.60	1.55	0.90	0.50	2.54 BSC

SOP8 (150mil)





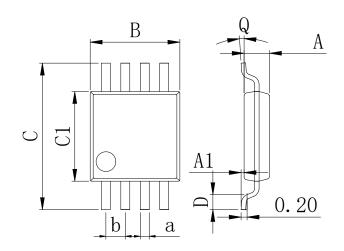
Dimensions In Millimeters(SOP8)

Symbol:	А	A1	В	С	C1	D	Q	а	b	
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1.27 BSC	
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	1.27 030	



Physical Dimensions

MSOP8



Dimensions In Millimeters(MSOP8)									
Symbol:	А	A1	В	С	C1	D	Q	а	b
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.25	0.65.000
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.35	0.65 BSC



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