International Rectifier

PD-95333A IRF3707ZPbF

IRF3707ZSPbF IRF3707ZLPbF

HEXFET® Power MOSFET

V _{DSS}	R _{DS(on)} max	Qg
30V	9.5m Ω	9.7nC

Applications

- High Frequency Synchronous Buck Converters for Computer Processor Power
- Lead-Free

Benefits

- Low R_{DS(on)} at 4.5V V_{GS}
- Ultra-Low Gate Impedance
- Fully Characterized Avalanche Voltage and Current



Absolute Maximum Ratings

	Parameter	Max.	Units V	
V _{DS}	Drain-to-Source Voltage	30		
V _{GS}	Gate-to-Source Voltage	± 20		
I _D @ T _C = 25°C	Continuous Drain Current, V _{GS} @ 10V ⑤	59⑦	А	
I _D @ T _C = 100°C	Continuous Drain Current, V _{GS} @ 10V ⑤	42 ⑦	1	
I _{DM}	Pulsed Drain Current ①	230	7	
P _D @T _C = 25°C	Maximum Power Dissipation	57	W	
P _D @T _C = 100°C	Maximum Power Dissipation	28		
	Linear Derating Factor	0.38	W/°C	
T _J	Operating Junction and	-55 to + 175	°C	
T _{STG}	Storage Temperature Range			
	Soldering Temperature, for 10 seconds	300 (1.6mm from case)		
	Mounting torque, 6-32 or M3 screw	10 lbf·in (1.1 N·m)		

Thermal Resistance

	Parameter	Тур.	Max.	Units
$R_{\theta JC}$	Junction-to-Case		2.653	°C/W
$R_{\theta CS}$	Case-to-Sink, Flat Greased Surface ③	0.50		
$R_{\theta JA}$	Junction-to-Ambient ③		62	
$R_{\theta JA}$	Junction-to-Ambient (PCB Mount) ®		40	

Notes ① through ② are on page 12

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Static @ T_J = 25°C (unless otherwise specified)

	Parameter	Min.	Тур.	Max.	Units	Conditions
BV _{DSS}	Drain-to-Source Breakdown Voltage	30			V	$V_{GS} = 0V, I_D = 250\mu A$
$\Delta BV_{DSS}/\Delta T_{J}$	Breakdown Voltage Temp. Coefficient		0.023		mV/°C	Reference to 25°C, I _D = 1mA
R _{DS(on)}	Static Drain-to-Source On-Resistance		7.5	9.5	mΩ	V _{GS} = 10V, I _D = 21A ③
			10	12.5		V _{GS} = 4.5V, I _D = 17A ③
V _{GS(th)}	Gate Threshold Voltage	1.35	1.80	2.25	V	$V_{DS} = V_{GS}$, $I_D = 25\mu A$
$\Delta V_{GS(th)}/\Delta T_{J}$	Gate Threshold Voltage Coefficient		-5.3		mV/°C	
I _{DSS}	Drain-to-Source Leakage Current	_		1.0	μΑ	$V_{DS} = 24V$, $V_{GS} = 0V$
				150		$V_{DS} = 24V, V_{GS} = 0V, T_{J} = 125^{\circ}C$
I _{GSS}	Gate-to-Source Forward Leakage	_		100	nA	V _{GS} = 20V
	Gate-to-Source Reverse Leakage			-100		V _{GS} = -20V
gfs	Forward Transconductance	81			S	$V_{DS} = 15V, I_D = 17A$
Q_g	Total Gate Charge		9.7	15		
Q _{gs1}	Pre-Vth Gate-to-Source Charge		2.8			$V_{DS} = 15V$
Q _{gs2}	Post-Vth Gate-to-Source Charge		1.0		nC	$V_{GS} = 4.5V$
Q_{gd}	Gate-to-Drain Charge		3.4			$I_D = 17A$
Q_{godr}	Gate Charge Overdrive		2.5			See Fig. 16
Q _{sw}	Switch Charge (Q _{gs2} + Q _{gd})		4.4			
Q _{oss}	Output Charge		6.2		nC	$V_{DS} = 16V$, $V_{GS} = 0V$
t _{d(on)}	Turn-On Delay Time		9.8			V _{DD} = 15V, V _{GS} = 4.5V ③
t _r	Rise Time		41			I _D = 17A
t _{d(off)}	Turn-Off Delay Time		12		ns	Clamped Inductive Load
t _f	Fall Time		3.6			
C _{iss}	Input Capacitance		1210			$V_{GS} = 0V$
C _{oss}	Output Capacitance		260		рF	$V_{DS} = 15V$
C _{rss}	Reverse Transfer Capacitance		130			f = 1.0MHz

Avalanche Characteristics

	Parameter	Тур.	Max.	Units
E _{AS}	Single Pulse Avalanche Energy ^②		40	mJ
I _{AR}	Avalanche Current ①		23	Α
E _{AR}	Repetitive Avalanche Energy ①		5.7	mJ

Diode Characteristics

	Parameter	Min.	Тур.	Max.	Units	Conditions
Is	Continuous Source Current			59⑦		MOSFET symbol
	(Body Diode)				Α	showing the
I _{SM}	Pulsed Source Current			230		integral reverse
	(Body Diode) ①					p-n junction diode.
V_{SD}	Diode Forward Voltage			1.0	V	$T_J = 25^{\circ}C, I_S = 17A, V_{GS} = 0V$ ③
t _{rr}	Reverse Recovery Time		14	21	ns	$T_J = 25^{\circ}C, I_F = 17A, V_{DD} = 15V$
Q _{rr}	Reverse Recovery Charge		5.2	7.8	nC	di/dt = 100A/µs ③

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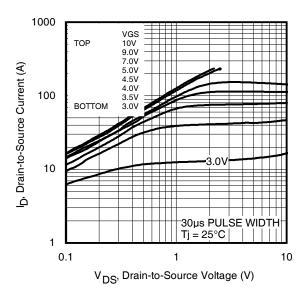


Fig 1. Typical Output Characteristics

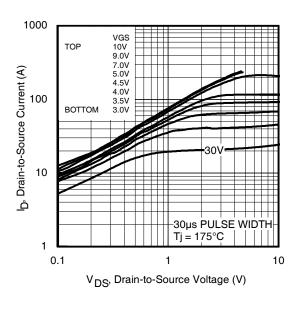


Fig 2. Typical Output Characteristics

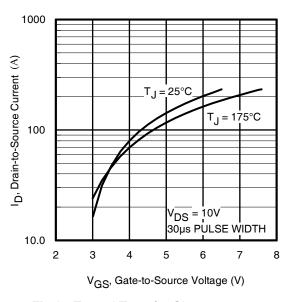


Fig 3. Typical Transfer Characteristics

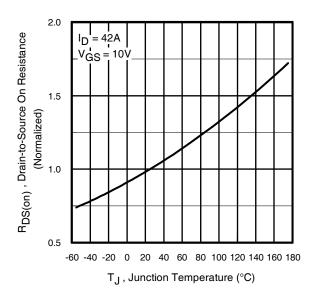


Fig 4. Normalized On-Resistance vs. Temperature

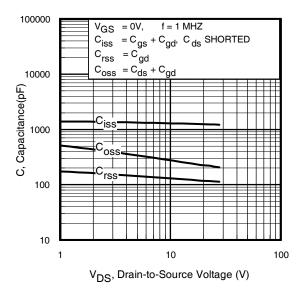


Fig 5. Typical Capacitance vs. Drain-to-Source Voltage

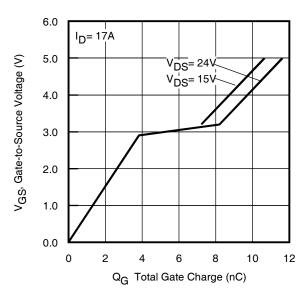


Fig 6. Typical Gate Charge vs. Gate-to-Source Voltage

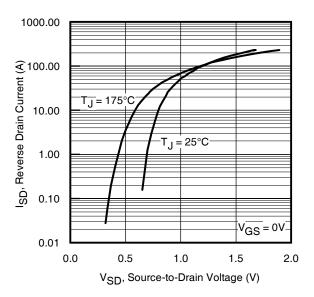


Fig 7. Typical Source-Drain Diode Forward Voltage

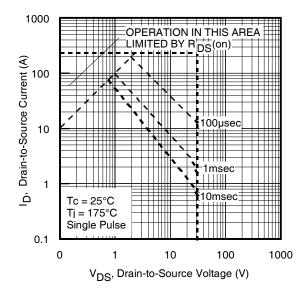
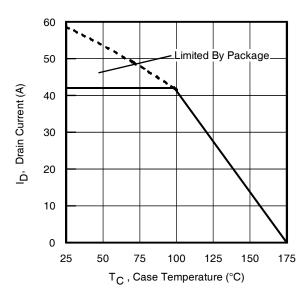


Fig 8. Maximum Safe Operating Area



2.5 (A) 0.5 (D) 0.5 (

Fig 9. Maximum Drain Current vs. Case Temperature

Fig 10. Threshold Voltage vs. Temperature

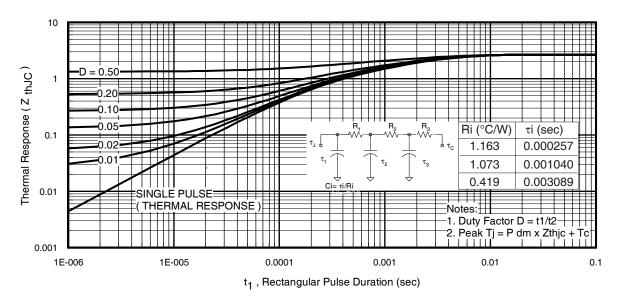


Fig 11. Maximum Effective Transient Thermal Impedance, Junction-to-Case

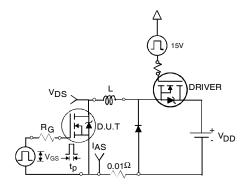


Fig 12a. Unclamped Inductive Test Circuit

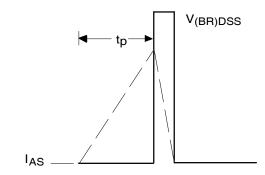


Fig 12b. Unclamped Inductive Waveforms

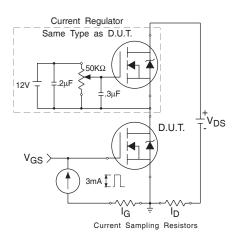


Fig 13. Gate Charge Test Circuit

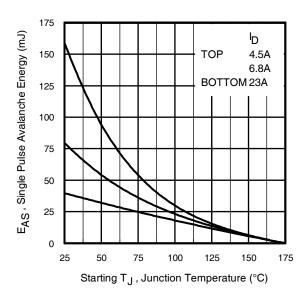


Fig 12c. Maximum Avalanche Energy vs. Drain Current

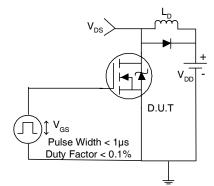


Fig 14a. Switching Time Test Circuit

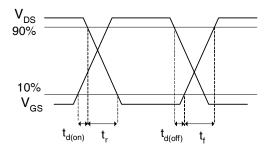


Fig 14b. Switching Time Waveforms www.irf.com

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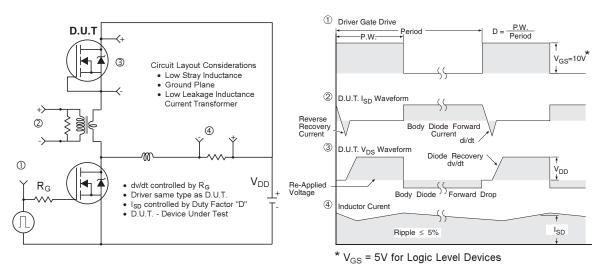


Fig 15. Peak Diode Recovery dv/dt Test Circuit for N-Channel HEXFET® Power MOSFETs

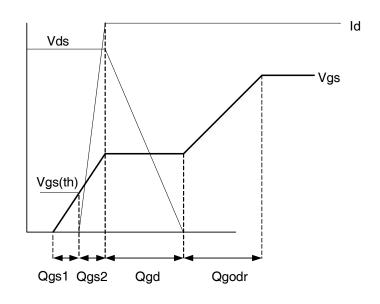


Fig 16. Gate Charge Waveform

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Power MOSFET Selection for Non-Isolated DC/DC Converters

Control FET

Special attention has been given to the power losses in the switching elements of the circuit - Q1 and Q2. Power losses in the high side switch Q1, also called the Control FET, are impacted by the $R_{\rm ds(on)}$ of the MOSFET, but these conduction losses are only about one half of the total losses.

Power losses in the control switch Q1 are given by;

$$P_{loss} = P_{conduction} + P_{switching} + P_{drive} + P_{output}$$

This can be expanded and approximated by;

$$\begin{split} P_{loss} &= \left(I_{rms}^{2} \times R_{ds(on)}\right) \\ &+ \left(I \times \frac{Q_{gd}}{i_{g}} \times V_{in} \times f\right) + \left(I \times \frac{Q_{gs2}}{i_{g}} \times V_{in} \times f\right) \\ &+ \left(Q_{g} \times V_{g} \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) \end{split}$$

This simplified loss equation includes the terms ${\rm Q_{gs2}}$ and ${\rm Q_{oss}}$ which are new to Power MOSFET data sheets.

 Q_{gs2} is a sub element of traditional gate-source charge that is included in all MOSFET data sheets. The importance of splitting this gate-source charge into two sub elements, Q_{gs1} and Q_{gs2} , can be seen from Fig 16.

 Q_{gs2} indicates the charge that must be supplied by the gate driver between the time that the threshold voltage has been reached and the time the drain current rises to I_{dmax} at which time the drain voltage begins to change. Minimizing Q_{gs2} is a critical factor in reducing switching losses in Q1.

 $\rm Q_{oss}$ is the charge that must be supplied to the output capacitance of the MOSFET during every switching cycle. Figure A shows how $\rm Q_{oss}$ is formed by the parallel combination of the voltage dependant (nonlinear) capacitance's $\rm C_{ds}$ and $\rm C_{dg}$ when multiplied by the power supply input buss voltage.

Synchronous FET

The power loss equation for Q2 is approximated by;

$$\begin{split} P_{loss} &= P_{conduction} + P_{drive} + P_{output}^* \\ P_{loss} &= \left(I_{rms}^2 \times R_{ds(on)}\right) \\ &+ \left(Q_g \times V_g \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) + \left(Q_{rr} \times V_{in} \times f\right) \end{split}$$

*dissipated primarily in Q1.

For the synchronous MOSFET Q2, $R_{\text{ds(on)}}$ is an important characteristic; however, once again the importance of gate charge must not be overlooked since it impacts three critical areas. Under light load the MOSFET must still be turned on and off by the control IC so the gate drive losses become much more significant. Secondly, the output charge Q_{oss} and reverse recovery charge Q_{rr} both generate losses that are transfered to Q1 and increase the dissipation in that device. Thirdly, gate charge will impact the MOSFETs' susceptibility to Cdv/dt turn on.

The drain of Q2 is connected to the switching node of the converter and therefore sees transitions between ground and $V_{\rm in}.$ As Q1 turns on and off there is a rate of change of drain voltage dV/dt which is capacitively coupled to the gate of Q2 and can induce a voltage spike on the gate that is sufficient to turn the MOSFET on, resulting in shoot-through current . The ratio of $Q_{\rm gd}/Q_{\rm gs1}$ must be minimized to reduce the potential for Cdv/dt turn on.

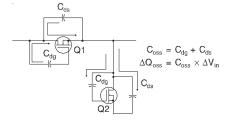


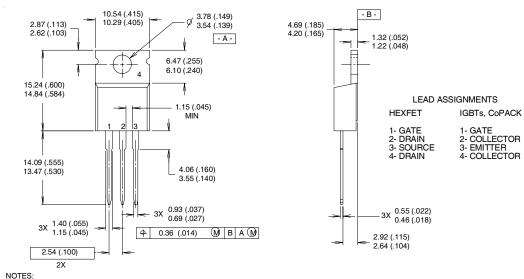
Figure A: Qoss Characteristic

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TO-220AB Package Outline

Dimensions are shown in millimeters (inches)



- 1 DIMENSIONING & TOLERANCING PER ANSI Y14.5M. 1982.
- 2 CONTROLLING DIMENSION: INCH

- 3 OUTLINE CONFORMS TO JEDEC OUTLINE TO-220AB.
- ${\tt 4\,\,\, HEATSINK\,\&\, LEAD\,\, MEASUREMENTS\,\, DO\,\, NOT\,\, INCLUDE\,\, BURRS.}$

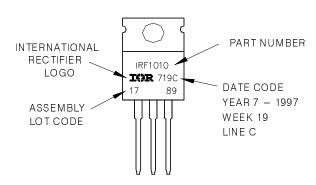
TO-220AB Part Marking Information

EXAMPLE: THIS IS AN IRF1010

LOT CODE 1789

ASSEMBLED ON WW 19, 1997 IN THE ASSEMBLY LINE "C"

Note: "P" in assembly line position indicates "Lead-Free"



Note: For the most current drawing please refer to IR website at http://www.irf.com/package/ www.irf.com

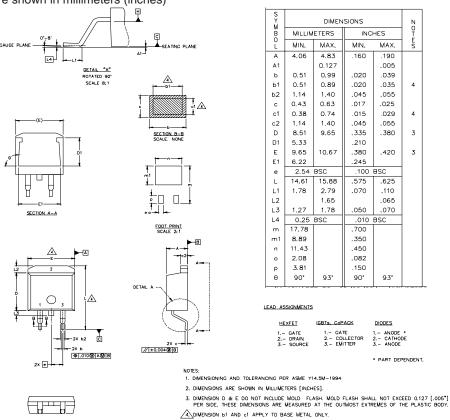
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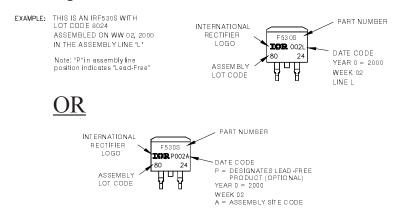
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D²Pak Package Outline

Dimensions are shown in millimeters (inches)



D²Pak Part Marking Information



5. CONTROLLING DIMENSION; INCH.

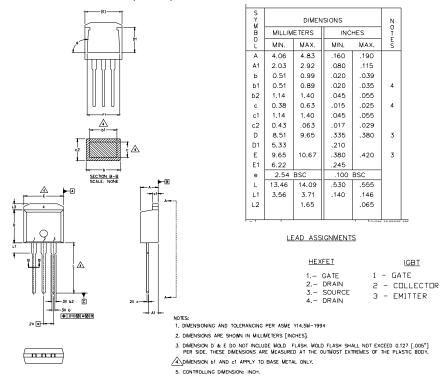
Note: For the most current drawing please refer to IR website at http://www.irf.com/package/

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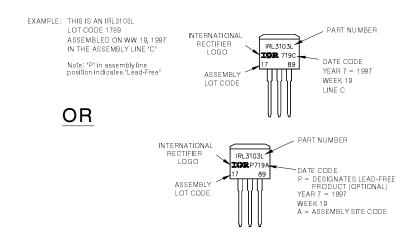
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TO-262 Package Outline

Dimensions are shown in millimeters (inches)



TO-262 Part Marking Information



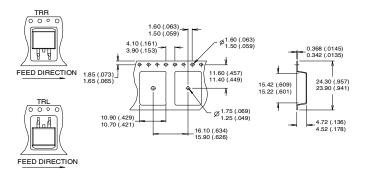
Note: For the most current drawing please refer to IR website at http://www.irf.com/package/ WWW.irf.com

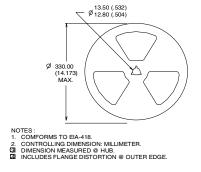
International

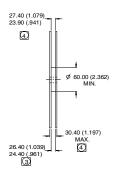
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D²Pak Tape & Reel Information

Dimensions are shown in millimeters (inches)







Notes:

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- ① Repetitive rating; pulse width limited by max. junction temperature.
- \bigcirc Starting T_J = 25°C, L = 0.15mH, R_G = 25Ω, I_{AS} = 23A.
- ③ Pulse width $\leq 400\mu s$; duty cycle $\leq 2\%$.
- $\ \ \,$ C_{oss} eff. is a fixed capacitance that gives the same charging time as C_{oss} while V_{DS} is rising from 0 to 80% V_{DSS}.
- ⑤ This is only applied to TO-220AB pakcage.
- ® This is applied to D²Pak, when mounted on 1" square PCB (FR-4 or G-10 Material). For recommended footprint and soldering techniques refer to application note #AN-994.
- O Calculated continuous current based on maximum allowable junction temperature. Package limitation current is 42A.

TO-220AB package is not recommended for Surface Mount Application.

Note: For the most current drawing please refer to IR website at http://www.irf.com/package/

Data and specifications subject to change without notice. This product has been designed and qualified for the Industrial market.

Qualification Standards can be found on IR's Web site.



IR WORLD HEADQUARTERS: 233 Kansas St., El Segundo, California 90245, USA Tel: (310) 252-7105

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单击下面可查看定价,库存,交付和生命周期等信息

>>Infineon(英飞凌)