

# 1A LOW DROPOUT POSITIVE ADJUSTABLE REGULATOR

#### **FEATURES**

- Guaranteed < 1.3V Dropout at Full Load Current
- Fast Transient Response
- 1% Voltage Reference Initial Accuracy
- Built-In Thermal Shutdown
- Available in SOT-223, D-Pak, Ultra Thin-Pak<sup>™</sup> and 8-Pin SOIC Surface-Mount Packages

#### **APPLICATIONS**

- VGA & Sound Card Applications
- Low Voltage High Speed Termination Applications
- Standard 3.3V Chip Set and Logic Applications

#### **DESCRIPTION**

The IRU1010 is a low dropout, three-terminal adjustable regulator with minimum of 1A output current capability. This product is specifically designed to provide well regulated supply for low voltage IC applications such as high speed bus termination and low current 3.3V logic supply. The IRU1010 is also well suited for other applications such as VGA and sound cards. The IRU1010 is guaranteed to have <1.3V dropout at full load current making it ideal to provide well regulated outputs of 2.5V to 3.6V with 4.75V to 7V input supply.

#### TYPICAL APPLICATION

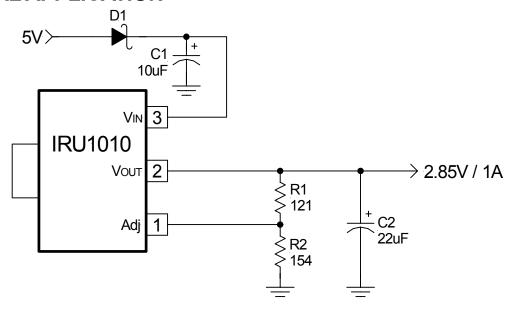


Figure 1 - Typical application of IRU1010 in a 5V to 2.85V SCSI termination regulator.

#### PACKAGE ORDER INFORMATION

#### **Basic Part (Non-Lead Free)**

T <sub>J</sub> (°C)	2-PIN PLASTIC	2-PIN PLASTIC	8-PIN PLASTIC	3-PIN PLASTIC
	TO-252 (D-Pak)	Ultra Thin-Pak™ (P)	SOIC (S)	SOT-223 (Y)
0 To 150	IRU1010CD	IRU1010CP	IRU1010CS	IRU1010CY

#### Leadfree Part

T <sub>J</sub> (°C)	2-PIN PLASTIC	2-PIN PLASTIC	8-PIN PLASTIC	3-PIN PLASTIC
	TO-252 (D-Pak)	Ultra Thin-Pak™ (P)	SOIC (S)	SOT-223 (Y)
0 To 150	Not available	Not available	IRU1010CSPbF	Not available

## IRU1010(PbF)

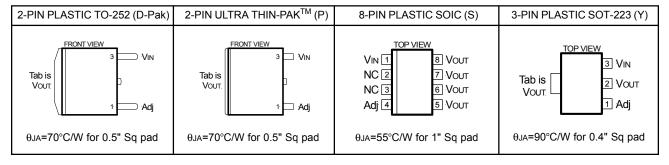


#### **ABSOLUTE MAXIMUM RATINGS**

Input Voltage (V<sub>IN</sub>) .......7V

Power Dissipation Internally Limited
Storage Temperature Range -65°C To 150°C
Operating Junction Temperature Range 0°C To 150°C

#### PACKAGE INFORMATION



#### **ELECTRICAL SPECIFICATIONS**

Unless otherwise specified, these specifications apply over  $C_{IN}=1\mu F$ ,  $C_{OUT}=10\mu F$ , and  $T_{J}=0$  to 150°C. Typical values refer to  $T_{J}=25$ °C.

PARAMETER	SYM	TEST CONDITION	MIN	TYP	MAX	UNITS
Reference Voltage	V <sub>REF</sub>	Io=10mA, T <sub>J</sub> =25°C, (V <sub>IN</sub> -Vo)=1.5V		1.250	1.262	V
		Io=10mA, (V <sub>IN</sub> -Vo)=1.5V	1.225	1.250	1.275	
Line Regulation		lo=10mA, 1.3V<(V <sub>IN</sub> -Vo)<7V			0.2	%
Load Regulation (Note 1)		V <sub>IN</sub> =3.3V, V <sub>ADJ</sub> =0, 10mA <lo<1a< td=""><td></td><td></td><td>0.4</td><td>%</td></lo<1a<>			0.4	%
Dropout Voltage (Note 2)	ΔVo	Note 2, Io=1A		1.1	1.3	V
Current Limit		V <sub>IN</sub> =3.3V, ΔVo=100mV	1.1			Α
Minimum Load Current (Note 3)		VIN=3.3V, VADJ=0V		5	10	mA
Thermal Regulation		30ms Pulse, V <sub>IN</sub> -Vo=3V, Io=1A		0.01	0.02	%/W
Ripple Rejection		f=120Hz, Co=25μF Tantalum,				
		Io=0.5A, V <sub>IN</sub> -Vo=3V	60	70		dB
Adjust Pin Current	Iadj	Io=10mA, V <sub>IN</sub> -Vo=1.5V, T <sub>J</sub> =25°C,				
		Io=10mA, V <sub>IN</sub> -Vo=1.5V		55	120	μΑ
Adjust Pin Current Change		Io=10mA, V <sub>IN</sub> -Vo=1.5V, T <sub>J</sub> =25°C		0.2	5	μΑ
Temperature Stability		Vin=3.3V, Vadj=0V, Io=10mA		0.5		%
Long Term Stability		T <sub>J</sub> =125°C, 1000Hrs		0.3	1	%
RMS Output Noise		T <sub>J</sub> =25°C, 10Hz <f<10khz< td=""><td></td><td>0.003</td><td></td><td>%Vo</td></f<10khz<>		0.003		%Vo

**Note 1:** Low duty cycle pulse testing with Kelvin connections is required in order to maintain accurate data.

**Note 2:** Dropout voltage is defined as the minimum differential voltage between  $V_{\text{IN}}$  and  $V_{\text{OUT}}$  required to maintain regulation at  $V_{\text{OUT}}$ . It is measured when the output voltage drops 1% below its nominal value.

**Note 3:** Minimum load current is defined as the minimum current required at the output in order for the output voltage to maintain regulation. Typically, the resistor dividers are selected such that it automatically maintains this current.



#### **PIN DESCRIPTIONS**

PIN#	PIN SYMBOL	PIN DESCRIPTION
1	Adj	A resistor divider from this pin to the V <sub>OUT</sub> pin and ground sets the output voltage.
2	Vоит	The output of the regulator. A minimum of $10\mu F$ capacitor must be connected from this pin to ground to insure stability.
3	VIN	The input pin of the regulator. Typically a large storage capacitor is connected from this pin to ground to insure that the input voltage does not sag below the minimum dropout voltage during the load transient response. This pin must always be 1.3V higher than Vout in order for the device to regulate properly.

### **BLOCK DIAGRAM**

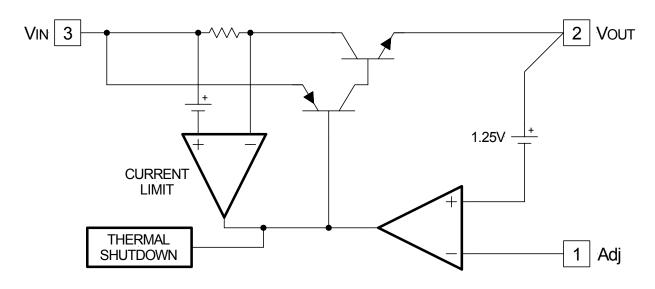


Figure 2 - Simplified block diagram of the IRU1010.

#### **APPLICATION INFORMATION**

#### Introduction

The IRU1010 adjustable Low Dropout (LDO) regulator is a three-terminal device which can easily be programmed with the addition of two external resistors to any voltages within the range of 1.25 to 5.5V. This regulator, unlike the first generation of the three-terminal regulators such as LM117 that required 3V differential between the input and the regulated output, only needs 1.3V differential to maintain output regulation. This is a key requirement for today's low voltage IC applications that typically need 3.3V supply and are often generated from the 5V supply. Other applications such as high speed

memory termination need to switch the load current from zero to full load in tens of nanoseconds at their pins, which translates to an approximately 300 to 500ns current step at the regulator. In addition, the output voltage tolerances are sometimes tight and they include the transient response as part of the specification.

The IRU1010 is specifically designed to meet the fast current transient needs as well as providing an accurate initial voltage, reducing the overall system cost with the need for fewer output capacitors.

#### **Output Voltage Setting**

The IRU1010 can be programmed to any voltages in the range of 1.25V to 5.5V with the addition of R1 and R2 external resistors according to the following formula:

$$V_{OUT} = V_{REF} \times \left(1 + \frac{R2}{R1}\right) + I_{ADJ} \times R2$$

Where:

V<sub>REF</sub> = 1.25V Typically I<sub>ADJ</sub> = 50μA Typically R1 and R2 as shown in Figure 3:

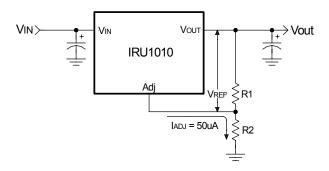


Figure 3 - Typical application of the IRU1010 for programming the output voltage.

The IRU1010 keeps a constant 1.25V between the output pin and the adjust pin. By placing a resistor R1 across these two pins a constant current flows through R1, adding to the IADJ current and into the R2 resistor producing a voltage equal to the  $(1.25/R1) \times R2 + I_{ADJ} \times R2$  which will be added to the 1.25V to set the output voltage. This is summarized in the above equation. Since the minimum load current requirement of the IRU1010 is 10mA, R1 is typically selected to be  $121\Omega$  resistor so that it automatically satisfies the minimum current requirement. Notice that since IADJ is typically in the range of 50 µA it only adds a small error to the output voltage and should only be considered when a very precise output voltage setting is required. For example, in a typical 3.3V application where R1=121 $\Omega$  and R2=200 $\Omega$  the error due to lad is only 0.3% of the nominal set point.

#### **Load Regulation**

Since the IRU1010 is only a three-terminal device, it is not possible to provide true remote sensing of the output voltage at the load. Figure 4 shows that the best load regulation is achieved when the bottom side of R2 is connected to the load and the top side of R1 resistor is connected directly to the case or the Vout pin of the regulator and not to the load. In fact, if R1 is connected

to the load side, the effective resistance between the regulator and the load is gained up by the factor of (1+ R2/R1), or the effective resistance will be  $R_{P(eff)}=R_P\times(1+R2/R1)$ . It is important to note that for high current applications, this can represent a significant percentage of the overall load regulation and one must keep the path from the regulator to the load as short as possible to minimize this effect.

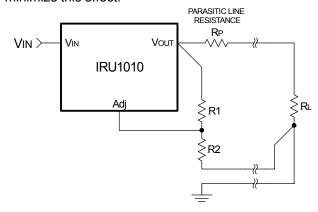


Figure 4 - Schematic showing connection for best load regulation.

#### **Stability**

The IRU1010 requires the use of an output capacitor as part of the frequency compensation in order to make the regulator stable. Typical designs for microprocessor applications use standard electrolytic capacitors with a typical ESR in the range of 50 to  $100m\Omega$  and an output capacitance of 500 to  $1000\mu\text{F}$ . Fortunately as the capacitance increases, the ESR decreases resulting in a fixed RC time constant. The IRU1010 takes advantage of this phenomena in making the overall regulator loop stable. For most applications a minimum of  $100\mu\text{F}$  aluminum electrolytic capacitor such as Sanyo MVGX series, Panasonic FA series as well as the Nichicon PL series insures both stability and good transient response.

#### **Thermal Design**

The IRU1010 incorporates an internal thermal shutdown that protects the device when the junction temperature exceeds the maximum allowable junction temperature. Although this device can operate with junction temperatures in the range of 150°C, it is recommended that the selected heat sink be chosen such that during maximum continuous load operation the junction temperature is kept below this number. The next example for a SCSI terminator application shows the steps in sellecting the proper regulator in a surface-mount package. (See IRU1015 for non-surface-mount packages)

Assuming the following specifications:

 $V_{IN}$  = 5V  $V_F$  = 0.5V  $V_{OUT}$  = 2.85V  $I_{OUT(MAX)}$  = 0.8A  $T_A$  = 35°C

Where: V<sub>F</sub> is the forward voltage drop of the D1 diode as shown in Figure 5.

The steps for selecting the right package with proper board area for heatsinking to keep the junction temperature below 135°C is given as:

1) Calculate the maximum power dissipation using:

$$P_D = I_{OUT} \times (V_{IN} - V_F - V_{OUT})$$
  
 $P_D = 0.8 \times (5 - 0.5 - 2.85) = 1.32W$ 

2) Calculate the maximum  $\theta_{JA}$  allowed for our example:

$$\theta_{\text{JA(MAX)}} = \frac{T_{\text{J}} - T_{\text{A}}}{P_{\text{D}}} = \frac{135 - 35}{1.32} = 75.6^{\circ}\text{C/W}$$

3) Select a package from the datasheet with lower  $\theta_{\text{JA}}$  than the one calculated in the previous step.

Selecting TO-252 (D-Pak) with at least 0.5" square of 0.062" FR4 board using 1 oz. copper has 70°C/W which is lower than the calculated number.

To set the output DC voltage, we need to select R1 and R2:

4) Assuming R1 =  $121\Omega$ , 1%:

$$R_2 = \left(\frac{V_{OUT}}{V_{REF}} - 1\right) \times R_1 = \left(\frac{2.85}{1.25} - 1\right) \times 121 = 154.8\Omega$$

Select R2 =  $154\Omega$ , 1%.

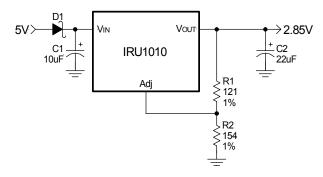
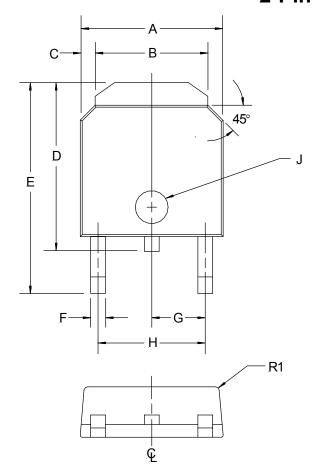


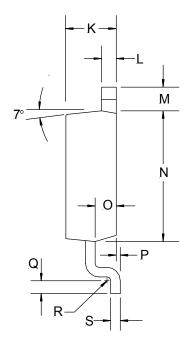
Figure 5 - Final Schematic for half of the GTL+ termination regulator.

#### **Layout Consideration**

The output capacitors must be located as close to the  $V_{\text{OUT}}$  terminal of the device as possible. It is recommended to use a section of a layer of the PC board as a plane to connect the  $V_{\text{OUT}}$  pin to the output capacitors to prevent any high frequency oscillation that may result due to excessive trace inductance.

(D) TO-252 Package 2-Pin

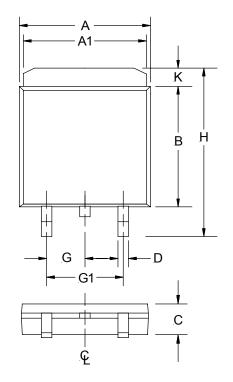


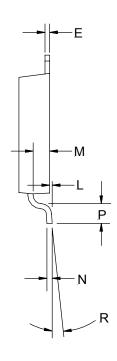


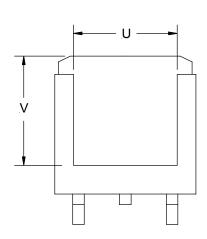
SYMBOL	MIN	MAX	
Α	6.477	6.731	
В	5.004	5.207	
С	0.686	0.838	
D	7.417	8.179	
Е	9.703	10.084	
F	0.635	0.889	
G	2.286	BSC	
Н	4.521	4.623	
J	Ø1.52	Ø1.62	
K	2.184	2.388	
L	0.762	0.864	
М	1.016	1.118	
N	5.969	6.223	
0	1.016	1.118	
Р	0	0.102	
Q	0.534	0.686	
R	R0.31 TYP		
R1	R0.51 TYP		
S	0.428	0.588	

NOTE: ALL MEASUREMENTS ARE IN MILLIMETERS.

# (P) Ultra Thin-Pak™ 2-Pin



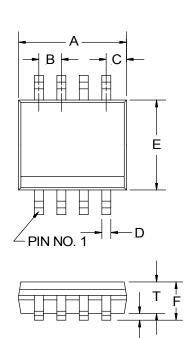




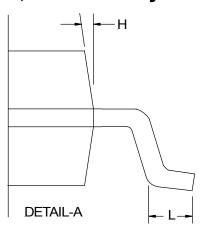
SYMBOL	MIN	MAX	
Α	5.91	6.17	
A1	5.54	5.79	
В	6.02	6.27	
С	1.70	2.03	
D	0.63	0.79	
Е	0.17	0.33	
G	2.16	2.41	
G1	4.45	4.70	
Η	9.42	9.68	
K	0.76	1.27	
L	0.02	0.13	
М	0.89	1.14	
N	0.25	0.25	
Р	0.94	1.19	
R	2°	6°	
U	2.92	3.30	
V	5.08 NOM		

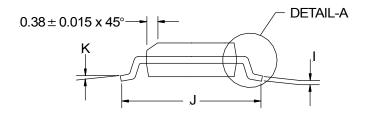
NOTE: ALL MEASUREMENTS ARE IN MILLIMETERS.

# (S) SOIC Package 8-Pin Surface Mount, Narrow Body



G

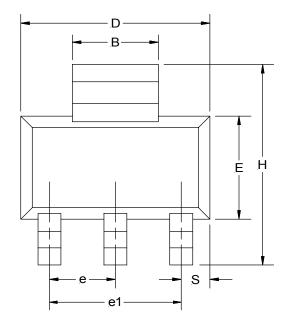




8-PIN					
SYMBOL	MIN	MAX			
Α	4.80	4.98			
В	1.27	BSC			
С	0.53	REF			
D	0.36	0.46			
Е	3.81	3.99			
F	1.52	1.72			
G	0.10	0.25			
Н	7° E	3SC			
I	0.19	0.25			
J	5.80	6.20			
K	0°	8°			
L	0.41	1.27			
Т	1.37	1.57			

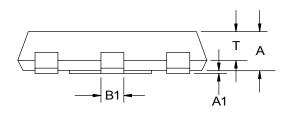
NOTE: ALL MEASUREMENTS ARE IN MILLIMETERS.

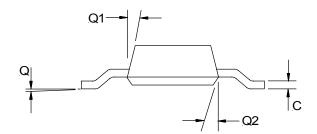
# (Y) SOT-223 Package 3-Pin



SYMBOL	MIN	MAX	
Α	1.498	1.702	
A1	0.02	0.11	
В	2.895	3.15	
B1	0.637	0.85	
С	0.239	0.381	
D	6.299	6.706	
Ш	3.30	3.708	
е	2.209	2.953	
e1	4.496	4.699	
Ι	6.70	7.30	
Q	0°	10°	
Q1	7°	16°	
Q2	7°	16°	
S	0.838	1.05	
T	1.092	1.30	

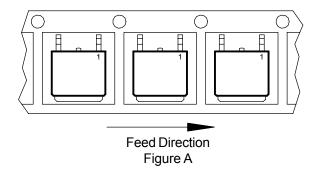
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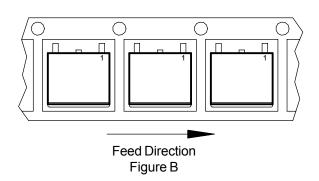


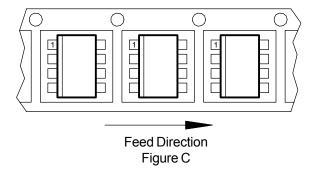


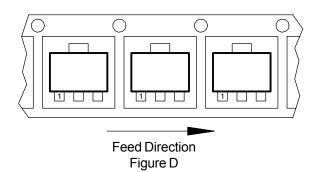
#### PACKAGE SHIPMENT METHOD

PKG	PACKAGE	PIN	PARTS	PARTS	T&R
DESIG	DESCRIPTION	COUNT	PER TUBE	PER REEL	Orientation
D	TO-252, (D-Pak)	2	75	2500	Fig A
Р	Ultra Thin-Pak™	2	75	2500	Fig B
S	SOIC, Narrow Body	8	95	2500	Fig C
Y	SOT-223	3	80	2500	Fig D









International
Rectifier

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