



High Performance, Constant Current Switching Regulator For White LED

General Description

The LP3302 is a high frequency, asynchronous boost converter for constant current white LED driver applications. The internal MOSFET can support up to 8 White LEDs for backlighting and OLED power application, and the internal soft start function can reduce the inrush current. The LED current is initially set with the external sense resistor. To improve efficiency, the feedback voltage is set to 250mV, which reduces the power dissipation in the current setting resistor.

The LP3302 implements a constant frequency 1MHz PWM control scheme. Optimized operation frequency can meet the requirement of small LC filters value. Highly integration and internal compensation network minimizes as 5 external component counts to provide the best solution for PCB space saving and total BOM cost. SOT23-6 packages.

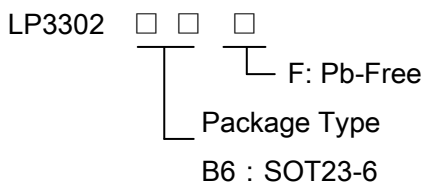
Features

- ◆ High Efficiency: 93%
- ◆ 1MHz Fixed-Frequency PWM Operation
- ◆ Maximum Output Voltage up to 29V
- ◆ Operating Range : 2.7V to 6V
- ◆ Shutdown Supply Current:<1uA
- ◆ Available in SOT23-6 Package
- ◆ Built-in Over Voltage Protection
- ◆ Minimize the External Component
- ◆ RoHS Compliant and 100% Lead Pb-Free

Applications

- ◇ WLED Backlight driver
- ◇ OLED Backlight driver
- ◇ PDA
- ◇ DSC
- ◇ Camera Flash WLED driver

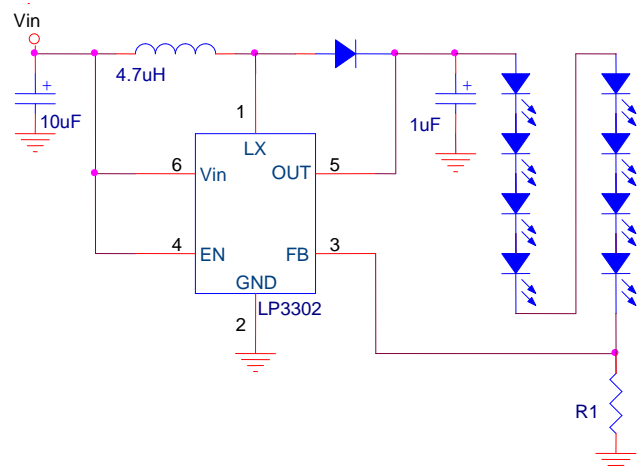
Order Information



Marking Information

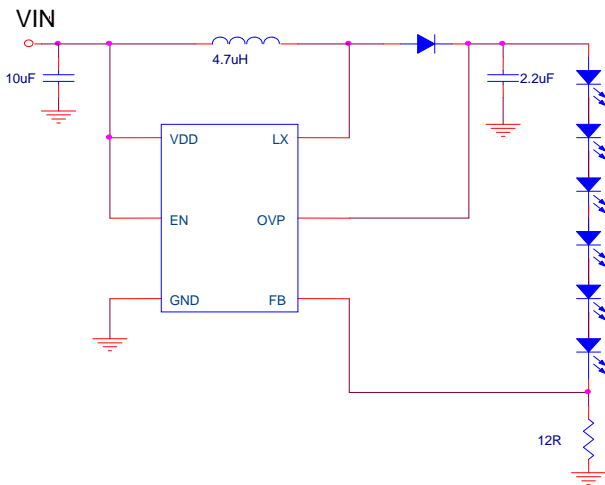
Device	Marking	Package	Shipping
LP3302B6F	LPS F3XXX	SOT23-6	3K/REEL

Typical Application Circuit

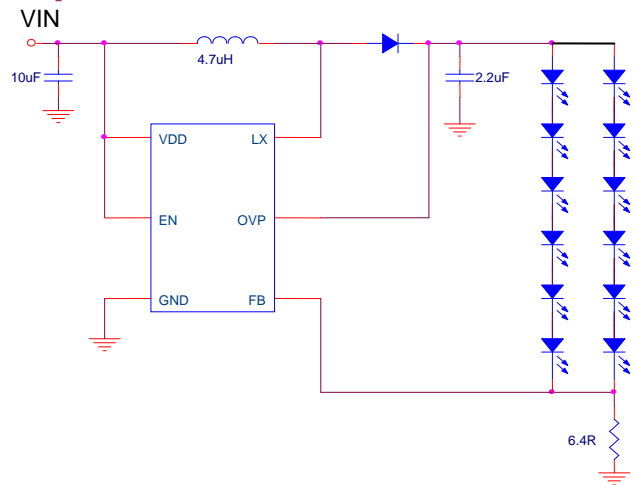




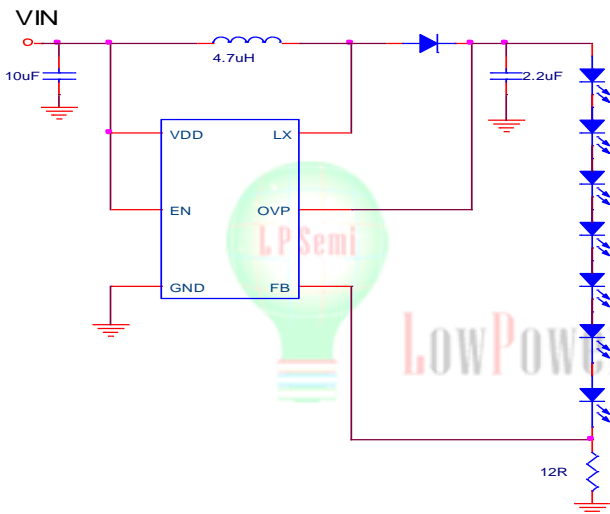
Application Circuits



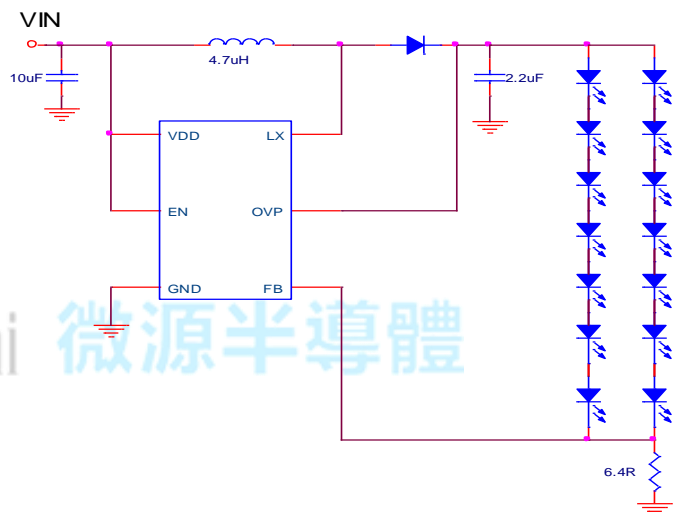
LP3302 for 6pcs Application Circuit



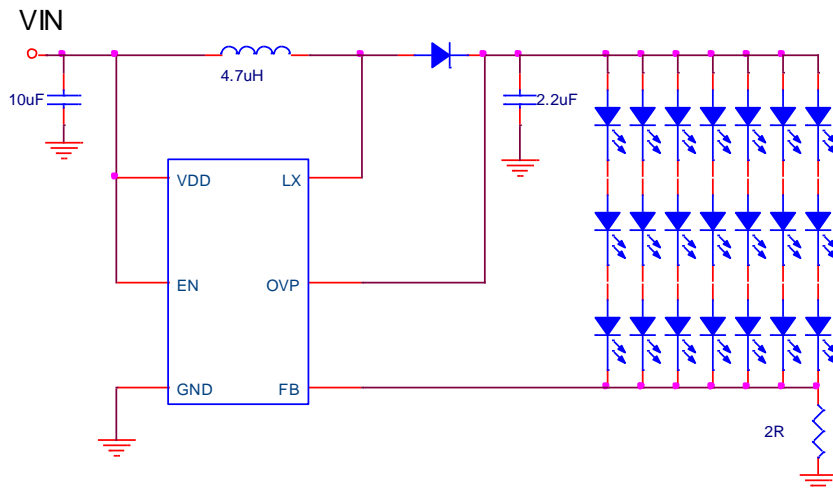
LP3302 for 6pcs×2 Application Circuit



LP3302 for 7pcs Application Circuit



LP3302 for 7pcs×2 Application Circuit



LP3302 for 3pcs×7 Application Circuit



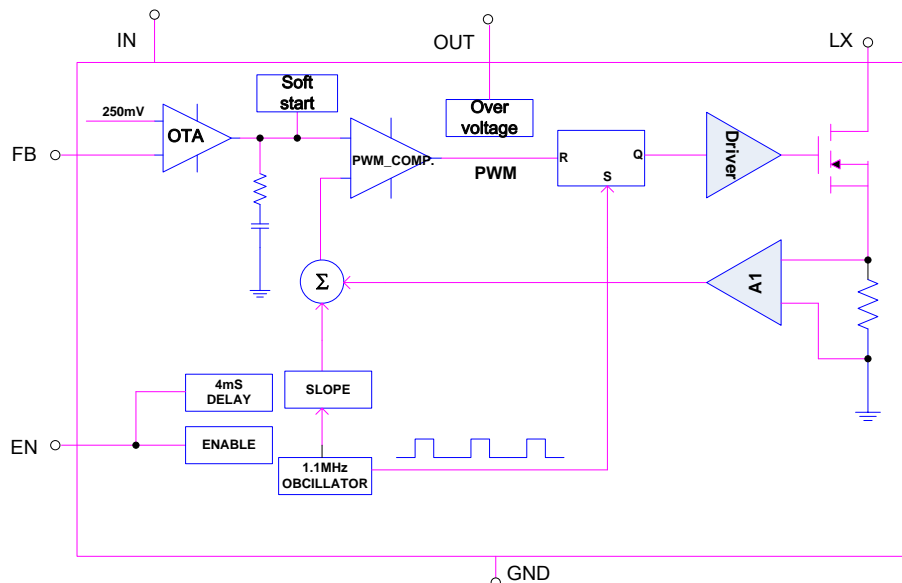
Functional Pin Description

Package Type	Pin Configurations
SOT23-6	

Pin Description

Pin	Name	Description
1	LX	Switch Pin. Connect this Pin to inductor and catch diode. Minimize the track area to reduce EMI.
2	GND	Ground Pin
3	FB	Feedback Reference Voltage Pin. Series connect a resistor between WLED and ground as a current sense. Sense the current feedback voltage to set the current rating.
4	EN	Chip Enable (Active High). Voltage sensing input to trigger the function of over voltage protection. Note that this pin is high impedance. There should be a pull low 100kΩ resistor connected to GND when the control signal is floating.
5	OUT	OVP Pin. Overvoltage Sense. When VOUT is greater than 29V, the internal N-channel MOSFET turns off until VOUT drops below 28V, then the IC reenters start. Connect a 2.2uF capacitor from OUT to GND.
6	VIN	Supply Input Voltage Pin. Bypass 10uF capacitor to GND to reduce the input noise.

Function Block Diagram





Absolute Maximum Ratings

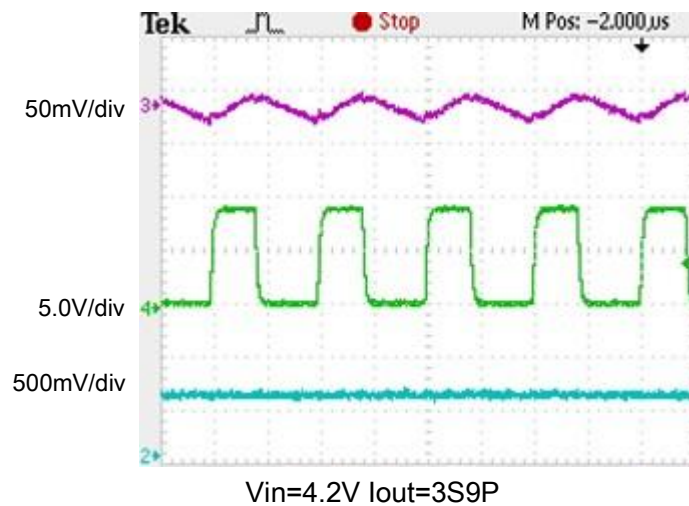
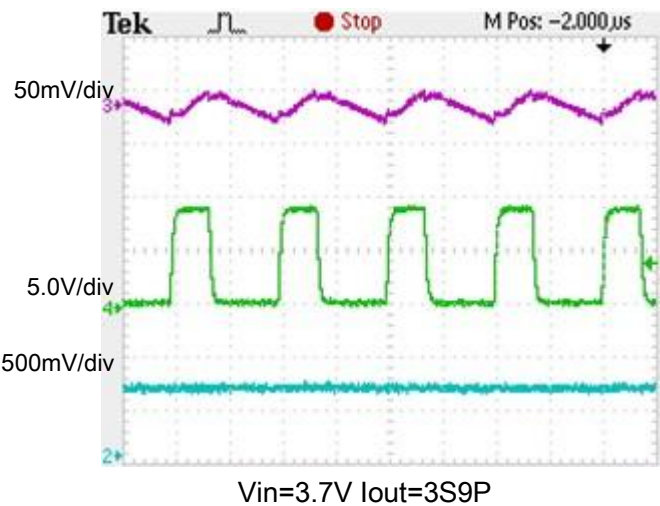
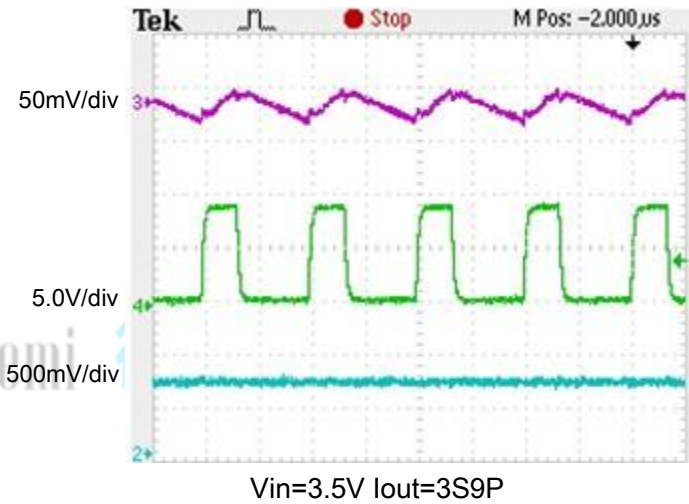
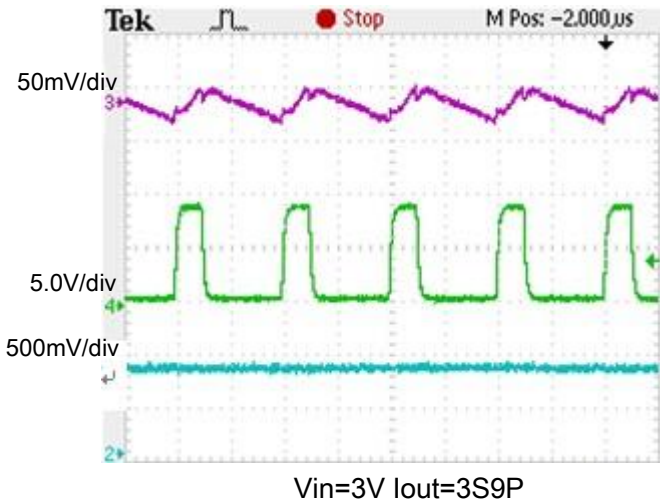
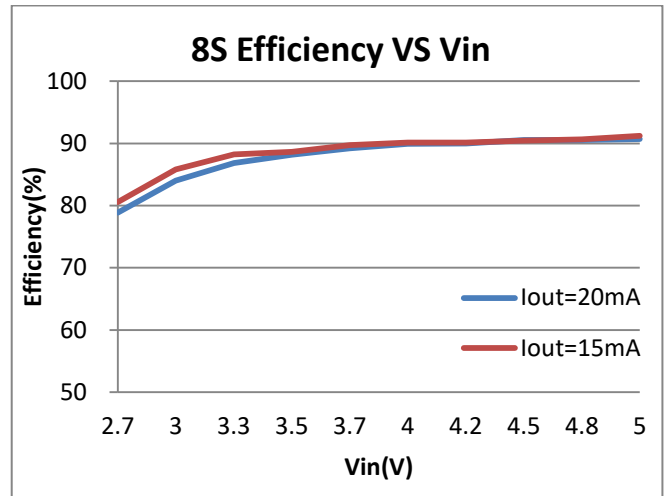
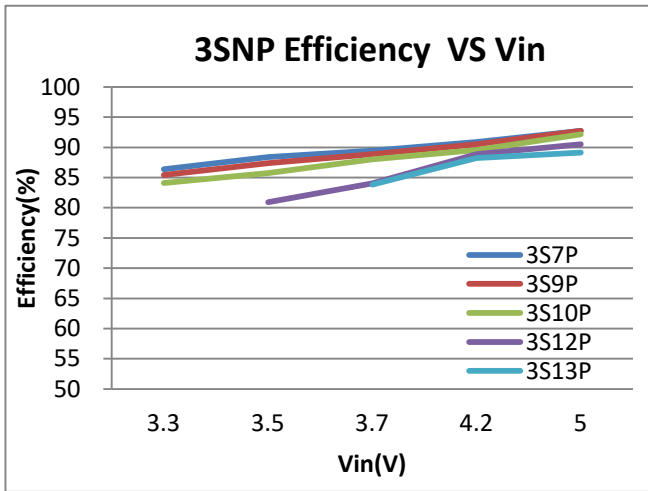
- ◇ IN Pin to GND----- -0.3V to 6.5V
- ◇ LX Pin to GND ----- -0.3V to 36V
- ◇ OVP Pin to GND ----- -0.3V to 36V
- ◇ The Other Pins ----- -0.3V to 5.5V
- ◇ Storage Temperature Range (TJ) ----- -65°C to 150°C
- ◇ Maximum Soldering Temperature (at leads, 10sec) ----- 260°C
- ◇ Maximum Power Dissipation (PD,TA<40°C) ----- 0.45W
- ◇ Thermal Resistance (JA) -----250°C/W
- ◇ Maximum Junction Temperature ----- 125°C
- ◇ Operating Junction Temperature Range (TJ) ----- -40°C to 85°C
- ◇ ESD Susceptibility HBM(Human Body Mode) ----- 2KV
- ◇ MM(Machine Mode) ----- 200V

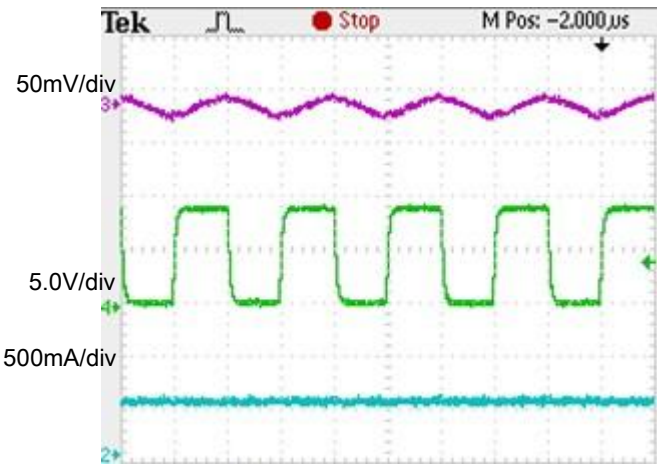
Electrical Characteristics

Parameter	Symbol	Test Condition	Min	Typ.	Max	Units
System Supply Input						
Operation voltage Range	VDD		2.7		6	V
Under Voltage Lock Out	VDD		2.2	2.4	2.6	V
Supply Current	IDD	NO LOAD		0.2	0.4	mA
Shut Down Current	IDD	VEN < 0.4V		0.1	1	uA
Oscillator						
Operation Frequency	FOSC			1		MHz
Maximum Duty Cycle			89	92	96	%
Dimming Frequency			100		1M	Hz
Feedback Voltage	LP3302		230	250	270	mV
MOSFET						
On Resistance of MOSFET	RDS(ON)			0.3		Ω
Protection						
OVP Threshold	VOVP		27	29	31	V
OVP Sink Current				5		μA
OCP				1100		mA
Shut Down Voltage	VEN				0.4	V
Enable Voltage	VEN		1.4			V

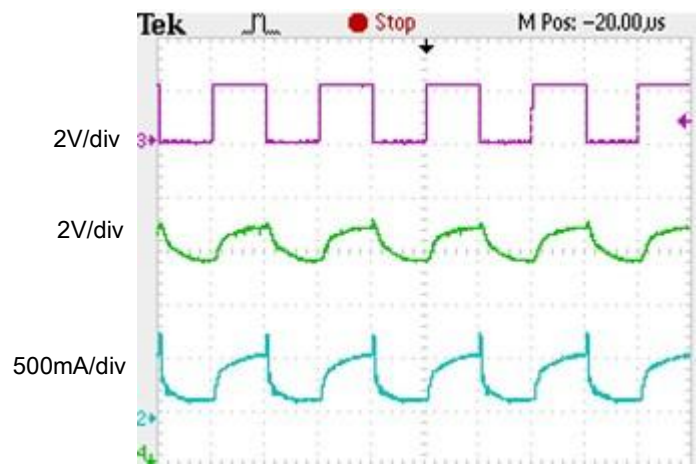


Typical Operating Characteristics

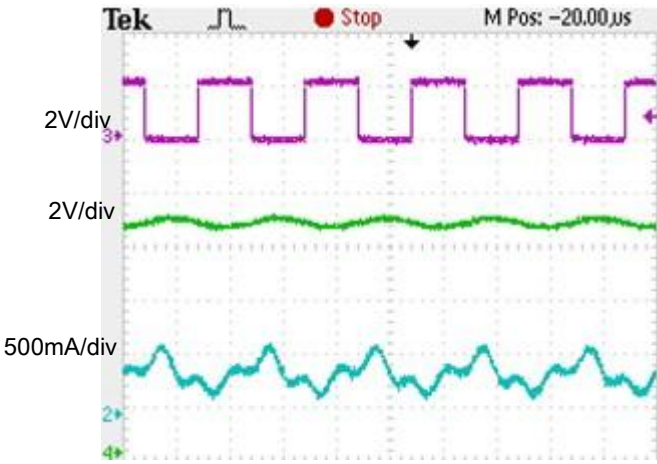




Vin=5V Iout=3S9P



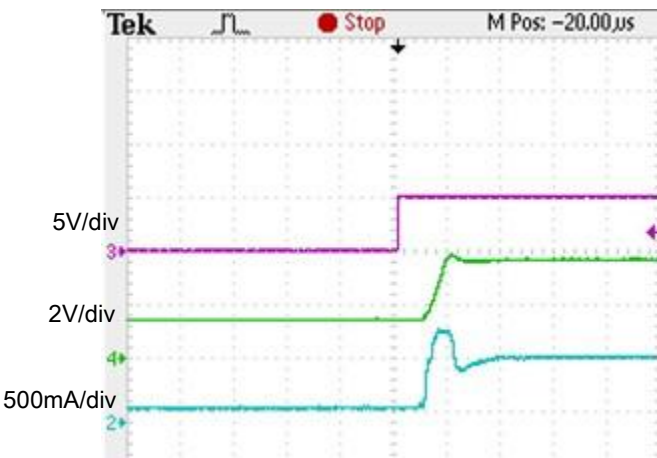
EN=1KHz



EN=10KHz



EN=100KHz





Applications Information

LED Current Control

The LP3302 regulates the LED current by setting the current sense resistor (R1) connecting to feedback and ground. The internal feedback reference voltage is 0.25V. The LED current can be set from following equation easily.

$$I_{LED} = 250mV / R1$$

In order to have an accurate LED current, precision resistors are preferred (1% is recommended). The table for R2 selection is shown below.

R1 Resistor Value selection

I _{LED} (mA)	R1(Ω)
5	50
10	25
12	21
20	12.5

Inductor Selection

The recommended value of inductor for 2 to 8WLEDs applications are 4.7 to 22μH. Small size and better efficiency are the major concerns for portable device, such as LP3302 used for mobile phone. The inductor should have low core loss at 1MHz and low DCR for better efficiency. To avoid inductor saturation current rating should be considered.

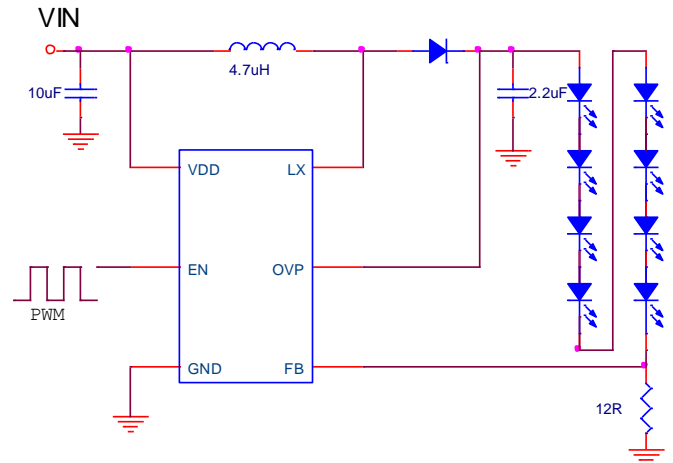
Dimming control

a. Using a PWM Signal to EN Pin

To control the brightness of LED, the LP3302 can perform the dimming control by applying a PWM signal to EN pin. The internal soft-start and wide range dimming frequency from 100Hz to 1MHz can

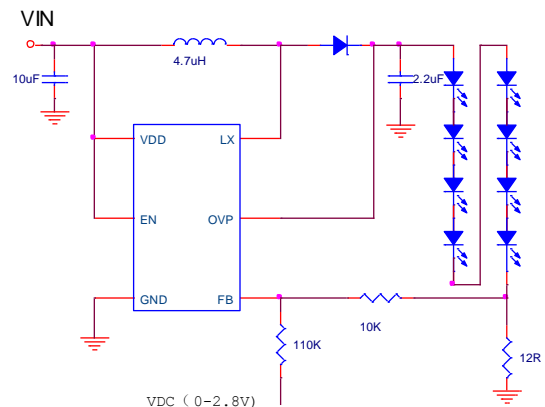
insignificantly reduce audio noise when dimming.

The average LED current is proportional to the PWM signal duty cycle. The magnitude of the PWM signal should be higher than the maximum enable voltage of EN pin, in order to let the dimming control perform correctly.



b. Using a DC Voltage

Using a variable DC voltage to adjust the brightness is a popular method in some applications. The dimming control using a DC voltage circuit is shown below. According to the Superposition Theorem, as the DC voltage increases, the voltage contributed to VFB increases and the voltage drop on R1 decreases, i.e. the LED current decreases. For example, if the VDC range is from 0V to 2.8V, the selection of resistors below sets dimming control of LED current from 20mA to 5.5mA.

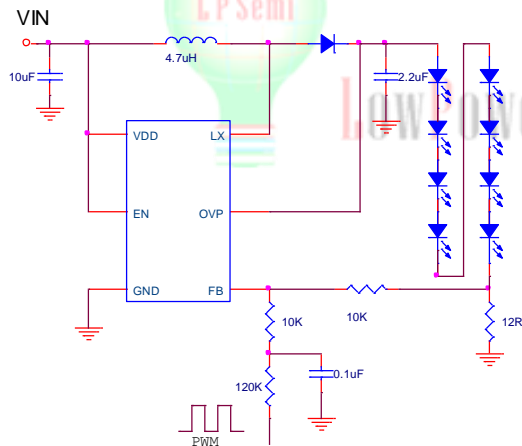




c. Using a Filtered PWM signal

Another common application is using a filtered PWM signal as an adjustable DC voltage for LED dimming control. A filtered PWM signal acts as the DC voltage to regulate the output current. The recommended application circuit is shown in the Figure . In this circuit, the output ripple depends on the frequency of PWM signal. For smaller output voltage ripple (<100mV), the recommended frequency of 2.8V PWM

signal should be above 2kHz. To fix the frequency of PWM signal and change the duty cycle of PWM signal can get different output current. According to the application circuit of Figure , output current is from 20.5mA to 5.5mA by adjusting the PWM duty cycle from 10% to 90%.



Thermal Considerations

For continuous operation, do not exceed absolute maximum operation junction temperature. The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surroundings airflow and temperature difference between junction to ambient.

The maximum power dissipation can be calculated by following formula :

$$PD(MAX) = (T_J(MAX) - T_A) / q_{JA}$$

Where T_{J(MAX)} is the maximum operation junction temperature, T_A is the ambient temperature and the q_{JA} is the junction to ambient thermal resistance. For the recommended operating conditions specification of LP3302, the maximum junction temperature of the die is 125 °C. The junction to ambient thermal resistance q_{JA} is layout dependent. The junction to ambient thermal resistance for TSOT-23-6 package is 250°C/W on the standard JEDEC51-3 single layer thermal test board. Themaximum power dissipation at T_A = 25°C can be calculated by following formula :

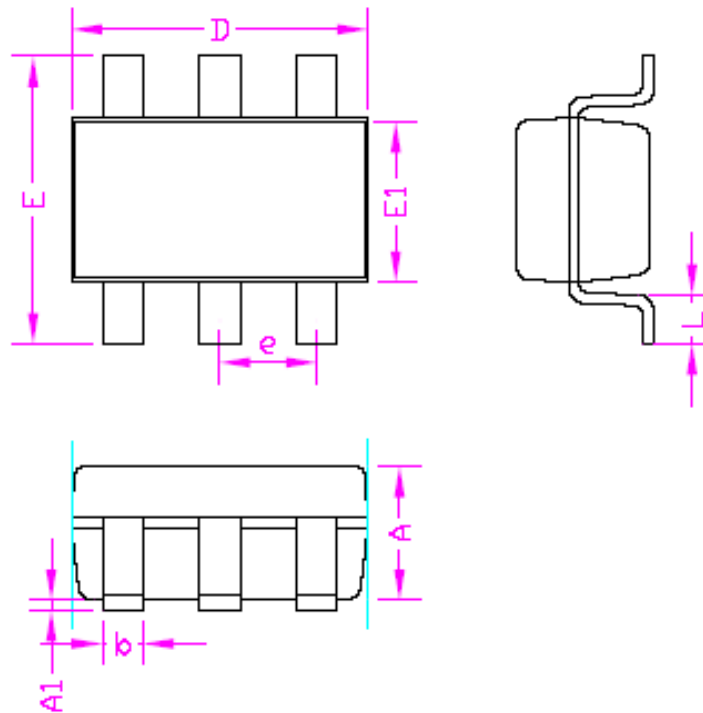
$$PD(MAX) = (125°C - 25°C) / (250°C/W) = 0.4W$$

The maximum power dissipation depends on operating ambient temperature for fixed T_{J(MAX)} and thermal resistance q_{JA}.



Packaging Information

SOT23-6



SYMBOLS	MILLIMETERS		INCHES	
	MIN.	MAX.	MIN.	MAX.
A	-	1.45	-	0.057
A1	0.00	0.15	0.000	0.006
b	0.30	0.50	0.012	0.020
D	2.90		0.114	
E1	1.60		0.063	
e	0.95		0.037	
E	2.60	3.00	0.102	0.118
L	0.3	0.60	0.012	0.024

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