## 340KHz, 2A Step-down Converter With Soft-Start

#### **General Description**

The LP6482 contains an independent 340KHz constant frequency, current mode, PWM step-down converters. The converter integrates a main switch and a synchronous rectifier for high efficiency without an external Schottky diode. The converter can supply 2A of load current. The LP6482 can also run at 98% duty cycle for low dropout applications. provides It fast transient response and cycle-by-cycle limit with current mode control.

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The LP6482 is available in a SOP8 package and is rated over the -40°C to 85°C temperature range.

#### **Order Information**

# **Features**

- Input Voltage Range: 4.5V to 30V
- Output Voltage Range: 0.6V to 12V
- 2A Load Current
- Up to 95% Efficiency
- 340KHz Switching Frequency
- Short Circuit Protection
- **Thermal Fault Protection**
- SOP8 Package

**Applications** 

RoHS Compliant and 100% Lead (Pb)-Free

LowPowerSem LP6482 🗆 🗆 🗆 - F: Pb-Free

Package Type

SO: SOP8

#### Portable Media Players

- Cellular and Smart mobile phone
- PDA/DSC ∻
- **GPS** Applications

#### **Marking Information**

Device	Marking	Package	Shipping			
LP6482SOF	LPS	SO:SOP8	4K/REEL			
	LP6482					
YWX						
Y: Year code. W: Week code. X: Batch numbers.						



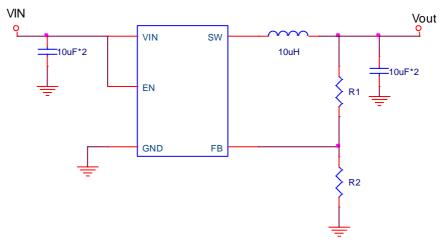
## **Functional Pin Description**

Package Type	Pin Configurations			
	NC 1	8	NC	
SOP-8	VIN 2	7	EN	
	SW 3	6	NC	
	GND 4	5	FB	

## **Pin Description**

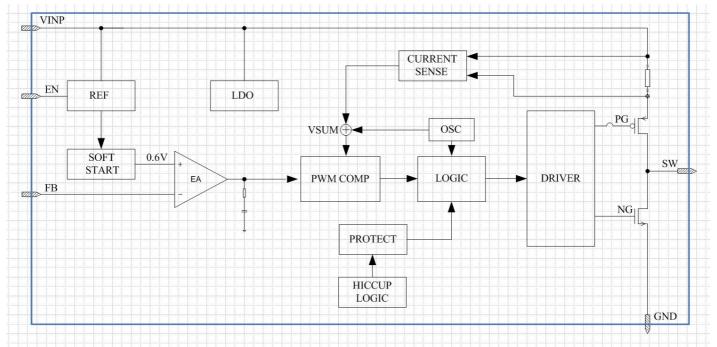
Pin	Name	Description
1/6/8	NC	No Connector.
2	VIN	Supply Input.
3	SW	Switch Mode Connection to Inductor. This pin connects to the drains of the internal main and synchronous power MOSFET switches.
4	GND	Ground. IN TOWER OCHI THE MASS
5	FB	Feedback Input. Connect FB to the center point of the external resistor divider. Normal voltage for this pin is 0.6V.
7	EN	Enable Control Input. Drive EN above 2.5V to turn on the Channel. Drive EN below 0.4V to turn it off (shutdown current < 0.1µA).

## **Typical Application Circuit**





## **Function Diagram**



## Absolute Maximum Ratings Note 1

$\diamond$	VIN/SW/EN to GND	36V
$\diamond$	FB to GND (VFB)	0.3V to 6V
$\diamond$	Junction Temperature	150°C
$\diamond$	Storage Temperature	65°C to 165°C
$\diamond$	Operating Ambient Temperature Range (T <sub>A</sub> )	40°C to 85°C
$\diamond$	Maximum Soldering Temperature (at leads, 10sec)	260°C

**Note 1.** Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## **Thermal Information**

$\diamond$	Maximum Power Dissipation (SOP8, P <sub>D</sub> ,T <sub>A</sub> =25°C)	1.5W
$\diamond$	Thermal Resistance (SOP8, θ <sub>JA</sub> ) 80	J°C/W

## **ESD Susceptibility**

$\diamond$	HBM(Human Body Mode)	2KV
$\diamond$	MM(Machine Mode)	200V



#### **Electrical Characteristics**

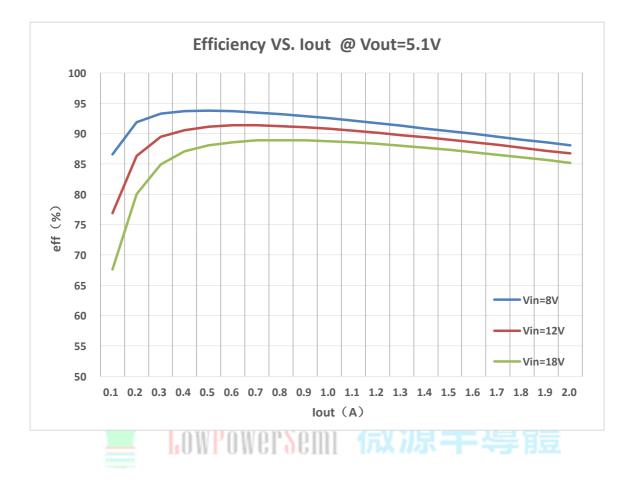
#### V<sub>IN</sub>=12V, V<sub>EN</sub>=5V, L=10uH, T<sub>A</sub>=25°C, unless otherwise noted

Symbol	Parameter	Condition	Min	Тур	Max	Units
VIN	Input Voltage		4.5		30	V
VIN-OVP	VIN OVP Voltage Lockout Threshold			32		
Vout	Output Voltage Range		0.6		12	V
lq	Quiescent Current	V <sub>IN</sub> =12V		10		mA
Ishdn	Shutdown Current	EN=GND		10		μA
Ilim	P-Channel Current Limit		3			А
Rds(on)_H	High-Side Switch On Resistance			80		mΩ
Rds(on)_L	Low-Side Switch On Resistance			60		mΩ
ILX_LEAK	LX Leakage Current	V <sub>EN</sub> =0V, V <sub>SW</sub> =0 or 5V, V <sub>IN</sub> =5V		1		μA
V <sub>FB</sub>	Feedback Threshold Voltage Accuracy	V <sub>IN</sub> =12V	0.588	0.6	0.612	V
I <sub>FB</sub>	FB Leakage Current	Vout=5.0V			30	nA
fosc	Oscillator Frequency			340		KHz
ts	Startup Time	From Enable to Output Regulation		2.7		mS
T <sub>SD</sub>	Over-Temperature Shutdown Threshold	rSemi 成源丰	导	150		°C
T <sub>HYS</sub>	Over-Temperature Shutdown Hysteresis			20		°C
Ven(L)	Enable Threshold Low				1	V
V <sub>EN(H)</sub>	Enable Threshold High		2.5			V
IEN	Enable Current	V <sub>IN</sub> =12V, V <sub>EN</sub> =12V		5		μA

Note: Output Voltage:  $V_{OUT} = V_{FB} \times (1 + R_1 / R_2)$  Volts;



## **Typical Operating Characteristics**



## **Application Information**

The LP6482 operates by a constant frequency, current mode architecture. The output voltage is set by an external divider returned to the FB pin. An error amplifier compares the divided output voltage with a reference voltage of 0.6V and adjusts the peak inductor current accordingly.

#### **Thermal Protection**

The total power dissipation in LP6482 is limited by a thermal protection circuit. When the device temperature rises to approximately 150°C, this circuit turns off the output, allowing the IC to cool. The thermal protection circuit can protect the device from being damaged by overheating in the event of fault conditions. Continuously running the LP6482 into thermal shutdown degrades device reliability.

#### Setting Output Voltage

The output voltage is set with a resistor divider from the output node to the FB pin. It is recommended to use divider resistors with 1% tolerance or better. To improve efficiency at very light loads consider using larger value resistors. If the values are too high the regulator is more susceptible to noise and voltage errors from the FB input current are noticeable. For most applications, a resistor in the 10k $\Omega$  to 1M $\Omega$ range is suggested for R<sub>UP</sub> and R<sub>DN</sub>. The output voltage is established by the following equations

$$V_{OUT}=0.6V \times (1+R_1/R_2)$$

Where  $V_{\text{REF}}$  is 0.6V.

#### **Inductor Selection**

The inductor is required to supply constant current to the output load while being driven by the switched input voltage. A larger value inductor results in less ripple current and lower output ripple voltage. However, the larger value inductor has a larger physical size, higher series resistance, and lower saturation current. Choose an inductor that does not saturate under the worst-case load conditions. A good rule for determining the inductance is to allow the peak-to-peak ripple current in the inductor to be approximately 30% of the maximum load current. The inductance value can be calculated by the equation:

$$L=(V_{OUT})\times(V_{IN}-V_{OUT})/(V_{IN}\times f\times \Delta I)$$

#### **Output Capacitor**

The output capacitor is required to maintain the DC output voltage. Low ESR capacitors are preferred to keep the output voltage ripple low. The characteristics of the output capacitor also affect the stability of the regulator control loop. Ceramic, tantalum, or low ESR electrolytic capacitors are recommended. In the case of ceramic capacitors, the impedance at the switching frequency is dominated by the capacitance. The output voltage ripple is estimated to be:

$$V_{RIPPLE}=1.4 \times V_{IN} \times (f_{LC}/f)^2$$

Where  $V_{RIPPLE}$  is the output ripple voltage,  $V_{IN}$  is the input voltage,  $f_{LC}$  is the resonant frequency of the LC filter, f is the switching frequency. In the case of tantalum or low ESR electrolytic capacitors, the ESR dominates the impedance at the switching frequency, and so the output ripple is calculated as:

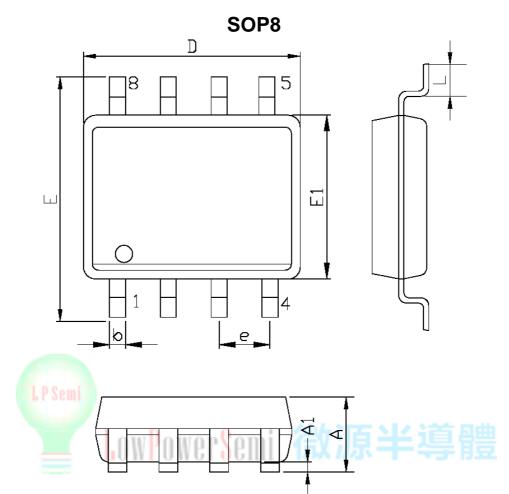
#### $V_{\text{RIPPLE}} \approx \Delta I \times R_{\text{ESR}}$

Where  $V_{RIPPLE}$  is the output voltage ripple,  $\Delta I$  is the inductor ripple current, and  $R_{ESR}$  is the equivalent series resistance of the output capacitors.



LP6482

## **Packaging Information**



SYMBOLS	MILLIMETERS		INCHES	
SIMBOLS	MIN.	MAX.	MIN.	MAX.
A	1.35	1.75	0.053	0.069
A1	0.10	0.25	0.004	0.010
D	4.90		0.193	
E	5.80	6.20	0.228	0.244
E1	3.90		0.153	
L	0.40	1.27	0.016	0.050
b	0.31	0.51	0.012	0.020
e	1.27		0.0	050



单击下面可查看定价,库存,交付和生命周期等信息

>>LOW POWER(微源半导体)