TJA1043

High-speed CAN transceiver Rev. 6 — 10 November 2017

Product data sheet

General description 1.

The TJA1043 high-speed CAN transceiver provides an interface between a Controller Area Network (CAN) protocol controller and the physical two-wire CAN bus. The transceiver is designed for high-speed CAN applications in the automotive industry, providing differential transmit and receive capability to (a microcontroller with) a CAN protocol controller.

The TJA1043 belongs to the third generation of high-speed CAN transceivers from NXP Semiconductors, offering significant improvements over first- and second-generation devices such as the TJA1041A. It offers improved ElectroMagnetic Compatibility (EMC) and ElectroStatic Discharge (ESD) performance, very low power consumption, and passive behavior when the supply voltage is turned off. Advanced features include:

- Low-power management controls the power supply throughout the node while supporting local and remote wake-up with wake-up source recognition
- Several protection and diagnostic functions including bus line short-circuit detection and battery connection detection
- Can be interfaced directly to microcontrollers with supply voltages from 3 V to 5 V

The TJA1043 implements the CAN physical layer as defined in ISO 11898-2:2016 and SAE J2284-1 to SAE J2284-5. This implementation enables reliable communication in the CAN FD fast phase at data rates up to 5 Mbit/s.

These features make the TJA1043 the ideal choice for high speed CAN networks containing nodes that need to be available all times, even when the internal V_{IO} and V_{CC} supplies are switched off.

Features and benefits 2.

2.1 General

- ISO 11898-2:2016 and SAE J2284-1 to SAE J2284-5 compliant
- Loop delay symmetry timing enables reliable communication at data rates up to 5 Mbit/s in the CAN FD fast phase
- Suitable for 12 V and 24 V systems
- Low ElectroMagnetic Emission (EME) and high ElectroMagnetic Immunity (EMI)
- V_{IO} input allows for direct interfacing with 3 V and 5 V microcontrollers
- SPLIT voltage output for stabilizing the recessive bus level
- Listen-only mode for node diagnosis and failure containment
- Available in SO14 and HVSON14 packages



New to NXP?

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- Leadless HVSON14 package (3.0 mm × 4.5 mm) with improved Automated Optical Inspection (AOI) capability
- AEC-Q100 qualified
- Dark green product (halogen free and Restriction of Hazardous Substances (RoHS) compliant)

2.2 Low-power management

- Very low current Standby and Sleep modes, with local and remote wake-up
- Capability to power down the entire node while supporting local, remote and host wake-up
- Wake-up source recognition
- Transceiver disengages from the bus (zero load) when V_{BAT} absent
- Functional behavior predictable under all supply conditions

2.3 Protection and diagnosis (detection and signalling)

- High ESD handling capability on the bus pins
- Bus pins and V_{BAT} protected against transients in automotive environments
- Transmit Data (TXD) dominant time-out function with diagnosis
- TXD-to-RXD short-circuit handler with diagnosis
- Thermal protection with diagnosis
- Undervoltage detection and recovery on pins V_{CC}, V_{IO} and V_{BAT}
- Bus line short-circuit diagnosis
- Bus dominant clamping diagnosis
- Cold start diagnosis (first battery connection)

3. Quick reference data

Table 1. Quick reference data

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{CC}	supply voltage		4.5	-	5.5	V
V _{IO}	supply voltage on pin V _{IO}		2.8	-	5.5	V
V _{uvd(VCC)}	undervoltage detection voltage on pin V _{CC}		3	3.5	4.3	V
$V_{uvd(VIO)}$	undervoltage detection voltage on pin V_{IO}	V _{BAT} or V _{CC} > 4.5 V	0.8	1.8	2.5	V
I _{CC}	supply current	Normal mode; bus dominant	30	48	65	mA
		Normal or Listen-only mode; bus recessive	3	6	9	mA
		Standby or Sleep mode	0	0.75	4.3 2.5 65 9 5 2	μΑ
I _{IO}	supply current on pin V _{IO}	Normal mode; V _{TXD} = 0 V (dominant)	-	150	500	μΑ
		Normal or Listen-only mode; $V_{TXD} = V_{IO}$ (recessive)	0	1		μΑ
		Standby or Sleep mode	0	1	4	μΑ
V _{ESD}	electrostatic discharge voltage	IEC 61000-4-2 at pins CANH and CANL	-8	-	+8	kV

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Table 1. Quick reference data ...continued

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
V _{CANH}	voltage on pin CANH		-58	-	+58	V
V_{CANL}	voltage on pin CANL		-58	-	+58	V
T _{vj}	virtual junction temperature		-40	-	+150	°C

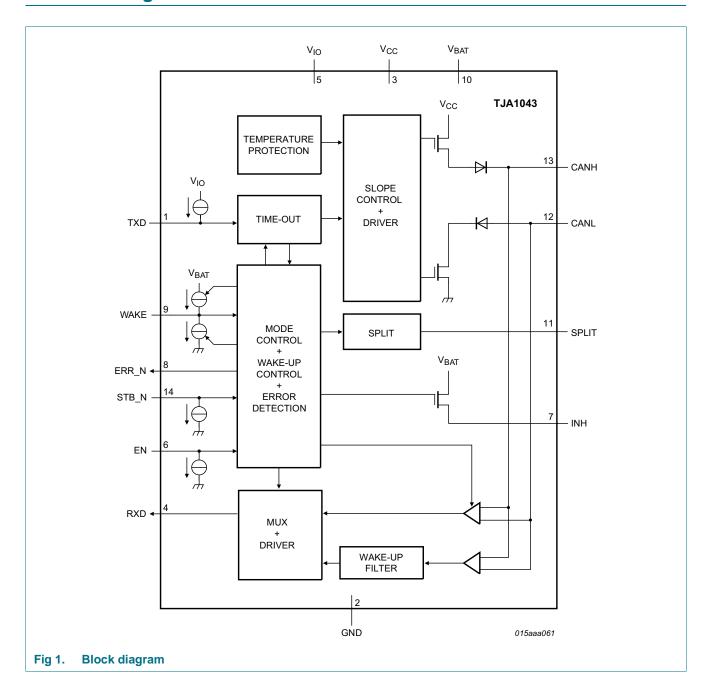
4. Ordering information

Table 2. Ordering information

Type number	Package	ackage				
	Name	Description	Version			
TJA1043T	SO14	plastic small outline package; 14 leads; body width 3.9 mm	SOT108-1			
TJA1043TK	HVSON14	plastic, thermal enhanced very thin small outline package; no leads; 14 terminals; body $3\times4.5\times0.85$ mm	SOT1086-2			

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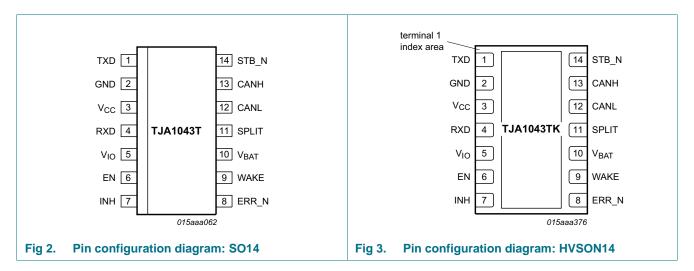
5. Block diagram



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6. Pinning information

6.1 Pinning



6.2 Pin description

Table 3. Pin description

Symbol	Pin	Description
TXD	1	transmit data input
GND[1]	2	ground supply
V _{CC}	3	transceiver supply voltage
RXD	4	receive data output; reads out data from the bus lines
V _{IO}	5	supply voltage for I/O level adaptor
EN	6	enable control input
INH	7	inhibit output for switching external voltage regulators
ERR_N	8	error and power-on indication output (active LOW)
WAKE	9	local wake-up input
V_{BAT}	10	battery supply voltage
SPLIT	11	common-mode stabilization output
CANL	12	LOW-level CAN bus line
CANH	13	HIGH-level CAN bus line
STB_N	14	standby control input (active LOW)

^[1] HVSON14 package die supply ground is connected to both the GND pin and the exposed center pad. The GND pin must be soldered to board ground. For enhanced thermal and electrical performance, it is recommended that the exposed center pad also be soldered to board ground.

7. Functional description

The TJA1043 is a stand-alone high-speed CAN transceiver with a number of operating modes, fail-safe features and diagnostic features that offer enhanced system reliability and advanced power management. The transceiver combines the functionality of the

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TJA1041A with improved EMC and ESD capability and quiescent current performance. Improved slope control and high DC handling capability on the bus pins provide additional application flexibility.

7.1 Operating modes

The TJA1043 supports five operating modes. Control pins STB_N and EN are used to select the operating mode. Switching between modes allows access to a number of diagnostics flags via pin ERR_N. Table 4 describes how to switch between modes. Figure 4 illustrates the mode transitions when V_{CC} , V_{IO} and V_{BAT} are valid.

Table 4. Operating mode selection

Internal fla	ags		Control pi	ns	Operating mode	Pin INH
UV _{NOM} [1]	UV _{BAT}	Wake ^[2]	STB_N[3]	EN		
From Norr	nal, Listen-	only, Standb	y and Go-to-	Sleep mod	des	'
set	X	X	X	X	Sleep mode	floating
cleared	set	X	HIGH	X	Standby mode	HIGH
cleared	X	set	LOW	X	Standby mode	HIGH
cleared	X	cleared	LOW	LOW	Standby mode	HIGH
cleared	X	cleared	LOW	HIGH	Go-to-Sleep mode[4]	HIGH[4]
cleared	cleared	Х	HIGH	LOW	Listen-only mode	HIGH
cleared	cleared	Х	HIGH	HIGH	Normal mode	HIGH
From Slee	p mode					
set	X	X	X	X	Sleep mode	floating
cleared	set	Х	HIGH	X	Standby mode	HIGH
cleared	X	set	LOW	X	Standby mode	HIGH
cleared	X	cleared	LOW	Х	Sleep mode	floating
cleared	cleared	Х	HIGH	LOW	Listen-only mode	HIGH
cleared	cleared	Х	HIGH	HIGH	Normal mode	HIGH

^[1] Setting the UV_{NOM} flag will clear the WAKE flag.

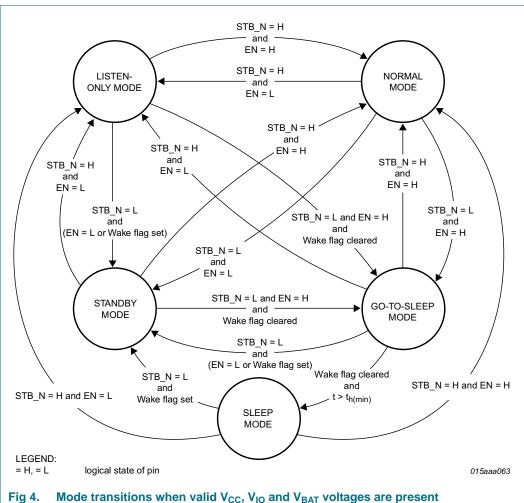
^[2] Setting the Wake flag will clear the UV_{NOM} flag.

^[3] A LOW-to-HIGH transition on pin STB_N will clear the UV_{NOM} flag

^[4] After the minimum hold time, in Go-to-Sleep mode, t_{h(min)}, the transceiver will enter Sleep mode and pin INH will be set floating.

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7.1.1 Normal mode

In Normal mode, the transceiver can transmit and receive data via the bus lines CANH and CANL (see Figure 1 for the block diagram). The differential receiver converts the analog data on the bus lines into digital data which is output to pin RXD. The slopes of the output signals on the bus lines are controlled internally and are optimized in a way that guarantees the lowest possible EME. The bus pins are biased to 0.5V_{CC} (via R_i). Pin INH is active, so voltage regulators controlled by pin INH (see Figure 9) will be active too.

7.1.2 Listen-only mode

In Listen-only mode, the transceiver's transmitter is disabled, effectively providing a transceiver listen-only feature. The receiver will still convert the analog bus signal on pins CANH and CANL into digital data, available for output on pin RXD. As in Normal mode, the bus pins are biased at $0.5V_{CC}$ and pin INH remains active.

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7.1.3 Standby mode

Standby mode is the TJA1043's first-level power saving mode, offering reduced current consumption. In Standby mode, the transceiver is unable to transmit or receive data and the low-power receiver is activated to monitor bus activity. The bus pins are biased at ground level (via R_i). Pin INH is still active, so voltage regulators controlled by this pin will also be active.

Pins RXD and ERR_N will reflect any active wake-up requests (provided that V_{IO} and V_{BAT} are present).

7.1.4 Go-to-Sleep mode

Go-to-Sleep mode is the controlled route for entering Sleep mode. In Go-to-Sleep mode, the transceiver behaves as in Standby mode, with the addition that a go-to-sleep command is issued to the transceiver. The transceiver will remain in Go-to-Sleep mode for the minimum hold time $(t_{h(min)})$ before entering Sleep mode. The transceiver will not enter Sleep mode if the state of pin STB_N or pin EN is changed or if the Wake flag is set before $t_{h(min)}$ has elapsed.

7.1.5 Sleep mode

Sleep mode is second-level power saving mode of the TJA1043. Sleep mode is entered via Go-to-Sleep mode, and also when the undervoltage detection time on either V_{CC} or V_{IO} elapses before the relevant voltage level has recovered. In Sleep mode, the transceiver behaves as described for Standby mode, with the exception that pin INH is set floating. Voltage regulators controlled by this pin will be switched off, and the current into pin V_{BAT} will be reduced to a minimum. Pins STB_N, EN and the Wake flag can be used to wake up a node from Sleep mode (see Table 4).

7.2 Internal flags

The TJA1043 makes use of seven internal flags for its fail-safe fallback mode control and system diagnosis support. Five of these flags can be polled by the controller via pin ERR_N. Which flag is available on pin ERR_N at any time depends on the active operating mode and on a number of other conditions. <u>Table 5</u> describes how to access these flags.

Table 5. Accessing internal flags via pin ERR_N

Internal flag	Flag is available on pin ERR_N[1]	Flag is cleared
UV _{NOM}	no	by setting the Pwon or Wake flags, by a LOW-to-HIGH transition on STB_N or when both $V_{\rm IO}$ and $V_{\rm BAT}$ have recovered.
UV _{BAT}	no	when V _{BAT} has recovered
Pwon	in Listen-only mode (coming from Standby mode, Go-to-Sleep mode, or Sleep mode)	on entering Normal mode
Wake	in Standby mode, Go-to-Sleep mode, and Sleep mode (provided that V_{IO} and V_{BAT} are present)	on entering Normal mode or by setting the UV _{NOM} flag

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Table 5. Accessing internal flags via pin ERR_Ncontinued
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Internal flag	Flag is available on pin ERR_N ^[1]	Flag is cleared
Wake-up source	in Normal mode (before the fourth dominant-to-recessive edge on pin TXD[2])	on leaving Normal mode
Bus failure	in Normal mode (after the fourth dominant-to-recessive edge on pin TXD[2])	on re-entering Normal mode or by setting the Pwon flag
Local failure	in Listen-only mode (coming from Normal mode)	on entering Normal mode or when RXD is dominant while TXD is recessive (provided that all local failures are resolved) or by setting the Pwon flag

^[1] Pin ERR_N is an active-LOW output, so a LOW-level indicates a set flag and a HIGH-level indicates a cleared flag. Allow pin ERR_N to stabilize for at least 8 μs after changing operating modes.

7.2.1 UV_{NOM} flag

 UV_{NOM} is the V_{CC} and V_{IO} undervoltage detection flag. The flag is set when the voltage on pin V_{CC} drops below the V_{CC} undervoltage detection voltage, $V_{uvd(VCC)}$, for longer than the undervoltage detection time, $t_{det(uv)}$, or when the voltage on pin V_{IO} drops below $V_{uvd(VIO)}$ for longer than $t_{det(uv)}$. When the UV_{NOM} flag is set, the transceiver enters Sleep mode to save power and to ensure the bus is not disturbed. In Sleep mode the voltage regulators connected to pin INH are disabled, avoiding any extra power consumption that might be generated as a result of a short-circuit condition.

Any wake-up request, setting the Pwon flag or a LOW-to-HIGH transition on STB_N will clear UV_{NOM} and the timers, allowing the voltage regulators to be reactivated (at least until UV_{NOM} is set again). UV_{NOM} will also be cleared if both V_{CC} and V_{IO} recover for longer than the undervoltage recovery time, $t_{rec(uv)}$. The transceiver will then switch to the operating mode indicated by the logic levels on pins STB_N and EN (see Table 4).

7.2.2 UV_{BAT} flag

 UV_{BAT} is the V_{BAT} undervoltage detection flag. This flag is set when the voltage on pin V_{BAT} drops below $V_{uvd(VBAT)}$. When UV_{BAT} is set, the transceiver will try to enter Standby mode to save power and will disengage from the bus (zero load). UV_{BAT} is cleared when the voltage on pin V_{BAT} recovers. The transceiver will then switch to the operating mode indicated by the logic levels on pins STB_N and EN (see Table 4).

7.2.3 Pwon flag

Pwon is the V_{BAT} power-on flag. This flag is set when the voltage on pin V_{BAT} recovers after previously dropping below $V_{uvd(VBAT)}$ (usually because the battery was disconnected). Setting the Pwon flag clears the UV_{NOM} flag and timers. The Wake and Wake-up source flags are set to ensure consistent system power-up under all supply conditions. In Listen-only mode the Pwon flag can be polled via pin ERR_N (see <u>Table 5</u>). The flag is cleared when the transceiver enters Normal mode.

7.2.4 Wake flag

The Wake flag is set when the transceiver detects a local or remote wake-up request. A local wake-up request is detected when the logic level on pin WAKE changes, and the new level remains stable for at least t_{wake} . The Wake flag can be set in Standby mode, Go-to-Sleep mode or Sleep mode. Setting the Wake flag clears the UV_{NOM} flag and

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^[2] Allow for a TXD dominant time of at least 4 μs per dominant-recessive cycle.

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timers. Once set, the Wake flag status is immediately available on pins ERR_N and RXD (provided V_{IO} and V_{BAT} are present). This flag is also set at power-on and cleared when the UV_{NOM} flag is set or the transceiver enters Normal mode.

7.2.5 Remote wake-up (via the CAN bus)

The TJA1043 wakes up from Standby or Sleep mode when a dedicated wake-up pattern (specified in ISO 11898-2: 2016) is detected on the bus. This filtering helps avoid spurious wake-up events. A spurious wake-up sequence could be triggered by, for example, a dominant clamped bus or by dominant phases due to noise or spikes on the bus.

The wake-up pattern consists of:

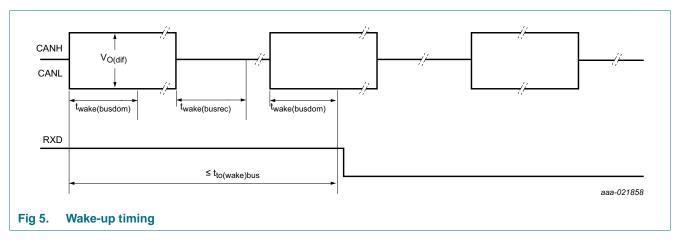
- a dominant phase of at least twake(busdom) followed by
- a recessive phase of at least t_{wake(busrec)} followed by
- a dominant phase of at least twake(busdom)

Dominant or recessive bits between the above mentioned phases that are shorter than $t_{\text{wake(busdom)}}$ and $t_{\text{wake(busrec)}}$ respectively are ignored.

The complete dominant-recessive-dominant pattern must be received within $t_{to(wake)bus}$ to be recognized as a valid wake-up pattern (see <u>Figure 5</u>). Otherwise, the internal wake-up logic is reset. The complete wake-up pattern will then need to be retransmitted to trigger a wake-up event. Pin RXD remains HIGH until the wake-up event has been triggered.

A wake-up event is not flagged on RXD if any of the following events occurs while a valid wake-up pattern is being received:

- The TJA1043 switches to Normal mode
- The complete wake-up pattern was not received within tto(wake)bus
- A V_{CC} or V_{IO} undervoltage is detected (UV_{NOM} flag set; see <u>Section 7.2.1</u>)



7.2.6 Wake-up source flag

Wake-up source recognition is provided via the Wake-up source flag, which is set when the Wake flag is set by a local wake-up request via the WAKE pin. The Wake-up source flag can be polled via the ERR_N pin in Normal mode (see <u>Table 5</u>). This flag is also set at power-on and cleared when the transceiver leaves Normal mode.

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7.2.7 Bus failure flag

The Bus failure flag is set if the transceiver detects a bus line short-circuit condition to V_{BAT} , V_{CC} or GND during four consecutive dominant-recessive cycles on pin TXD, while trying to drive the bus lines dominant. The Bus failure flag can be polled via the ERR_N pin in Normal mode (see <u>Table 5</u>). This flag is cleared at power-on or when the transceiver re-enters Normal mode.

7.2.8 Local failure flag

In Normal and Listen-only modes, the transceiver can distinguish four different local failure events, any of which will cause the Local failure flag to be set. The four local failure events are: TXD dominant clamping, TXD-to-RXD short circuit, bus dominant clamping and an overtemperature event. The nature and detection of these local failures is described in Section 7.3. The Local failure flag can be polled via the ERR_N pin in Listen-only mode (see Table 5). This flag is cleared at power-on, when entering Normal mode or when RXD is dominant while TXD is recessive, provided that all local failures have been resolved.

7.3 Local failures

The TJA1043 can detect four different local failure conditions. Any of these failures will set the Local failure flag, and in most cases the transmitter of the transceiver will be disabled.

7.3.1 TXD dominant time-out function

A permanent LOW level on pin TXD (due to a hardware or software application failure) would drive the CAN bus into a permanent dominant state, blocking all network communications. The TXD dominant time-out function prevents such a network lock-up by disabling the transmitter if pin TXD remains LOW for longer than the TXD dominant time-out time $t_{to(dom)TXD}$. The $t_{to(dom)TXD}$ timer defines the minimum possible bit rate of 40 kbit/s. The transmitter remains disabled until the Local failure flag has been cleared.

7.3.2 TXD-to-RXD short-circuit detection

A short-circuit between pins RXD and TXD would lock the bus in a permanent dominant state once it had been driven dominant, because the low-side driver of RXD is typically stronger than the high-side driver of the controller connected to TXD. TXD-to-RXD short-circuit detection prevents such a network lock-up by disabling the transmitter. The transmitter remains disabled until the Local failure flag has been cleared.

7.3.3 Bus dominant time-out function

A CAN bus short circuit (to V_{BAT} , V_{CC} or GND) or a failure in one of the other network nodes could result in a differential voltage on the bus high enough to represent a bus dominant state. Because a node will not start transmission if the bus is dominant, the normal bus failure detection will not detect this failure, but the bus dominant clamping detection will. The Local failure flag is set if the dominant state on the bus persists for longer than $t_{to(dom)bus}$. By checking this flag, the controller can determine if a clamped bus is blocking network communications. There is no need to disable the transmitter. Note that the Local failure flag does not retain a bus dominant clamping failure, and is released as soon as the bus returns to recessive state.

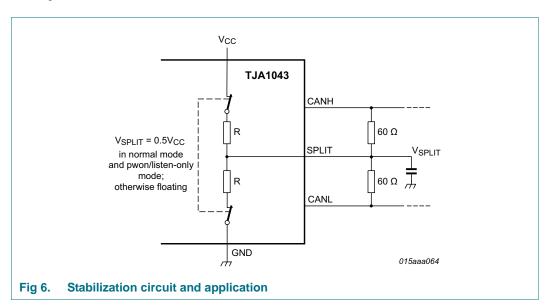
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7.3.4 Overtemperature detection

If the junction temperature becomes excessive, the transmitter will shut down in time to protect the output drivers from overheating without compromising the maximum operating temperature. The transmitter will remain disabled until the Local failure flag has been cleared.

7.4 SPLIT pin

Using the SPLIT pin on the TJA1043 in conjunction with a split termination network (see Figure 6 and Figure 9) can help to stabilize the recessive voltage level on the bus. This will reduce EME in networks with DC leakage to ground (e.g. from deactivated nodes with poor bus leakage performance). In Normal and Listen-only modes, pin SPLIT delivers a DC output voltage of 0.5V_{CC}. In Standby, Go-to-Sleep and Sleep modes, pin SPLIT is floating.



7.5 V_{IO} supply pin

Pin V_{IO} should be connected to the microcontroller supply voltage (see <u>Figure 9</u>). This will cause the signal levels of pins TXD, RXD, STB_N, EN and ERR_N to be adjusted to the I/O levels of the microcontroller, facilitating direct interfacing without the need for glue logic.

7.6 WAKE pin

A local wake-up event is triggered by a LOW-to-HIGH or HIGH-to-LOW transition on the WAKE pin, allowing for maximum flexibility when designing a local wake-up circuit. To minimize current consumption, the internal bias voltage will follow the logic state on the pin after a delay of t_{wake} . A HIGH level on pin WAKE is followed by an internal pull-up to V_{BAT} . A LOW level on pin WAKE is followed by an internal pull-down towards GND. In applications that don't make use of the local wake-up facility, it is recommended that the WAKE pin be connected to V_{BAT} or GND to ensure optimal EMI performance.

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8. Limiting values

Table 6. Limiting values

In accordance with the Absolute Maximum Rating System (IEC 60134).

Symbol	Parameter	Conditions		Min	Max	Unit
V _{BAT}	battery supply voltage[1]			-0.3	+58	V
		load dump		-	58	V
V _x	voltage on pin x[1]	on pins CANH, CANL and SPLIT		-58	+58	V
		on pins INH and WAKE		-0.3	+58	V
		on pins V_{CC} , V_{IO} , TXD, RXD, STB_N, EN, ERR_N		-0.3	+7	V
V _(CANH-CANL)	voltage between pin CANH and pin CANL			-20	+20	V
I _{WAKE}	current on pin WAKE			-	-15	mA
V _{trt}	transient voltage	on pins CANH, CANL, SPLIT and V _{BAT}	[2]			
		pulse 1		-100	-	V
		pulse 2a		-	75	V
		pulse 3a		-150	-	V
		pulse 3b		-	100	V
V_{ESD}	electrostatic discharge	IEC 61000-4-2 (150 pF, 330 Ω)	[3]			
	voltage	at pins CANH and CANL		-8	+8	kV
		Human Body Model (HBM); 100 pF, 1.5 k Ω	[4]			
		at pins CANH and CANL		-8	+8	kV
		at any other pin		-4	+4	kV
		Machine Model (MM); 200 pF, 0.75 μ H, 10 Ω	[5]			
		at any pin		-300	+300	V
		Charged Device Model (CDM); field Induced charge; 4 pF	[6]			
		at corner pins		-750	+750	V
		at any pin		-500	+500	V
T _{vj}	virtual junction temperature		[7]	-40	+150	°C
T _{stg}	storage temperature			-55	+150	°C

^[1] The device can sustain voltages up to the specified values over the product lifetime, provided applied voltages (including transients) never exceed these values.

- [2] According to IEC TS 62228 (2007), Section 4.2.4; parameters for standard pulses defined in ISO7637 part 2: 2004-06.
- [3] According to IEC TS 62228 (2007), Section 4.3; DIN EN 61000-4-2.
- [4] According to AEC-Q100-002.
- [5] According to AEC-Q100-003.
- [6] According to AEC-Q100-011 Rev-C1. The classification level is C4B.
- [7] In accordance with IEC 60747-1. An alternative definition of virtual junction temperature is: $T_{vj} = T_{amb} + P \times R_{th(vj-a)}$, where $R_{th(vj-a)}$ is a fixed value to be used for the calculation of T_{vj} . The rating for T_{vj} limits the allowable combinations of power dissipation (P) and ambient temperature (T_{amb}).

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9. Thermal characteristics

Table 7. Thermal characteristics

Value determined for free convection conditions on a JEDEC 2S2P board.

Symbol	Parameter	Conditions	Тур	Unit
R _{th(vj-a)}	thermal resistance from virtual junction to ambient	SO14 package; in free air	68	K/W
		HVSON14 package; in free air	44	K/W

10. Static characteristics

Table 8. Static characteristics

 $V_{CC}=4.5~V$ to 5.5 V; $V_{IO}=2.8~V$ to V_{CC} ; $V_{BAT}=4.5~V$ to 40 V; $V_{L}=60~\Omega$; $V_{L}=60~\Omega$; $V_{L}=60~\Omega$; $V_{CC}=4.5~V$ to 5.5 V; $V_{IO}=2.8~V$ to $V_{CC}=4.5~V$ to 40 V; $V_{L}=60~\Omega$; $V_{L}=60~$

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Supply pin	V _{CC}		·		·	
V _{CC}	supply voltage		4.5	-	5.5	V
$V_{uvd(VCC)}$	undervoltage detection voltage on pin V_{CC}	V _{BAT} > 4.5 V	3	3.5	4.3	V
I _{CC}	supply current	Normal mode; dominant; V _{TXD} = 0 V	30	48	65	mA
		Normal or Listen-only mode; recessive; $V_{TXD} = V_{IO}$	3	6	9	mA
		Standby or Sleep mode; V _{BAT} > V _{CC}	0	0.75	2	μΑ
		Normal mode; dominant; $V_{TXD} = 0 \text{ V}$; short circuit on bus lines; $-3 \text{ V} < (V_{CANH} = V_{CANL}) < +18 \text{ V}$	3	79	109	mA
I/O level ad	dapter supply; pin V _{IO}					
V _{IO}	supply voltage on pin V _{IO}		2.8	-	5.5	V
$V_{uvd(VIO)}$	undervoltage detection voltage on pin V _{IO}	V_{BAT} or $V_{CC} > 4.5 \text{ V}$	0.8	1.8	2.5	V
I _{IO}	supply current on pin V _{IO}	Normal mode; V _{TXD} = 0 V (dominant)	-	150	500	μΑ
		Normal or Listen-only mode; V _{TXD} = V _{IO} (recessive)	0	1	4	μА
		Standby or Sleep mode	0	1	4	μΑ
Supply pin	V _{BAT}		,			
V_{BAT}	battery supply voltage		4.5	-	40	V
$V_{uvd(VBAT)}$	undervoltage detection voltage on pin V _{BAT}		3	3.5	4.3	V
I _{BAT}	battery supply current	Normal or Listen-only mode	15	40	70	μΑ
		Standby mode; $V_{CC} > 4.5 \text{ V}$; $V_{INH} = V_{WAKE} = V_{BAT}$	5	18	30	μА
		Sleep mode; $V_{INH} = V_{CC} = V_{IO} = 0 V$; $V_{WAKE} = V_{BAT}$	5	18	30	μА
CAN trans	mit data input; pin TXD					
V _{IH}	HIGH-level input voltage		0.7V _{IO}	-	$V_{IO} + 0.3$	V
V _{IL}	LOW-level input voltage		-0.3	-	+0.3V _{IO}	V
I _{IH}	HIGH-level input current	$V_{TXD} = V_{IO}$	-5	0	+5	μΑ

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 Table 8.
 Static characteristics ...continued

 $V_{CC} = 4.5 \text{ V}$ to 5.5 V; $V_{IO} = 2.8 \text{ V}$ to V_{CC} ; $V_{BAT} = 4.5 \text{ V}$ to V_{CC} ; $V_{RL} = 60 \Omega$; $V_{CC} = 4.5 \text{ V}$ to $V_{CC} = 4.5 \text{ V}$ to V

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I _{IL}	LOW-level input current	Normal mode; V _{TXD} = 0 V	-300	-200	-30	μΑ
C _i	input capacitance	not tested	-	5	10	pF
CAN receive	e data output; pin RXD	1				
I _{OH}	HIGH-level output current	$V_{RXD} = V_{IO} - 0.4 \text{ V}; V_{IO} = V_{CC}$	-12	-6	-1	mA
I _{OL}	LOW-level output current	$V_{RXD} = 0.4 \text{ V}; V_{TXD} = V_{IO}; \text{ bus dominant}$	2	6	14	mΑ
Standby and	d enable control inputs; pi	ns STB_N and EN		1	1	
V _{IH}	HIGH-level input voltage		0.7V _{IO}	-	V _{IO} +0.3	V
V _{IL}	LOW-level input voltage		-0.3	-	0.3V _{IO}	V
I _{IH}	HIGH-level input current	V_{STB_N} or $V_{EN} \ge 0.7 V_{IO}$	1	4	10	μΑ
I _{IL}	LOW-level input current	$V_{STB_N} = V_{EN} = 0 V$	-1	0	+1	μΑ
Error and p	ower-on indication output	; pin ERR_N		1	1	
I _{OH}	HIGH-level output current	$V_{ERR_N} = V_{IO} - 0.4 \text{ V}; V_{IO} = V_{CC}$	-50	-20	-4	μΑ
I _{OL}	LOW-level output current	V _{ERR N} = 0.4 V	0.1	0.5	2	mA
Local wake-	-up input; pin WAKE	-				-
I _{IH}	HIGH-level input current	$V_{\text{WAKE}} = V_{\text{BAT}} - 1.9 \text{ V}$	-10	-5	-1	μΑ
I _{IL}	LOW-level input current	V _{WAKE} = V _{BAT} - 3.1 V	1	5	10	μА
V _{th}	threshold voltage	V _{STB_N} = 0 V	V _{BAT} – 3	V _{BAT} – 2.5	V _{BAT} – 2	V
Inhibit outp	ut; pin INH					
ΔV_{H}	HIGH-level voltage drop	$I_{INH} = -0.18 \text{ mA}$	0	0.25	0.8	V
IL	leakage current	Sleep mode	-2	0	+2	μА
Bus lines; p	oins CANH and CANL	-				
V _{O(dom)}	dominant output voltage	$V_{TXD} = 0 \text{ V}; \text{ t} < \text{t}_{to(dom)TXD}$				
,		pin CANH; $R_L = 50 \Omega$ to 65Ω	2.75	3.5	4.5	V
		pin CANL; $R_L = 50 \Omega$ to 65Ω	0.5	1.5	2.25	V
V _{dom(TX)sym}	transmitter dominant voltage symmetry	$V_{\text{dom(TX)sym}} = V_{\text{CC}} - V_{\text{CANH}} - V_{\text{CANL}}$	-400	-	+400	mV
V _{TXsym}	transmitter voltage symmetry	$V_{TXsym} = V_{CANH} + V_{CANL}; C_{SPLIT} = 4.7 \text{ nF};$ [2] $f_{TXD} = 250 \text{ kHz}, 1 \text{ MHz and } 2.5 \text{ MHz};$ [3] $V_{CC} = 4.75 \text{ V to } 5.25 \text{ V}$	0.9V _{CC}	-	1.1V _{CC}	V
$V_{O(dif)}$	differential output voltage	dominant; Normal mode; $V_{TXD} = 0 \text{ V}$; $t < t_{to(dom)TXD}$; $V_{CC} = 4.75 \text{ V}$ to 5.25 V				
		$R_L = 45 \Omega$ to 65Ω	1.5	-	3	V
		$R_L = 45 \Omega$ to 70Ω	1.5	-	3.3	V
		$R_L = 2240 \Omega$	1.5	-	5	V
		recessive; no load				
		Normal or Listen-only mode; V _{TXD} = V _{IO}	-50	-	+50	mV
		Standby or Sleep mode	-0.2	-	+0.2	V
V _{O(rec)}	recessive output voltage	recessive; no load				
J (. 55)	,	,	0	0.51/	2	V
		Normal or Listen-only mode; $V_{TXD} = V_{IO}$	2	$0.5V_{CC}$	3	V

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 Table 8.
 Static characteristics ...continued

 $V_{CC} = 4.5 \text{ V}$ to 5.5 V; $V_{IO} = 2.8 \text{ V}$ to V_{CC} ; $V_{BAT} = 4.5 \text{ V}$ to 40 V; $V_{L} = 60 \Omega$; $V_{CC} = 4.5 \text{ V}$ to $V_{CC} = 4.5 \text{ V}$ to

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
I _{O(sc)dom}	dominant short-circuit	$V_{TXD} = 0 \text{ V}; \text{ t} < \text{t}_{\text{to(dom)TXD}}; \text{ V}_{CC} = 5 \text{ V}$				
	output current	pin CANH; $V_{CANH} = -15 \text{ V to } +40 \text{ V}$	-100	-70	-40	mΑ
		pin CANL; $V_{CANL} = -15 \text{ V to } +40 \text{ V}$	40	70	100	mΑ
I _{O(sc)rec}	recessive short-circuit output current	Normal mode; $V_{TXD} = V_{IO}$; $V_{CANH} = V_{CANL} = -27 \text{ V to } +32 \text{ V}$	-3	-	+3	mA
$V_{th(RX)dif}$	differential receiver threshold voltage	$-30 \text{ V} \le \text{V}_{\text{CANL}} \le +30 \text{ V};$ $-30 \text{ V} \le \text{V}_{\text{CANH}} \le +30 \text{ V}$				
		Normal or Listen-only mode	0.5	0.7	0.9	V
		Standby or Sleep mode	0.4	0.7	1.15	V
V _{rec(RX)}	receiver recessive voltage	$-30 \text{ V} \le \text{V}_{\text{CANL}} \le +30 \text{ V};$ $-30 \text{ V} \le \text{V}_{\text{CANH}} \le +30 \text{ V}$				
		Normal or Listen-only mode	-4	-	0.5	V
Sta		Standby or Sleep mode	-4	-	0.4	V
$V_{dom(RX)}$	receiver dominant voltage	$-30 \text{ V} \le \text{V}_{CANL} \le +30 \text{ V};$ $-30 \text{ V} \le \text{V}_{CANH} \le +30 \text{ V}$				
		Normal or Listen-only mode	0.9	-	9.0	V
		Standby or Sleep mode	1.15	-	9.0	V
V _{hys(RX)dif}	differential receiver hysteresis voltage	Normal or Listen-only mode; $-30 \text{ V} \le \text{V}_{\text{CANL}} \le +30 \text{ V};$ $-30 \text{ V} \le \text{V}_{\text{CANH}} \le +30 \text{ V}$	50	120	400	mV
IL	leakage current	V _{CC} = 0 V; V _{CANH} = V _{CANL} = 5 V	100	170	250	μΑ
		V _{BAT} = 0 V; V _{CANH} = V _{CANL} = 5 V	-2	-	+2	μΑ
		$V_{CC} = V_{IO} = V_{BAT} = 0 \text{ V or}$ $V_{CC} = V_{IO} = V_{BAT} = \text{shorted to ground via}$ $47 \text{ k}\Omega; V_{CANH} = V_{CANL} = 5 \text{ V}$	-2	-	+2	μА
R _i	input resistance	$-2 \text{ V} \le \text{V}_{CANL} \le +7 \text{ V};$ $-2 \text{ V} \le \text{V}_{CANH} \le +7 \text{ V}$	2] 9	15	28	kΩ
ΔR_i	input resistance deviation	$ 0 \ V \le V_{CANL} \le +5 \ V; $ $ 0 \ V \le V_{CANH} \le +5 \ V $	2] -3	0	+3	%
$R_{i(dif)}$	differential input resistance	$-2 \text{ V} \le \text{V}_{CANL} \le +7 \text{ V};$ $-2 \text{ V} \le \text{V}_{CANH} \le +7 \text{ V}$	2 19	30	52	kΩ
$C_{i(cm)}$	common-mode input capacitance	$V_{TXD} = V_{CC}$	2] _	-	20	pF
$C_{i(dif)}$	differential input capacitance	$V_{TXD} = V_{CC}$	2] _	-	10	pF
Common-r	mode stabilization output; p	in SPLIT	'			
V _O	output voltage	Normal or Listen-only mode; –500 μA < I _{SPLIT} < 500 μA	0.3V _{CC}	0.5V _{CC}	0.7V _{CC}	V
		Normal or Listen-only mode $R_L = 1 M\Omega$	0.45V _{CC}	0.5V _{CC}	0.55V _{CC}	V
lL	leakage current	Standby or Sleep mode; -58 V < V _{SPLIT} < +58 V	-3	0	+3	μΑ
Temperatu	re detection		1	1	1	
T _{j(sd)}	shutdown junction temperature		2] _	190	-	°C

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- [1] All parameters are guaranteed over the virtual junction temperature range by design. Factory testing uses correlated test conditions to cover the specified temperature and power supply voltage range.
- [2] Not tested in production; guaranteed by design.
- [3] The test circuit used to measure the bus output voltage symmetry (which includes C_{SPLIT}) is shown in Figure 12.

11. Dynamic characteristics

Table 9. Dynamic characteristics;

 $V_{CC} = 4.5 \text{ V to } 5.5 \text{ V}$; $V_{IO} = 2.8 \text{ V to } V_{CC}$; $V_{BAT} = 4.5 \text{ V to } 40 \text{ V}$; $R_L = 60 \Omega$; $T_{Vj} = -40 ^{\circ}\text{C}$ to +150 $^{\circ}\text{C}$; unless otherwise specified; all voltages are defined with respect to ground; positive currents flow into the device.

Symbol	Parameter	Conditions		Min	Тур	Max	Unit
Timing chara	acteristics; <u>Figure 7</u>						-
t _{d(TXD-busdom)}	delay time from TXD to bus dominant	Normal mode		-	70	-	ns
t _{d(TXD-busrec)}	delay time from TXD to bus recessive	Normal mode		-	90	-	ns
t _{d(busdom-RXD)}	delay time from bus dominant to RXD	Normal or Listen-only mode		-	60	-	ns
t _{d(busrec-RXD)}	delay time from bus recessive to RXD	Normal or Listen-only mode		-	70	-	ns
t _{d(TXDL-RXDL)}	delay time from TXD LOW to RXD LOW	Normal mode		40	-	240	ns
t _{d(TXDH-RXDH)}	delay time from TXD HIGH to RXD HIGH	Normal mode		40	-	240	ns
t _{det(uv)}	undervoltage detection time			100	-	350	ms
t _{rec(uv)}	undervoltage recovery time			1	-	5	ms
t _{bit(bus)}	transmitted recessive bit width	$t_{bit(TXD)} = 500 \text{ ns}$	[2]	435	-	530	ns
		$t_{bit(TXD)} = 200 \text{ ns}$	[2]	155	-	210	ns
t _{bit(RXD)}	bit time on pin RXD	$t_{bit(TXD)} = 500 \text{ ns}$	[2]	400	-	550	ns
		$t_{bit(TXD)} = 200 \text{ ns}$	[2]	120	-	220	ns
$\Delta t_{\sf rec}$	receiver timing symmetry	$t_{bit(TXD)} = 500 \text{ ns}$		-65	-	+40	ns
		$t_{bit(TXD)} = 200 \text{ ns}$		-45	-	+15	ns
t _{to(dom)TXD}	TXD dominant time-out time	$V_{TXD} = 0 V$	[3]	0.4	0.6	1.5	ms
t _{to(dom)bus}	bus dominant time-out time	$V_{O(dif)} > 0.9 V$		0.4	0.6	1.5	ms
t _h	hold time	from issuing go-to-sleep command to entering Sleep mode		20	35	50	μS
t _{wake(busdom)}	bus dominant wake-up time	Standby or Sleep mode; V _{BAT} = 12 V		0.5	1.75	3	μS
t _{wake(busrec)}	bus recessive wake-up time	Standby or Sleep mode; V _{BAT} = 12 V		0.5	1.75	3	μS
t _{to(wake)bus}	bus wake-up time-out time			0.5	-	2	ms
t _{wake}	wake-up time	in response to a falling or rising edge on pin WAKE; Standby or Sleep mode		5	25	50	μS

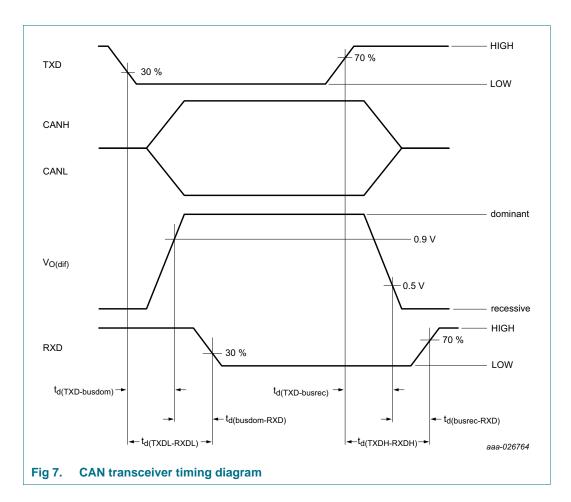
^[1] All parameters are guaranteed over the virtual junction temperature range by design. Factory testing uses correlated test conditions to cover the specified temperature and power supply voltage range.

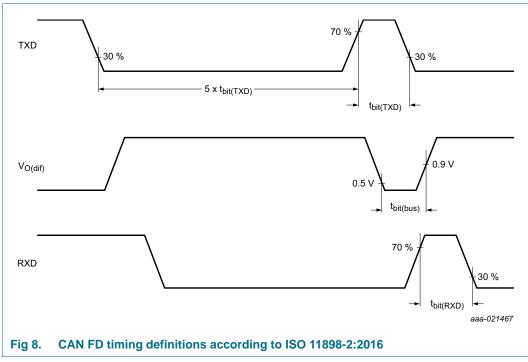
- [2] See Figure 8.
- [3] Minimum value of 0.8ms required according to SAE J2284; 0.3ms is allowed according to ISO11898-2:2016 for legacy devices.

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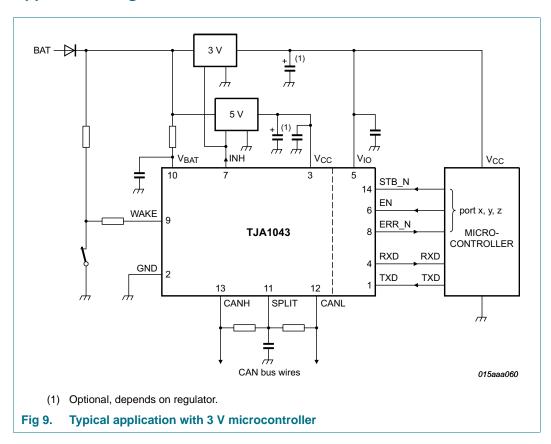
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12. Application information

12.1 Application diagram

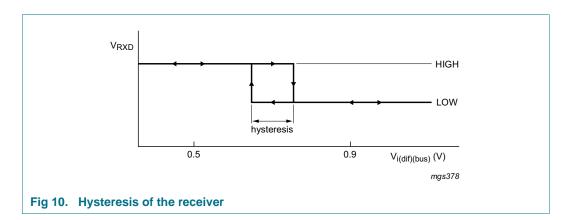


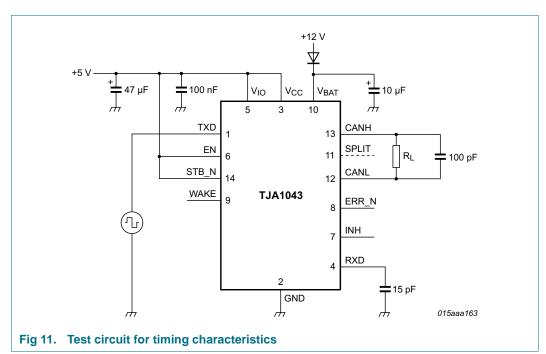
12.2 Application hints

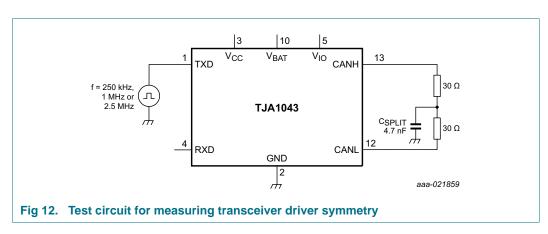
Further information on the application of the TJA1043 can be found in NXP application hints AH1014 'Application Hints - Standalone high speed CAN transceiver TJA1042/TJA1043/TJA1048/TJA1051'.

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13. Test information







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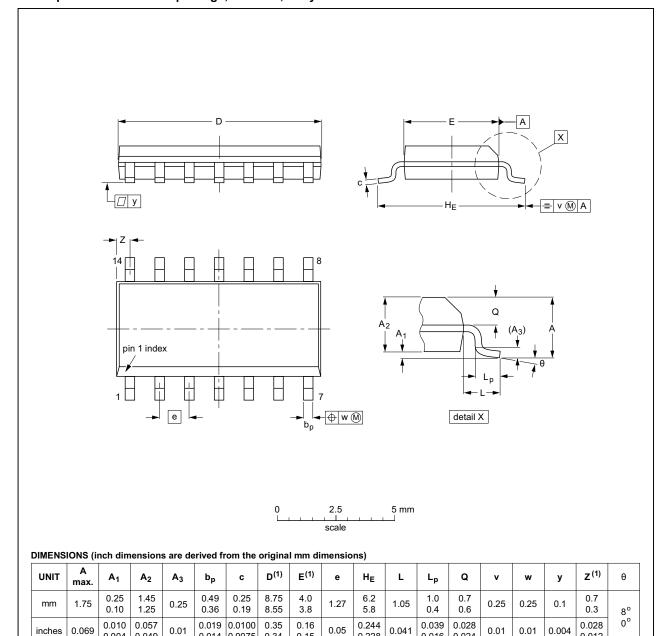
13.1 Quality information

This product has been qualified in accordance with the Automotive Electronics Council (AEC) standard *Q100 Rev-G - Failure mechanism based stress test qualification for integrated circuits*, and is suitable for use in automotive applications.

14. Package outline

SO14: plastic small outline package; 14 leads; body width 3.9 mm

SOT108-1



Note

1. Plastic or metal protrusions of 0.15 mm (0.006 inch) maximum per side are not included.

0.014 0.0075

0.34

0.15

OUTLINE		REFER	ENCES	EUROPEAN	ISSUE DATE	
VERSION	IEC	JEDEC	JEITA	PROJECTION	ISSUE DATE	
SOT108-1	076E06	MS-012			99-12-27 03-02-19	

0.228

0.016

0.024

Fig 13. Package outline SOT108-1 (SO14)

0.004

0.049

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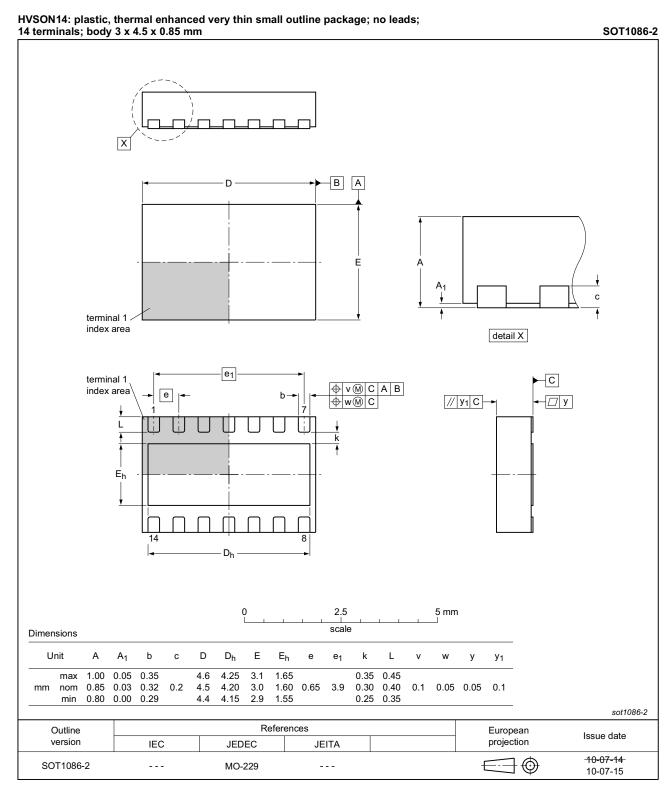


Fig 14. Package outline SOT1086 (HVSON14)

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15. Handling information

All input and output pins are protected against ElectroStatic Discharge (ESD) under normal handling. When handling ensure that the appropriate precautions are taken as described in *JESD625-A* or equivalent standards.

16. Soldering of SMD packages

This text provides a very brief insight into a complex technology. A more in-depth account of soldering ICs can be found in Application Note *AN10365* "Surface mount reflow soldering description".

16.1 Introduction to soldering

Soldering is one of the most common methods through which packages are attached to Printed Circuit Boards (PCBs), to form electrical circuits. The soldered joint provides both the mechanical and the electrical connection. There is no single soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and Surface Mount Devices (SMDs) are mixed on one printed wiring board; however, it is not suitable for fine pitch SMDs. Reflow soldering is ideal for the small pitches and high densities that come with increased miniaturization.

16.2 Wave and reflow soldering

Wave soldering is a joining technology in which the joints are made by solder coming from a standing wave of liquid solder. The wave soldering process is suitable for the following:

- Through-hole components
- Leaded or leadless SMDs, which are glued to the surface of the printed circuit board

Not all SMDs can be wave soldered. Packages with solder balls, and some leadless packages which have solder lands underneath the body, cannot be wave soldered. Also, leaded SMDs with leads having a pitch smaller than ~0.6 mm cannot be wave soldered, due to an increased probability of bridging.

The reflow soldering process involves applying solder paste to a board, followed by component placement and exposure to a temperature profile. Leaded packages, packages with solder balls, and leadless packages are all reflow solderable.

Key characteristics in both wave and reflow soldering are:

- Board specifications, including the board finish, solder masks and vias
- Package footprints, including solder thieves and orientation
- The moisture sensitivity level of the packages
- Package placement
- Inspection and repair
- Lead-free soldering versus SnPb soldering

16.3 Wave soldering

Key characteristics in wave soldering are:

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- Process issues, such as application of adhesive and flux, clinching of leads, board transport, the solder wave parameters, and the time during which components are exposed to the wave
- Solder bath specifications, including temperature and impurities

16.4 Reflow soldering

Key characteristics in reflow soldering are:

- Lead-free versus SnPb soldering; note that a lead-free reflow process usually leads to higher minimum peak temperatures (see <u>Figure 15</u>) than a SnPb process, thus reducing the process window
- Solder paste printing issues including smearing, release, and adjusting the process window for a mix of large and small components on one board
- Reflow temperature profile; this profile includes preheat, reflow (in which the board is heated to the peak temperature) and cooling down. It is imperative that the peak temperature is high enough for the solder to make reliable solder joints (a solder paste characteristic). In addition, the peak temperature must be low enough that the packages and/or boards are not damaged. The peak temperature of the package depends on package thickness and volume and is classified in accordance with Table 10 and 11

Table 10. SnPb eutectic process (from J-STD-020D)

Package thickness (mm)	Package reflow temperature (°C)			
	Volume (mm³)			
	< 350	≥ 350		
< 2.5	235	220		
≥ 2.5	220	220		

Table 11. Lead-free process (from J-STD-020D)

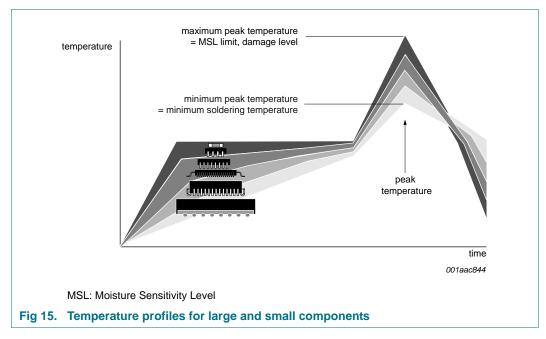
Package thickness (mm)	Package reflow temperature (°C)						
	Volume (mm³)						
	< 350	350 to 2000	> 2000				
< 1.6	260	260	260				
1.6 to 2.5	260	250	245				
> 2.5	250	245	245				

Moisture sensitivity precautions, as indicated on the packing, must be respected at all times.

Studies have shown that small packages reach higher temperatures during reflow soldering, see Figure 15.

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For further information on temperature profiles, refer to Application Note *AN10365* "Surface mount reflow soldering description".

17. Soldering of HVSON packages

<u>Section 16</u> contains a brief introduction to the techniques most commonly used to solder Surface Mounted Devices (SMD). A more detailed discussion on soldering HVSON leadless package ICs can found in the following application notes:

- AN10365 'Surface mount reflow soldering description"
- AN10366 "HVQFN application information"

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18. Appendix: ISO 11898-2:2016 parameter cross-reference list

Table 12. ISO 11898-2:2016 to NXP data sheet parameter conversion

ISO 11898-2:2016		NXP data sheet			
Parameter	Notation	Symbol	Parameter		
HS-PMA dominant output characteristics					
Single ended voltage on CAN_H	V _{CAN_H}	$V_{O(dom)}$	dominant output voltage		
Single ended voltage on CAN_L	V _{CAN_L}				
Differential voltage on normal bus load	V_{Diff}	$V_{O(dif)}$	differential output voltage		
Differential voltage on effective resistance during arbitration					
Optional: Differential voltage on extended bus load range	_				
HS-PMA driver symmetry					
Driver symmetry	V_{SYM}	V_{TXsym}	transmitter voltage symmetry		
Maximum HS-PMA driver output current					
Absolute current on CAN_H	I _{CAN_H}	I _{O(sc)dom}	dominant short-circuit output		
Absolute current on CAN_L	I _{CAN_L}		current		
HS-PMA recessive output characteristics, bus biasing ac	ctive/inactiv	/e			
Single ended output voltage on CAN_H	V _{CAN_H}	V _{O(rec)}	recessive output voltage		
Single ended output voltage on CAN_L	V _{CAN_L}				
Differential output voltage	V_{Diff}	$V_{O(dif)}$	differential output voltage		
Optional HS-PMA transmit dominant timeout					
Transmit dominant timeout, long	t _{dom}	t _{to(dom)TXD}	TXD dominant time-out time		
Transmit dominant timeout, short					
HS-PMA static receiver input characteristics, bus biasing	g active/ina	ctive			
Recessive state differential input voltage range Dominant state differential input voltage range	V_{Diff}	V _{th(RX)dif}	differential receiver threshold voltage		
		$V_{\text{rec}(RX)}$	receiver recessive voltage		
		$V_{dom(RX)}$	receiver dominant voltage		
HS-PMA receiver input resistance (matching)					
Differential internal resistance	R _{Diff}	R _{i(dif)}	differential input resistance		
Single ended internal resistance	R _{CAN_H} R _{CAN_L}	R _i	input resistance		
Matching of internal resistance	MR	ΔR_i	input resistance deviation		
HS-PMA implementation loop delay requirement					
Loop delay	t _{Loop}	t _{d(TXDH-RXDH)}	delay time from TXD HIGH to RXD HIGH		
		t _{d(TXDL-RXDL)}	delay time from TXD LOW to RXD LOW		
Optional HS-PMA implementation data signal timing requ 2 Mbit/s and above 2 Mbit/s up to 5 Mbit/s	uirements f	or use with bit	rates above 1 Mbit/s up to		
Transmitted recessive bit width @ 2 Mbit/s / @ 5 Mbit/s, intended	t _{Bit(Bus)}	t _{bit(bus)}	transmitted recessive bit width		
1	4		hit times are min DVD		
Received recessive bit width @ 2 Mbit/s / @ 5 Mbit/s	t _{Bit(RXD)}	t _{bit(RXD)}	bit time on pin RXD		

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Table 12. ISO 11898-2:2016 to NXP data sheet parameter conversion

ISO 11898-2:2016		NXP data she	et
Parameter	Notation	Symbol	Parameter
HS-PMA maximum ratings of V _{CAN_H} , V _{CAN_L} and V _{Diff}			
Maximum rating V _{Diff}	V_{Diff}	V _(CANH-CANL)	voltage between pin CANH and pin CANL
General maximum rating V _{CAN_H} and V _{CAN_L}	V _{CAN_H}	V _x	voltage on pin x
Optional: Extended maximum rating VCAN_H and VCAN_L			
HS-PMA maximum leakage currents on CAN_H and CAN	L, unpow	ered	
Leakage current on CAN_H, CAN_L	I _{CAN_H} I _{CAN_L}	IL	leakage current
HS-PMA bus biasing control timings			
CAN activity filter time, long	t _{Filter}	t _{wake(busdom)} [1]	bus dominant wake-up time
CAN activity filter time, short		twake(busrec)[1]	bus recessive wake-up time
Wake-up timeout, short	t _{Wake}	t _{to(wake)bus}	bus wake-up time-out time
Wake-up timeout, long			
Timeout for bus inactivity	t _{Silence}	t _{to(silence)}	bus silence time-out time
Bus Bias reaction time	t _{Bias}	t _{d(busact-bias)}	delay time from bus active to bias

^[1] $t_{fltr(wake)bus}$ - bus wake-up filter time, in devices with basic wake-up functionality

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19. Revision history

Table 13. Revision history

Document ID	Release date	Data sheet status	Change notice	Supersedes			
TJA1043 v.6.01	20171110	Product data sheet	-	TJA1043 v.5.01			
Modifications:	Updated to comply with ISC	D 11898-2:2016 and SAE	J22884-1 through SAE	J2284-5 specifications:			
	 Section 1: text amended 	d (2nd last paragraph)					
	 Section 2.1: text amend 	ed (1st entry)					
	 Section 7.2.5: text amended (1st paragraph) 						
	 <u>Table 8</u>: conditions added to parameters R_i, ΔR_i and R_{i(dif)}; values/conditions changed for parameters I_{CC}, V_{O(dif)}, V_{O(rec)}, I_{O(sc)dom}, V_{rec(RX)}, V_{dom(RX)}, V_{TXsym} 						
	 <u>Table 8</u>: Additional mea <u>Figure 12</u> 	asurements taken at f _{TXD}	= 1 MHz and 2.5 MHz fo	r parameter V _{TXsym} ; see			
	- Table 9: Table note 3 ac	lded					
	- Figure 8: title changed						
	• Amended Figure 7, Figure 9 and Figure 12						
	Section 12.2: reference upon	dated					
TJA1043 v.5.01	20160523	Product data sheet	-	TJA1043 v.4			
TJA1043 v.4	20150119	Product data sheet	-	TJA1043 v.3			
TJA1043 v.3	20130424	Product data sheet	-	TJA1043 v.2			
TJA1043 v.2	20110620	Product data sheet	-	TJA1043 v.1			
TJA1043 v.1	20100330	Product data sheet	-	-			

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