

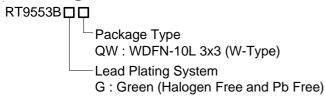
# High Accuracy Current/Voltage Sensing and Protection IC

# **General Description**

The RT9553B is designed for over-current detection and over-/under-voltage protection.

It provides users to set over-current threshold by the ILIM pin to compare that senses dropping voltage between CSP pin and CSN pin. Moreover, users can set over-/under-voltage threshold by the OVSET pin/UVSET pin to compare with the CSP voltage. The RT9553B is available in the WDFN-10L 3x3 package.

### **Ordering Information**



#### Note:

Richtek products are:

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

### **Features**

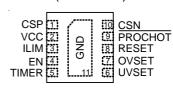
- Common Mode Input Range up to 24V
- VCC Operating Current : 200μA
- VCC Shutdown Current: 10µA (under S3/S4/S5)
- Programmable Over-Current Level
- One Shot PROCHOT Signal when OCP
- Pull Low RESET Signal when OVP/UVP
- OVP/UVP with 8% Recover Hysteresis Range
- Programmable OVP/UVP De-glitch Time

## **Applications**

Notebook

### **Pin Configuration**

(TOP VIEW)



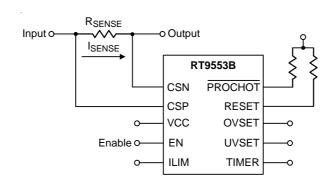
WDFN-10L 3x3

## **Marking Information**



MK=: Product Code YMDNN: Date Code

## **Simplified Application Circuit**

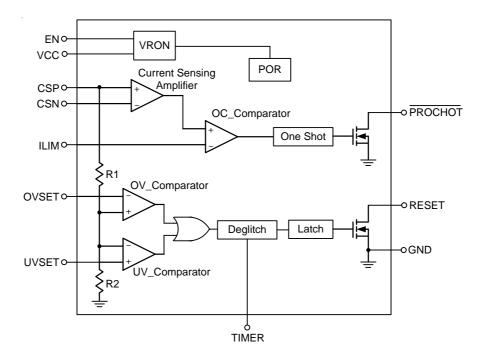




# **Functional Pin Description**

Pin No.	Pin Name	Pin Function				
1	CSP	Positive current sense input.				
2	VCC	Power supply input. Connect this pin to 5V and place a minimum $0.1\mu F$ decoupling capacitor. The decoupling capacitor should be placed to this pin as close as possible.				
3	ILIM	Over-current trip point setting input. The setting range is from 0.4V to 2V.				
4	EN	Enable control input.				
5	TIMER	Deglitch time setting of OV/UV protection. Connect a resistor from this pin tied to GND to set deglitch time. Do not parallel any filter capacitor to this pin.				
6	UVSET	Under-voltage trip point setting input. The setting range is from 0.55V to 3V.				
7	OVSET	Over-voltage trip point setting input. The setting range is from 0.55V to 3V.				
8	RESET	Open-drain output. Connect to an external resistor to pull high. When OV/UV occurs, this pin will be pulled low.				
9	PROCHOT	Open-drain output. Connect to an external resistor to pull high. When OC occurs, this pin will be pulled low.				
10	CSN	Negative current sense input.				
11 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.				

# **Functional Block Diagram**





## **Operation**

The RT9553B consists of a current sensing amplifier, and two comparators, and it provides the following functions: over-current protection, over-voltage protection and undervoltage protection.

#### **Over-Current Protection**

With m $\Omega$  order of resistor shunts between CSP and CSN, the current sensing amplifier multiplies the voltage between CSP and CSN by 100 and compares the result with the ILIM voltage. If the output voltage of current sensing amplifier is larger than the ILIM voltage, the  $\overline{\text{PROCHOT}}$  voltage is pulled low.

### **OVP/UVP Deglitch Time**

The output signal of OV\_comparator or UV\_comparator is delayed by deglitch timer. The delayed time is decided by the resistor (see spec.) connected to the TIMER pin.

### Over-/Under-Voltage Protection

The OVSET voltage is set by users and it compares with the voltage which is 1/9 of CSP voltage with OV\_comparator. The UVSET voltage is set by users and it also compares with the voltage which is 1/9 of CSP voltage with UV\_comparator. Either the output of OV\_comparator or UV\_comparator is high, the RESET voltage is pulled low.



## Absolute Maximum Ratings (Note 1)

CSP/CSN to GND	–0.3V to 28V
• VCC, ILIM, EN, UVSET, OVSET, RESET, PROCHOT to GND	-0.3V to 6V
<ul> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> </ul>	
WDFN-10L 3x3	3.27W
Package Thermal Resistance (Note 2)	
WDFN-10L 3x3, $\theta_{JA}$	30.5°C/W
WDFN-10L 3x3, $\theta_{JC}$	7.5°C/W
• Lead Temperature (Soldering, 10 sec.)	
Junction Temperature	
Storage Temperature Range	–65°C to 150°C
ESD Susceptibility (Note 3)	
HBM (Human Body Model)	2kV
Recommended Operating Conditions (Note 4)	
High-Side Voltage, VCSP/VCSN	5V to 24V

### **Electrical Characteristics**

( $V_{CC}$  = 5V,  $T_A$  = 25°C, unless otherwise specified)

Parameter		Symbol	Test Conditions	Min	Тур	Max	Unit
CSN CSP Inpu	ut	•		•			
Input Voltage F	Range	V <sub>CSP</sub> , V <sub>CSN</sub>		5		24	V
ICSN + ICSP			VCC = 5V, EN high		50		- Λ
			VCC = 5V, EN low			5	μΑ
VCC Input							
VCC Operating	g Current	Ivcc			200		μΑ
VCC Shutdown Current		I <sub>VCC_shd</sub>				10	μΑ
V00 D0D D: :		V <sub>IN_POR</sub>	Rising	2.8		3.7	V
VCC POR RISI	VCC POR Rising Voltage		Hysteresis		400		mV
Enable		•		•			
Enable Input Logic-Hi		ViH		0.7			.,
Voltage	Logic-Low	VIL				0.3	V
Over-Current	Protection	•					
System Response Time					50		μS
Open-Drain Output Duration		ТРПОСНОТ		10		18	ms
Open-Drain Ou	utput Ron	RON_PROCHOT				10	Ω

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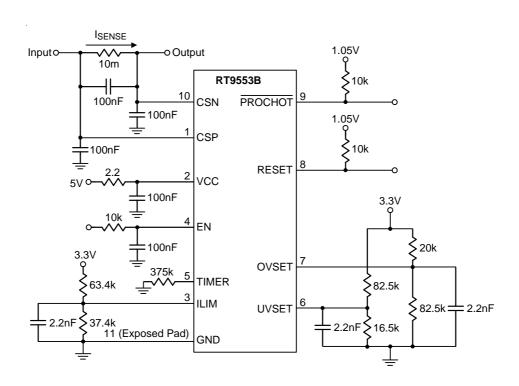


Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Input Current Sensing Accuracy	OCacc	V <sub>ILIM</sub> = 0.4V to 2V			6.5	%
PROCHOT Leakage Current	I <sub>leak_PROCHOT</sub>	EN Low			5	μΑ
ILIM Leakage Current	I <sub>leak_ILIM</sub>	EN Low			5	μΑ
ILIM Operation Range	VILIM		0.4		2	V
Over-/Under-Voltage Prot	ection		•			
OVP Accuracy	OVPacc	Vovset = 0.55V to 3V			5	%
OVP Recover Hysteresis Range		V <sub>O</sub> VSET = 0.55V to 3V		8		%
UVP Accuracy	UVP <sub>acc</sub>	V <sub>UVSET</sub> = 0.55V to 3V			5	%
UVP Recover Hysteresis Range		V <sub>UVSET</sub> = 0.55V to 3V		8		%
Open Drain Output Ron	Ron_reset	I <sub>Sink</sub> = 10mA			10	Ω
RESET Leakage Current	I <sub>leak_RESET</sub>	EN Low			5	μΑ
OVD/UVD Do glitch Time	<b>t</b>	R <sub>Timer</sub> = 375k		15		0
OVP/UVP De-glitch Time	t <sub>Timer</sub>	R <sub>Timer</sub> = 125k		5		μS

- **Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- **Note 2.**  $\theta_{JA}$  is measured under natural convection (still air) at  $T_A = 25^{\circ}\text{C}$  with the component mounted on a high effective-thermal-conductivity four-layer test board on a JEDEC 51-7 thermal measurement standard.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.

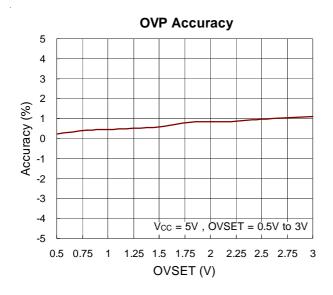


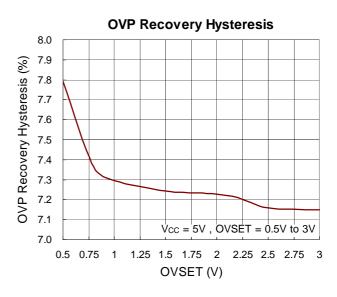
# **Typical Application Circuit**

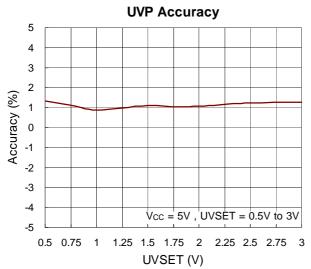


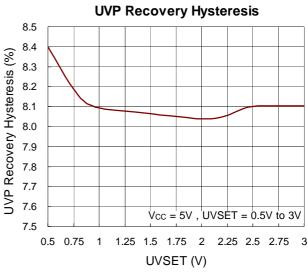


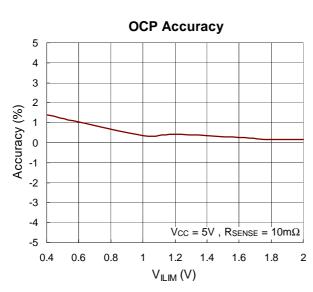
# **Typical Operating Characteristics**

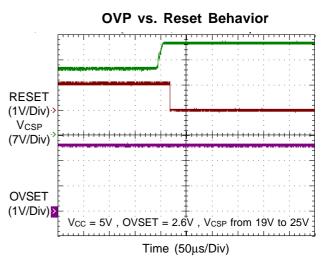










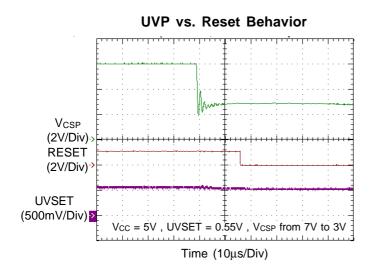


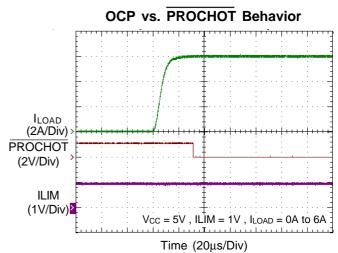
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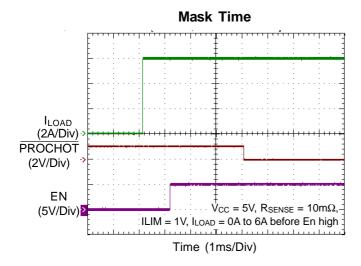
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### **Application Information**

The RT9553B provides OVP, UVP, OCP protection functions with RESET, PROCHOT indicator to inform system. It can be operated minimize external components of switching power supply systems to achieve protection. For easily using and increasing PCB space utilization, the RT9553B is used in notebook applications.

### **Over-Voltage Protection (OVP)**

The CSP pin voltage can be continuously monitored for over-voltage condition.

When the CSP pin voltage exceeds the setting threshold at the OVSET pin, the over-voltage protection will be triggered, and the RESET pin will be pull low after a deglitch time, the RESET is resumed when the CSP pin voltage down to 8% hysteresis.

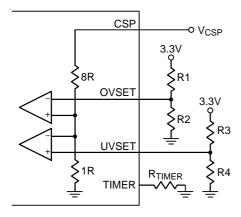
For more details, please refer to the following information.

The equation of over-voltage setting threshold is shown as below:

$$V_{OVSET} = 3.3V \times \frac{R2}{R1 + R2}$$
$$V_{CSP} = 3.3V \times \frac{R2}{R1 + R2} \times 9$$

We suggest that the total resistance of divider network should be higher than  $100k\Omega$ .

V<sub>CSP</sub> is over-voltage protection trigger point.



Fighre 1. Setting Network of Over-Voltage Protection and Under-Voltage Protection

### **Under-Voltage Protection (UVP)**

The principle of under-voltage protection is the same as over-voltage protection function. The under-voltage protection threshold is set by the UVSET pin.

When the CSP pin voltage is lower the setting threshold at the UVSET pin, the under-voltage protection will be triggered, and the RESET pin will be pull low after a deglitch time, the RESET is resumed when the CSP pin voltage up to 8% hysteresis.

### **Over-Current Protection (OCP)**

As an industry standard, high accuracy current sense amplifier is used to monitor the input current that flows through current sense resistor. The RT9553B detects CSP-CSN differential voltage across the current sense resistor to monitor input current. The equation of overcurrent protection is shown as below:

$$\begin{split} &V_{ILIM} = 3.3V \times \frac{R6}{R5 + R6} \\ &(I_{SENSE} \times R_{SENSE}) \times 100 = V_{ILIM} \\ &I_{SENSE} = \frac{V_{ILIM}}{100 \times R_{SENSE}} \end{split}$$

We suggest that the total resistance of divider network should be higher than  $100k\Omega$ .

I<sub>SENSE</sub> is over-current protection trigger point.

The 100 is amplification of internal error amplifier.

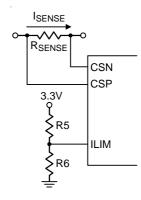


Figure 2. Setting Network of Over-Current Protection



#### **RESET/Timer**

The RESET pin is an open-drain output, and requires to be connected to a pull-up resistor. When over-voltage or under-voltage occurs, the RESET pin will be pulled low and resume to high with a hysteresis.

Users can set the deglitch time of RESET by adjusting the TIMER pin, and there are two choices of deglitch time, 15  $\mu s$  and 5  $\mu s$ . When the  $R_{TIMER}$  is  $375 k\Omega$ , the deglitch time is 15  $\mu s$ ; when the  $R_{TIMER}$  is  $125 k\Omega$ , the deglitch time is 5  $\mu s$ . Users can choose a suitable deglitch time to avoid protection false triggering.

### **PROCHOT**

The PROCHOT pin is an open-drain output, and requires to be connected to a pull-up resistor. When over-current is detected, PROCHOT will be pulled low within 50µs for 15ms. If over-current condition keeps longer than 15ms, PROCHOT maintains low status until over-current protection is released.

Figure 3, Figure 4, Figure5 illustrates RESET and PROCHOT indicator signals.

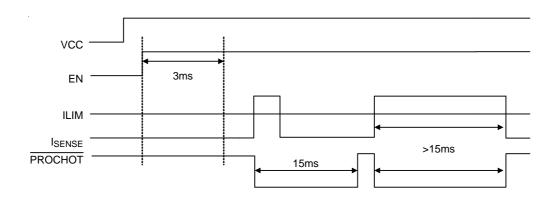


Figure 3. RESET and PROCHOT Indicator Illustration

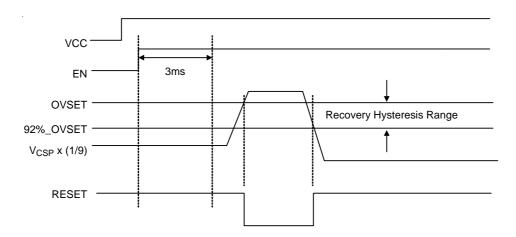


Figure 4. OVP and RESET Indicator Illustrator

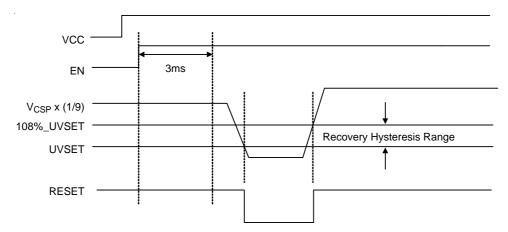


Figure 5. UVP and RESET Indicator Illustrator

#### **Thermal Considerations**

The junction temperature should never exceed the absolute maximum junction temperature  $T_{J(MAX)}$ , listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula :

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance,  $\theta_{JA}$ , is highly package dependent. For a WDFN-10L 3x3 package, the thermal resistance,  $\theta_{JA}$ , is 30.5°C/W on a standard JEDEC 51-7 high effective-thermal-conductivity four-layer test board. The maximum power dissipation at  $T_A=25^\circ\text{C}$  can be calculated as below :

 $P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (30.5^{\circ}C/W) = 3.27W$  for a WDFN-10L 3x3 package.

The maximum power dissipation depends on the operating ambient temperature for the fixed  $T_{J(MAX)}$  and the thermal resistance,  $\theta_{JA}$ . The derating curves in Figure 6 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

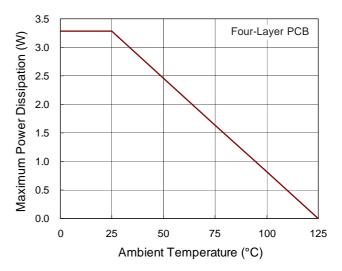


Figure 6. Derating Curve of Maximum Power Dissipation

### **Layout Considerations**

Layout is very important in the RT9553B, if designed improperly, the PCB may radiate excessive noise. Certain points must be considered before starting a layout for the RT9553B.

- ▶ Connect a RC low pass filter to VCC. The recommended BOM of RC filter is  $2.2\Omega$  and  $0.1\mu F$ . Place the filter capacitor close to the IC.
- Current sense connections must always be made using Kelvin connections to ensure an accurate signal with the current limit resistor located at the device.
- All sensitive analog traces and components such as CSP, CSN, VCC, EN, PROCHOT, RESET and ILIM should be placed away form high voltage switching nodes to avoid coupling.

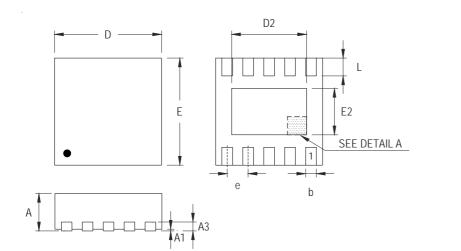
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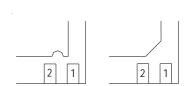
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# **Outline Dimension**





**DETAIL A** 

Pin #1 ID and Tie Bar Mark Options

Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

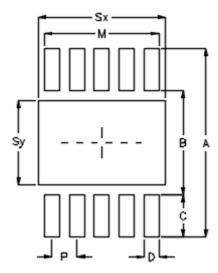
Symbol	Dimensions I	n Millimeters	Dimensions In Inches			
	Min	Max	Min	Max		
А	0.700	0.800	0.028	0.031		
A1	0.000	0.050	0.000	0.002		
А3	0.175	0.250	0.007	0.010		
b	0.180	0.300	0.007	0.012		
D	2.950	3.050	0.116	0.120		
D2	2.300	2.650	0.091	0.104		
Е	2.950	3.050	0.116	0.120		
E2	1.500	1.750	0.059	0.069		
е	0.5	500	0.020			
L	0.350	0.450	0.014	0.018		

W-Type 10L DFN 3x3 Package

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# **Footprint Information**



Package	Number of	Footprint Dimension (mm)						Tolerance		
	Pin	Р	Α	В	С	D	Sx	Sy	М	Tolerance
V/W/U/X/ZDFN3*3-10	10	0.50	3.80	2.10	0.85	0.30	2.55	1.70	2.30	±0.05

### **Richtek Technology Corporation**

14F, No. 8, Tai Yuen 1<sup>st</sup> Street, Chupei City Hsinchu, Taiwan, R.O.C.

Tel: (8863)5526789

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