

# **High Efficiency Backlight LED Driver**

# **General Description**

The RT4501/A is a high-efficiency LED driver for backlight applications. An asynchronous boost converter with an internal Schottky diode and a current source driver are designed to support 6LED/channel with wide input voltage range from 2.5V to 5.5V.

An I<sup>2</sup>C interface can provide easy backlight control in fast and high speed mode. The RT4501 supports linear mappings with 256 steps to setup the brightness of backlight LEDs. It also supports PWM dimming to adjust the brightness.

For brightness dimming, the RT4501A supports 128 steps pulse dimming which determines the LED current, and RT4501 support 256 steps PWM dimming which determines the LED current.

The RT4501/A provides complete protection functions such as input under-voltage lockout, over-current, output overvoltage and over-temperature protection. The OVP threshold voltage can be set at 16V and 25.5V for different applications. The RT4501/A is available in the WL-CSP-10B 0.87x2.07 (BSC) package.

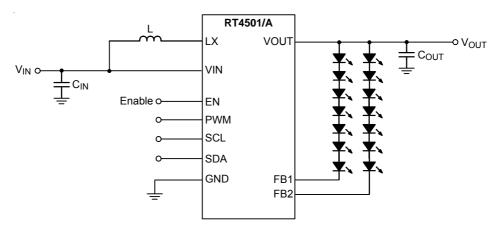
### **Features**

- Input Voltage Range: 2.5V to 5.5V
- Internal Schottky Diode
- Fast-speed mode I<sup>2</sup>C Compatible Interface
- Drive Up to 7 WLEDs in 2 String
- External PWM/Pulse Brightness Control
- 550k/1.1MHz Switching Frequency
- Built-in Internal Soft-Start
- PWM Dimming/Pulse Dimming Resolution 256/128
- I<sup>2</sup>C Programmable 256 Steps Linear Current Regulation
- Up to 85% Efficiency with Small Magnetics
- Programmable 16V/25.5V OVP
- Current Accuracy ±5% and Current Balance ±3%
- UVLO, OVP, OCP, OTP Protection
- Shutdown Current : < 1μA
- Temperature Range : -40°C to 85°C
- RoHS Compliant and Halogen Free

# **Applications**

- Cellular Phones
- Digital Cameras
- PDAs and Smart Phones
- Portable Instruments

# Simplified Application Circuit





# **Ordering Information**

RT4501/A 🗖

Package Type

WSC: WL-CSP-10B 0.87x2.07 (BSC)

RT4501: PWM Dimming RT4501A: Pulse Dimming

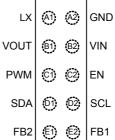
Note:

Richtek products are:

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

# **Pin Configurations**

(TOP VIEW)



WL-CSP-10B 0.87x2.07(BSC)

# **Marking Information**

RT4501WSC



1T: Product Code

W: Date Code

RT4501AWSC



28 : Product Code

W: Date Code

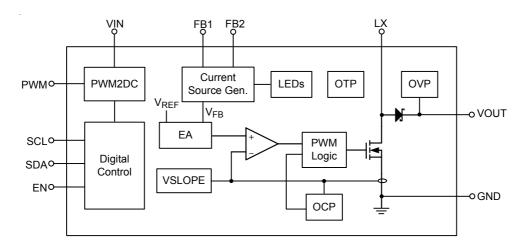
**Functional Pin Description** 

Pin No.	Pin Name	Pin Function			
A1	LX	Switch Node of Boost Converter. Connect an inductor between LX and VIN.			
A2	GND	Ground.			
B1	VOUT	Power Output of the Asynchronous Boost Converter for Backlight LE Connect a $1\mu F$ or larger ceramic capacitor from VOUT to ground.			
B2	VIN	Power Input. Connect this pin to the input power supply volta. Connect a $10\mu F$ or larger ceramic capacitor from the VIN to ground.			
C1	PWM	PWM Dimming Input for Backlight LED.			
C2	EN	Enable Control Input (Active High). The chip is in shutdown mode when the EN pin is low.			
D1	SDA	I <sup>2</sup> C Serial Data Input/Output. An external pull-up resistor is required.			
D2	SCL	I <sup>2</sup> C Serial Clock Input. An external pull-up resistor is required.			
E1	FB2	Single Output 2 for Backlight LED.			
E2	FB1	Single Output 1 for Backlight LED.			

DS4501/A-00 September 2014



# **Function Block Diagram**



# **Operation Description**

The RT4501/A is a high efficiency solution with 14 WLEDs in 2 parallels 7 series for backlight applications. The RT4501/A optimizes the feedback regulation voltage to provide up to 85% high efficiency with as high as 8bits resolution application.

### **Linear Brightness Dimming**

The RT4501/A is built-in a I<sup>2</sup>C 8-bit resolution brightness control with maximum 20mA/30mA selection. Reg0x02<3> corresponds to full-scale LED current control. Reg0x04 sets 8bits resolution brightness dimming.

### **PWM Brightness Dimming**

Besides programmable built-in I<sup>2</sup>C backlight LED current control, the RT4501 features a built-in PWM dimming current control by setting Reg0x02<6> to 1, offering a linear current dimming by external clock source. In order to guarantee the PWM dimming resolution, recommending dimming frequency have to be operated at range of 400Hz to 20kHz.

### **Pulse Brightness Dimming**

The EN pin features a simple digital interface to allow digital brightness control. Using the one-wire dimming brightness control can achieve as high as 128 steps resolution, recommending dimming pulse is larger than 0.2µs. RT4501/A keeps shunt down status, when EN pin is pulled low keeping 1ms.

#### **OCP Protection**

The RT4501/A features a 1A current limitation. Once detecting current level over current limitation, the RT4501/A's LX witching will be forced off to avoid large current damage.

### **OTP Protection**

The over-temperature protection function will be latched at shutdown status when the junction temperature exceeds 150°C for 2ms. After re-power on sequence, the converter will automatically resume switching.

### **OVP Protection**

The over-voltage protection function monitors the output voltage via the VOUT pin voltage. The OVP threshold voltage is 25.5V/16V by selection Reg0x02<7>, Once the LED is open, the output voltage will be limited at OVP protection level to avoid device breakdown. RT4501 is shunt-down latched, by triggering OVP event over 40 times.



# **Absolute Maximum Ratings** (Note 1)

Recommended Operating Conditions (Note 4)  • Supply Input Voltage, VIN	
MM (Machine Model)	200V
HBM (Human Body Model)	2kV
• ESD Susceptibility (Note 3)	
Storage Temperature Range	65°C to 150°C
• Junction Temperature	
• Lead Temperature (Soldering, 10 sec.)	260°C
WL-CSP-10B 0.87x2.07 (BSC), $\theta_{JA}$	99.6°C/W
Package Thermal Resistance (Note 2)	
WL-CSP-10B 0.87X2.07 (BSC)	1W
<ul> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> </ul>	
• Other Pins, EN, PWM, SCL, SDA	0.3V to 6V
Current Source Voltage, FB1, FB2	0.3V to 27V
• Switching Voltage, LX	0.3V to 27V
Boost Output Voltage, VOUT	0.3V to 27V
• Supply Input Voltage, VIN	0.3V to 6V

## **Electrical Characteristics**

(V<sub>IN</sub> = 3.6V, C<sub>IN</sub> =  $10\mu F$ , C<sub>OUT</sub> =  $1\mu F$ , L =  $10\mu H$ , T<sub>A</sub> =  $25^{\circ}C$ , unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit	
Input Power Supply	•			•			
Input Supply Voltage	V <sub>IN</sub>		2.5		5.5	V	
VIN Quiescent Current	IQ	PWM, No Switching		0.6		mA	
Shutdown Current	I <sub>SHDN</sub>	V <sub>IN</sub> = 4.2V, EN = GND		1	3	μΑ	
Under-Voltage Lockout Threshold	e Lockout V <sub>UVLO</sub> V <sub>IN</sub> Falling, Check I <sub>Q</sub> < 200μA				2.3	٧	
Under-Voltage Lockout Hysteresis	$\Delta V_{UVLO}$	After UVLO, V <sub>IN</sub> Rising, Until I <sub>Q</sub> > 200µA		200		mV	
Backlight LED Current Source							
Accuracy of Output Current	I <sub>LED_ACC</sub>	FB1, FB2 = 0.15V, I <sub>FB1</sub> = I <sub>FB2</sub> = 20mA	-5		5	%	
Matching of Output Current	I <sub>LED_MAT</sub>	FB1, FB2 = 0.15V, I <sub>FB1</sub> = I <sub>FB2</sub> = 20mA	-3		3	%	
Oscillator and Timing							
Operating Frequency	perating Frequency   f <sub>SW</sub>   Reg0x03 [6] = 1		0.88	1.1	1.32	MHz	
Maximum Duty Cycle	D <sub>MAX</sub>	FB1 = FB2 = 0V, check MAX duty	90	95		%	
Brightness Ramp Rate T <sub>RAMP</sub> Reg0x03 [5		Reg0x03 [5:0] = 111111	209.72	262.14	314.58	ms	

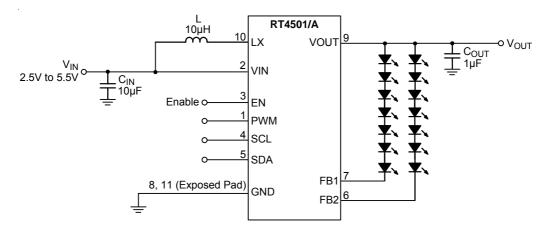


Parameter		Symbol	Test Conditions	Min	Тур	Max	Unit	
Power Switch								
N-MOSFET On-Resistance		R <sub>DS(ON)_N</sub>	V <sub>IN</sub> = 3.6V	90	175	350	mΩ	
N-MOSFET Cu	urrent Limit	IOCP		0.8	1	1.2	Α	
N-MOSFET Le Current	akage	I <sub>LEAK</sub>	LX = 24V, No Switching			1	μА	
Protection Fu	nction							
Over Veltage F	0 1/4 5 / 4		Reg0x02 [7] = 0	14	16	18	V	
Over Voltage F	Totection	OVP	Reg0x02 [7] = 1	24.5	25.5	26.5	V	
Thermal Shutd Threshold	own	T <sub>SD</sub>	Thermal latch		150		°C	
Logic Control								
EN Input	Logic-High	VENH		1.3			.,	
Voltage	Logic-Low	V <sub>ENL</sub>				0.4	V	
PWM Input	Logic-High	V <sub>PWML</sub>		1.3			V	
Voltage	Logic-Low	V <sub>PWML</sub>				0.4		
SCL Input	Logic-High	V <sub>SCLH</sub>		1.3			V	
Voltage	Logic-Low	V <sub>SCLL</sub>				0.4	V	
SDA Input	Logic-High	VSDAH		1.3			V	
Voltage	Logic-Low	V <sub>SDAL</sub>				0.4	V	
EN Pull-Down	Resistance				400		kΩ	
PWM Pull-Dow	n Resistance				400		kΩ	
Clock Frequen	Clock Frequency of SCL					400	kHz	
Pulse Dimmin	g Control							
EN Minimum Shut Down Pulse Width		toff		1			ms	
EN Minimum Logic High Pulse Width		t <sub>High_Min</sub>		0.2			μS	
EN Minimum Logic Low Pulse Width		t <sub>Low</sub>		0.2		500	μS	

- **Note 1.** Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.  $\theta_{JA}$  is measured at  $T_A = 25^{\circ}C$  on a high effective thermal conductivity four-layer test board per JEDEC 51-7.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.

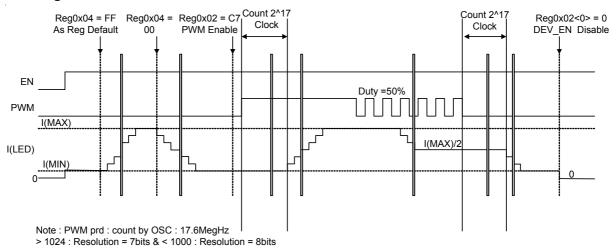


# **Typical Application Circuit**

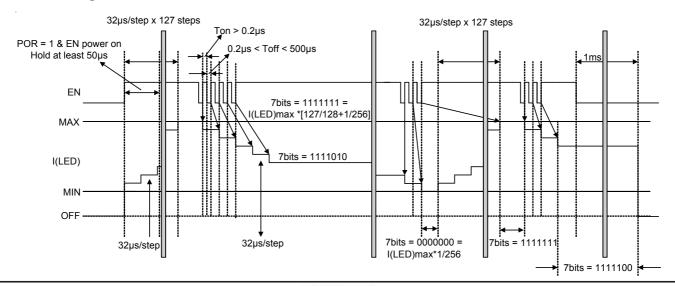


# **Timing Diagram**

### **PWM Dimming**



### **Pulse Dimming**



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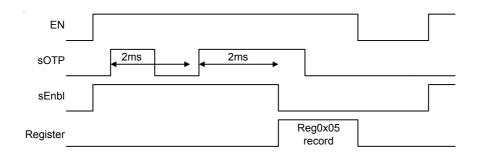
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DS4501/A-00 September 2014

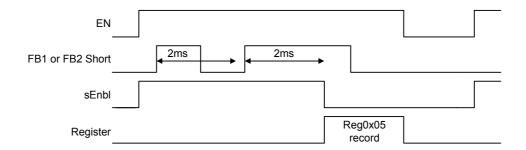


## **Protection Timing**

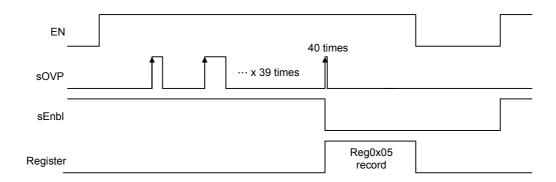
**OTP** 



### **LED Short**

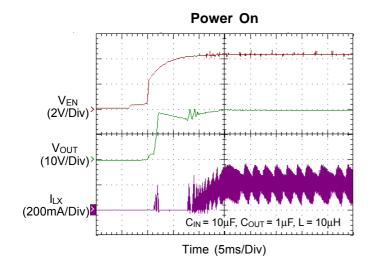


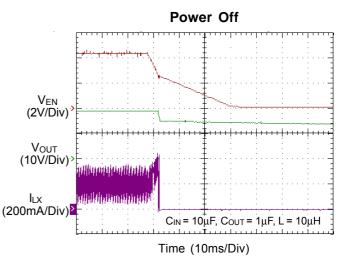
## **LED Open**

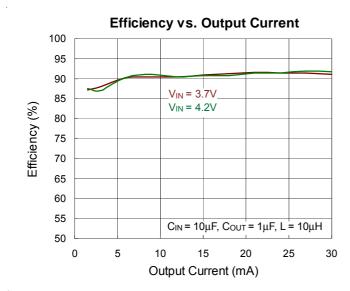


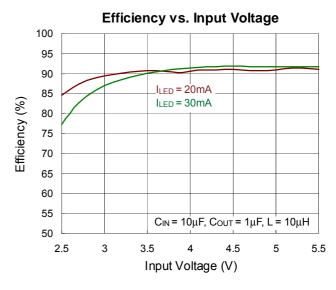


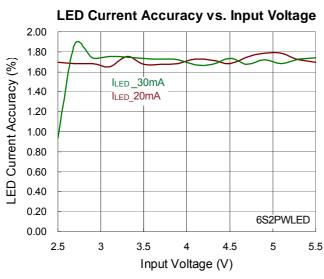
# **Typical Operating Characteristics**

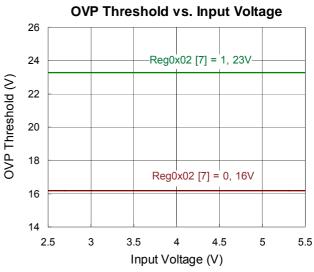






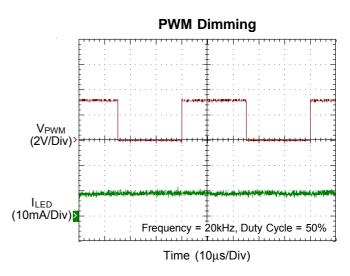


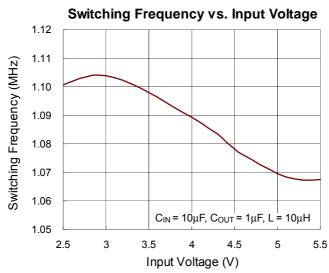


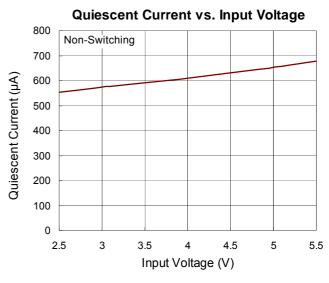


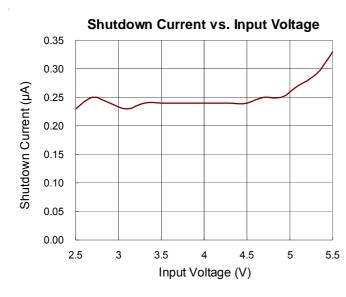


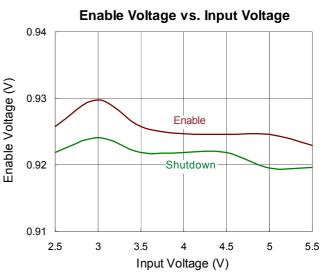














## Table 1. Pulse Dimming Register Map

### Slave address =0100010x

Address	Address Name	ВІТ	LABEL	Default (Reset Value)	Description
0x00	Device ID	7:0	DEV_ID		
0v01	Ovota Manufat	7:2	REV	000000	Revision number
UXUT	0x01 Manufacture		VID	11	Vendor ID : Richtek
		7	OVPsel	1	OVP threshold ([0] 16V, [1] 25.5V)
		6	Reserved		
		5	Reserved		
0x02	Confinal	4	Reserved		
0x02	Config1	3	MAX_Current	0	[0] 20mA, [1] 30mA,
		2	LED1_EN	1	Backlight LED1 : [0] OFF, [1] ON
		1	LED2_EN	1	Backlight LED2 : [0] OFF, [1] ON
		0	DEV_EN	1	CHIP enable : [0] OFF, [1] ON
		7	RST_SW	0	Software reset : [0] Disable(Auto), [1] Reset all registers
0x03	Timing	6	FSW	1	Switching frequency ([0]550kHz, [1]1.1MHz)
		5:0	Reserved		
	FLAG	7:4	Reserved	0	
		3	LED1_FT	0	LED1 short : [0] Normal, [1] Fault
0x05		2	LED2_FT	0	LED2 short : [0] Normal, [1] Fault
		1	OVP	0	Output over voltage : [0] Normal, [1] Fault
		0	ОТР	0	Over temperature : [0] Normal, [1] Fault



## Table 2. PWM Dimming Register Map

### Slave address =0100010x

Address	Address Name	BIT	LABEL	Default (Reset Value)	Description
0x00	Device ID	7:0	DEV_ID		
0x01	Manufactura	7:2	REV	000000	Revision number
	Manufacture	1:0	VID	11	Vendor ID : Richtek
		7	OVPsel	1	OVP threshold ([0] 16V, [1] 25.5V)
		6	PWM_EN	0	PWM enable : [0] Ignored, [1] Enable
		5	PWM_SET	0	PWM active setup : [0] High active, [1] Low active
0x02	Config1	4	Reserved		
	9	3	MAX_Current	0	[0] 20mA, [1] 30mA,
		2	LED1_EN	1	Backlight LED1 : [0] OFF, [1] ON
		1	LED2_EN	1	Backlight LED2 : [0] OFF, [1] ON
		0	DEV_EN	1	CHIP enable : [0] OFF, [1] ON
	Timing	7	RST_SW	0	Software reset : [0] Disable(Auto), [1] Reset all registers
0x03		6	FSW	1	Switching frequency ([0]550kHz, [1]1.1MHz)
		5:3	UP_RATE	000	Brightness ramp-up rate : [000] 32µs, [001] 4.096ms, [010] 8.192ms, [011] 16.383ms, [100] 32.768ms, [101] 65.536ms, [110] 131.072ms, [111] 262.144ms
		2:0	DN_RATE	000	Brightness ramp-down rate : [000] 32µs, [001] 4.096ms, [010] 8.192ms, [011] 16.383ms, [100] 32.768ms, [101] 65.536ms, [110] 131.072ms, [111] 262.144ms
0x04	Linear Brightness Control	7:0	BRIGHT_LIN	11111111	[00000000] 0.39%, [00000001] 0.39%, [11111111] 100%
		7:4	Reserved	0	
	FLAG	3	LED1_FT	0	LED1 short : [0] Normal, [1] Fault
0x05		2	LED2_FT	0	LED2 short : [0] Normal, [1] Fault
		1	OVP	0	Output over voltage : [0] Normal, [1] Fault
		0	ОТР	0	Over temperature : [0] Normal, [1] Fault



# **Application Information**

#### **LED Short Protection**

LED short protection prevents abnormal connection to cause IC damage avoiding FB1/FB2 connecting power supply. And, If unbalanced LEDs series (cause FB1 or FB2 >12V) is different between channel1 and channel2, IC will also occur LED short event. As LED short event occur more than deglitch time 2ms, IC will shut-down latch until IC is reset by EN pin.

#### Soft-Start

The RT4501/A includes a soft-start function to avoid high inrush current during start-up. The soft-start function is achieved by clamping the output voltage of the error amplifier with another voltage source that is increased slowly from zero to near VIN during the soft-start period.

#### **OCP Protection**

The RT4501/A features a 1A current limitation. The current flowing through the inductor during a charging period is detected by a current sensing circuit. If the value exceeds the current limit, the N-MOSFET will be turned off. The inductor will then be forced to leave charging stage and enter discharging stage. Therefore, the inductor current will not increase to reach current limit.

The over-temperature protection function will be latched at shutdown status when the junction temperature exceeds 165°C for 2ms. After re-power on sequence, the converter will automatically resume switching.

#### **OVP Protection**

The chip provides over-voltage protection function to limit the output voltage in abnormal conditions. The OVP threshold voltage is 25.5V/16V by selecting Reg0x02 <7>. Once the LED is open, the output voltage will be limited at OVP protection level to avoid device breakdown.

#### **Under-Voltage Lockout**

An under-voltage lockout circuit prevents the operation of the device at input voltages below under-voltage threshold (2.3V maximum). When the input voltage is below the threshold, the device is shut down. If the input voltage rises over under-voltage lockout hysteresis (200mV typical), the IC restarts.

### **Linear Brightness Dimming**

The chip is built-in an I<sup>2</sup>C 8-bit resolution brightness control with maximum 20mA/30mA selection. Reg0x02 <3> corresponds to full-scale LED current control. Reg0x04 <7:0> sets 8bits resolution brightness dimming control.

$$I_{LED} = \frac{Code}{255} \times I_{LED, Full}$$

Where

I<sub>LED, Full</sub>: the full-scale LED current set by Reg0x02 <3>.

Code : the 8bit brightness code Reg0x04 <7:0> programmed by  $I^2C$  interface.

### **PWM Brightness Dimming**

Besides programmable built-in I<sup>2</sup>C backlight LED current control, the RT4501/A features a built-in PWM dimming current control by setting Reg0x02 <6> to 1, offering a linear current dimming by external clock source. In order to guarantee the PWM dimming resolution (7 bit at > 15kHz application), recommending dimming frequency have to be operated at range of 400Hz to 20kHz.

### **Pulse Brightness Dimming**

The EN pin features a simple digital interface to allow digital brightness control. Using the one-wire dimming brightness control can achieve as high as 128 steps resolution, recommending dimming pulse is larger than  $0.2\mu s$ . RT4501/A keeps shunt down status, when EN pin is pulled low keeping 1ms.

$$I_{LED} = \left(\frac{Code}{128} + \frac{1}{256}\right) \times I_{LED, Full}$$

Where

I<sub>LED, Full</sub>: the full-scale LED current set by Reg0x02 <3>.

#### **Inductor Selection**

The recommended inductor value for dual-channel 6 WLEDs applications is  $10\mu H$ . When selecting the inductor, the inductor rated saturation current should be higher than the peak current at maximum load. Small size and better efficiency are major concerns for portable devices. The inductor should have low core loss at 1.1MHz and low DCR for better efficiency.

### **Capacitor Selection**

 $10\mu F$  input ceramic capacitor and  $1\mu F$  output ceramic capacitor are recommended for driving dual-channel 6 WLEDs applications. For better voltage filtering, ceramic capacitors with low ESR are recommended. X5R and X7R types are suitable because of their wide voltage and temperature ranges.

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum junction temperature. The maximum power dissipation depends on the thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For WL-CSP-10B 0.87x2.07 (BSC) package, the thermal resistance,  $\theta_{JA}$ , is 99.6°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at  $T_A$  = 25°C can be calculated by the following formula:

 $P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (99.6^{\circ}C/W) = 1W$  for WL-CSP-10B 0.87x2.07 (BSC) package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal

resistance,  $\theta_{JA}$ . The derating curve in Figure 1 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

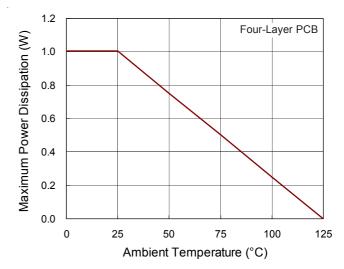


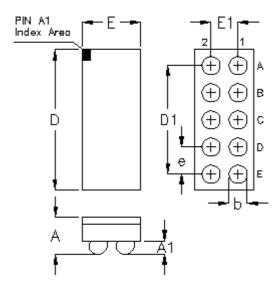
Figure 1. Derating Curve of Maximum Power Dissipation

#### **Layout Consideration**

As for all switching power supplies, the layout is an important step in the design, especially at high peak currents and switching frequencies. If the layout is not carefully done, the regulator might expose noise problems and duty cycle jitter. Therefore, use wide and short traces for high current paths. The input capacitor should be placed as close as possible to the input pin for good input voltage filtering. The inductor should be placed as close as possible to the switch pin to minimize the noise coupling into other circuits. The output capacitor needs to be placed directly from the VOUT pin to GND rather than across the LEDs. This reduces the ripple current in the trace to the LEDs. When doing the PCB layout, the bold traces should be routed first, as well as placement of the inductor, and input and output capacitors.



## **Outline Dimension**



Symbol	Dimensions I	n Millimeters	Dimensions In Inches		
	Min.	Max.	Min.	Max.	
Α	0.500	0.600	0.020	0.024	
A1	0.170	0.230	0.007	0.009	
b	0.240	0.300	0.009	0.012	
D	2.020	2.120	0.080	0.083	
D1	1.6	600	0.0	063	
E	0.820	0.920	0.032	0.036	
E1	0.4	100	0.016		
е	0.4	100	0.0	016	

10B WL-CSP 0.87x2.07 Package (BSC)

## **Richtek Technology Corporation**

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