

SGM8604-2XTDU8G Itput Drive, High Precision,

15MHz, High Output Drive, High Precision, Low Noise Operational Amplifier

GENERAL DESCRIPTION

The SGM8604-2XTDU8G is a dual, low noise, high precision CMOS operational amplifier that provides a high output current of 235mA, rail-to-rail output operation from a range of 2.7V to 5.5V single supply.

The SGM8604-2XTDU8G offers low input offset voltage, low input offset voltage drift and high output current drive. The device also can achieve a high 15MHz gain-bandwidth product and a high 7V/µs slew rate.

The SGM8604-2XTDU8G is specifically designed to drive high current load, such as 32Ω headset, V_{BIAS} of RF power amplifier, etc.

The SGM8604-2XTDU8G is available in a Green TDFN-3×3-8AL package. It operates over an ambient temperature range of -40°C to +125°C.

APPLICATIONS

 V_{BIAS} of RF Power Amplifiers Portable Stereo Headphone Driver s (32 Ω) Battery-Powered Equipment Audio System Optical Module DAC Buffer Industrial Equipment

FEATURES

Output Drive Capability: 235mA

Low Input Offset Voltage: 10μV (MAX)

• Low Input Offset Voltage Drift: 20nV/°C (TYP)

Low Noise: 28nV/√Hz at 1kHz

Gain-Bandwidth Product: 15MHz

High Slew Rate: 7V/µs

High Open-Loop Gain (R_L = 2kΩ): 145dB

Power Supply Rejection Ratio: 120dB

• Over-Temperature Protection

No Phase Reversal for Overdriven Inputs

• Rail-to-Rail Input and Output

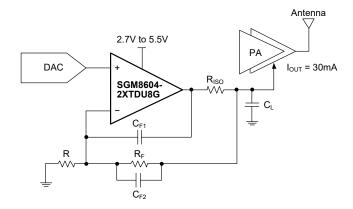
Supply Voltage Range: 2.7V to 5.5V

• Quiescent Current: 2.3mA (TYP)

-40°C to +125°C Operating Temperature Range

Available in a Green TDFN-3×3-8AL Package

TYPICAL APPLICATION

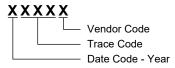


PACKAGE/ORDERING INFORMATION

MODEL	PACKAGE DESCRIPTION	SPECIFIED TEMPERATURE RANGE	ORDERING NUMBER	PACKAGE MARKING	PACKING OPTION
SGM8604-2XTDU8G	TDFN-3×3-8AL	-40°C to +125°C	SGM8604-2XTDU8G/TR	SGM RC7DU XXXXX	Tape and Reel, 4000

MARKING INFORMATION

NOTE: XXXXX = Date Code, Trace Code and Vendor Code.



Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, +V _S to -V _S	6V
All Other Pins	$(-V_S)$ - 0.3V to $(+V_S)$ + 0.3V
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (Soldering, 1	0s)+260°C
ESD Susceptibility	
HBM	7000V
CDM	1000\/

RECOMMENDED OPERATING CONDITIONS

Operating Tempera	ature Range	40°C to +125°C
Operating Supply V	/oltage Range	2.7V to 5.5V

OVERSTRESS CAUTION

Stresses beyond those listed in Absolute Maximum Ratings may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect reliability. Functional operation of the device at any conditions beyond those indicated in the Recommended Operating Conditions section is not implied.

ESD SENSITIVITY CAUTION

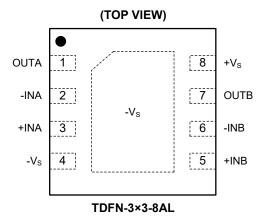
This integrated circuit can be damaged if ESD protections are not considered carefully. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle

performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because even small parametric changes could cause the device not to meet the published specifications.

DISCLAIMER

SG Micro Corp reserves the right to make any change in circuit design, or specifications without prior notice.

PIN CONFIGURATION



NOTE: The exposed pad must be connected to -V $_{\rm S}$ or left floating. Connect it to -V $_{\rm S}$ plane can maximize thermal performance.

ELECTRICAL CHARACTERISTICS

 $(V_S = 2.7V \text{ to } 5V, -V_S = 0V, V_{CM} = V_S/2, V_{OUT} = V_S/2, R_L \text{ connected to } V_S/2, \text{ Full } = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}, \text{ typical values are at } T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted.})$

PARAMETER	CONDITIONS		TEMP	MIN	TYP	MAX	UNITS	
Input Characteristics								
	V _S = 2.7V		+25°C		2	8	\/	
Input Offset Voltage (Vos)	V _S = 5V		+25°C		3	10	μV	
	V _S = 2.7V		+25°C		24	126	1400	
Input Offset Voltage Drift ($\Delta V_{OS}/\Delta T$)	V _S = 5V		+25°C		20	130	nV/°C	
Input Bias Current (I _B)	V _S = 5V		+25°C		300	520	pА	
Input Offset Current (I _{OS})	V _S = 5V		+25°C		500	1000	pА	
Common Mode Voltage Range (V _{CM})			Full	(-V _S) - 0.1		(+V _S) + 0.1	V	
Common Mode Rejection Ratio (CMRR)	(-V _S) - 0.1V < V _{CM} < (+V _S) + 0.1V	,	+25°C	104	116		dB	
(OWITH)			Full		4.40			
		$R_L = 2k\Omega$		(-V _S) - 0.1	140			
	$V_S = 2.7V$, $(-V_S) + 0.2V < V_{OUT} < (+V_S) - 0.2V$				405			
	(-Vs) 1 0.2V \ Vout \ (1 Vs) - 0.2V	$R_L = 200\Omega$			135			
Open-Loop Voltage Gain (A _{OL})							dB	
		$R_1 = 2k\Omega$	+25°C	115 112 110	145			
	V _S = 5V,		Full	112				
	$(-V_S) + 0.2V < V_{OUT} < (+V_S) - 0.2V$	B 2000	+25°C	110	140			
		11[- 20012	Full	108				
Output Characteristics								
		$R_L = 32\Omega$ $+25^{\circ}C$ Full		245	300			
	R _L = 20		Full			370		
			+25°C		45	68		
		$R_L = 200\Omega$	Full			73		
	V _S = 2.7V		+25°C		4	12		
		$R_L = 2k\Omega$	Full			14		
		$R_{L} = 2k\Omega \qquad \qquad Full \qquad 1$ $R_{L} = 200\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 2k\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 200\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 32\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 32\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 200\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$ $R_{L} = 200\Omega \qquad \frac{+25^{\circ}C}{Full} \qquad 1$		62	100			
		$I_{OUT} = 10mA$		C C C C C C C C C C		122		
Output Voltage Swing from Rail			+25°C		360	485	mV	
		$R_L = 32\Omega$				585		
					65	95		
		$R_L = 200\Omega$				113		
	V _S = 5V				7	15		
		$R_L = 2k\Omega$				18		
					54	85		
	I _{OUT} = 10mA				J 4	102		
			+25°C	Q.F.	120	102		
	V _S = 2.7V				120		-	
Output Short-Circuit Current (I _{SC})	V _S = 5V		Full		007		- mA	
			+25°C		235			
			Full	154				

ELECTRICAL CHARACTERISTICS (continued)

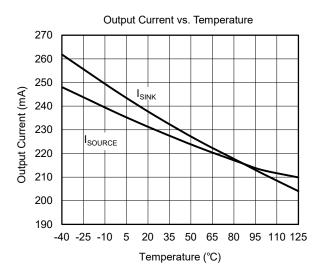
 $(V_S = 2.7V \text{ to } 5V, -V_S = 0V, V_{CM} = V_S/2, V_{OUT} = V_S/2, R_L \text{ connected to } V_S/2, \text{ Full } = -40^{\circ}\text{C} \text{ to } +125^{\circ}\text{C}, \text{ typical values are at } T_A = +25^{\circ}\text{C}, \text{ unless otherwise noted.})$

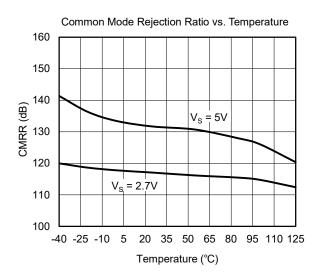
PARAMETER	CONDITIONS	TEMP	MIN	TYP	MAX	UNITS
Power Supply		•			•	•
Specified Voltage Range (V _S)		Full	2.7		5.5	V
Power Supply Rejection Ratio	V _S = 2.7V to 5.5V	+25°C	102	120		dB
(PSRR)	V _S = 2.7 V to 5.5 V	Full	94			ub ub
Quincoant Current (L.)	1 -00	+25°C		2.3	3	mA
Quiescent Current (IQ)	$I_{OUT} = 0A$	Full			3.64	mA
Frequency Response						
Gain-Bandwidth Product	$R_L = 10k\Omega$, $C_L = 100pF$	+25°C		15		MHz
Slew Rate	G = +1, V _{OUT} = 2V _{P-P}	+25°C		7		V/µs
Total Harmonic Distortion + Noise	$V_S = 5V$, $R_L = 32\Omega$, $f = 10kHz$, $BW = 10Hz$ to $90kHz$, $V_{OUT} = 2V_{P-P}$, $A_{VCL} = 1V/V$	+25°C		0.017		%
Input Capacitance		+25°C		20		pF
Channel-to-Channel Isolation	$f = 1kHz$, $R_L = 100k\Omega$	+25°C		125		dB
Capacitive-Load Stability	A _{VCL} = 1V/V, no sustained oscillations	+25°C		780		pF
Power-Up Time	V _S = 5V	+25°C		50		μs
Noise	•					
Input Voltage Noise	f = 0.1Hz to 10Hz	+25°C		0.5		μV _{P-P}
Input Valtage Naige Density	f = 1kHz	+25°C		28		> // / II=
Input Voltage Noise Density	f = 10kHz	+25°C		25		nV/√Hz

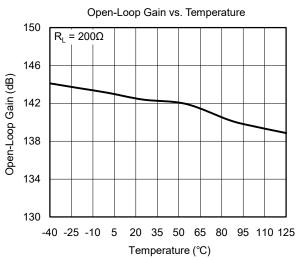


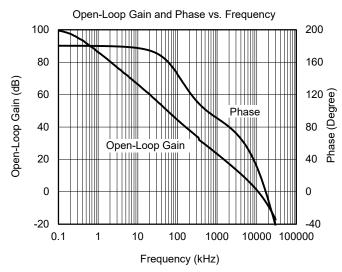
TYPICAL PERFORMANCE CHARACTERISTICS

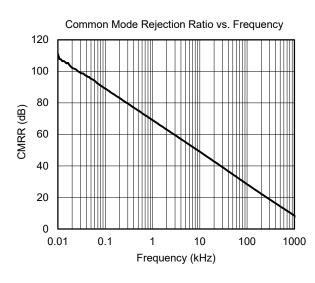
At T_A = +25°C, V_S = 5V, unless otherwise noted.

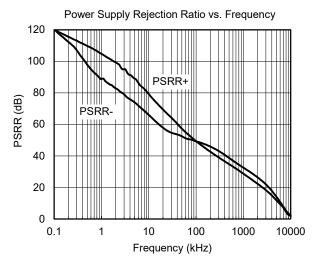






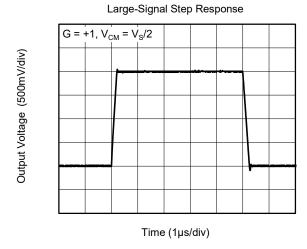


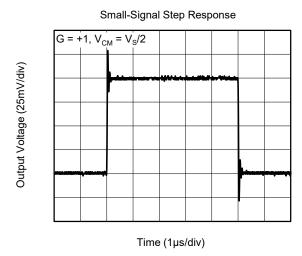


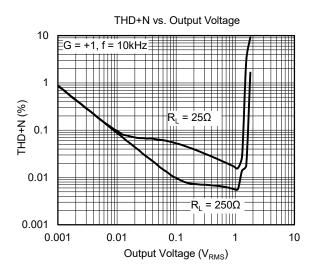


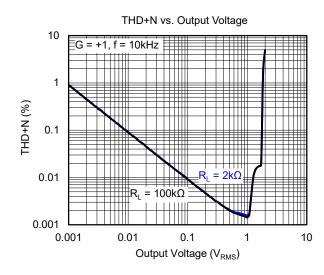
TYPICAL PERFORMANCE CHARACTERISTICS (continued)

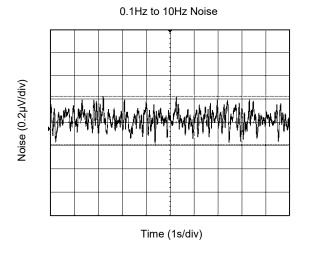
At $T_A = +25$ °C, $V_S = 5V$, unless otherwise noted.

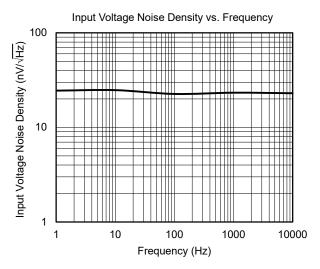






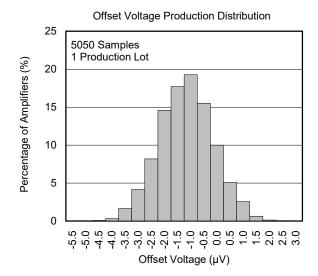






TYPICAL PERFORMANCE CHARACTERISTICS (continued)

At $T_A = +25$ °C, $V_S = 5V$, unless otherwise noted.



APPLICATION INFORMATION

Single-Supply Stereo Headphone Driver

A single-supply stereo headphone driver is shown in Figure 1 as an example to explain the simplified design procedure.

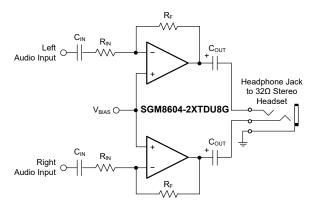


Figure 1. Stereo Headphone Driver

In this circuit, C_{IN} and R_{IN} form a high-pass filter, the DC bias is removed from the incoming signal. The -3dB point of the high-pass filter is using Equation 1:

$$f_{-3dB} = \frac{1}{2\pi R_{IN} C_{IN}}$$
 (1)

The gain of driver is -R_F/R_{IN}. The C_{OUT} and the load impedance form a high-pass filter with the -3dB point determined by Equation 2:

$$f_{-3dB} = \frac{1}{2\pi R_L C_{OUT}}$$
 (2)

Bridge Amplifier

A bridge amplifier circuit which can provide 200mW at 3V is shown in Figure 2. Due to differential output, this structure eliminates the large coupling capacitors in Figure 1. The voltage gain is 10V/V and the gain can be changed by changing R_2 .

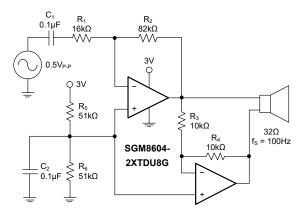


Figure 2. 200mW Bridge Amplifier at 3V

APPLICATIONS INFORMATION (continued)

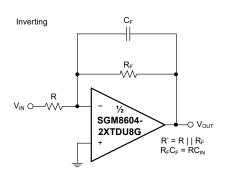
Cancel Input Capacitance

The C_{IN} (20pF TYP) at inverting input pin will generate a pole at frequency $(2\pi R'C_{IN})^{-1}$, where R' is the parallel combination of the gain-setting resistor for the inverting or non-inverting amplifier in Figure 3. If the pole-frequency is less than or comparable to the unity-gain bandwidth (15MHz), the phase margin will be reduced, ringing in the step response or sustained oscillation will be generated. To cancel this pole, C_F is used to compensate C_{IN} in Figure 3. Equation 3 gives the C_F feedback capacitance.

$$C_F = 8 \times (R/R_F) pF$$
 (3)

where:

 R_F is the feedback resistor. R is the gain-setting resistor.



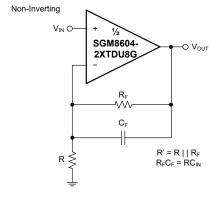


Figure 3. Inverting and Non-Inverting Amplifiers with C_{F} to Compensate C_{IN}

Input Current-Limit Protection

For ESD diode clamping protection, when the current flowing through ESD diode exceeds the maximum rating value, the ESD diode and amplifier will be damaged, so current-limit protection will be added in some applications. One resistor is selected to limit the current not to exceed the maximum rating value. In Figure 4, a series input resistor is used to limit the input current to less than 10mA, but the drawback of this current-limit resistor is to contribute thermal noise at the amplifier input. If this resistor must be added, its value must be selected as small as possible.

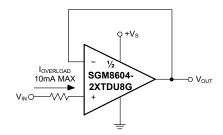


Figure 4. Input Current-Limit Protection

Rail-to-Rail Output

The SGM8604-2XTDU8G supports rail-to-rail output operation. In single power supply application, for example, when $+V_S = 5V$, $-V_S = GND$, $2k\Omega$ load resistor is tied from OUT pin to $V_S/2$, the typical output swing range is from 0.007V to 4.993V.

APPLICATIONS INFORMATION (continued)

Driving Capacitive Loads

The SGM8604-2XTDU8G is designed for unity-gain stable for capacitive load up to 780pF. In Figure 5, it shows the transient response with capacitive load (C_L). If greater capacitive load must be driven in application, the circuit in Figure 6 can be used. In this circuit, the IR drop voltage generated by $R_{\rm ISO}$ is compensated by feedback loop.

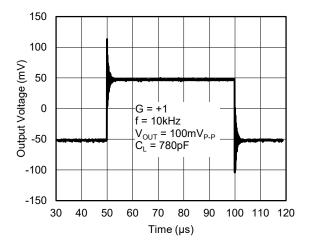


Figure 5. Small-Signal Transient Response (Capacitive Load)

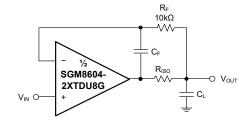


Figure 6. Circuit to Drive Capacitive Load

Power Supply Decoupling and Layout

A clean and low noise power supply is very important in amplifier circuit design, besides of input signal noise, the power supply is one of important source of noise to the amplifiers through $+V_S$ and $-V_S$ pins. Power supply bypassing is an effective method to clear up the noise at power supply, and the low impedance path to ground of decoupling capacitor will bypass the noise to GND. In application, $10\mu F$ ceramic capacitor paralleled with $0.1\mu F$ or $0.01\mu F$ ceramic capacitor is used in Figure 7. The ceramic capacitors should be placed as close as possible to $+V_S$ and $-V_S$ power supply pins.

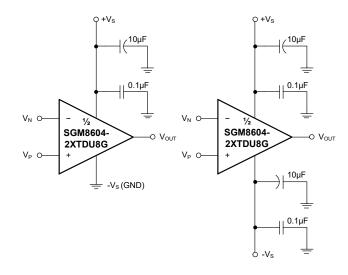


Figure 7. Amplifier Power Supply Bypassing

15MHz, High Output Drive, High Precision, **Low Noise Operational Amplifier**

SGM8604-2XTDU8G

REVISION HISTORY

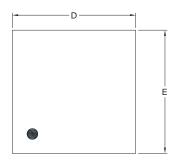
NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

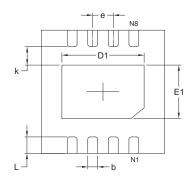
Changes from Original (AUGUST 2021) to REV.A

Page



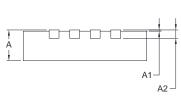
PACKAGE OUTLINE DIMENSIONS TDFN-3×3-8AL

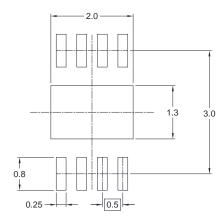




TOP VIEW







SIDE VIEW

RECOMMENDED LAND PATTERN (Unit: mm)

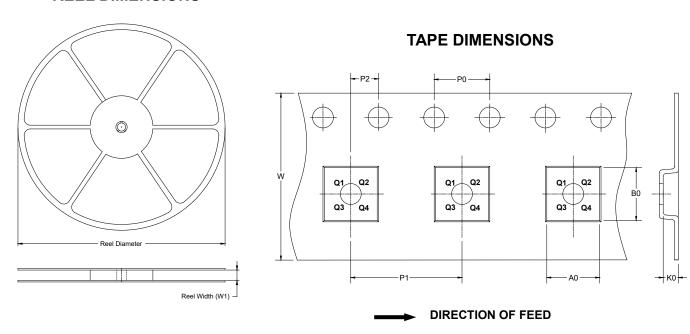
Symbol	_	nsions meters	Dimensions In Inches		
	MIN	MAX	MIN	MAX	
А	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A2	0.203	REF	0.008	REF	
D	2.900	3.100	0.114	0.122	
D1	1.900	2.100	0.075	0.083	
Е	2.900	3.100	0.114	0.122	
E1	1.200	1.400	0.047	0.055	
k	0.350) REF	0.014	REF	
b	0.200	0.300	0.008	0.012	
е	0.500) TYP	0.020	TYP	
L	0.400	0.600	0.016	0.024	

NOTE: This drawing is subject to change without notice.



TAPE AND REEL INFORMATION

REEL DIMENSIONS

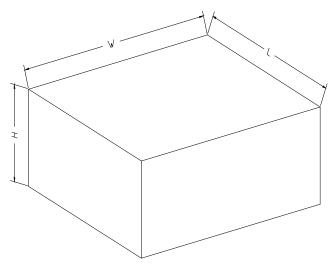


NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

Package Type	Reel Diameter	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P0 (mm)	P1 (mm)	P2 (mm)	W (mm)	Pin1 Quadrant
TDFN-3×3-8AL	13"	12.4	3.35	3.35	1.13	4.0	8.0	2.0	12.0	Q2

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

Reel Type	Length (mm)	Width (mm)	Height (mm)	Pizza/Carton
13"	386	280	370	5

单击下面可查看定价,库存,交付和生命周期等信息

>>SGMICRO(圣邦微电子)