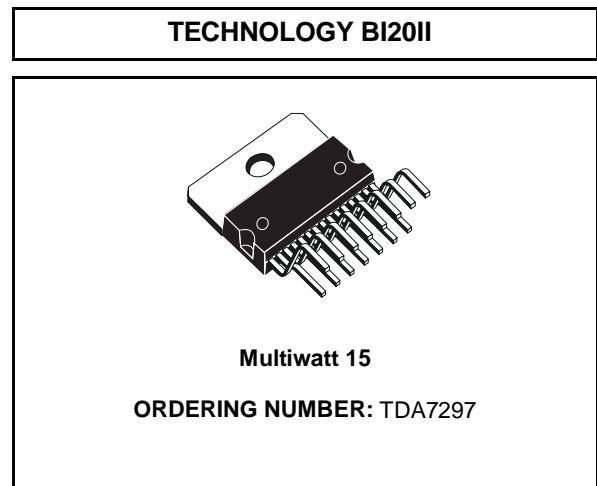




TDA7297

15+15W DUAL BRIDGE AMPLIFIER

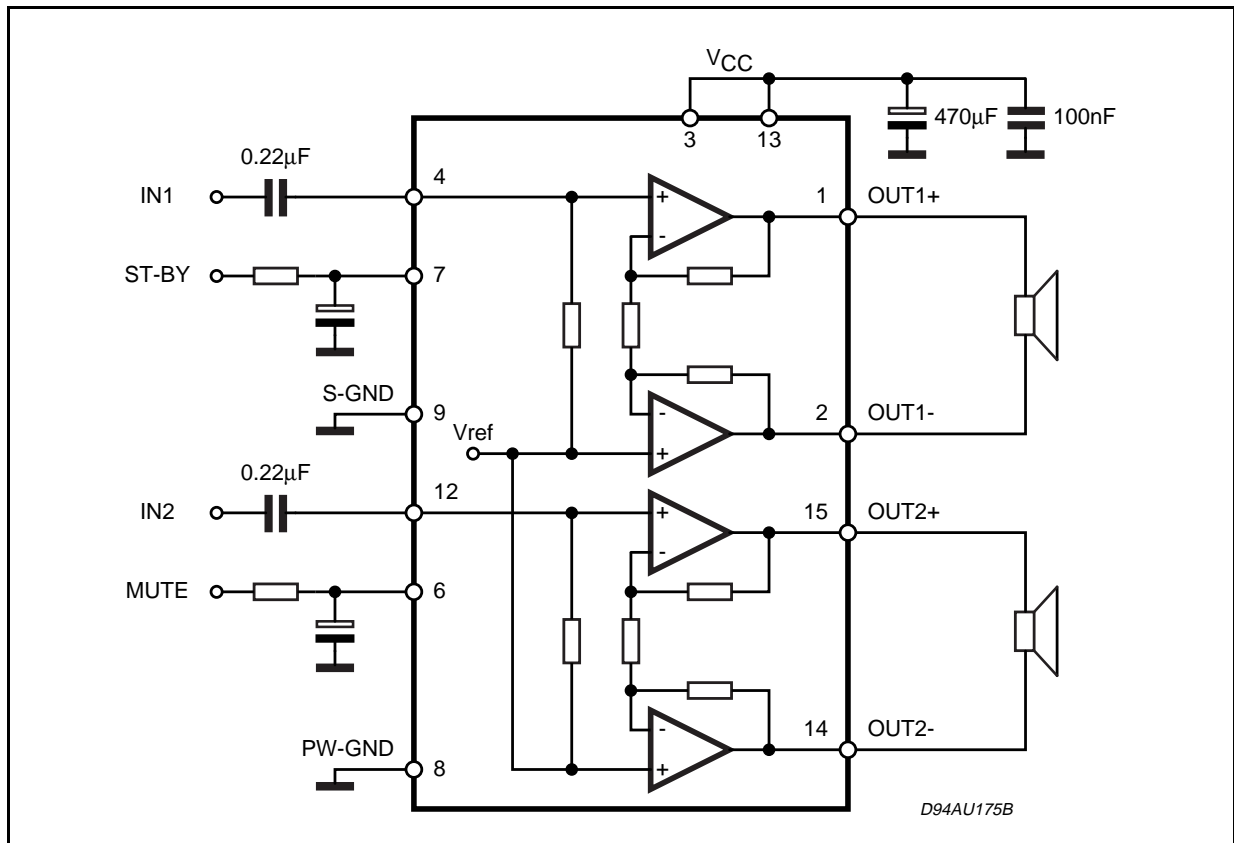
- WIDE SUPPLY VOLTAGE RANGE (6V -18V)
- MINIMUM EXTERNAL COMPONENTS
 - NO SVR CAPACITOR
 - NO BOOTSTRAP
 - NO BOUCHEROT CELLS
 - INTERNALLY FIXED GAIN
- STAND-BY & MUTE FUNCTIONS
- SHORT CIRCUIT PROTECTION
- THERMAL OVERLOAD PROTECTION



DESCRIPTION

The TDA7297 is a dual bridge amplifier specially designed for TV and Portable Radio applications.

BLOCK AND APPLICATION DIAGRAM



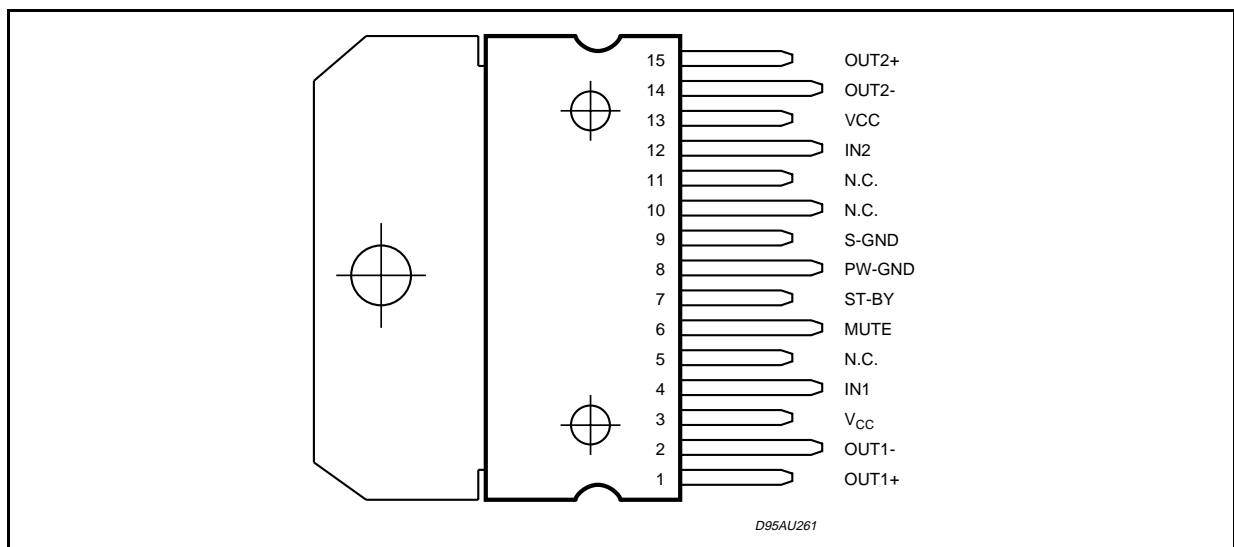
ABSOLUTE MAXIMUM RATINGS

| Symbol | Parameter | Value | Unit |
|----------------|--|-------------|-------------|
| V_S | Supply Voltage | 20 | V |
| I_o | Output Peak Current (internally limited) | 2 | A |
| P_{tot} | Total Power Dissipation ($T_{case} = 70^{\circ}C$) | 33 | W |
| T_{op} | Operating Temperature | 0 to 70 | $^{\circ}C$ |
| T_{stg}, T_j | Storage and Junction Temperature | -40 to +150 | $^{\circ}C$ |

THERMAL DATA

| Symbol | Description | Value | Unit |
|------------------|-------------------------------------|--------------------|---------------|
| $R_{th\ j-case}$ | Thermal Resistance Junction to case | Typ. 1.4 Max. 2 | $^{\circ}C/W$ |

PIN CONNECTION (Top view)



ELECTRICAL CHARACTERISTICS ($V_{CC} = 16.5V$, $R_L = 8\Omega$, $f = 1kHz$, $T_{amb} = 25^{\circ}C$ unless otherwise specified.)

| Symbol | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|--------------|---------------------------|--|------|------|------|-------------|
| V_{CC} | Supply Range | | 6.5 | | 18 | V |
| I_q | Total Quiescent Current | $R_L = \infty$ | | 50 | 65 | mA |
| V_{OS} | Output Offset Voltage | | | | 120 | mV |
| P_O | Output Power | THD = 10% | 13 | 15 | | W |
| THD | Total Harmonic Distortion | $P_O = 1W$ | | 0.1 | 0.3 | % |
| | | $P_O = 0.1W$ to $5W$ $f = 100Hz$ to $15kHz$ | | | 1 | % |
| SVR | Supply Voltage Rejection | $f = 100Hz$ $V_R = 0.5V$ | 40 | 56 | | dB |
| CT | Crosstalk | | 46 | 60 | | dB |
| A_{MUTE} | Mute Attenuation | | 60 | 80 | | dB |
| T_W | Thermal Threshold | | | 150 | | $^{\circ}C$ |
| G_V | Closed Loop Voltage Gain | | 31 | 32 | 33 | dB |
| ΔG_V | Voltage Gain Matching | | | | 0.5 | dB |
| R_i | Input Resistance | | 25 | 30 | | $K\Omega$ |

ELECTRICAL CHARACTERISTICS (Continued)

| Symbol | Parameter | Test Condition | Min. | Typ. | Max. | Unit |
|-----------------|----------------------------------|---|------|------------|------|--------------------------------|
| $V_{T_{MUTE}}$ | Mute Threshold | $V_O = -30\text{dB}$ | 2.3 | 2.9 | 4.1 | V |
| $V_{T_{ST-BY}}$ | St-by Threshold | | 0.8 | 1.3 | 1.8 | V |
| I_{ST-BY} | ST-BY current $V_6 = \text{GND}$ | | | | 100 | μA |
| e_N | Total Output Noise Voltage | A curve $f = 20\text{Hz to } 20\text{kHz}$ | | 150 220 | 500 | μV μV |

APPLICATION SUGGESTION

STAND-BY AND MUTE FUNCTIONS

(A) Microprocessor Application

In order to avoid annoying "Pop-Noise" during Turn-On/Off transients, it is necessary to guarantee the right St-by and mute signals sequence. It is quite simple to obtain this function using a microprocessor (Fig. 1 and 2).

At first St-by signal (from μP) goes high and the voltage across the St-by terminal (Pin 7) starts to increase exponentially. The external RC network is intended to turn-on slowly the biasing circuits of

the amplifier, this to avoid "POP" and "CLICK" on the outputs.

When this voltage reaches the St-by threshold level, the amplifier is switched-on and the external capacitors in series to the input terminals (C3, C5) start to charge.

It's necessary to maintain the mute signal low until the capacitors are fully charged, this to avoid that the device goes in play mode causing a loud "Pop Noise" on the speakers.

A delay of 100-200ms between St-by and mute signals is suitable for a proper operation.

Figure 1: Microprocessor Application

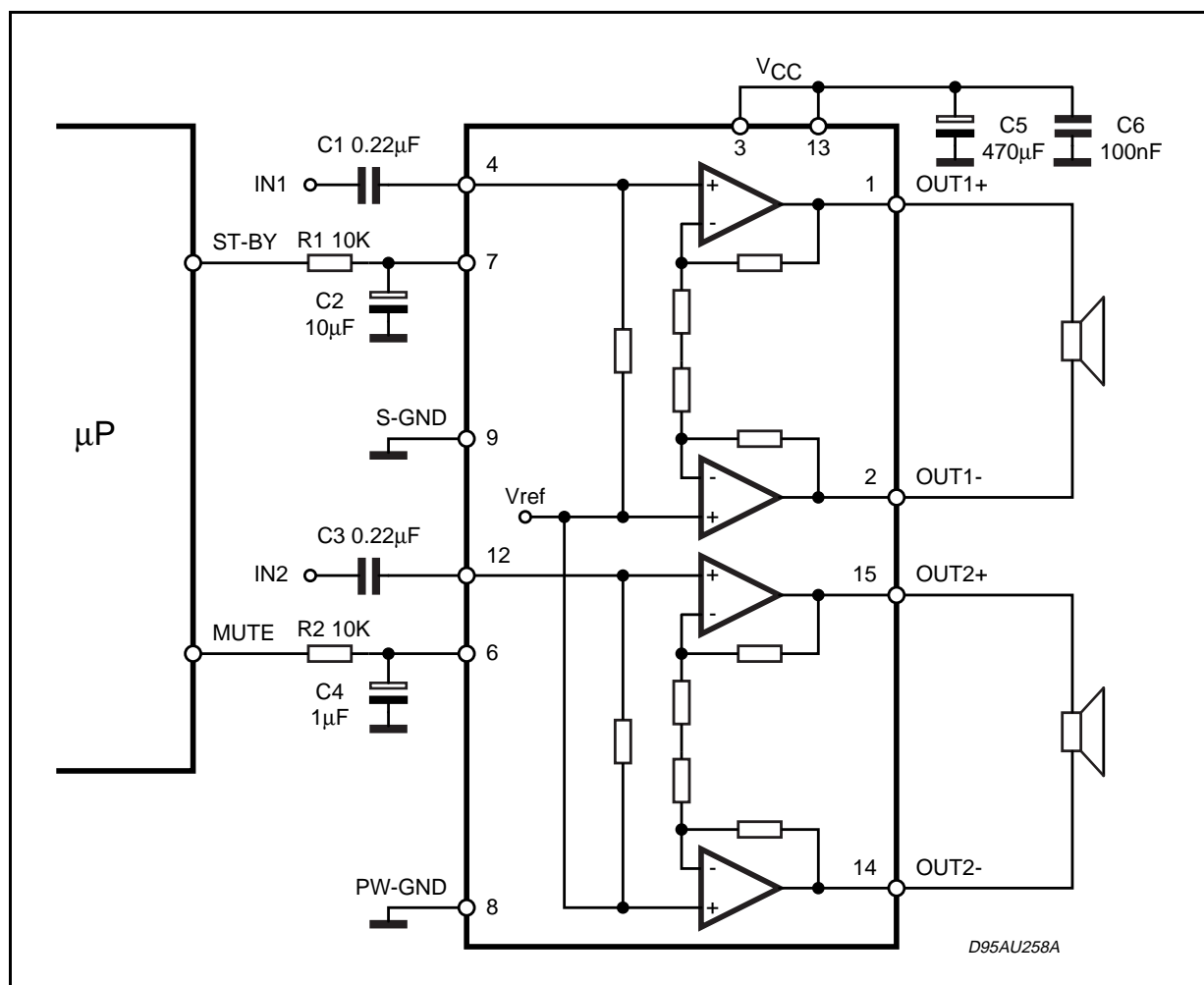
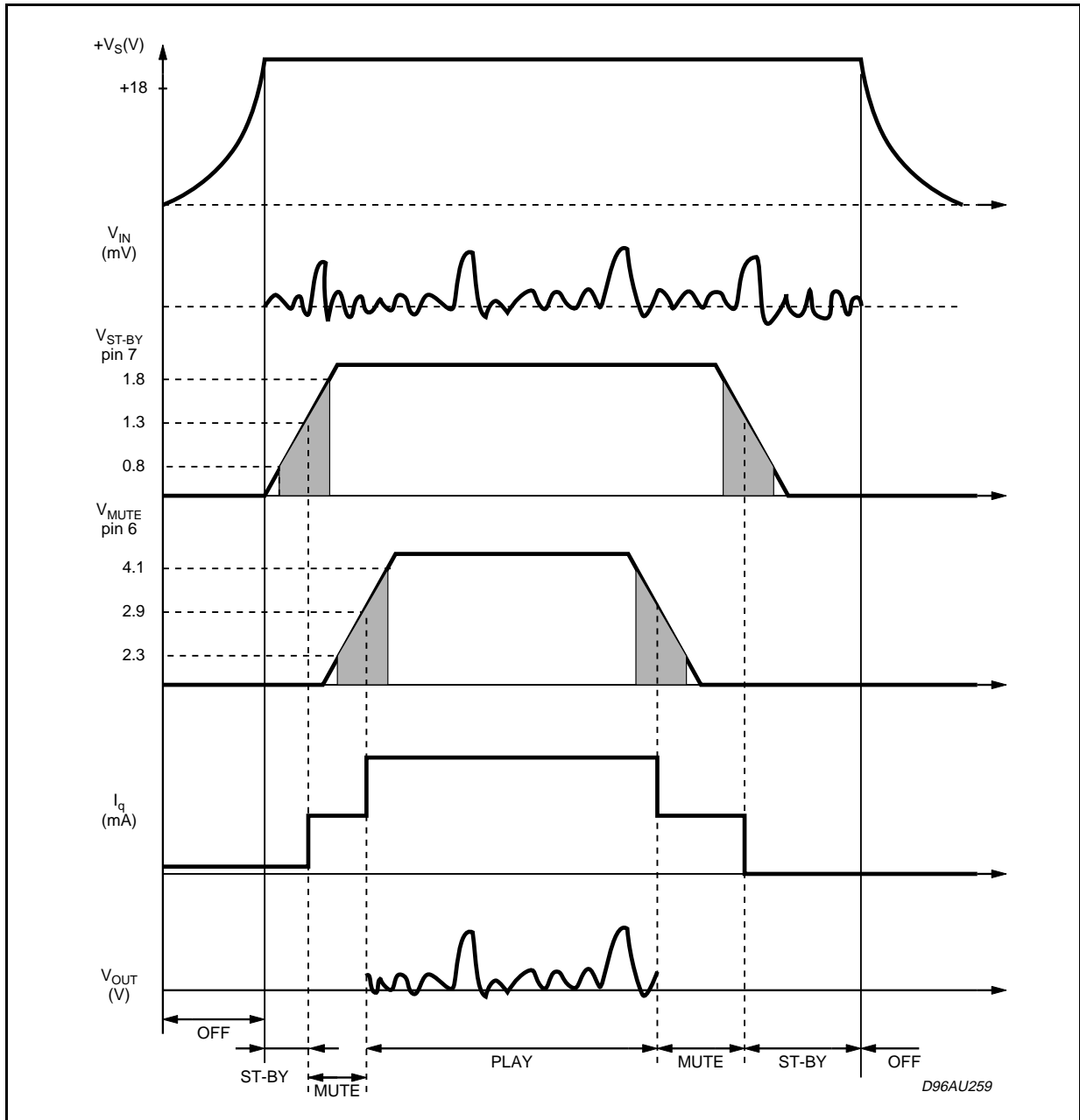


Figure 2: Microprocessor Driving Signals.



(B) Low Cost Application

In low cost applications where the mP is not present, the suggested circuit is shown in fig.3.

The St-by and mute terminals are tied together and they are connected to the supply line via an

external voltage divider.

The device is switched-on/off from the supply line and the external capacitor C4 is intended to delay the St-by and mute threshold exceeding, avoiding "Popping" problems.

Figure 3: Stand-alone Low-cost Application.

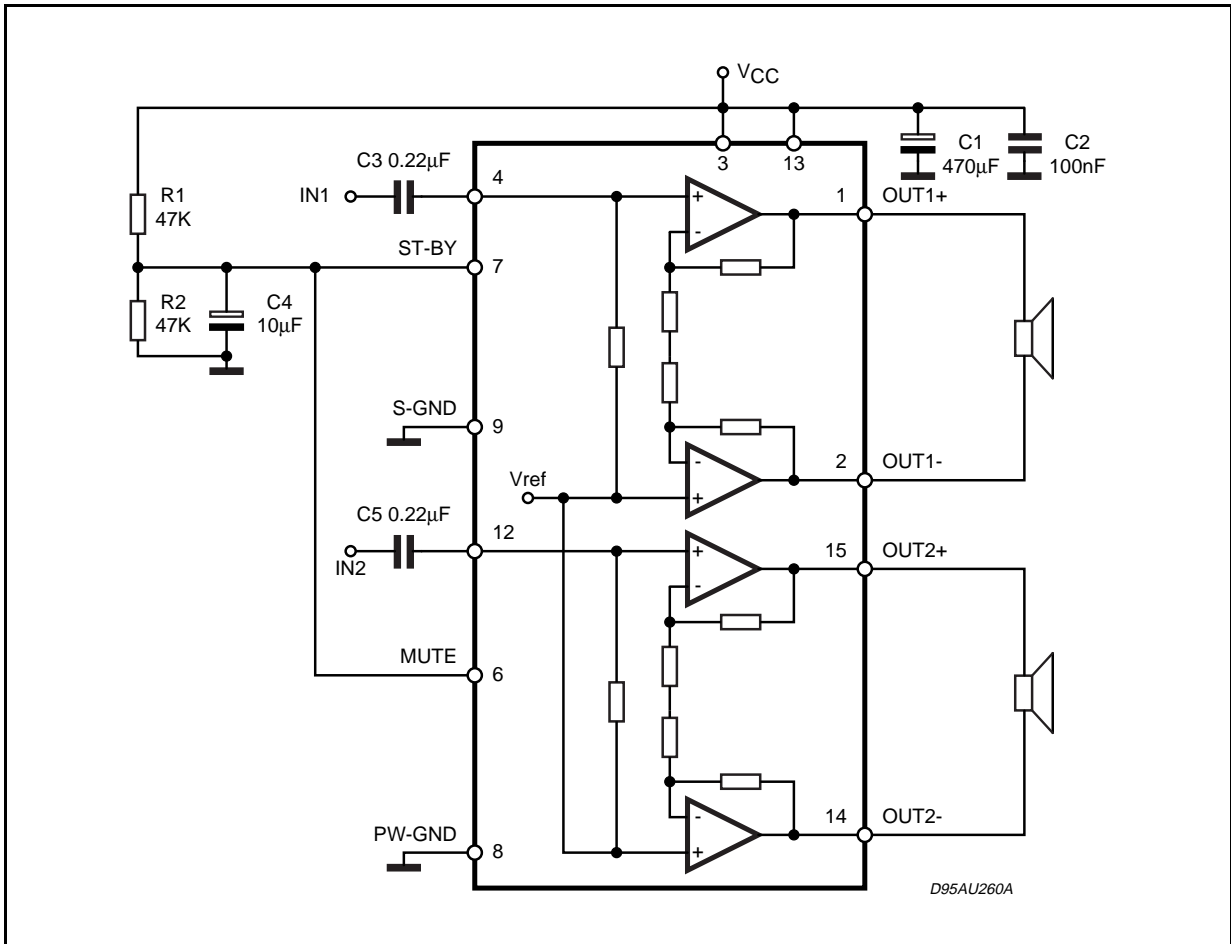


Figure 3b: PCB and Component Layout of the Application Circuit (Fig. 1).

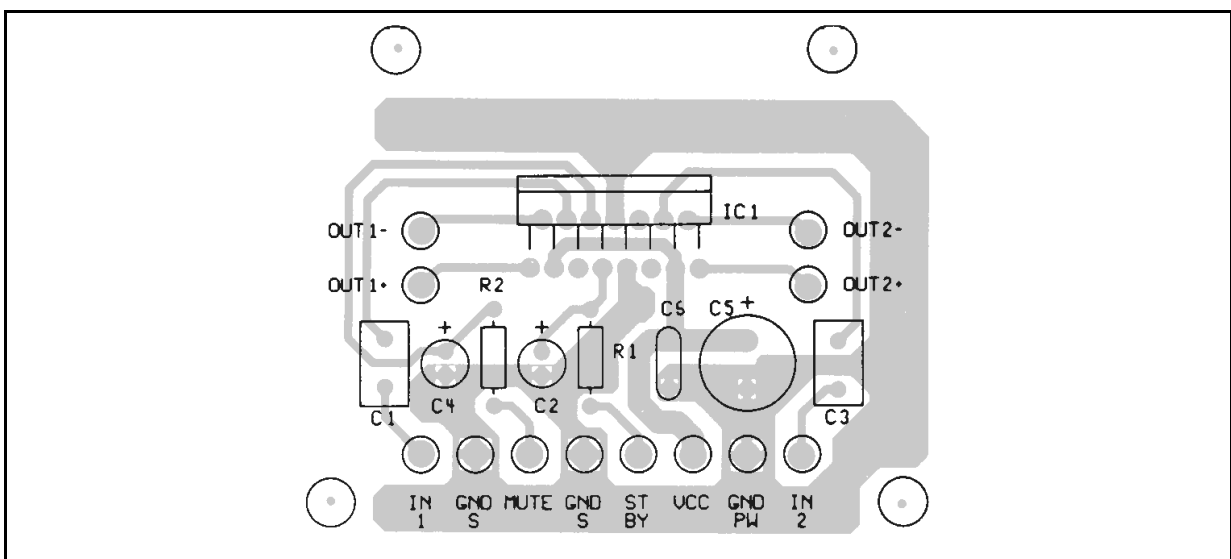


Figure 4: Distortion vs Output Power

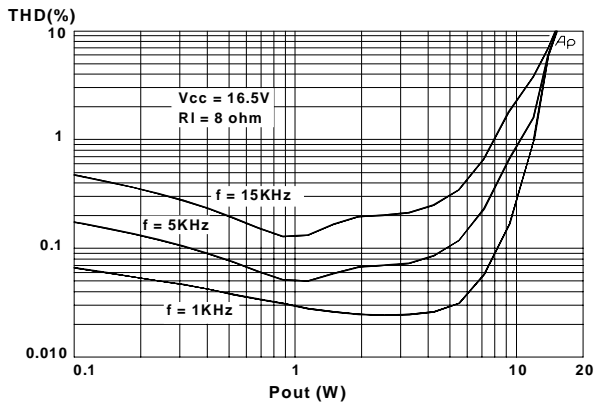


Figure 5: Distortion vs Output Power

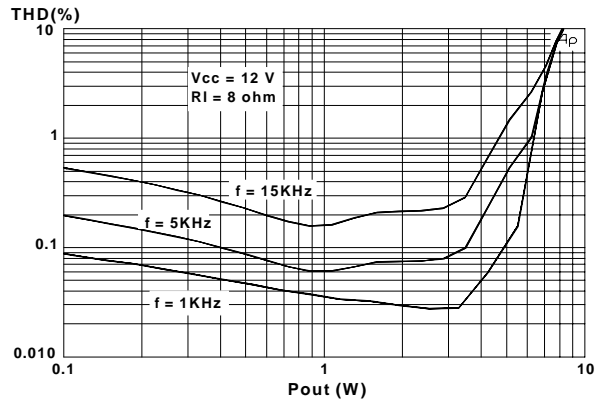


Figure 6: Distortion vs Frequency

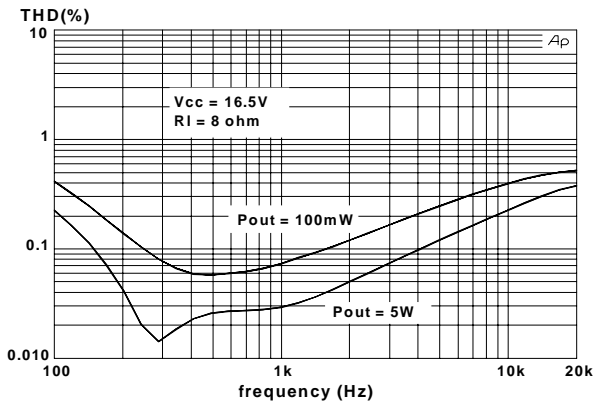


Figure 7: Frequency Response

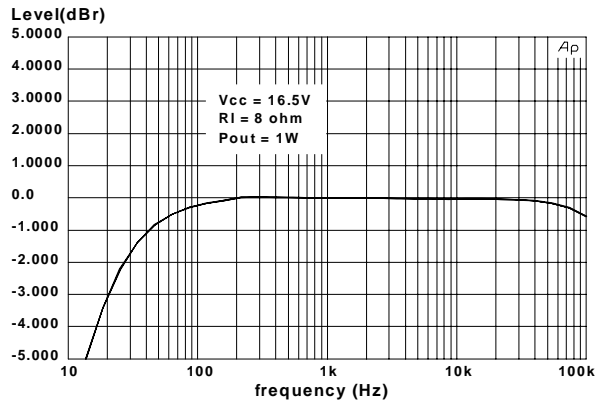


Figure 8: Output Power vs Supply Voltage

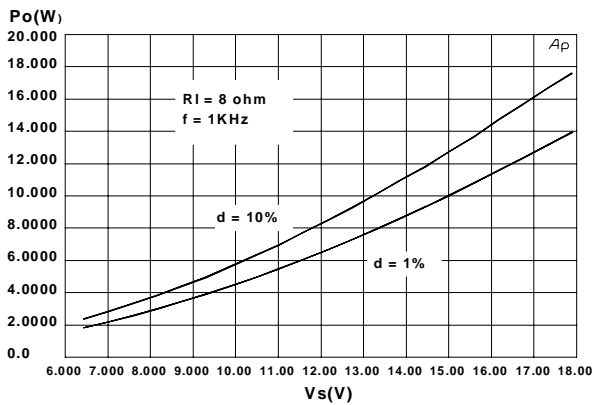


Figure 9: Total Power Dissipation & Efficiency vs Output Power

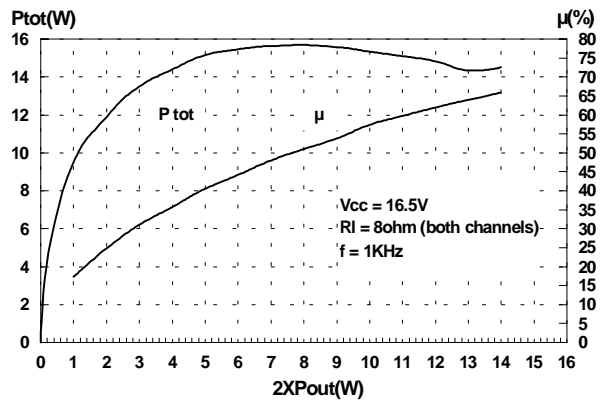


Figure 10: Mute Attenuation vs. V pin.6

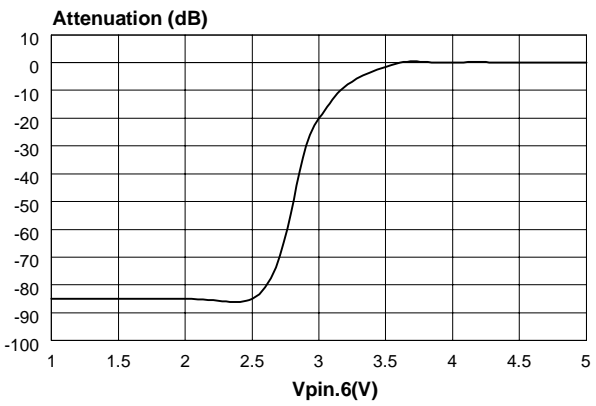


Figure 11: Stand-By Attenuation vs Vpin.7

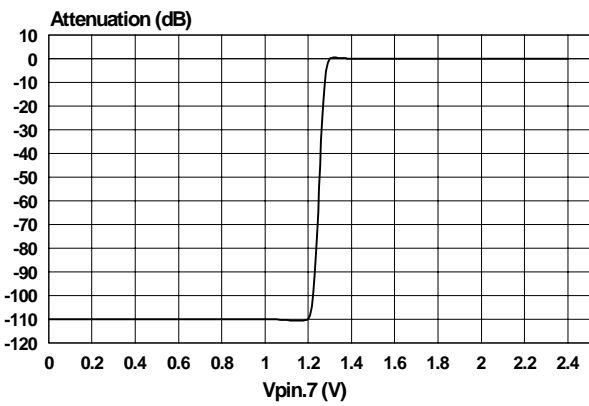
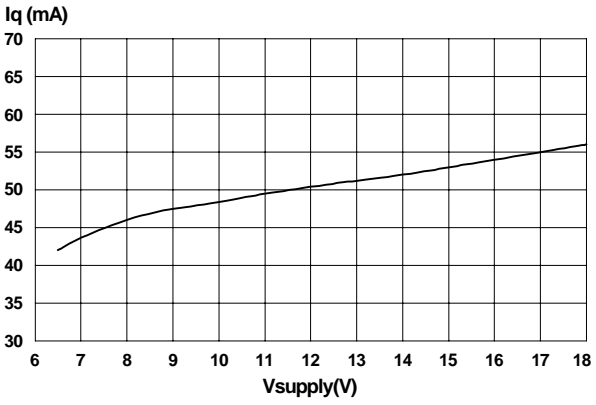
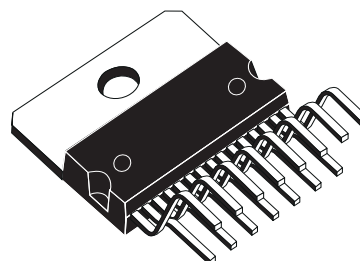


Figure 12: Quiscent Current vs. Supply Voltage

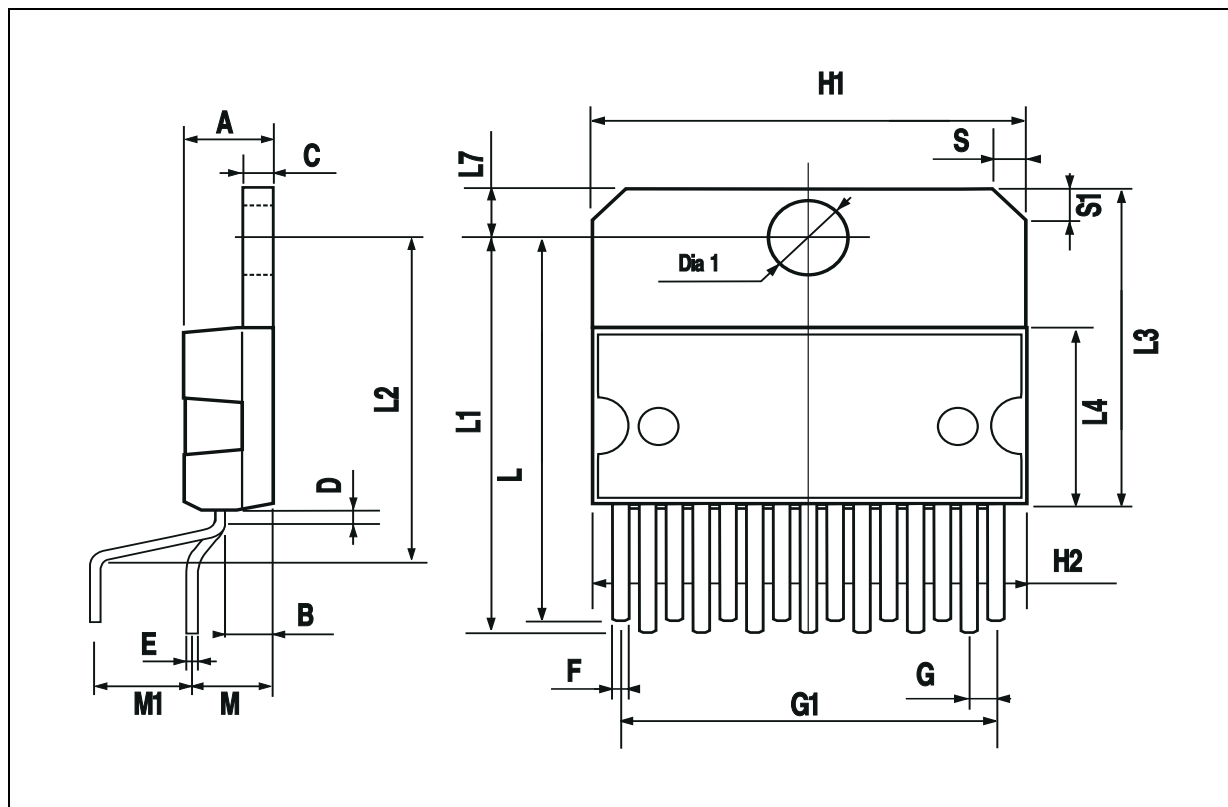


| DIM. | mm | | | inch | | |
|------|-------|-------|-------|-------|-------|-------|
| | MIN. | TYP. | MAX. | MIN. | TYP. | MAX. |
| A | | | 5 | | | 0.197 |
| B | | | 2.65 | | | 0.104 |
| C | | | 1.6 | | | 0.063 |
| D | | 1 | | | 0.039 | |
| E | 0.49 | | 0.55 | 0.019 | | 0.022 |
| F | 0.66 | | 0.75 | 0.026 | | 0.030 |
| G | 1.02 | 1.27 | 1.52 | 0.040 | 0.050 | 0.060 |
| G1 | 17.53 | 17.78 | 18.03 | 0.690 | 0.700 | 0.710 |
| H1 | 19.6 | | | 0.772 | | |
| H2 | | | 20.2 | | | 0.795 |
| L | 21.9 | 22.2 | 22.5 | 0.862 | 0.874 | 0.886 |
| L1 | 21.7 | 22.1 | 22.5 | 0.854 | 0.870 | 0.886 |
| L2 | 17.65 | | 18.1 | 0.695 | | 0.713 |
| L3 | 17.25 | 17.5 | 17.75 | 0.679 | 0.689 | 0.699 |
| L4 | 10.3 | 10.7 | 10.9 | 0.406 | 0.421 | 0.429 |
| L7 | 2.65 | | 2.9 | 0.104 | | 0.114 |
| M | 4.25 | 4.55 | 4.85 | 0.167 | 0.179 | 0.191 |
| M1 | 4.63 | 5.08 | 5.53 | 0.182 | 0.200 | 0.218 |
| S | 1.9 | | 2.6 | 0.075 | | 0.102 |
| S1 | 1.9 | | 2.6 | 0.075 | | 0.102 |
| Dia1 | 3.65 | | 3.85 | 0.144 | | 0.152 |

OUTLINE AND MECHANICAL DATA



Multiwatt15 V



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