

# OPAx377-Q1 Low-Noise, Low Quiescent Current, Precision Automotive Grade Operational Amplifier

## 1 Features

- Qualified for Automotive Applications
- AEC-Q100 Qualified with the Following Results:
  - Device Temperature Grade 1: –40°C to +125°C Ambient Operating Temperature Range
  - Device HBM ESD Classification Level 3A
  - Device CDM ESD Classification Level C6
- Low Noise: 7.5 nV/ $\sqrt{\text{Hz}}$  at 1 kHz
- 0.1-Hz to 10-Hz Noise: 0.8  $\mu\text{V}_{\text{PP}}$
- Quiescent Current: 760  $\mu\text{A}$  (typical)
- Low Offset Voltage: 250  $\mu\text{V}$  (typical)
- Gain Bandwidth Product: 5.5 MHz
- Rail-to-Rail Input and Output
- Single-Supply Operation
- Supply Voltage: 2.2 V to 5.5 V
- Space-Saving Packages:
  - SOT-23, VSSOP, TSSOP

## 2 Applications

- Active Cruise Control
- Park Assist
- Tire Pressure Monitoring
- Infotainment
- Active Filtering
- Sensor Signal Conditioning

## 3 Description

The OPAx377-Q1 family of operational amplifiers are wide-bandwidth CMOS amplifiers that provide very low noise, low input bias current, and low offset voltage while operating on a low quiescent current of 0.76 mA (typical).

The OPAx377-Q1 op amps are optimized for low-voltage, single-supply applications. The exceptional combination of ac and dc performance make them ideal for a wide range of applications, including small signal conditioning, audio, and active filters. In addition, these parts have a wide supply range with excellent PSRR, making them attractive for applications that run directly from batteries without regulation.

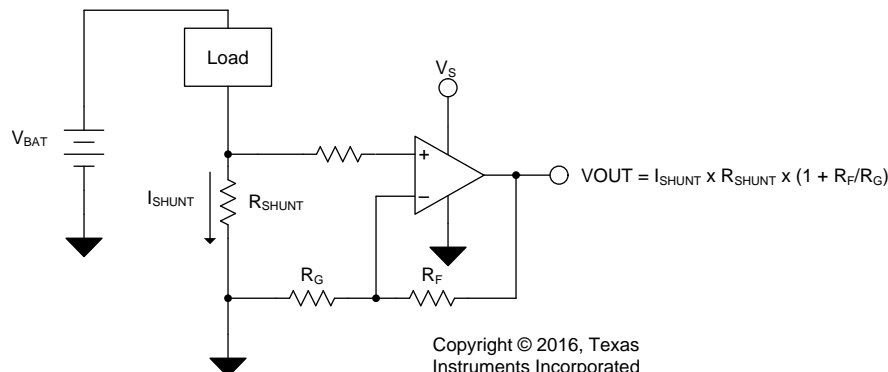
The OPA377-Q1 is available in the SOT23-5 package. The dual, OPA2377-Q1, is offered in the MSOP-8 package and the quad OPA4377-Q1 is offered in the TSSOP-14 package. All versions are specified for operation from –40°C to +125°C.

### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
OPA377-Q1	SOT-23 (5)	2.90 mm x 1.60 mm
OPA2377-Q1	VSSOP (8)	3.00 mm x 3.00 mm
OPA4377-Q1	TSSOP (14)	5.00 mm x 4.40 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

### Low-Side Current Sense Amplifier



## Table of Contents

<b>1 Features</b> .....	<b>1</b>	7.3 Feature Description.....	<b>14</b>
<b>2 Applications</b> .....	<b>1</b>	7.4 Device Functional Modes.....	<b>15</b>
<b>3 Description</b> .....	<b>1</b>	<b>8 Application and Implementation</b> .....	<b>16</b>
<b>4 Revision History</b> .....	<b>2</b>	8.1 Application Information.....	<b>16</b>
<b>5 Pin Configuration and Functions</b> .....	<b>3</b>	8.2 Typical Application .....	<b>16</b>
<b>6 Specifications</b> .....	<b>6</b>	<b>9 Power Supply Recommendations</b> .....	<b>18</b>
6.1 Absolute Maximum Ratings .....	<b>6</b>	<b>10 Layout</b> .....	<b>18</b>
6.2 ESD Ratings .....	<b>6</b>	10.1 Layout Guidelines .....	<b>18</b>
6.3 Recommended Operating Conditions.....	<b>6</b>	10.2 Layout Example .....	<b>18</b>
6.4 Thermal Information: OPA377-Q1 .....	<b>6</b>	<b>11 Device and Documentation Support</b> .....	<b>20</b>
6.5 Thermal Information: OPA2377-Q1 .....	<b>6</b>	11.1 Device Support .....	<b>20</b>
6.6 Thermal Information: OPA4377-Q1 .....	<b>7</b>	11.2 Documentation Support .....	<b>21</b>
6.7 Electrical Characteristics: $V_S = 2.2\text{ V to }5.5\text{ V}$ .....	<b>7</b>	11.3 Community Resources.....	<b>21</b>
6.8 Typical Characteristics .....	<b>9</b>	11.4 Trademarks .....	<b>21</b>
<b>7 Detailed Description</b> .....	<b>13</b>	11.5 Electrostatic Discharge Caution.....	<b>21</b>
7.1 Overview .....	<b>13</b>	11.6 Glossary .....	<b>21</b>
7.2 Functional Block Diagram .....	<b>13</b>	<b>12 Mechanical, Packaging, and Orderable Information</b> .....	<b>21</b>

## 4 Revision History

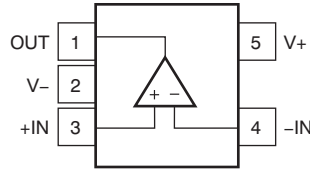
### Changes from Original (May 2016) to Revision A

Page

• Changed device status from <i>Product Preview</i> to <i>Production Data</i> .....	<b>1</b>
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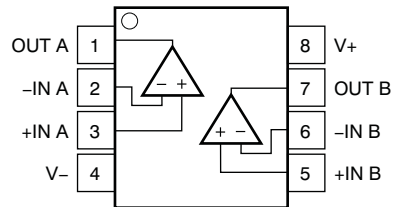
## 5 Pin Configuration and Functions

**OPA377-Q1: DBV Package  
5-Pin SOT23  
Top View**



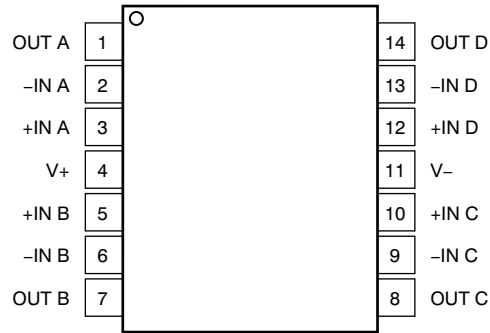
**Pin Functions: OPA377-Q1**

PIN		I/O	DESCRIPTION
NAME	NO. DBV		
+IN	3	I	Noninverting input
-IN	4	I	Inverting input
NC	—	—	No internal connection (can be left floating)
OUT	1	O	Output
V-	2	—	Negative (lowest) power supply
V+	5	—	Positive (highest) power supply

**OPA2377-Q1: DGK Package  
8-Pin VSSOP and SOIC  
Top View**

**Pin Functions: OPA2377-Q1**

PIN		I/O	DESCRIPTION
NAME	NO. DGK		
-IN A	2	I	Inverting input, channel A
-IN B	6	I	Inverting input, channel B
+IN A	3	I	Noninverting input, channel A
+IN B	5	I	Noninverting input, channel B
OUT A	1	O	Output, channel A
OUT B	7	O	Output, channel B
V-	4	—	Negative (lowest) power supply
V+	8	—	Positive (highest) power supply

**OPA4377-Q1: PW Package  
14-Pin TSSOP  
Top View**



**Pin Functions: OPA4377-Q1**

PIN		I/O	DESCRIPTION
NAME	NO. PW		
-IN A	2	I	Inverting input, channel A
-IN B	6	I	Inverting input, channel B
-IN C	9	I	Inverting input, channel C
-IN D	13	I	Inverting input, channel D
+IN A	3	I	Noninverting input, channel A
+IN B	5	I	Noninverting input, channel B
+IN C	10	I	Noninverting input, channel C
+IN D	12	I	Noninverting input, channel D
OUT A	1	O	Output, channel A
OUT B	7	O	Output, channel B
OUT C	8	O	Output, channel C
OUT D	14	O	Output, channel D
V-	11	—	Negative (lowest) power supply
V+	4	—	Positive (highest) power supply

## 6 Specifications

### 6.1 Absolute Maximum Ratings

 over operating free-air temperature range (unless otherwise noted)<sup>(1)</sup>

		MIN	MAX	UNIT
$V_S = (V+) - (V-)$	Supply voltage		7	V
	Signal input terminal voltage <sup>(2)</sup>	(V-) – 0.5	(V+) + 0.5	V
	Signal input terminal current <sup>(2)</sup>	–10	10	mA
	Output short-circuit current <sup>(3)</sup>	Continuous		
$T_A$	Operating temperature	–40	150	°C
$T_J$	Junction temperature		150	°C
$T_{stg}$	Storage temperature	–65	150	°C

- Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5 V beyond the supply rails should be current limited to 10 mA or less.
- Short-circuit to ground, one amplifier per package.

### 6.2 ESD Ratings

			VALUE	UNIT
$V_{(ESD)}$	Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 <sup>(1)</sup>	±4000	V
		Charged-device model (CDM), per AEC Q100-011	±1000	

- AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

### 6.3 Recommended Operating Conditions

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT
	Supply voltage	2.2	5.5	V
$T_A$	Operating temperature	–40	150	°C

### 6.4 Thermal Information: OPA377-Q1

THERMAL METRIC <sup>(1)</sup>		OPA377-Q1	UNIT
		DBV (SOT23)	
		5 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	273.8	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	126.8	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	85.9	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	10.9	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	84.9	°C/W
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	n/a	°C/W

- For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

### 6.5 Thermal Information: OPA2377-Q1

THERMAL METRIC <sup>(1)</sup>		OPA2377-Q1	UNIT
		DGK (VSSOP)	
		8 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	171.2	°C/W
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	63.9	°C/W

- For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

**Thermal Information: OPA2377-Q1 (continued)**

THERMAL METRIC <sup>(1)</sup>		OPA2377-Q1	UNIT
		DGK (VSSOP)	
		8 PINS	
R <sub>θJB</sub>	Junction-to-board thermal resistance	92.8	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	9.2	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	91.2	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	°C/W

**6.6 Thermal Information: OPA4377-Q1**

THERMAL METRIC <sup>(1)</sup>		OPA4377-Q1	UNIT
		PW (TSSOP)	
		14 PINS	
R <sub>θJA</sub>	Junction-to-ambient thermal resistance	107.8	°C/W
R <sub>θJC(top)</sub>	Junction-to-case (top) thermal resistance	29.6	°C/W
R <sub>θJB</sub>	Junction-to-board thermal resistance	52.6	°C/W
Ψ <sub>JT</sub>	Junction-to-top characterization parameter	1.5	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	51.6	°C/W
R <sub>θJC(bot)</sub>	Junction-to-case (bottom) thermal resistance	n/a	°C/W

(1) For more information about traditional and new thermal metrics, see the *Semiconductor and IC Package Thermal Metrics* application report, [SPRA953](#).

**6.7 Electrical Characteristics: V<sub>S</sub> = 2.2 V to 5.5 V**

At T<sub>A</sub> = 25°C, R<sub>L</sub> = 10 kΩ connected to V<sub>S</sub>/2, V<sub>CM</sub> = V<sub>S</sub>/2, and V<sub>OUT</sub> = V<sub>S</sub>/2, unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
<b>OFFSET VOLTAGE</b>						
V <sub>OS</sub>	Input offset voltage	V <sub>S</sub> = 5 V		0.25	1	mV
	Input offset voltage versus temperature	At T <sub>A</sub> = –40°C to +125°C, V <sub>S</sub> = 2.2 V to 5.5 V, V <sub>CM</sub> < (V+) – 1.3 V		5		μV/V
dV <sub>OS</sub> /dT	Input offset voltage versus drift	At T <sub>A</sub> = –40°C to +125°C		0.32	2	μV/°C
PSRR	Input offset voltage versus power supply	At T <sub>A</sub> = 25°C, V <sub>S</sub> = 2.2 V to 5.5 V, V <sub>CM</sub> < (V+) – 1.3 V		5	28	μV/V
	Channel separation, dc (dual, quad)			0.5		μV/V
<b>INPUT BIAS CURRENT</b>						
I <sub>IB</sub>	Input bias current			±0.2	±10	pA
	Input bias current versus temperature			See <a href="#">Typical Characteristics</a>		pA
I <sub>OS</sub>	Input offset current			±0.2	±10	pA
<b>NOISE</b>						
	Input voltage noise	f = 0.1 Hz to 10 Hz		0.8		μV <sub>PP</sub>
e <sub>n</sub>	Input voltage noise density	f = 1 kHz		7.5		nV/√Hz
i <sub>n</sub>	Input current noise density	f = 1 kHz		2		fA/√Hz
<b>INPUT VOLTAGE RANGE</b>						
V <sub>CM</sub>	Common-mode voltage range		(V-) – 0.1		(V+) + 0.1	V
CMRR	Common-mode rejection ratio	(V-) < V <sub>CM</sub> < (V+) – 1.3 V	70	90		dB
<b>INPUT CAPACITANCE</b>						
	Differential			6.5		pF
	Common-mode			13		pF
<b>OPEN-LOOP GAIN</b>						
A <sub>OL</sub>	Open-loop voltage gain	50 mV < V <sub>O</sub> < (V+) – 50 mV, R <sub>L</sub> = 10 kΩ	112	134		dB
		100 mV < V <sub>O</sub> < (V+) – 100 mV, R <sub>L</sub> = 2 kΩ		126		dB
<b>FREQUENCY RESPONSE, V<sub>S</sub> = 5.5 V</b>						

**Electrical Characteristics:  $V_S = 2.2\text{ V to }5.5\text{ V}$  (continued)**

 At  $T_A = 25^\circ\text{C}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_{S/2}$ ,  $V_{CM} = V_{S/2}$ , and  $V_{OUT} = V_{S/2}$ , unless otherwise noted.

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
GBW	Gain-bandwidth product			5.5		MHz
SR	Slew rate	$G = +1$		2		V/ $\mu\text{s}$
$t_s$	Settling time	At 0.1%, 2-V step, $G = +1$		1.6		$\mu\text{s}$
		At 0.01%, 2-V step, $G = +1$		2		$\mu\text{s}$
	Overload recovery time	$V_{IN} \times \text{Gain} > V_S$		0.33		$\mu\text{s}$
THD+N	Total harmonic distortion + noise	$V_O = 1\text{ V}_{RMS}$ , $G = +1$ , $f = 1\text{ kHz}$ , $R_L = 10\text{ k}\Omega$		0.00027%		
<b>OUTPUT</b>						
	Voltage output swing from rail	At $T_A = 25^\circ\text{C}$ , $R_L = 10\text{ k}\Omega$		10	20	mV
		At $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$ , $R_L = 10\text{ k}\Omega$			40	mV
$I_{SC}$	Short-circuit current			+30/-50		mA
$C_{LOAD}$	Capacitive load drive			See <a href="#">Typical Characteristics</a>		
$R_O$	Open-loop output impedance			150		$\Omega$
<b>POWER SUPPLY</b>						
$V_S$	Specified voltage		2.2		5.5	V
$I_Q$	Quiescent current (per amplifier)	At $T_A = 25^\circ\text{C}$ , $I_O = 0$ , $V_S = 5.5\text{ V}$		0.76	1.05	mA
		At $T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$			1.2	mA
<b>TEMPERATURE</b>						
	Specified temperature		-40		+125	$^\circ\text{C}$



### 6.8 Typical Characteristics

At  $T_A = 25^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

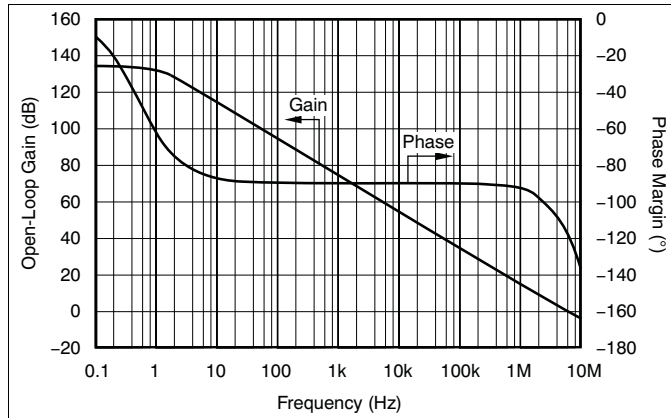


Figure 1. Open-Loop Gain and Phase vs Frequency

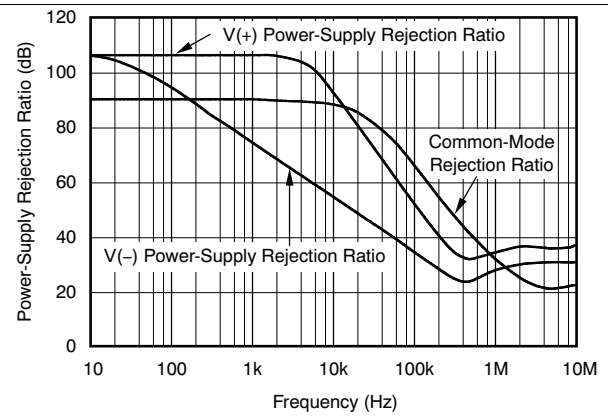


Figure 2. Power-Supply and Common-Mode Rejection Ratio vs Frequency

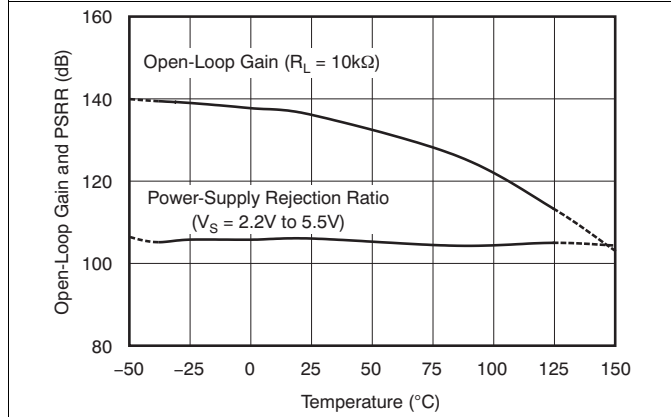


Figure 3. Open-Loop Gain and Power-Supply Rejection Ratio vs Temperature

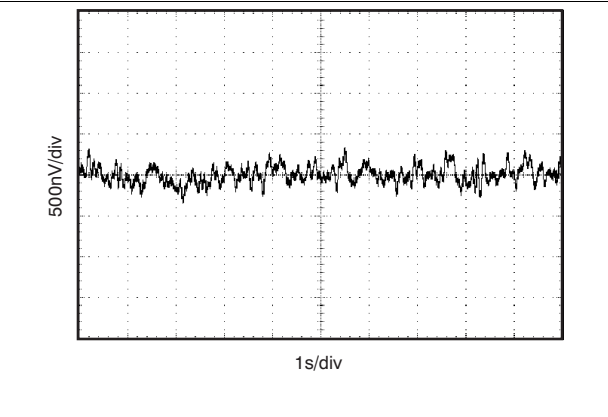


Figure 4. 0.1-Hz to 10-Hz Input Voltage Noise

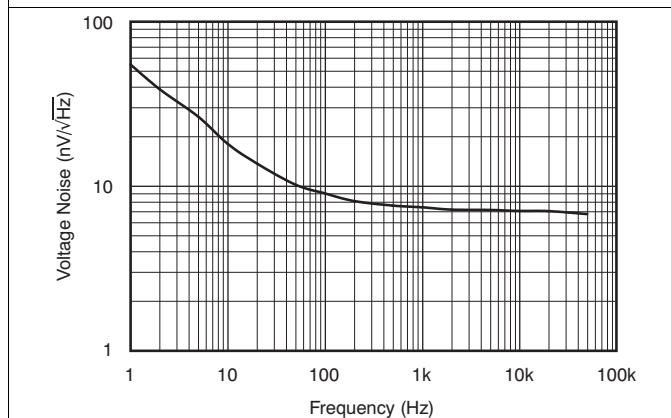


Figure 5. Input Voltage Noise Spectral Density

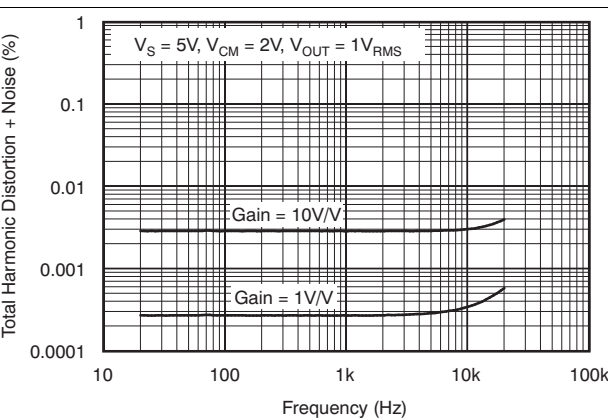


Figure 6. Total Harmonic Distortion and Noise vs Frequency

### Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

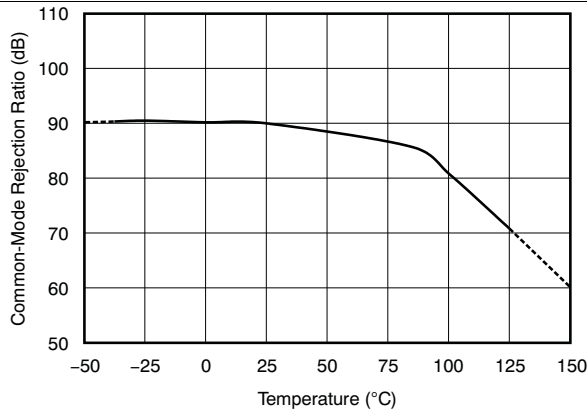


Figure 7. Common-Mode Rejection Ratio vs Temperature

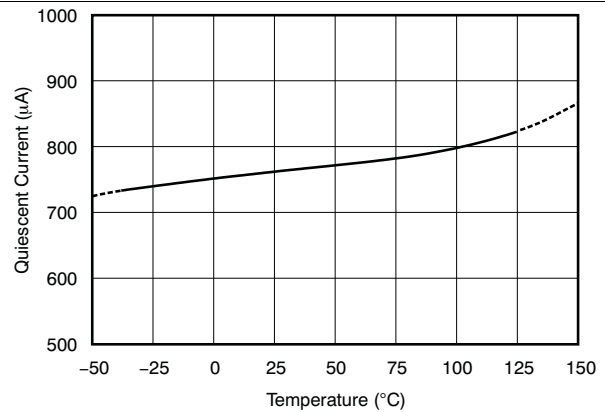


Figure 8. Quiescent Current vs Temperature

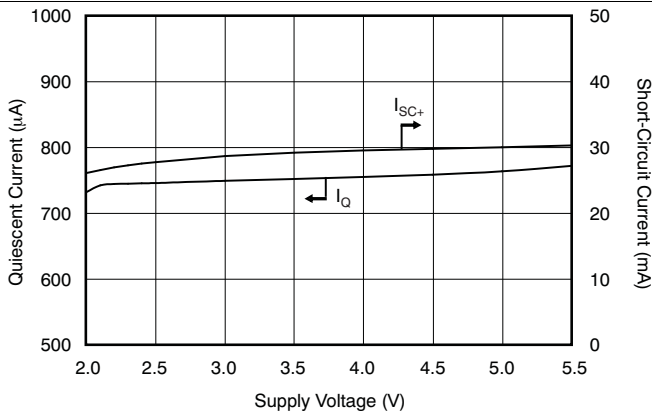


Figure 9. Quiescent and Short-Circuit Current vs Supply Voltage

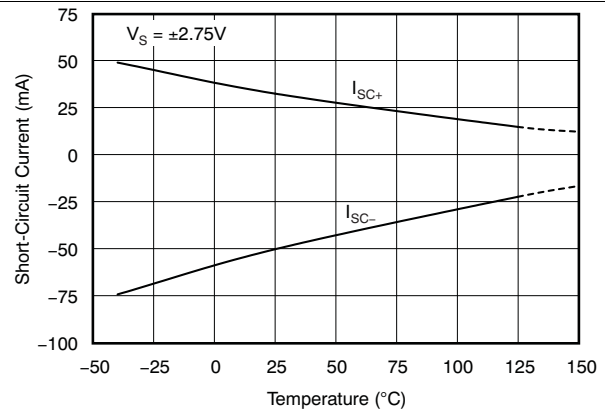


Figure 10. Short-Circuit Current vs Temperature

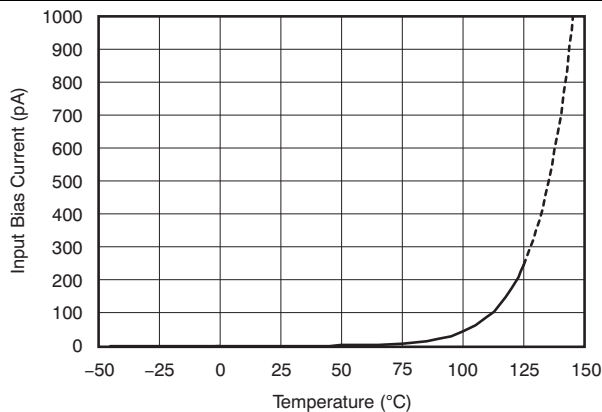


Figure 11. Input Bias Current vs Temperature

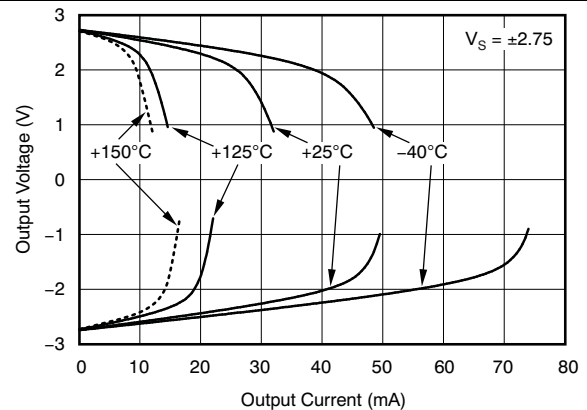


Figure 12. Output Voltage vs Output Current

Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

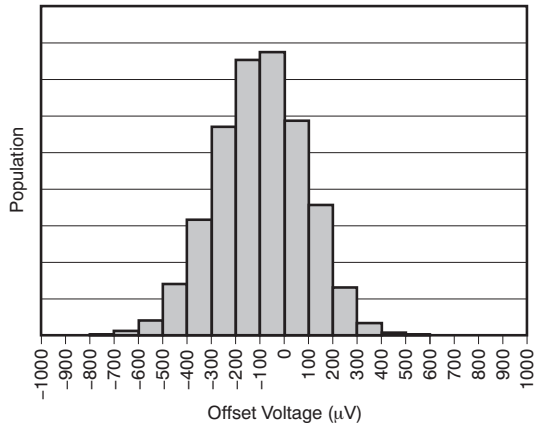


Figure 13. Offset Voltage Production Distribution

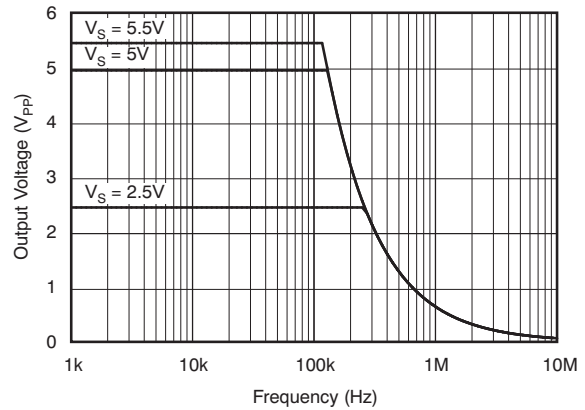


Figure 14. Maximum Output Voltage vs Frequency

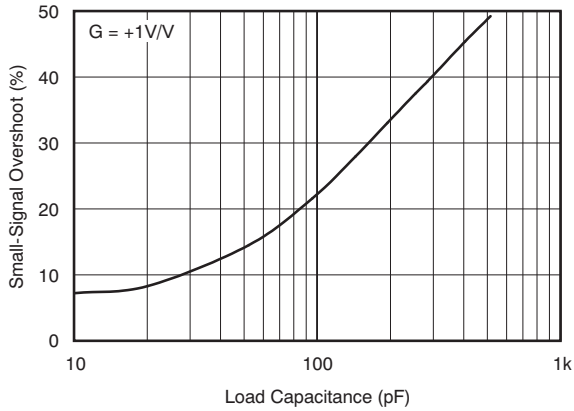


Figure 15. Small-Signal Overshoot vs Load Capacitance

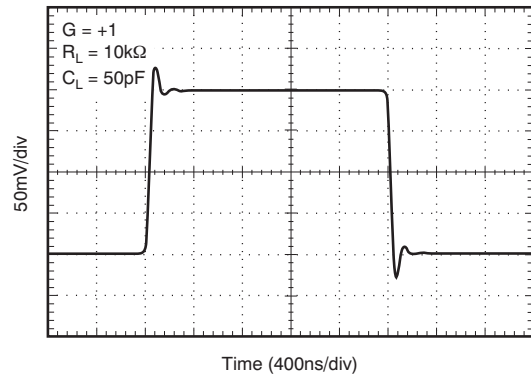


Figure 16. Small-Signal Pulse Response

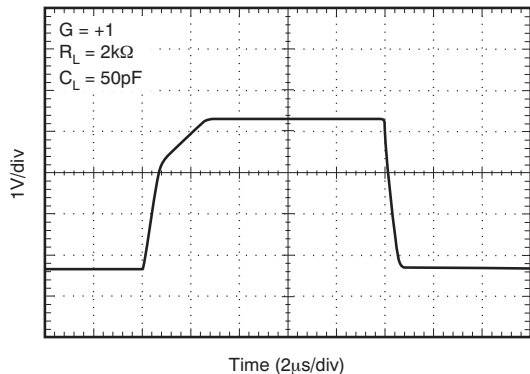


Figure 17. Large-Signal Pulse Response

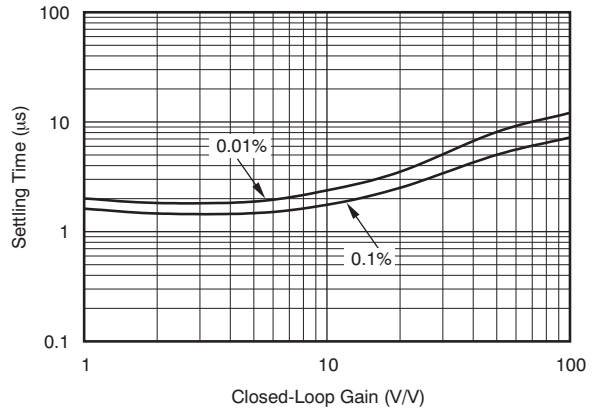


Figure 18. Settling Time vs Closed-Loop Gain

Typical Characteristics (continued)

At  $T_A = 25^\circ\text{C}$ ,  $V_S = 5\text{ V}$ ,  $R_L = 10\text{ k}\Omega$  connected to  $V_S/2$ ,  $V_{CM} = V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.

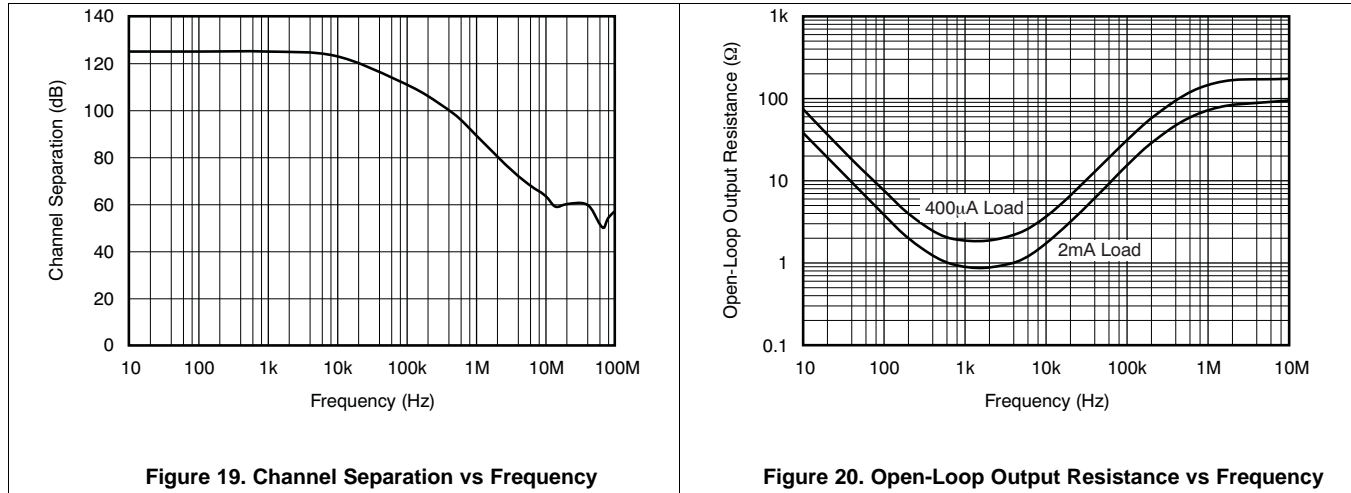


Figure 19. Channel Separation vs Frequency

Figure 20. Open-Loop Output Resistance vs Frequency

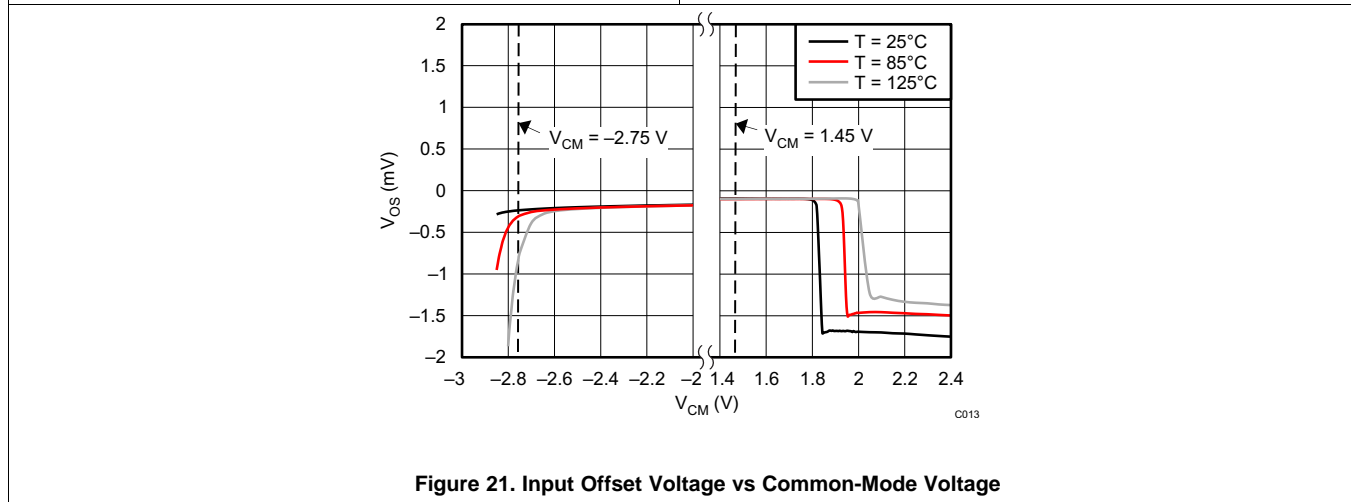


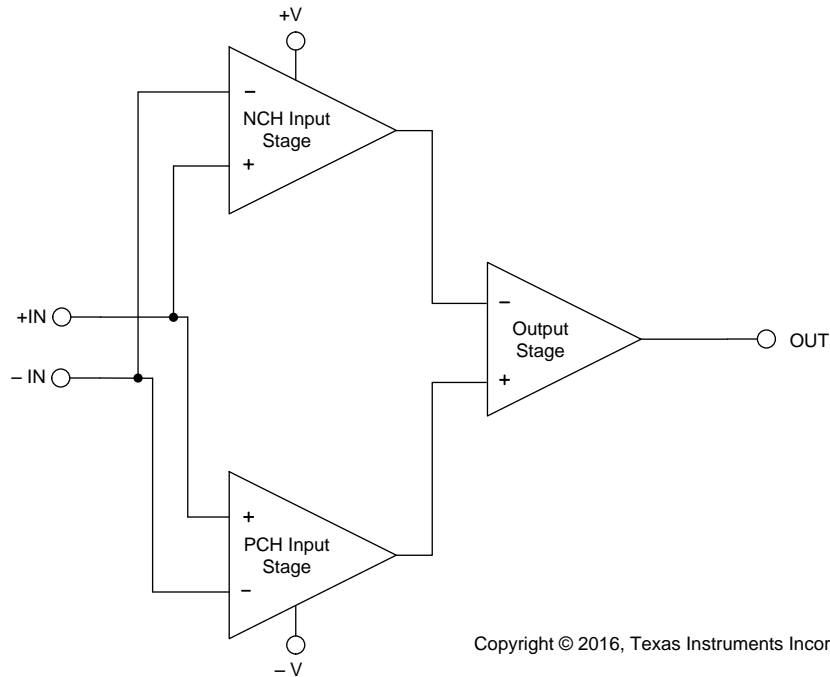
Figure 21. Input Offset Voltage vs Common-Mode Voltage

## 7 Detailed Description

### 7.1 Overview

The OPAx377-Q1 family belongs to a new generation of low-noise operational amplifiers, giving customers outstanding dc precision and ac performance. Low noise, rail-to-rail input and output, and low offset, drawing a low quiescent current, make these devices ideal for a variety of precision and portable applications. In addition, this device has a wide supply range with excellent PSRR, making it a suitable option for applications that are battery-powered without regulation.

### 7.2 Functional Block Diagram



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## 7.3 Feature Description

### 7.3.1 Operating Characteristics

The OPAx377-Q1 family of amplifiers has parameters that are fully specified from 2.2 V to 5.5 V ( $\pm 1.1$  V to  $\pm 2.75$  V). Many of the specifications apply from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in the *Typical Characteristics* section.

### 7.3.2 Common-Mode Voltage Range

The input common-mode voltage range of the OPAx377-Q1 series extends 100 mV beyond the supply rails. The offset voltage of the amplifier is low, from approximately  $(V-)$  to  $(V+) - 1$  V, as shown in Figure 22. The offset voltage increases as common-mode voltage exceeds  $(V+) - 1$  V. Common-mode rejection is specified from  $(V-) - 1.3$  V.

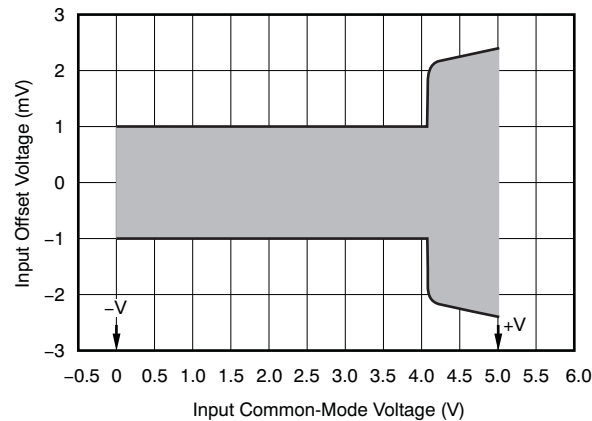
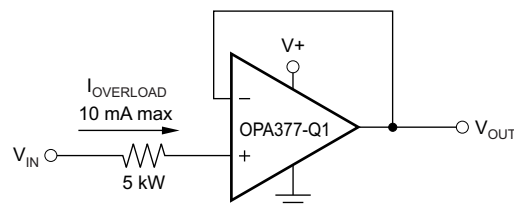


Figure 22. Offset and Common-Mode Voltage

### 7.3.3 Input and ESD Protection

The OPAx377-Q1 family incorporates internal electrostatic discharge (ESD) protection circuits on all pins. In the case of input and output pins, this protection primarily consists of current steering diodes connected between the input and power-supply pins. These ESD protection diodes also provide in-circuit, input overdrive protection, as long as the current is limited to 10 mA as stated in the *Absolute Maximum Ratings* table.

Figure 23 shows how a series input resistor may be added to the driven input to limit the input current. The added resistor contributes thermal noise at the amplifier input and its value must be kept to a minimum in noise-sensitive applications.



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Figure 23. Input Current Protection

## Feature Description (continued)

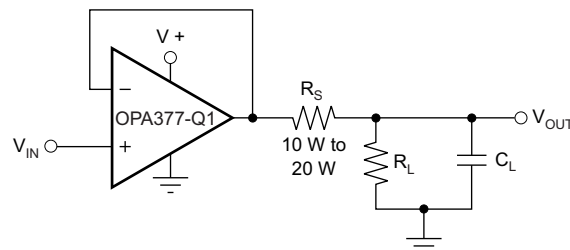
### 7.3.4 EMI Susceptibility and Input Filtering

Operational amplifiers vary in susceptibility to electromagnetic interference (EMI). If conducted EMI enters the operational amplifier, the dc offset observed at the amplifier output may shift from the nominal value while the EMI is present. This shift is a result of signal rectification associated with the internal semiconductor junctions. While all amplifier pin functions can be affected by EMI, the input pins are likely to be the most susceptible. The OPAx377-Q1 operational amplifier family incorporates an internal input low-pass filter that reduces the amplifier response to EMI. Both common-mode and differential mode filtering are provided by the input filter. The filter is designed for a cutoff frequency of approximately 75 MHz (–3 dB), with a roll-off of 20 dB per decade.

### 7.3.5 Capacitive Load and Stability

The OPAx377-Q1 series of amplifiers may be used in applications where driving a capacitive load is required. As with all op amps, there may be specific instances where the OPAx377-Q1 can become unstable, leading to oscillation. The particular op amp circuit configuration, layout, gain, and output loading are some of the factors to consider when establishing whether an amplifier will be stable in operation. An op amp in the unity-gain (1 V/V) buffer configuration and driving a capacitive load exhibits a greater tendency to be unstable than an amplifier operated at a higher noise gain. The capacitive load, in conjunction with the op amp output resistance, creates a pole within the feedback loop that degrades the phase margin. The degradation of the phase margin increases as the capacitive loading increases.

The OPAx377-Q1 in a unity-gain configuration can directly drive up to 250-pF pure capacitive load. Increasing the gain enhances the ability of the amplifier to drive greater capacitive loads; see the typical characteristic plot, [Figure 15](#). In unity-gain configurations, capacitive load drive can be improved by inserting a small (10-Ω to 20-Ω) resistor,  $R_S$ , in series with the output, as shown in [Figure 24](#). This resistor significantly reduces ringing while maintaining dc performance for purely capacitive loads. However, if there is a resistive load in parallel with the capacitive load, a voltage divider is created, introducing a gain error at the output and slightly reducing the output swing. The error introduced is proportional to the ratio  $R_S/R_L$ , and is generally negligible at low output current levels.



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**Figure 24. Improving Capacitive Load Drive**

## 7.4 Device Functional Modes

The OPAx377-Q1 has a single functional mode and is operational when the power-supply voltage is greater than 2.2 V ( $\pm 1.1$  V). The maximum power supply voltage for the OPAx376-Q1 is 5.5 V ( $\pm 2.75$  V).

## 8 Application and Implementation

### NOTE

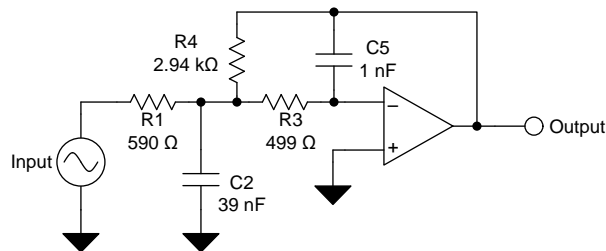
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The OPAx377-Q1 family of operational amplifiers is built on a precision analog CMOS technology featuring low noise and low offset voltage. The OPAx377-Q1 family delivers excellent offset voltage (250  $\mu\text{V}$ , typical). Additionally, the amplifier boasts a fast slew rate, low drift, low noise, and excellent PSRR and  $A_{OL}$ . These 5.5-MHz CMOS op amps operate on 760  $\mu\text{A}$  (typical) quiescent current.

### 8.2 Typical Application

Low-pass filters are commonly employed in signal processing applications to reduce noise and prevent aliasing. The OPA377-Q1 is ideally suited to construct high-speed, high-precision active filters. Figure 25 shows a second-order, low-pass filter commonly encountered in signal processing applications.



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**Figure 25. Typical Application Schematic**

#### 8.2.1 Design Requirements

Use the following parameters for this design example:

- Gain = 5 V/V (inverting gain)
- Low-pass cutoff frequency = 25 kHz
- Second-order Chebyshev filter response with 3-dB gain peaking in the passband

#### 8.2.2 Detailed Design Procedure

The infinite-gain multiple-feedback circuit for a low-pass network function is shown in Figure 25. Use Equation 1 to calculate the voltage transfer function.

$$\frac{\text{Output}}{\text{Input}}(s) = \frac{-1/R_1 R_3 C_2 C_5}{s^2 + (s/C_2)(1/R_1 + 1/R_3 + 1/R_4) + 1/R_3 R_4 C_2 C_5} \quad (1)$$

This circuit produces a signal inversion. For this circuit, the gain at dc and the low-pass cutoff frequency are calculated by Equation 2:

$$\text{Gain} = \frac{R_4}{R_1}$$

$$f_c = \frac{1}{2\pi} \sqrt{(1/R_3 R_4 C_2 C_5)} \quad (2)$$

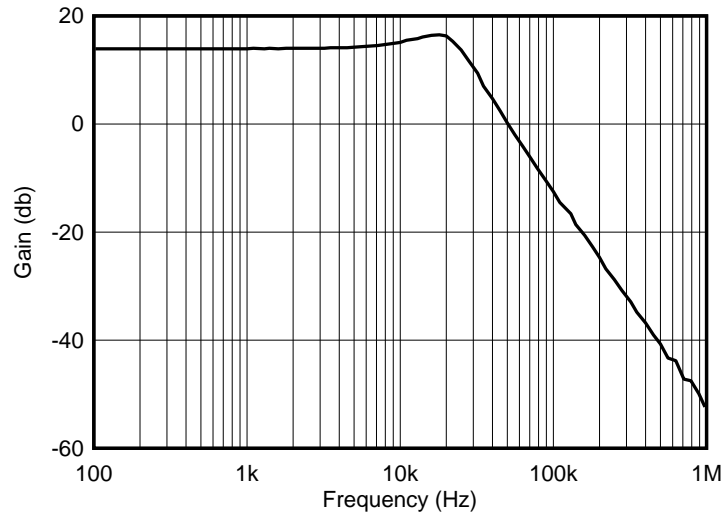
Software tools are readily available to simplify filter design. WEBENCH® Filter Designer is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.



**Typical Application (continued)**

Available as a web-based tool from the WEBENCH® Design Center, [WEBENCH® Filter Designer](#) allows to design, optimize, and simulate complete multi-stage active filter solutions within minutes.

**8.2.3 Application Curve**



**Figure 26. Low-Pass Filter Transfer Function**

## 9 Power Supply Recommendations

The OPAx377-Q1 family of devices is specified for operation from 2.2 V to 5.5 V ( $\pm 1.1$  V to  $\pm 2.75$  V); many specifications apply from  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ . Parameters that can exhibit significant variance with regard to operating voltage or temperature are presented in the [Typical Characteristics](#) section.

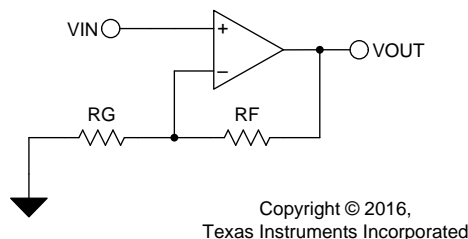
## 10 Layout

### 10.1 Layout Guidelines

For best operational performance of the device, use good printed circuit board (PCB) layout practices, including:

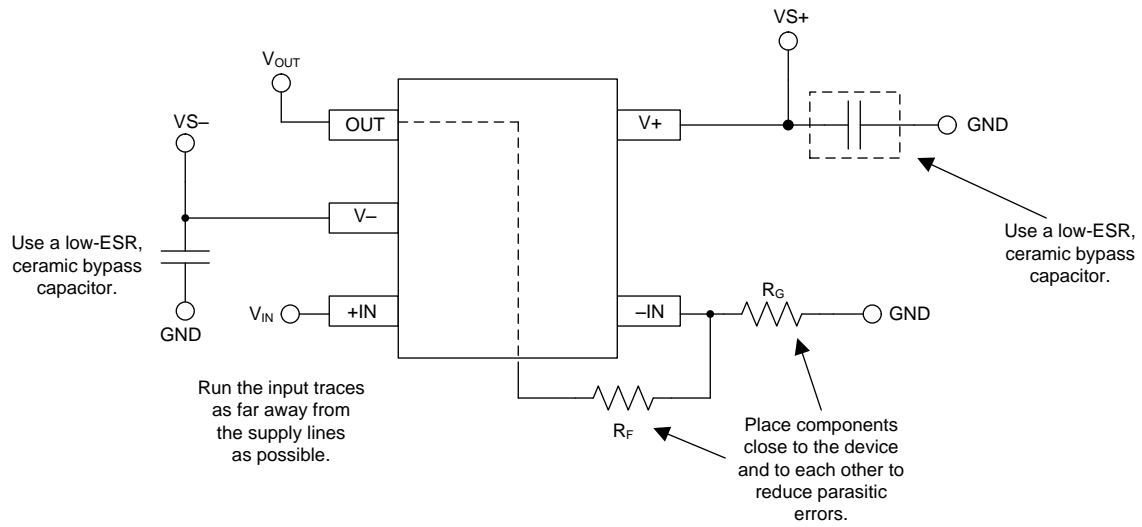
- Noise can propagate into analog circuitry through the power pins of the circuit as a whole and op amp itself. Bypass capacitors are used to reduce the coupled noise by providing low-impedance power sources local to the analog circuitry.
  - Connect low-ESR, 0.1- $\mu\text{F}$  ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from  $V+$  to ground is applicable for single-supply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes. A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital and analog grounds paying attention to the flow of the ground current. For more detailed information refer to the application report, *Circuit Board Layout Techniques*, [SLOA089](#).
- In order to reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If these traces cannot be kept separate, crossing the sensitive trace perpendicular is much better as opposed to in parallel with the noisy trace.
- Place the external components as close to the device as possible. As shown in [Figure 28](#), keeping  $R_F$  and  $R_G$  close to the inverting input minimizes parasitic capacitance.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.
- Cleaning the PCB following board assembly is recommended for best performance.
- Any precision integrated circuit may experience performance shifts due to moisture ingress into the plastic package. Following any aqueous PCB cleaning process, baking the PCB assembly is recommended to remove moisture introduced into the device packaging during the cleaning process. A low temperature, post cleaning bake at  $85^{\circ}\text{C}$  for 30 minutes is sufficient for most circumstances.

### 10.2 Layout Example



**Figure 27. Typical Schematic for PCB Layout Example**

**Layout Example (continued)**



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**Figure 28. Typical PCB Layout Example**

## 11 Device and Documentation Support

### 11.1 Device Support

#### 11.1.1 Development Support

##### 11.1.1.1 TINA-TI™ (Free Software Download)

TINA™ is a simple, powerful, and easy-to-use circuit simulation program based on a SPICE engine. TINA-TI™ is a free, fully-functional version of the TINA software, preloaded with a library of macro models in addition to a range of both passive and active models. TINA-TI provides all the conventional dc, transient, and frequency domain analysis of SPICE, as well as additional design capabilities.

Available as a [free download](#) from the Analog eLab Design Center, TINA-TI offers extensive post-processing capability that allows users to format results in a variety of ways. Virtual instruments offer the ability to select input waveforms and probe circuit nodes, voltages, and waveforms, creating a dynamic quick-start tool.

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#### NOTE

These files require that either the TINA software (from DesignSoft™) or TINA-TI software be installed. Download the free TINA-TI software from the [TINA-TI folder](#).

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##### 11.1.1.2 DIP Adapter EVM

The [DIP Adapter EVM](#) tool provides an easy, low-cost way to prototype small surface mount ICs. The evaluation tool these TI packages: D or U (SOIC-8), PW (TSSOP-8), DGK (MSOP-8), DBV (SOT23-6, SOT23-5 and SOT23-3), DCK (SC70-6 and SC70-5), and DRL (SOT563-6). The DIP Adapter EVM may also be used with terminal strips or may be wired directly to existing circuits.

##### 11.1.1.3 Universal Op Amp EVM

The [Universal Op Amp EVM](#) is a series of general-purpose, blank circuit boards that simplify prototyping circuits for a variety of IC package types. The evaluation module board design allows many different circuits to be constructed easily and quickly. Five models are offered, with each model intended for a specific package type. PDIP, SOIC, MSOP, TSSOP and SOT23 packages are all supported.

---

#### NOTE

These boards are unpopulated, so users must provide their own ICs. TI recommends requesting several op amp device samples when ordering the Universal Op Amp EVM.

---

##### 11.1.1.4 TI Precision Designs

TI Precision Designs are analog solutions created by TI's precision analog applications experts and offer the theory of operation, component selection, simulation, complete PCB schematic and layout, bill of materials, and measured performance of many useful circuits. TI Precision Designs are available online at <http://www.ti.com/ww/en/analog/precision-designs/>.

##### 11.1.1.5 WEBENCH® Filter Designer

[WEBENCH® Filter Designer](#) is a simple, powerful, and easy-to-use active filter design program. The WEBENCH Filter Designer lets you create optimized filter designs using a selection of TI operational amplifiers and passive components from TI's vendor partners.

Available as a web-based tool from the WEBENCH® Design Center, [WEBENCH® Filter Designer](#) allows you to design, optimize, and simulate complete multistage active filter solutions within minutes.

## 11.2 Documentation Support

### 11.2.1 Related Documentation

For related documentation see the following:

- *Circuit Board Layout Techniques*, [SLOA089](#)
- *Operational Amplifier Gain stability, Part 3: AC Gain-Error Analysis*, [SLYT383](#)
- *Operational Amplifier Gain Stability, Part 2: DC Gain-Error Analysis*, [SLYT374](#)
- *Op Amp Performance Analysis*, [SBOS054](#)
- *Shelf-Life Evaluation of Lead-Free Component Finishes*, [SZZA046](#)
- *Single-Supply Operation of Operational Amplifiers*, [SBOA059](#)
- *Tuning in Amplifiers*, [SBOA067](#)
- *Using Infinite-Gain, MFB Filter Topology in Fully Differential Active Filters*, [SLYT343](#)

### 11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At [e2e.ti.com](#), you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

### 11.4 Trademarks

TINA-TI, E2E are trademarks of Texas Instruments.

WEBENCH is a registered trademark of Texas Instruments.

TINA, DesignSoft are trademarks of DesignSoft, Inc.

### 11.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### 11.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
OPA2377QDGKRQ1	ACTIVE	VSSOP	DGK	8	2500	RoHS & Green	NIPDAUAG	Level-2-260C-1 YEAR	-40 to 125	2377	<a href="#">Samples</a>
OPA377QDBVRQ1	ACTIVE	SOT-23	DBV	5	3000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	377Q	<a href="#">Samples</a>
OPA4377AQPWRQ1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-2-260C-1 YEAR	-40 to 125	4377Q1	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**OTHER QUALIFIED VERSIONS OF OPA2377-Q1, OPA377-Q1, OPA4377-Q1 :**

- Catalog: [OPA2377](#), [OPA377](#), [OPA4377](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product

**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
OPA2377QDGKRQ1	VSSOP	DGK	8	2500	330.0	12.4	5.3	3.4	1.4	8.0	12.0	Q1
OPA377QDBVRQ1	SOT-23	DBV	5	3000	178.0	9.0	3.3	3.2	1.4	4.0	8.0	Q3
OPA4377AQPWRQ1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1



## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
OPA2377QDGKRQ1	VSSOP	DGK	8	2500	366.0	364.0	50.0
OPA377QDBVRQ1	SOT-23	DBV	5	3000	180.0	180.0	18.0
OPA4377AQPWRQ1	TSSOP	PW	14	2000	356.0	356.0	35.0

# DBV0005A



# PACKAGE OUTLINE

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



4214839/J 02/2024

**NOTES:**

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. Reference JEDEC MO-178.
4. Body dimensions do not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.25 mm per side.
5. Support pin may differ or may not be present.

# EXAMPLE BOARD LAYOUT

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:15X



SOLDER MASK DETAILS

4214839/J 02/2024

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

# EXAMPLE STENCIL DESIGN

DBV0005A

SOT-23 - 1.45 mm max height

SMALL OUTLINE TRANSISTOR



SOLDER PASTE EXAMPLE  
BASED ON 0.125 mm THICK STENCIL  
SCALE:15X

4214839/J 02/2024

NOTES: (continued)

8. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.
9. Board assembly site may have different recommendations for stencil design.

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



4040064-3/G 02/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
  - E. Falls within JEDEC MO-153

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
  - This drawing is subject to change without notice.
  - Publication IPC-7351 is recommended for alternate designs.
  - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

DGK (S-PDSO-G8)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 per end.
  - D. Body width does not include interlead flash. Interlead flash shall not exceed 0.50 per side.
  - E. Falls within JEDEC MO-187 variation AA, except interlead flash.



- NOTES:
- A. All linear dimensions are in millimeters.
  - B. This drawing is subject to change without notice.
  - C. Publication IPC-7351 is recommended for alternate designs.
  - D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
  - E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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