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[Reference](http://www.ti.com/lit/ug/tidu584/tidu584.pdf?dcmp=dsproject&hqs=rd) Design

[DRV110](http://www.ti.com/product/drv110?qgpn=drv110)

SLVSBA8G –MARCH 2012–REVISED MARCH 2018

DRV110 120- and 230-V AC, 6- to 48-V DC Current Controller for Solenoids, Relays, and Valves

1 Features

- Internal Zener Diode on Supply Pin for High-Voltage Operation
	- 120- and 230-V AC Supply Through Rectifier and R_S Resistor
	- $-$ 24-V, 48-V, and Higher DC Supply Through R_S Resistor
- • Drives an External MOSFET With PWM to Control Solenoid Current
	- External Sense Resistor for Regulating Solenoid Current
- Fast Ramp-Up of Solenoid Current to Ensure Activation
- Solenoid Current is Reduced in Hold Mode for Lower Power and Thermal Dissipation
- Ramp Peak Current, Keep Time at Peak Current, Hold Current, and PWM Clock Frequency Can Be Set Externally. They Can Also Be Operated at Nominal Values Without External Components.
- **Protection**
	- Thermal Shutdown
	- Undervoltage Lockout (UVLO)
- Optional STATUS Output
- Operating Temperature Range: –40ºC to +125ºC
- • 8-Pin and 14-Pin TSSOP Package Options

2 Applications

- Electromechanical Drivers: Solenoids, Valves, Relays, Contactors, Switchgear, Pneumatics
- • White Goods, Solar, Transportation, Smart Grid, Power Distribution

3 Description

The DRV110 device is a PWM current controller for solenoids. The device is designed to regulate the current with a well-controlled waveform to reduce power dissipation. The solenoid current is ramped up fast to ensure opening of the valve or relay. After initial ramping, the solenoid current is kept at a peak value to ensure correct operation, after which the current is reduced to a lower hold level to avoid thermal problems and reduce power dissipation.

The peak current duration is set with an external capacitor. The peak and hold levels of the current ramp, as well as the PWM frequency, can independently be set with external resistors. External setting resistors can also be omitted if the default values for the corresponding parameters are suitable for the application.

The DRV110 device has an internal Zener diode that limits the supply at VIN to V_{ZENER} for applications that require a higher supply voltage. Using the internal Zener, the DRV110 can be powered from 120-V and 230-V AC supplies through a rectifier and currentlimiting resistor. High DC voltages such as 48-V can also be accommodated this way.

Device Information[\(1\)](#page-0-0)

(1) For all available packages, see the orderable addendum at the end of the data sheet.

DRV110 Supplied by Power Line Voltage

Product Folder Links: *[DRV110](http://www.ti.com/product/drv110?qgpn=drv110)*

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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision F (March 2017) to Revision G Page

Changes from Revision E (November 2016) to Revision F Page

Changes from Revision D (June 2016) to Revision E Page • Changed the title of the document to include V AC and V DC values ... [1](#page-0-4) • Revised *Features* and *Applications* lists .. [1](#page-0-5) • Changed first page graphic to schematic .. [1](#page-0-6) • Revised table notes for *Recommended Operating Conditions* table.. [5](#page-4-8) **Changes from Revision C (April 2016) to Revision D Page**

Changes from Revision B (July 2015) to Revision C Page

Changes from Revision A (January 2013) to Revision B Page

EXAS NSTRUMENTS

5 Pin Configuration and Functions

Pin Functions

(1) In the 8-pin package, the HOLD pin is not bonded out. For this package, the HOLD mode is configured to default (internal) settings.

6 Specifications

6.1 Absolute Maximum Ratings

See (1) and (2)

(1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions* is not implied. Exposure to absolute–maximum–rated conditions for extended periods may affect device reliability.

(2) All voltage values are with respect to network ground terminal.

6.2 ESD Ratings

(1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

(2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

 -40 °C ≤ T_A ≤ 125°C (unless otherwise noted)

(1) The device regulates the supply with an internal Zener diode. The device sinks up to 3 mA with the added supply current. See [Equation](#page-11-1) 5 to find appropriate value for the R_S resistor.

The maximum input voltage of the device depends on the clamping voltage of the internal Zener diode, which changes over temperature. A current-limiting resistor is required to limit current to the Zener diode if the input voltage (V_{IN}) is greater than V_{ZENER} . For more information on resistor sizing see the *Detailed [Description](#page-7-0)* section and *Application and [Implementation](#page-13-0)* section.

For V_S voltages less than V_{ZENER} , $V_{IN} = V_S$. For V_S voltages greater than V_{ZENER} , $V_{IN} = V_{ZENER}$.

(4) 4.7-µF input capacitor and full wave rectified 230-Vrms AC supply results in approximately 500-mV supply ripple.

6.4 Thermal Information

(1) For more information about traditional and new thermal metrics, see the *[Semiconductor](http://www.ti.com/lit/pdf/spra953) and IC Package Thermal Metrics* application [report.](http://www.ti.com/lit/pdf/spra953)

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6.5 Electrical Characteristics

6.6 Typical Characteristics

Figure 1. Solenoid Current, EN, and PWM vs Time

Texas **NSTRUMENTS**

7 Detailed Description

7.1 Overview

The DRV110 device provides a PWM current controller for use with solenoids. The device provides a quick ramp to a high peak current value in order to ensure opening of the valve or relay. The current is held for a programmable time and then lowered to the hold current value to maintain the open state of the valve or relay while reducing the total current consumption. Peak current duration, peak current amount, hold current amount (in the 14-pin package), and PWM frequency can all be controlled by external components or used at default levels by omitting these components (except peak current duration).

Enable and disable of the switch is controlled by the EN pin. The EN pin contains an internal resistor network to set the pin to logic HIGH when the EN pin is floating. This feature can be used for situations where a control signal is not required and the solenoid is only energized when a supply voltage is present. Such applications could be valves or contactors.

The DRV110 also features a wide VIN range with an internal bypass regulator to maintain VIN at an acceptable level. Finally, the 14-pin package features an open-drain pull-down path on the STATUS pin which is enabled as long as undervoltage lockout or thermal shutdown has not triggered.

7.2 Functional Block Diagram

(1) Available only in the 14-pin package.

8

7.3 Feature Description

The DRV110 controls the current through the solenoid as shown in [Figure](#page-8-1) 2. Activation starts when EN pin voltage is pulled high either by an external driver or internal pullup. In the beginning of activation, DRV110 allows the solenoid current to ramp up to the peak value I_{PEAK} and it regulates it at the peak value for the time, t_{KEEP} , before reducing it to I_{HOLD} . The solenoid current is regulated at the hold value as long as the EN pin is kept high. The initial current ramp-up time depends on the inductance and resistance of the solenoid. Once EN pin is driven to GND, DRV110 allows the solenoid current to decay to zero.

Figure 2. Typical Current Waveform Through the Solenoid

7.3.1 Keep Time

The keep time, t_{KEEP} , is set externally by connecting a capacitor to the KEEP pin. A constant current is sourced from the KEEP pin that is driven into an external capacitor resulting in a linear voltage ramp. When the KEEP pin voltage reaches 100 mV, the current regulation reference voltage, V_{REF} , is switched from V_{PEAK} to V_{HOLD} . The internal current source is switched off, and the capacitor is grounded for discharge. The dependency of t_{KFFP} from the external capacitor size can be calculated with [Equation](#page-8-2) 1.

$$
t_{KEEP}\left[s\right] = C_{KEEP}\left[F\right]\cdot 10^5 \left[\frac{s}{F}\right]
$$

(1)

7.3.2 PWM Current Control

The current control loop regulates, cycle-by-cycle, the solenoid current by sensing voltage at the SENSE pin and controlling the external switching device gate through the OUT pin. During the ON-cycle, the OUT pin voltage is driven and kept high (equal to VIN voltage) allowing current to flow through the external switch as long as the voltage at the SENSE pin is less than V_{REF} . As soon as the voltage at the SENSE pin is above V_{REF} , the OUT pin voltage is immediately driven low and kept low until the next ON-cycle is triggered by the internal PWM clock signal. In the beginning of each ON-cycle, the OUT pin voltage is driven high and kept high for at least the time determined by the minimum PWM signal duty cycle, D_{MIN} .

Because the current sense is done by comparing the voltage at the SENSE pin to a reference voltage, the DRV110 device acts like a hysteresis controller. When the device acts like a hysteresis controller, it can make the PWM frequency and duty cycle appear uneven for some solenoids (see [Figure](#page-9-2) 3).

Feature Description (continued)

(1) The DRV110 device measures the voltage at the SENSE node (V_{SENSE}) . This voltage is compared against the reference voltage (V_{REF}) each clock cycle. The voltage at the output node (V_{OUT}) becomes low when V_{SENSE} \geq V_{REF}. The duty cycle (D) of the output voltage varies from 8% to 100%. In summary, the SENSE voltage is sampled after each rising edge of the PWM CLK signal (PWM_{CLK}) and goes low when V_{SENSE} ≥ V_{REF} at a minimum duty cycle of 8%.

Figure 3. DRV110 Current Control with Varying OUT Duty Cycle

7.3.3 Configuring Peak and Hold Currents

I_{PEAK} and I_{HOLD} depend on fixed resistance values R_{PEAK} and R_{HOLD} as shown in [Figure](#page-10-4) 4. If the PEAK pin or HOLD pin is connected to ground or R_{PEAK} or R_{HOLD} is less than 43.33 kΩ (typical), then I_{PEAK} is at its default value of 300 mA for I_{PFAK} and 50 mA for I_{HOL} .

The I_{PEAK} value can alternatively be set by connecting an external resistor to ground from the PEAK pin. For example, if a 60-kΩ (= R_{PEAK}) resistor is connected between PEAK and GND, and R_{SENSE} = 1 Ω, then the externally set I_{PEAK} level will be 900 mA. If R_{PEAK} = 200 kΩ and R_{SENSE} = 1 Ω, then the externally set I_{PEAK} level will be 300 mA. TI does not recommend using a resistor from 30 kΩ and 55 kΩ to avoid the I_{PEAK} or I_{HOLD} current slipping from the maximum current setting to the default setting.

In case R_{SENSE} = 2 Ω instead of 1 Ω, then I_{PEAK} = 450 mA (when R_{PEAK} = 55 kΩ) and I_{PEAK} = 150 mA (when R_{PEAK} = 200 kΩ). In the 8-pin package, the HOLD reference uses the internal V_{REF} setting of 50 mV. In the 14pin package, external setting of the HOLD current, I_{HOLD} , works in the same way as the external setting for I_{PEAK} but the current levels are $1/6$ of the I_{PFAK} levels for the same resistor setting.

External settings for I_{PEAK} and I_{HOLD} are independent of each other. If R_{PEAK} or R_{HOLD} is decreased below 33.33 kΩ (typical value), then the reference is clamped to the internal setting of 300 mV for PEAK and 50 mV for HOLD. Use [Equation](#page-10-1) 2 and Equation 3 to calculate the values for I_{PEAK} and I_{HOLD} respectively.

The currents and resistor values should be chosen such that the voltage across the sense resistor is more than 30 mV.

Feature Description (continued)

Figure 4. **I**_{PEAK} and **I**_{HOLD} settings for $R_{\text{SENSE}} = 1 \Omega$

7.3.4 Configuring the PWM Frequency

Frequency of the internal PWM clock signal, PWM_{CLK}, that triggers each OUT pin ON-cycle can be adjusted by external resistor, R_{OSC} , connected between OSC and GND. Frequency as a function of resistor value is shown in [Figure](#page-10-6) 5. Default frequency is used when OSC is connected to GND directly. Use [Equation](#page-10-7) 4 to calculate the PWM frequency as a function of the external fixed adjustment resistor value (greater than 160 kΩ).

$$
f_{\text{PWM}} = \frac{60 \text{ kHz}}{R_{\text{OSC}}} \times 66.67 \text{ k}\Omega; \ 160 \text{ k}\Omega < R_{\text{OSC}} < 2 \text{ M}\Omega \tag{4}
$$
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Figure 5. PWM Clock Frequency Setting

7.3.5 Voltage Supply and Integrated Zener Diode

Voltage at the OUT pin, that is the gate voltage of an external switching device, is equal to VIN voltage during the ON-cycle. The voltage is driven to ground during the OFF-cycle. VIN voltages below V_{ZENER} can be supplied directly from an external voltage source. Supply voltages of at least 6 V are supported.

Feature Description (continued)

The DRV110 is able to regulate VIN voltage from a higher external supply voltage, V_s , by an internal bypass regulator that replicates the function of an ideal Zener diode. This requires that the supply current is sufficiently limited by an external resistor between V_S and the VIN pin. An external capacitor connected to the VIN pin is used to store enough energy to charge the external switch gate capacitance at the OUT pin. A range of current limiting resistor sizes ($R_{S,min}$ and $R_{S,max}$) can be calculated with [Equation](#page-11-2) 5 and Equation 6. This range keeps the VIN current within the recommended operating conditions.

$$
R_{S,max} = \frac{V_{S,minDC} - V_{ZENER}}{1 mA + I_{Gate,AVE}}
$$

where

 $I_{\text{Gate,AVE}}$ is the current flowing to the external switch. For a MOSFET, $I_{\text{Gate,AVE}}$ is equal to the external FET gate charge multiplied by f_{PWM} . (5)

$$
R_{S,min} = \frac{V_{S,maxDC} - V_{ZENER}}{3 mA + I_{Gate,AVE}}
$$
(6)

Ideally, the DRV110 device clamps the input voltage to 15 V. For configurations that do not use the EN pin (force the pin high or leave it floating), the DRV110 device clamps at 15 V (V_{ZENER} = 15 V) across the temperature range of the device. If the EN pin is set to 0, then refer to the values in [Table](#page-11-3) 1 to find the V_{ZENER} used when calculating the value of R_s , based on the temperature range of the application. Because the V_{ZENER} changes when the EN state changes, select a value for R_S that meets the current requirements at both V_{ZENER} voltages.

Table 1. VZENER Value

The open-drain pulldown path at the STATUS pin is deactivated if the undervoltage lockout or thermal shutdown blocks have triggered or if the EN pin is low.

7.4 Device Functional Modes

7.4.1 Normal Mode

The DRV110 transitions through three different states in normal mode:

- **OFF state** In the OFF state, the EN pin is low and the PWM output is off.
- **PEAK state** The PEAK state begins when the EN pin is set high, and ends when the t_{KEEP} time has been reached. During this state, the PWM operates to reach the I_{PEAK} current set by the R_{PEAK} resistor.
- HOLD state In the HOLD state, the t_{KEEP} time has been reached, and the PWM continues to operate but at the I_{HOLD} level. This continues until the EN pin is set low again and the PWM turns off.

7.4.2 Shutdown

The DRV110 turns off the gate driver in undervoltage lockout (VIN < 4.6 V) or thermal shutdown ($T_J > 160^{\circ}$ C). If temperature shutdown is activated, the DRV110 resumes operation when the junction temperature is below 140°C. The shutdown conditions are expressed by the STATUS pin going to the high-impedance state. A pullup resistor can be connected to the STATUS pin so these conditions may be observed by a microcontroller. [Table](#page-12-2) 2 provides an explanation of this operation.

Table 2. Shutdown Operation

8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

8.1 Application Information

The DRV110 device is designed to operate a solenoid valve or relay. For detailed information on using the DRV110 with 230 V AC solenoids, see *Current Controlled Driver for 230V AC Solenoids [Reference](http://www.ti.com/lit/pdf/TIDU584) Design*. A typical DC input design will be outlined in *Typical [Application](#page-13-2)*. Approximate resistor and capacitor values for the peak current, hold current, sense, and keep time will be derived for a sample application.

8.2 Typical Application

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Figure 6. DRV110 Powered by a Rectified AC Power Source

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Typical Application (continued)

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Figure 7. DRV110 Powered by a DC Power Source Greater than 15 V

8.2.1 Design Requirements

The key elements to identify here are the system input voltage, peak current, hold current, and peak keep time values required for the solenoid or relay being used. With these values, approximate R_S , R_{PEAK} , R_{HOLD} (for 14-pin package), C_{KFFP} , and R_{SFNSF} values can be determined and the proper FET and diode can be identified. R_{OSC} can be varied in order to tune the circuit to the chosen solenoid or relay.

8.2.2 Detailed Design Procedure

8.2.2.1 Current Limiting Resistor Selection

The temperature range, input voltage, and enable state must be considered when selecting the current limiting resistor. These values must be considered because the Zener clamping voltage of the DRV110 device starts dropping from its ideal 15 V at temperatures greater than 45°C when the EN pin is pulled low. Applications that leave the EN pin floating or pulled high at all times only require a current-limiting resistor when the input voltage is greater than 15 V across all temperature.

While using a current-limiting resistor is not required when the supply voltage (V_S) is less than the Zener clamping voltage, V_{ZENER}, TI recommends populating a small resistor in case of possible input voltage transients during operation. At the very least, TI recommends placing a resistor footprint jumped by a 0-Ω resistor. [Table](#page-15-0) 3 lists recommended resistor values for voltages close to V_{ZENER} and common voltages greater than V_{ZENER} for different enable states.

Table 3. Recommended Resistor Values

8.2.2.2 Passive Component Selection

With the selected peak current, hold current, and peak keep time values, the values of R_{PEAK}, R_{HOLD} (for 14-pin package), C_{KEEP} , and R_{SENSE} can be determined. [Table](#page-15-1) 4 lists the example values and results from calculation.

VARIABLE	VALUE	14-PIN VALUES	8-PIN VALUES	CALCULATED FROM
Peak current	150 mA	$R_{\text{PFAK}} = 400 \text{ k}\Omega$	R_{PEAK} = 400 kΩ	Equation 2
Hold current	50 mA	R_{HOLD} = 200 k Ω or connect HOLD to ground	Default	Equation 3
Keep time	100 ms	$C_{KFFP} = 1 \mu F$	C_{KEEP} = 1 µF	Equation 1
PWM frequency	20 kHz	R_{OSC} = Shorted to ground	R_{OSC} = Shorted to ground	Equation 4

Table 4. Example Application Values When RSENSE = 1 Ω

Use [Equation](#page-10-1) 2 and Equation 3 to calculate the values of the ${\sf R}_{\sf PEAK}$ resistor and ${\sf R}_{\sf HOLD}$ (if applicable) resistor. For the sample values, the R_{PEAK} resistor is set to 400 kΩ and the R_{HOLD} resistor is shorted to GND. TI recommends using a 0-Ω resistor for prototyping in case changes to this value are desired.

Next, select the value of the C_{KEEP} capacitor based on [Equation](#page-8-2) 1. For the sample value, the C_{KEEP} capacitor is set to 1 µF. The R_{OSC} resistor is initially be shorted to GND, but a 0- Ω resistor is also recommended for prototyping. Additionally, a low-pass filter on the SENSE line can be added in a high-noise environment and is recommended for prototyping. The typical value for the low pass filter resistor is 1 kΩ and the typical value for the filter capacitor is 100 pF.

The value of sense resistor can be selected based on the preference of the designer. The only restriction is that the voltage across the sense resistor (found by the R_{SENSE} resistance times the I_{HOLD} current) must be greater than 30 mV for reliable operation.

The external FET and current recirculation diode must be selected based on the current values defined in [Table](#page-15-1) 4 and the supply voltage. The current recirculation diode should be a fast recovery diode.

8.2.3 Application Curve

Figure 8. I_{SOLENOID}, EN, and V_{IN} vs Time

9 Power Supply Recommendations

The input supply range must be at least 6 V, and needs a current-limiting resistor above V_{ZENER} . An input capacitor of 4.7 µF (typical) is required as well. I_Q max is 3 mA, but additional current will be required to operate the solenoid or relay.

10 Layout

10.1 Layout Guidelines

Routing for the SENSE pin should be careful to avoid noise sources. Routing for the output node and sense node should be minimized. The trace for the solenoid or relay current should be wide in order to prevent any unexpected voltage drop.

10.2 Layout Example

Figure 9. Layout Schematic

11 Device and Documentation Support

11.1 Documentation Support

11.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, *Current [Controlled](http://www.ti.com/lit/pdf/TIDU578) Driver for 24-V DC Solenoid With Plunger Fault Detection* reference [design](http://www.ti.com/lit/pdf/TIDU578)
- Texas Instruments, *Current Controlled Driver for 230V AC Solenoids [Reference](http://www.ti.com/lit/pdf/TIDU584) Design*
- Texas Instruments, *DRV110 and DRV120 [Evaluation](http://www.ti.com/lit/pdf/SLVU686) Modules (EVM)* user's guide

11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms](http://www.ti.com/corp/docs/legal/termsofuse.shtml) of [Use.](http://www.ti.com/corp/docs/legal/termsofuse.shtml)

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11.5 Electrostatic Discharge Caution

This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

11.6 Glossary

[SLYZ022](http://www.ti.com/lit/pdf/SLYZ022) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

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PACKAGING INFORMATION

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures. "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

TEXAS

ISTRUMENTS

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE

Pack Materials-Page 1

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PACKAGE MATERIALS INFORMATION

*All dimensions are nominal

Pack Materials-Page 2

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE

B. This drawing is subject to change without notice.

 $\hat{\mathbb{C}}$ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.

 $\hat{\mathbb{D}}$ Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.

E. Falls within JEDEC MO-153

PACKAGE OUTLINE

PW0008A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

NOTES:

1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.

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INCTDI IMENTC

- 2. This drawing is subject to change without notice.
- 3. This dimension does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.15 mm per side.
- 4. This dimension does not include interlead flash. Interlead flash shall not exceed 0.25 mm per side.
- 5. Reference JEDEC registration MO-153, variation AA.

EXAMPLE BOARD LAYOUT

PW0008A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

NOTES: (continued)

6. Publication IPC-7351 may have alternate designs.

7. Solder mask tolerances between and around signal pads can vary based on board fabrication site.

EXAMPLE STENCIL DESIGN

PW0008A TSSOP - 1.2 mm max height

SMALL OUTLINE PACKAGE

NOTES: (continued)

9. Board assembly site may have different recommendations for stencil design.

^{8.} Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.

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