

LM2902-Q1, LM2902B-Q1, and LM2902BA-Q1 Industry-Standard Quad Operational Amplifiers for Automotive Applications

1 Features

- AEC Q-100 qualified for automotive applications
 - Temperature grade 1: –40°C to +125°C
 - Device HBM ESD classification 2
 - Device CDM ESD classification C5
- Wide supply range of:
 - 3 V to 36 V (LM2902B-Q1 and LM2902BA-Q1)
 - 3 V to 32 V (LM2902KV and LM2902KAV)
 - 3 V to 26 V (all other products)
- Input offset voltage maximum at 25°C of:
 - 2 mV (LM2902BA-Q1 and LM2902KAV)
 - 3 mV (LM2902B-Q1)
 - 7 mV (all other products)
- Internal RF and EMI filter (LM2902B-Q1 and LM2902BA-Q1)
- Supply-current of 175 µA per channel, typical
- Unity-gain bandwidth of 1.2 MHz
- Common-mode input voltage range includes V–
- Differential input voltage range equal to maximum-rated supply voltage

2 Applications

- [Automotive lighting](#)
- [Body electronics](#)
- [Automotive head unit](#)
- [Telematics control unit](#)
- [Emergency call \(eCall\)](#)
- [Passive safety: brake system](#)
- Electric vehicle / hybrid electric:
 - [Inverter and motor control](#)
 - [On-board \(OBC\) and wireless charger](#)
 - [Battery management system \(BMS\)](#)

3 Description

This device consists of four independent high-gain frequency-compensated operational amplifiers that are designed specifically to operate from a single supply over a wide range of voltages. Operation from split supplies is possible when the difference between the two supplies is 3 V to 36 V (for B-version devices), 3 V to 32 V (for V-version devices) or 3 V to 26 V (for all other devices), and V_{CC} is at least 1.5 V more positive than the input common-mode voltage. The low supply-current drain is independent of the magnitude of the supply voltage.

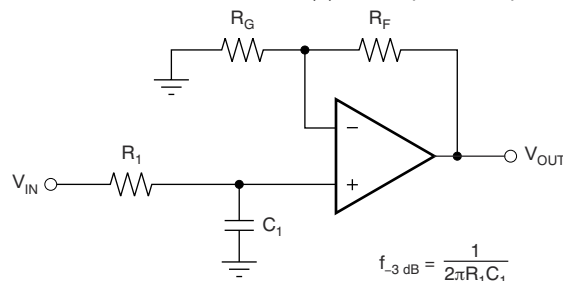
Applications include transducer amplifiers, dc amplification blocks, and all the conventional operational-amplifier circuits that now can be more easily implemented in single-supply-voltage systems. For example, the LM2902 can be operated directly from the standard 5-V supply that is used in digital systems and easily provides the required interface electronics without requiring additional ± 15 -V supplies.

Device Information

PART NUMBER ⁽¹⁾	PACKAGE	BODY SIZE (NOM)
LM2902B-Q1 ⁽²⁾	SOIC	8.65 mm × 3.91 mm
	TSSOP (14)	5.00 mm × 4.40 mm
LM2902BA-Q1 ⁽²⁾	SOIC (14)	8.65 mm × 3.91 mm
	TSSOP (14)	5.00 mm × 4.40 mm
LM2902-Q1	SOIC (14)	8.65 mm × 3.91 mm
	TSSOP (14)	5.00 mm × 4.40 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

(2) This product is preview only.



$$\frac{V_{OUT}}{V_{IN}} = \left(1 + \frac{R_F}{R_G}\right) \left(\frac{1}{1 + sR_1C_1}\right)$$

Single-Pole, Low-Pass Filter



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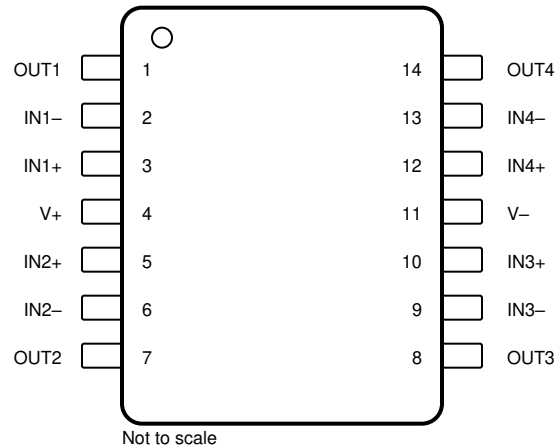
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4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Revision E (April 2008) to Revision F (May 2022)	Page
• Changed the name of the data sheet.....	1
• Revised <i>Features</i> section to include LM2902B-Q1 and LM2902BA-Q1.....	1
• Added <i>Applications</i> section.....	1
• Added LM2902B-Q1 and LM2902BA-Q1 to the <i>Device Information</i> table.....	1
• Added LM2902B-Q1 and LM2902BA-Q1 to the <i>Description</i> section.....	1
• Updated <i>Pin Configurations and Functions</i> section to include <i>Pin Functions</i> table.....	3
• Added LM2902B-Q1 and LM2902BA-Q1 to the <i>Absolute Maximum Ratings</i> table.....	4
• Added <i>ESD Ratings</i> table with LM2902B-Q1 and LM2902BA-Q1.....	4
• Added LM2902B-Q1 and LM2902B-Q1 to <i>Recommended Operating Conditions</i> section.....	4
• Added LM2902B-Q1 and LM2902BA-Q1 to <i>Thermal Information</i> section.....	5
• Added <i>Overview</i> section to the data sheet.....	11
• Added <i>Feature Description</i> section.....	11
• Added <i>Input Common Mode Range</i> section to <i>Feature Description</i> section.....	11
• Added <i>Device Functional Modes</i> information for LM2902B-Q1 and LM2902BA-Q1.....	12
• Added <i>Application and Implementation</i> section for LM2902B-Q1 and LM2902BA-Q1.....	13
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• Added <i>Power Supply Recommendations</i> section to data sheet.....	15
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• Added <i>Device and Documentation Support</i> section to data sheet.....	17
• Added <i>Mechanical, Packaging, and Orderable Information</i> section to data sheet.....	18

5 Pin Configurations and Functions



**Figure 5-1. D and PW Package
14-Pin SOIC and TSSOP
(Top View)**

Table 5-1. Pin Functions

PIN		I/O	DESCRIPTION
NAME	NO.		
IN1–	2	I	Inverting input, channel 1
IN1+	3	I	Noninverting input, channel 1
IN2–	6	I	Inverting input, channel 2
IN2+	5	I	Noninverting input, channel 2
IN3–	9	I	Inverting input, channel 3
IN3+	10	I	Noninverting input, channel 3
IN4–	13	I	Inverting input, channel 4
IN4+	12	I	Noninverting input, channel 4
NC	—	—	No internal connection
OUT1	1	O	Output, channel 1
OUT2	7	O	Output, channel 2
OUT3	8	O	Output, channel 3
OUT4	14	O	Output, channel 4
V–	11	—	Negative (lowest) supply or ground (for single-supply operation)
V+	4	—	Positive (highest) supply

6 Specifications

6.1 Absolute Maximum Ratings

For $T_A = 25^\circ\text{C}$ (unless otherwise noted)⁽¹⁾

	LM2902B-Q1, LM2902BA-Q1	LM2902-Q1	LM2902KV-Q1	UNIT
Supply voltage, V_{CC} ⁽²⁾	40	26	32	V
Differential input voltage, V_{ID} ⁽³⁾	± 40	± 26	± 32	V
Input voltage, V_I	-0.3 to 40	-0.3 to 26	-0.3 to 32	V
Duration of output short circuit (one amplifier) to ground at (or below) $T_A = 25^\circ\text{C}$, $V_{CC} \leq 15\text{ V}$ ⁽⁴⁾	Unlimited	Unlimited	Unlimited	
Operating virtual junction temperature, T_J	150	142	142	$^\circ\text{C}$
Storage temperature range, T_{stg}	-65 to 150	-65 to 150	-65 to 150	$^\circ\text{C}$

- (1) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to network ground terminal GND.
- (3) Differential voltages are at IN+ with respect to IN-.
- (4) Short circuits from outputs to V_{CC} can cause excessive heating and eventual destruction.

6.2 ESD Ratings

		VALUE	UNIT
LM2902B-Q1, LM2902BA-Q1, LM2902KV-Q1, and LM2902KAV-Q1			
$V_{(ESD)}$ Electrostatic discharge	Human-body model (HBM), per AEC Q100-002 ⁽¹⁾	± 2000	V
	Charged-device model (CDM), per AEC Q100-011	± 2000	
LM2902-Q1			
$V_{(ESD)}$ Electrostatic discharge	Charged-device model (CDM), per AEC Q100-011	± 1500	V

- (1) AEC Q100-002 indicates that HBM stressing shall be in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

6.3 Recommended Operating Conditions

over operating ambient temperature range (unless otherwise noted)

		MIN	MAX	UNIT
V_S Supply voltage, $V_S = ([V+] - [V-])$	LM2902B-Q1, LM2902BA-Q1	3	36	V
	LM2902KV-Q1, LM2902KAV-Q1	3	30	
	LM2902-Q1	3	26	
V_{CM} Common-mode voltage		V-	(V+) - 2	V
T_A Operating ambient temperature		-40	125	$^\circ\text{C}$

6.4 Thermal Information

THERMAL METRIC ⁽¹⁾	LM2902-Q1, LM2902KV-Q1, LM2902KAV-Q1		LM2902B-Q1, LM2902BA-Q1		UNIT
	D (SOIC)	PW (TSSOP)	D (SOIC)	PW (TSSOP)	
	14 PINS	14 PINS	14 PINS	14 PINS	
$R_{\theta JA}$ Junction-to-ambient thermal resistance	101	86	TBD	TBD	°C/W
$R_{\theta JC}$ Junction-to-case (top) thermal resistance	—	—	TBD	TBD	°C/W
$R_{\theta JB}$ Junction-to-board thermal resistance	—	—	TBD	TBD	°C/W
ψ_{JT} Junction-to-top characterization parameter	—	—	TBD	TBD	°C/W
ψ_{JB} Junction-to-board characterization parameter	—	—	TBD	TBD	°C/W

- (1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC package thermal metrics](#) application report.

6.5 Electrical Characteristics - LM2902B-Q1 and LM2902BA-Q1

For $V_S = (V+) - (V-) = 5\text{ V to }36\text{ V}$ ($\pm 2.5\text{ V to } \pm 18\text{ V}$), at $T_A = 25^\circ\text{C}$, $V_{CM} = V_{OUT} = V_S / 2$, and $R_L = 10\text{k}$ connected to $V_S / 2$ (unless otherwise noted)

PARAMETER	TEST CONDITIONS		MIN	TYP	MAX	UNIT	
OFFSET VOLTAGE							
V_{OS}	Input offset voltage	LM2902B-Q1		± 0.3	± 3.0	mV	
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$		± 4.0		
	LM2902BA-Q1		± 0.3	± 2			
		$T_A = -40^\circ\text{C to }125^\circ\text{C}$		2.5			
dV_{OS}/dT	Input offset voltage drift	$R_S = 0\ \Omega$	$T_A = -40^\circ\text{C to }125^\circ\text{C}$	± 7		$\mu\text{V}/^\circ\text{C}$	
PSRR	Input offset voltage versus power supply		65	100		dB	
	Channel separation	$f = 1\text{ kHz to }20\text{ kHz}$		120		dB	
INPUT VOLTAGE RANGE							
V_{CM}	Common-mode voltage range	$V_S = 3\text{ V to }36\text{ V}$		V_-	$(V+) - 1.5$	V	
		$V_S = 5\text{ V to }36\text{ V}$	$T_A = -40^\circ\text{C to }125^\circ\text{C}$	V_-	$(V+) - 2$		
CMRR	Common-mode rejection ratio	$(V_-) \leq V_{CM} \leq (V+) - 1.5\text{ V}$	$V_S = 3\text{ V to }36\text{ V}$	70	80	dB	
		$(V_-) \leq V_{CM} \leq (V+) - 2\text{ V}$	$V_S = 5\text{ V to }36\text{ V}$	$T_A = -40^\circ\text{C to }125^\circ\text{C}$	65		80
INPUT BIAS CURRENT							
I_B	Input bias current			-10	-35	nA	
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$		-50		
dI_{OS}/dT	Input offset current drift		$T_A = -40^\circ\text{C to }125^\circ\text{C}$	10		$\text{pA}/^\circ\text{C}$	
I_{OS}	Input offset current			± 0.5	± 4	nA	
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$		± 5		
dI_{OS}/dT	Input offset current drift		$T_A = -40^\circ\text{C to }125^\circ\text{C}$	10		$\text{pA}/^\circ\text{C}$	
NOISE							
E_N	Input voltage noise	$f = 0.1\text{ to }10\text{ Hz}$		3		μV_{PP}	
e_N	Input voltage noise density	$R_S = 100\ \Omega, V_I = 0\text{ V}, f = 1\text{ kHz}$ (see Figure 8)		35		$\text{nV}/\sqrt{\text{Hz}}$	
INPUT IMPEDANCE							
Z_{ID}	Differential			$10 \parallel 0.1$		$\text{M}\Omega \parallel \text{pF}$	
Z_{ICM}	Common-mode			$4 \parallel 1.5$		$\text{G}\Omega \parallel \text{pF}$	
OPEN-LOOP GAIN							
A_{OL}	Open-loop voltage gain	$V_S = 15\text{ V}, V_O = 1\text{ V to }11\text{ V}, R_L \geq 2\text{ k}\Omega$, connected to (V_-)		50	100	V/mV	
			$T_A = -40^\circ\text{C to }125^\circ\text{C}$	25			
FREQUENCY RESPONSE							
GBW	Gain-bandwidth product	$R_L = 1\text{ M}\Omega, C_L = 20\text{ pF}$ (see Figure 7)		1.2		MHz	
SR	Slew rate	$R_L = 1\text{ M}\Omega, C_L = 30\text{ pF}, V_I = \pm 10\text{ V}$ (see Figure 7)		0.5		V/ μs	
θ_m	Phase margin	$G = +1, R_L = 10\text{k}\Omega, C_L = 20\text{ pF}$		56		$^\circ$	
t_s	Settling time	To 0.1%, $V_S = 5\text{ V}, 2\text{-V Step}, G = +1, C_L = 100\text{ pF}$		4		μs	
	Overload recovery time	$V_{IN} \times \text{gain} > V_S$		10		μs	
THD+N	Total harmonic distortion + noise	$G = +1, f = 1\text{ kHz}, V_O = 3.53\text{ }V_{RMS}, V_S = 36\text{ V}, R_L = 100\text{k}, I_{OUT} \leq 50\ \mu\text{A}, \text{BW} = 80\text{ kHz}$		0.001%			
OUTPUT							
V_O	Voltage output swing from rail	Positive Rail (V+)		$I_{OUT} = -50\ \mu\text{A}$	1.35	1.5	V
V_O				$I_{OUT} = -1\text{ mA}$	1.4	1.6	V
V_O				$I_{OUT} = -5\text{ mA}$	1.5	1.75	V
V_O		Negative Rail (V-)		$I_{OUT} = 50\ \mu\text{A}$	100	150	mV
V_O				$I_{OUT} = 1\text{ mA}$	0.75	1	V
V_O			$V_S = 5\text{ V}, R_L \leq 10\text{ k}\Omega$ connected to (V_-)	$T_A = -40^\circ\text{C to }125^\circ\text{C}$	5	20	mV
I_O	Output current	$V_S = 15\text{ V}; V_O = V_-; V_{ID} = 1\text{ V}$	Source		-20	-30	mA
				$T_A = -40^\circ\text{C to }125^\circ\text{C}$	-10		mA
		$V_S = 15\text{ V}; V_O = V_+; V_{ID} = -1\text{ V}$	Sink		10	20	mA
				$T_A = -40^\circ\text{C to }125^\circ\text{C}$	5		mA
		$V_{ID} = -1\text{ V}; V_O = (V_-) + 200\text{ mV}$		50	85	μA	
I_{SC}	Short-circuit current	$V_S = 20\text{ V}, (V_+) = 10\text{ V}, (V_-) = -10\text{ V}, V_O = 0\text{ V}$		± 40	± 60	mA	
C_{LOAD}	Capacitive load drive			100		pF	

6.5 Electrical Characteristics - LM2902B-Q1 and LM2902BA-Q1 (continued)

For $V_S = (V+) - (V-) = 5\text{ V to }36\text{ V}$ ($\pm 2.5\text{ V to } \pm 18\text{ V}$), at $T_A = 25^\circ\text{C}$, $V_{CM} = V_{OUT} = V_S / 2$, and $R_L = 10\text{k}$ connected to $V_S / 2$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		MIN	TYP	MAX	UNIT
R_O	Open-loop output impedance	$f = 1\text{ MHz}, I_O = 0\text{ A}$			300		Ω
POWER SUPPLY							
I_Q	Quiescent current per amplifier	$V_S = 5\text{ V}; I_O = 0\text{ A}$	$T_A = -40^\circ\text{C to }125^\circ\text{C}$		175	300	μA
		$V_S = 36\text{ V}; I_O = 0\text{ A}$	$T_A = -40^\circ\text{C to }125^\circ\text{C}$		350	750	μA

6.6 Electrical Characteristics: LM2902-Q1, LM2902KV-Q1, LM2902KAV-Q1

For $V_S = (V_+) - (V_-) = 5\text{ V}$, at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS		TA ⁽¹⁾	MIN	TYP ⁽²⁾	MAX	UNIT
V _{IO}	Input offset voltage	V _{CC} = 5 V to 26 V, V _{IC} = V _{ICRmin} , V _O = 1.4 V		25°C		3	7	mV
				Full Range			10	
I _{IO}	Input offset current	V _O = 1.4 V		25°C		2	50	nA
				Full Range			300	
I _{IB}	Input bias current	V _O = 1.4 V		25°C		-20	-250	nA
				Full Range			-500	
V _{ICR}	Common-mode input voltage range	V _{CC} = 5 V to 26 V		25°C	V-	(V+) - 1.5		V
				Full Range		V-	(V+) - 2	
V _{OH}	High-level output voltage	R _L = 10 kΩ		25°C	(V+) - 1.5			V
		V _{CC} = 26 V, R _L = 2 kΩ	Full Range	22				
		V _{CC} = 26 V, R _L ≥ 10 kΩ	Full Range	23	24			
V _{OL}	Low-level output voltage	R _L ≤ 10 kΩ		Full Range		5	20	mV
A _{VD}	Large-signal differential voltage amplification	V _{CC} = 15 V, V _O = 1 V to 11 V, R _L ≥ 2 kΩ		25°C		100		V/mV
				Full Range		15		
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}		25°C	50	80		dB
k _{SVR}	Supply-voltage rejection ratio (ΔV _{CC} / ΔV _{IO})			25°C	50	100		dB
V _{O1} / V _{O2}	Crosstalk attenuation	f = 1 kHz to 20 kHz		25°C		120		dB
I _O	Output current	V _{CC} = 15 V, V _O = 0 V _{ID} = 1 V,		25°C	-20	-30	-60	mA
				Full Range		-10		
		V _{CC} = 15 V, V _O = 15 V V _{ID} = -1 V,		25°C	10	20		
				Full Range		5		
		V _{ID} = -1 V V _O = 200 mV	25°C		30		μA	
I _{OS}	Short-circuit output current	V _{CC} at 5 V, GND at -5 V	V _O = 0	25°C		±40	±60	mA
I _{CC}	Supply current (four amplifiers)	V _O = 2.5 V	No load	Full Range		0.7	1.2	mA
		V _{CC} = 26 V, V _O = 0.5 V _{CC}	No load	Full Range		1.4	3	
V _{IO}	Input offset voltage	V _{CC} = 5 V to 32 V, V _{IC} = V _{ICRmin} , V _O = 1.4 V		Non-A devices	25°C	3	7	mV
					Full Range			
		A-suffix devices	25°C	1	2			
			Full Range			4		
ΔV _{IO} /ΔT	Temperature drift	R _S = 0 Ω		Full Range		7		μV/°C
I _{IO}	Input offset current	V _O = 1.4 V		25°C		2	50	nA
				Full Range			150	
ΔI _{IO} /ΔT	Temperature drift			Full Range		10		pA/°C
I _{IB}	Input bias current	V _O = 1.4 V		25°C		-20	-250	nA
				Full Range			-500	
V _{ICR}	Common-mode input voltage range	V _{CC} = 5 V to 32 V		25°C	0 to V _{CC} - 1.5			V
				Full Range	0 to V _{CC} - 2			
V _{OH}	High-level output voltage	R _L = 10 kΩ		25°C	V _{CC} - 1.5			V
		V _{CC} = 32 V, R _L = 2 kΩ	Full Range	26				
		V _{CC} = 32 V, R _L ≥ 10 kΩ	Full Range	27				
V _{OL}	Low-level output voltage	R _L ≤ 10 kΩ		Full Range		5	20	mV

6.6 Electrical Characteristics: LM2902-Q1, LM2902KV-Q1, LM2902KAV-Q1 (continued)

For $V_S = (V_+) - (V_-) = 5\text{ V}$, at $T_A = 25^\circ\text{C}$ (unless otherwise noted)

PARAMETER		TEST CONDITIONS	TA ⁽¹⁾	MIN	TYP ⁽²⁾	MAX	UNIT
A _{VD}	Large-signal differential voltage amplification	V _{CC} = 15 V, V _O = 1 V to 11 V, R _L ≥ 2 kΩ	25°C	25	100		V/mV
			Full Range	15			
	Amplifier-to-amplifier coupling ⁽³⁾	f = 1 kHz to 20 kHz, input referred	25°C		120		dB
CMRR	Common-mode rejection ratio	V _{IC} = V _{ICRmin}	25°C	60	80		dB
k _{SVR}	Supply-voltage rejection ratio (ΔV _{CC} /ΔV _{IO})		25°C	60	100		dB
V _{O1} /V _{O2}	Crosstalk attenuation	f = 1 kHz to 20 kHz	25°C		120		dB
I _O	Output current	V _{CC} = 15, V _O = 0 V _{ID} = 1 V	25°C	-20	-30	-60	mA
			Full Range	-10			
		V _{CC} = 15, V _O = 15 V V _{ID} = -1 V	25°C	10	20		
			Full Range	5			
V _{ID} = -1 V V _O = 200 mV	25°C	12	40		μA		
I _{OS}	Short-circuit output current	V _{CC} at 5 V, GND at -5 V V _O = 0	25°C		±40	±60	mA
I _{CC}	Supply current (four amplifiers)	V _O = 2.5 V No load	Full Range		0.7	1.2	mA
		V _{CC} = 32 V, V _O = 0.5 V _{CC} No load	Full Range		1.4	3	

(1) Full range is -40°C to 125°C.

(2) All typical values are at T_A = 25°C

(3) Due to proximity of external components, ensure that coupling is not originating via stray capacitance between these external parts. Typically, this can be detected, as this type of coupling increases at higher frequencies.

6.7 Operating Conditions: LM2902-Q1, LM2902KV-Q1, LM2902KAV-Q1

For $V_S = (V_+) - (V_-) = 15\text{ V}$, at $T_A = 25^\circ\text{C}$

PARAMETER		TEST CONDITIONS	TYP	UNIT
SR	Slew rate at unity gain	R _L = 1 MΩ, C _L = 30 pF, V _I = ±10 V (see Figure 7-1)	0.5	V/μs
B ₁	Unity-gain bandwidth	R _L = 1 MΩ, C _L = 20 pF (see Figure 7-1)	1.2	MHz
V _N	Equivalent input noise voltage	R _S = 100 Ω, V _I = 0 V, f = 1 kHz (see Figure 7-2)	35	nV/√Hz

7 Parameter Measurement Information

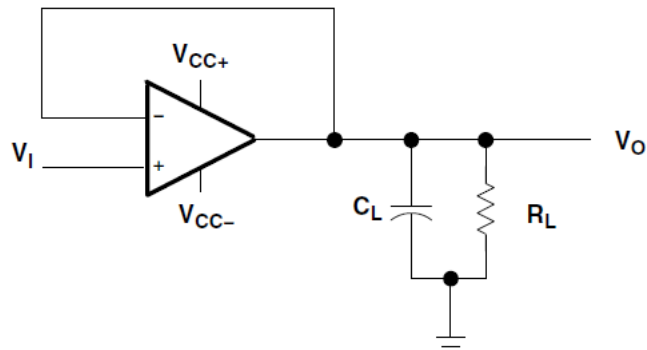


Figure 7-1. Unity-Gain Amplifier

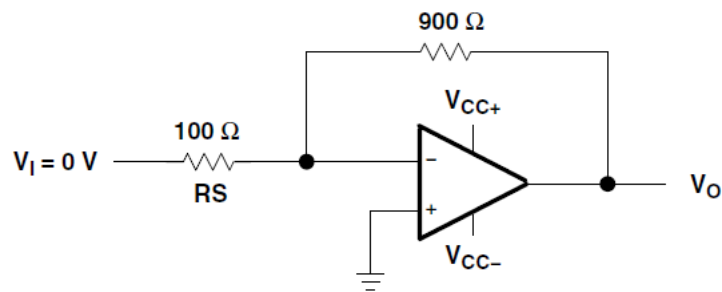


Figure 7-2. Noise-Test Circuit

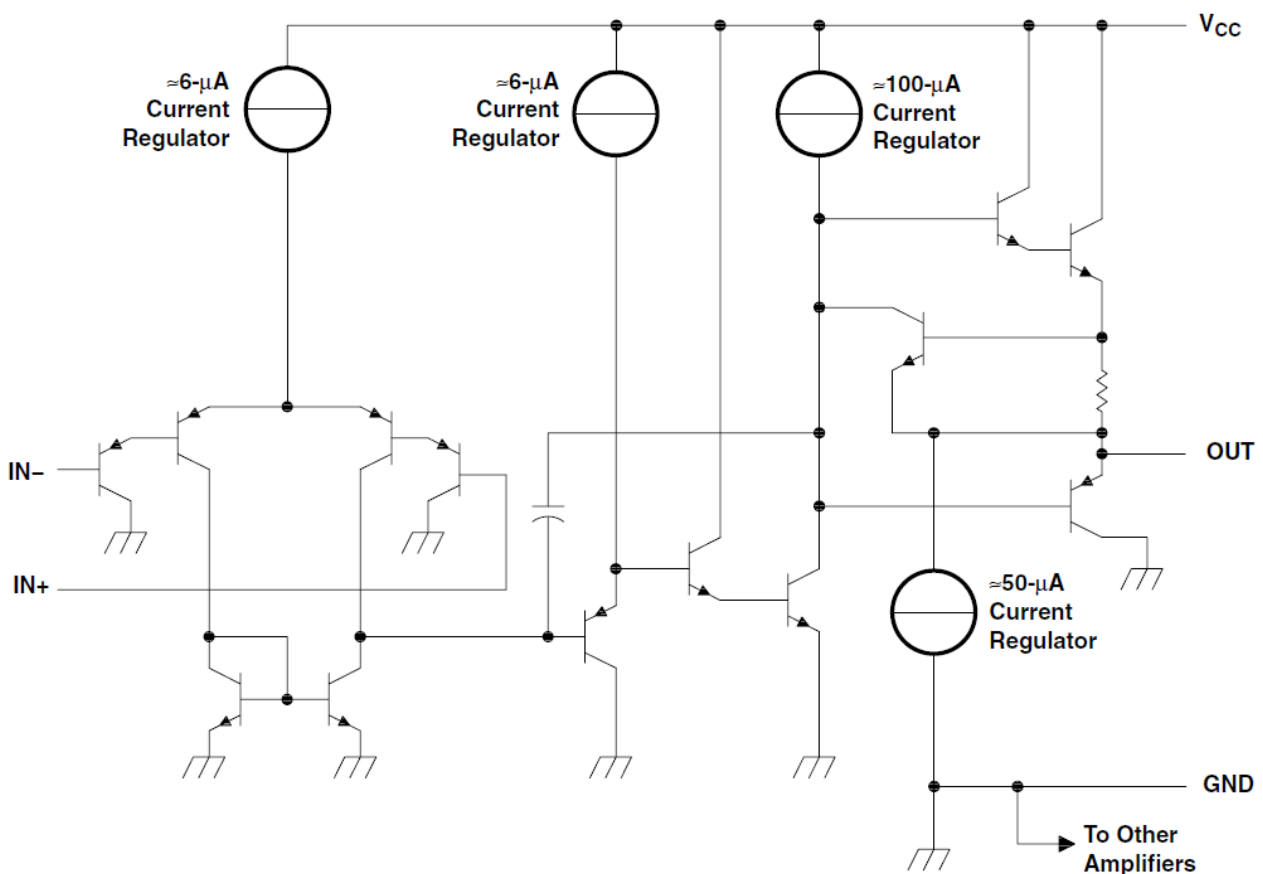
8 Detailed Description

8.1 Overview

The LM2902-Q1, LM2902B-Q1, and LM2902BA-Q1 devices consist of four independent, high-gain frequency-compensated operational amplifiers designed to operate from a single supply over a wide range of voltages. Operation from split supplies also is possible if the difference between the two supplies is within the supply voltage range, and V_S is at least 1.5 V more positive than the input common-mode voltage. The low supply-current drain is independent of the magnitude of the supply voltage.

Applications include transducer amplifiers, DC amplification blocks, and all the conventional operational amplifier circuits that now can be implemented more easily in single-supply-voltage systems. For example, these devices can be operated directly from the standard 5-V supply used in digital systems and easily can provide the required interface electronics without additional ± 5 -V supplies.

8.2 Functional Block Diagram



Schematic (Each Amplifier)

8.3 Feature Description

8.3.1 Input Common Mode Range

The valid common mode range is from device ground to $V_S - 1.5$ V ($V_S - 2$ V across temperature). Inputs may exceed V_S up to the maximum V_S without device damage. At least one input must be in the valid input common-mode range for the output to be the correct phase. If both inputs exceed the valid range, then the output phase is undefined. If either input more than 0.3 V below $V-$ then input current should be limited to 1 mA and the output phase is undefined.

8.4 Device Functional Modes

The LM2902-Q1, LM2902B-Q1, and LM2902BA-Q1 devices are powered on when the supply is connected. This device can be operated as a single-supply operational amplifier or dual-supply amplifier, depending on the application.

9 Application and Implementation

Note

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes, as well as validating and testing their design implementation to confirm system functionality.

9.1 Application Information

The LM2902-Q1, LM2902B-Q1, LM2902BA-Q1 operational amplifiers are useful in a wide range of signal conditioning applications. Inputs can be powered before V_S for flexibility in multiple supply circuits. For full application design guidelines related to this family of devices, please refer to the application report [Application design guidelines for LM324/LM358 devices](#).

9.2 Typical Application

A typical application for an operational amplifier is an inverting amplifier. This amplifier takes a positive voltage on the input, and makes it a negative voltage of the same magnitude. In the same manner, it also makes negative voltages positive.

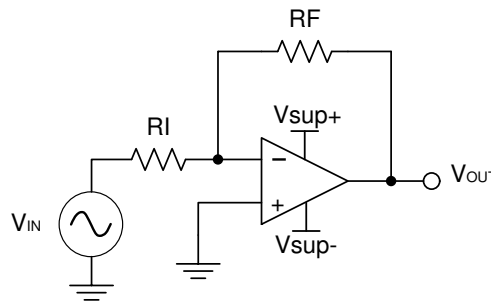


Figure 9-1. Application Schematic

9.3 Design Requirements

The supply voltage must be chosen such that it is larger than the input voltage range and output range. For instance, this application scales a signal of ± 0.5 V to ± 1.8 V. Setting the supply at ± 12 V is sufficient to accommodate this application.

9.4 Detailed Design Procedure

Determine the gain required by the inverting amplifier using [Equation 1](#) and [Equation 2](#):

$$A_V = \frac{V_{OUT}}{V_{IN}} \quad (1)$$

$$A_V = \frac{1.8}{-0.5} = -3.6 \quad (2)$$

Once the desired gain is determined, choose a value for R_I or R_F . Choosing a value in the kilohm range is desirable because the amplifier circuit uses currents in the milliampere range. This ensures the part does not draw too much current. This example uses 10 k Ω for R_I which means 36 k Ω is used for R_F . This was determined by [Equation 3](#).

$$A_V = -\frac{R_F}{R_I} \quad (3)$$

9.5 Application Curve

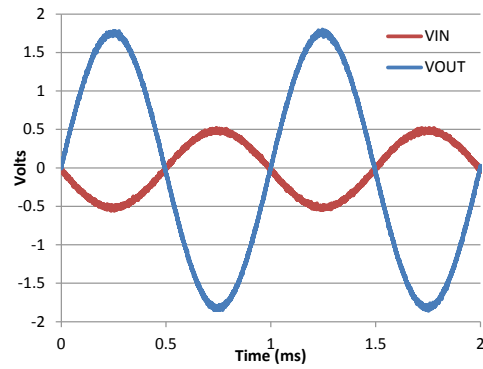


Figure 9-2. Input and Output Voltages of the Inverting Amplifier

10 Power Supply Recommendations

CAUTION

Supply voltages larger than specified in the recommended operating region can permanently damage the device (see [Section 6.1](#)).

Place 0.1- μ F bypass capacitors close to the power-supply pins to reduce errors coupling in from noisy or high-impedance power supplies. For more detailed information on bypass capacitor placement, see [Section 11](#).

11 Layout

11.1 Layout Guidelines

For best operational performance of the device, use good PCB layout practices, including:

- Noise can propagate into analog circuitry through the power pins of the circuit as a whole, as well as the operational amplifier. Bypass capacitors are used to reduce the coupled noise by providing low-impedance power sources local to the analog circuitry.
 - Connect low-ESR, 0.1- μF ceramic bypass capacitors between each supply pin and ground, placed as close to the device as possible. A single bypass capacitor from V_+ to ground is applicable for single-supply applications.
- Separate grounding for analog and digital portions of circuitry is one of the simplest and most-effective methods of noise suppression. One or more layers on multilayer PCBs are usually devoted to ground planes. A ground plane helps distribute heat and reduces EMI noise pickup. Make sure to physically separate digital and analog grounds, paying attention to the flow of the ground current.
- To reduce parasitic coupling, run the input traces as far away from the supply or output traces as possible. If it is not possible to keep them separate, it is much better to cross the sensitive trace perpendicular as opposed to in parallel with the noisy trace.
- Place the external components as close to the device as possible. Keeping R_F and R_G close to the inverting input minimizes parasitic capacitance, as shown in Section 11.2.
- Keep the length of input traces as short as possible. Always remember that the input traces are the most sensitive part of the circuit.
- Consider a driven, low-impedance guard ring around the critical traces. A guard ring can significantly reduce leakage currents from nearby traces that are at different potentials.

11.2 Layout Example

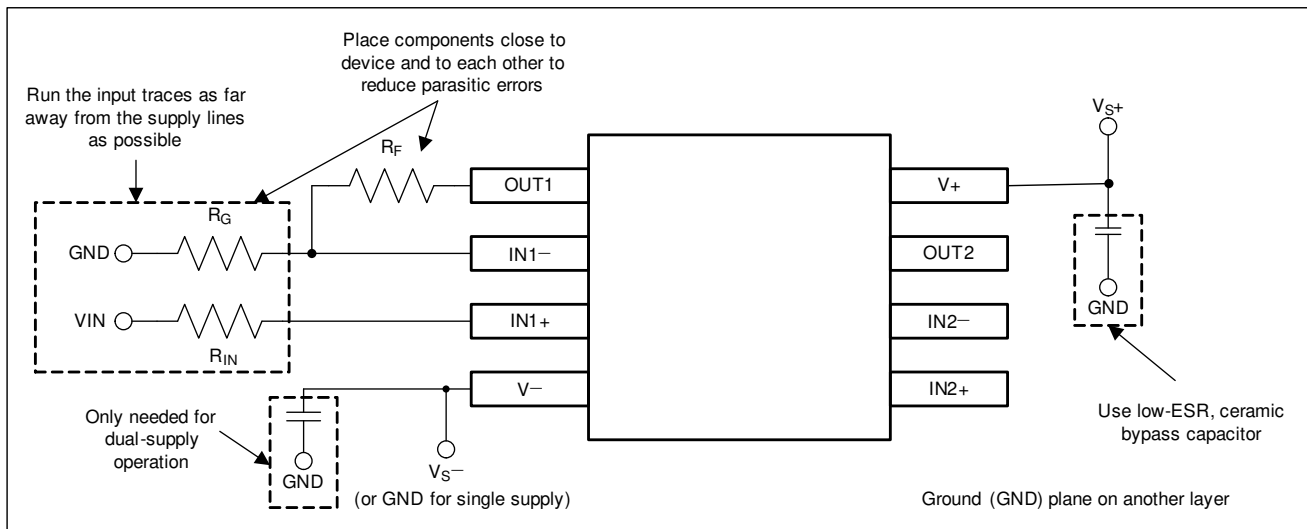


Figure 11-1. Operational Amplifier Board Layout for Noninverting Configuration

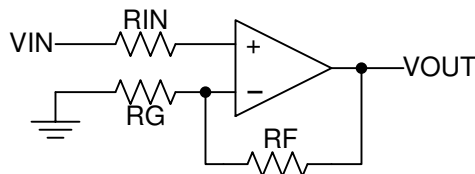


Figure 11-2. Operational Amplifier Schematic for Noninverting Configuration

12 Device and Documentation Support

12.1 Documentation Support

12.1.1 Related Documentation

For related documentation, see the following:

- Texas Instruments, [Application Design Guidelines for LM324/LM358 Devices application note](#)

12.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on [ti.com](#). Click on *Subscribe to updates* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

12.3 Support Resources

[TI E2E™ support forums](#) are an engineer's go-to source for fast, verified answers and design help — straight from the experts. Search existing answers or ask your own question to get the quick design help you need.

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12.4 Trademarks

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12.5 Electrostatic Discharge Caution



This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

12.6 Glossary

[TI Glossary](#) This glossary lists and explains terms, acronyms, and definitions.

13 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LM2902KAVQDRQ1	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KAQ	Samples
LM2902KAVQPWRG4Q1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KAQ	Samples
LM2902KAVQPWRQ1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KAQ	Samples
LM2902KVQDRQ1	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KVQ	Samples
LM2902KVQPWRG4Q1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KVQ	Samples
LM2902KVQPWRQ1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902KVQ	Samples
LM2902QDRG4Q1	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902Q1	Samples
LM2902QDRQ1	ACTIVE	SOIC	D	14	2500	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902Q1	Samples
LM2902QPWRG4Q1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902Q1	Samples
LM2902QPWRQ1	ACTIVE	TSSOP	PW	14	2000	RoHS & Green	NIPDAU	Level-1-260C-UNLIM	-40 to 125	2902Q1	Samples
PLM2902BQPWRQ1	ACTIVE	TSSOP	PW	14	3000	TBD	Call TI	Call TI	-40 to 125		Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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OTHER QUALIFIED VERSIONS OF LM2902-Q1, LM2902B-Q1 :

- Catalog : [LM2902](#), [LM2902B](#)
- Enhanced Product : [LM2902-EP](#)

NOTE: Qualified Version Definitions:

- Catalog - TI's standard catalog product
- Enhanced Product - Supports Defense, Aerospace and Medical Applications

TAPE AND REEL INFORMATION

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LM2902KAVQPWRG4Q1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
LM2902KAVQPWRQ1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
LM2902KVQPWRG4Q1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
LM2902KVQPWRQ1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
LM2902QPWRG4Q1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1
LM2902QPWRQ1	TSSOP	PW	14	2000	330.0	12.4	6.9	5.6	1.6	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LM2902KAVQPWRG4Q1	TSSOP	PW	14	2000	367.0	367.0	35.0
LM2902KAVQPWRQ1	TSSOP	PW	14	2000	367.0	367.0	35.0
LM2902KVQPWRG4Q1	TSSOP	PW	14	2000	367.0	367.0	35.0
LM2902KVQPWRQ1	TSSOP	PW	14	2000	356.0	356.0	35.0
LM2902QPWRG4Q1	TSSOP	PW	14	2000	356.0	356.0	35.0
LM2902QPWRQ1	TSSOP	PW	14	2000	356.0	356.0	35.0

D (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- A. All linear dimensions are in inches (millimeters).
 - B. This drawing is subject to change without notice.
 - $\triangle C$ Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0.006 (0,15) each side.
 - $\triangle D$ Body width does not include interlead flash. Interlead flash shall not exceed 0.017 (0,43) each side.
 - E. Reference JEDEC MS-012 variation AB.

D (R-PDSO-G14)

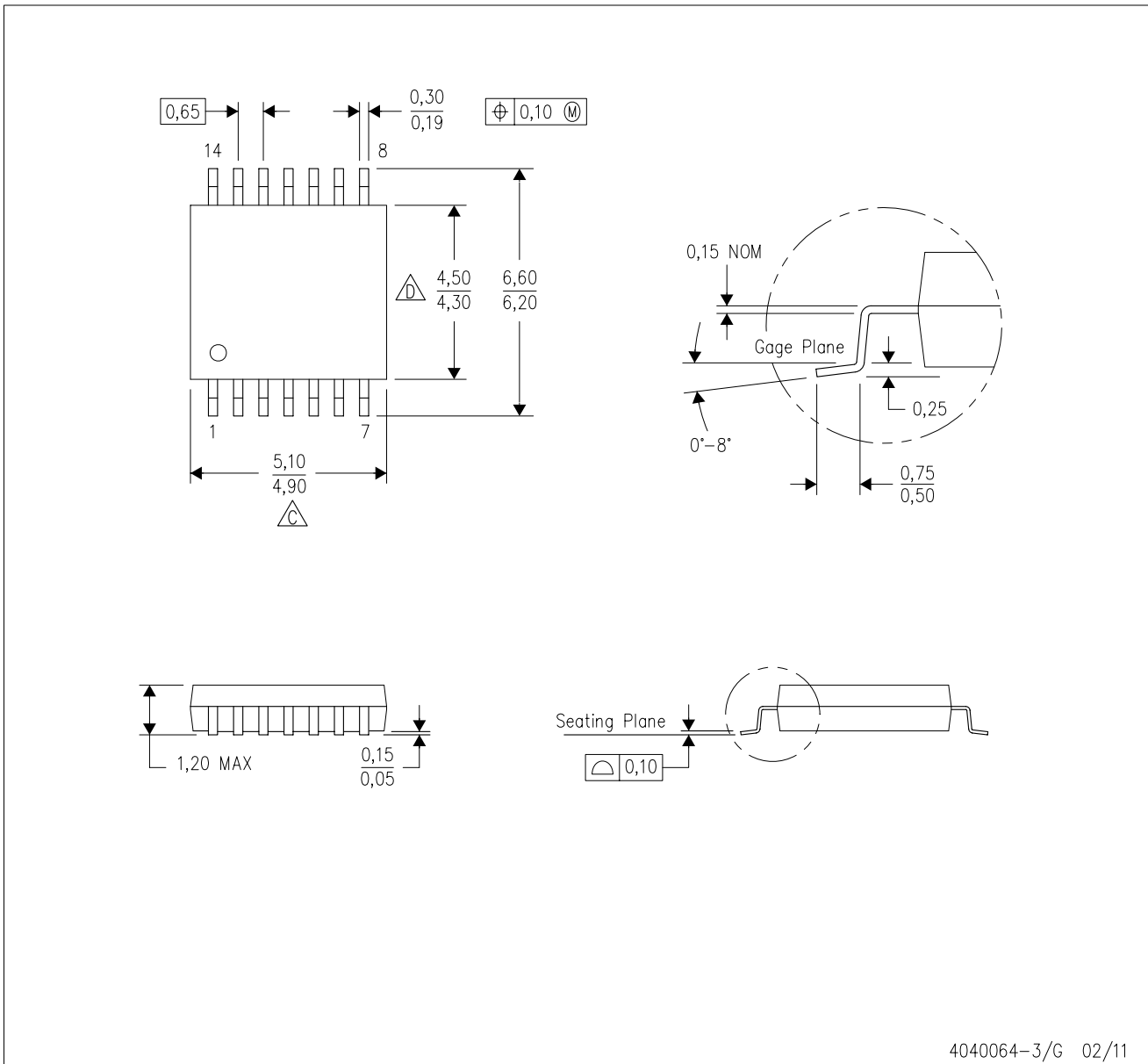
PLASTIC SMALL OUTLINE





- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



4040064-3/G 02/11

- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 -  Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
 -  Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
 - E. Falls within JEDEC MO-153

PW (R-PDSO-G14)

PLASTIC SMALL OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
 - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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