### 18V Operation 3.0A Synchronous Step-Down DC/DC Converters

### ■ GENERAL DESCRIPTION

The XCA204A0K1MR is 18V bootstrap synchronous step-down DC/DC converter with built-in Nch-Nch driver FET, the operating voltage range is 4.5V to 18.0V, designed to allow the use of ceramic capacitors.

0.765V reference voltage source is incorporated in the IC, and the output voltage can be set to a value from 0.8V to 0.65 × VIN or 7.0V (Max) using external resistors (R1, R2).

Switching frequency is 700kHz. In PWM/PFM automatic switchover control, IC can change the control method between PWM and PFM based on the output current requirement and as a result IC can achieve high efficiency over the full load range.

XCA204A0K1MR has a fixed internal soft-start time which is 1.3ms, with the built-in UVLO function, the driver FET is forced OFF when input voltage goes down to 4.1V or lower. Over current protection and thermal shutdown are embedded and they secure a safety operation.

### ■ APPLICATIONS

- Set Top Box
- Portable TV
- LCD TV

• AP Router

### ■FEATURES

Input Voltage Range Output Voltage Range FB Voltage Output Current Oscillation Frequency Control Method Efficiency Internal Power MOSFET Soft-start Time Protection

Low ESR Ceramic Capacitor Operating Ambient Temperature Package Environmentally Friendly

■TYPICAL APPLICATION CIRCUIT





Name	Number	Value	Model number
CIN	2	22uF	TMK212BBJ226MG (TAIYO YUDEN)
CL	2	22uF	LMK212BBJ226MG (TAIYO YUDEN)
L	1	2.2uH	CLF6045NIT-2R2N (TDK)
	1	3.3uH	CLF6045NIT-3R3N (TDK)



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4.5V ~ 18V (Absolute Max 20V)

- : 0.8V ~ 0.65  $\times$  VIN or 7V Max
- : 0.765V ±3.0%
- : 3.0A
- : 700kHz
  - PWM/PFM Automatic Up to 92% High-side : 85mΩ Low-side : 45mΩ 1.3ms
  - : UVLO Current Limit Thermal Shutdown
  - : Ceramic Capacitor
  - -40°C ∼ 85°C
- : SOT23-6
- : EU RoHS Compliant

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### BLOCK DIAGRAM



### ■ PRODUCT CLASSIFICATION

#### Ordering Information

Product Name	Package Name	Order Unit	Shipment
XCA204A0K1MR	SOT23-6	3,000pcs/Reel	Tape & Reel

#### Selection Guide

Chip Enable	Over Voltage Thermal Protection Shutdown		Soft Start	UVLO	Current Limit
YES	YES	YES	YES	YES	YES

### ■ PIN CONFIGURATION



### ■ PIN ASSIGNMENT

PIN NUMBER	PIN NAME	FUNCTION		
1		Ground. This pin is the voltage reference for the regulated output voltage.		
I	0ND	For this reason care must be taken in its layout.		
2	SW	Switch Output. Connect this pin to the switching end of the inductor.		
		Power Supply Input. Drive 4.5V to 18V voltage to this pin to power on this chip.		
3	VIN	Connecting a $10\mu F$ ~ $22\mu F$ ceramic bypass capacitor between VIN and GND to		
		eliminate noise.		
1	ER	Feedback. An external resistor divider from the output to GND, tapped to the FB pin		
4	ГD	sets the output voltage.		
5	EN	ON/OFF Control Input. Pull EN above 1.8V to turn the device on.		
6	PC	Bootstrap. A 100nF capacitor is connected between SW and BS pins to drive		
	62	the power switch's gate above the supply voltage.		

### ■FUNCTION CHART

PIN NAME	SIGNAL	STATUS	
	L or OPEN	Stand-by	
	Н	Active	

### ■ABSOLUTE MAXIMUM RATINGS<sup>(\*1)</sup>

			Ta=25℃
PARAMETER	SYMBOL	RATINGS	UNITS
Input Supply Voltage Pin	V <sub>IN</sub>	-0.3 ~ 20	V
EN Voltage Pin	VEN	-0.3 ~ 20	V
SW Voltage Pin	Vsw	-0.3 ~ 20	V
Boost Voltage Pin	V <sub>BS</sub>	Vsw - 0.3 ~ Vsw + 5.8	V
FB Voltage Pin	V <sub>FB</sub>	-0.3 ~ 5.6	V
Maximum Junction Temperature	Tj	150	°C
Thermal resistance(Junction-Air)	$ heta_{JA}$	88	°C/W
Thermal resistance(Junction-Case)	$\theta_{\rm JC}$	45	°C/W
Operating Ambient Temperature	Topr	-40 ~ 85	°C
Storage Temperature	Tstg	-65 ~ 150	°C

<sup>(\*1)</sup> Stresses exceed those ratings may damage the device.

### ■ELECTRICAL CHARACTERISTICS

						Ta=25°C
PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Input Voltage Range	VIN		4.5	-	18.0	V
Supply Current (Quiescent)	lq	V <sub>EN</sub> = V <sub>IN</sub> , No switching	-	300	-	μA
Supply Current (Shutdown)	ISTBY	V <sub>EN</sub> = 0V	-	3	10	μA
Feedback Voltage	VFB		0.742	0.765	0.788	V
Feedback Current	I <sub>FB</sub>	V <sub>FB</sub> = 1V	-0.1	0	0.1	μA
Switch-On High-Side Resistance (*1)	R <sub>DSH(ON)</sub>		-	85	-	mΩ
Switch-On Low-Side Resistance (*1)	R <sub>DSL(ON)</sub>		-	45	-	mΩ
Switch Leakage	I <sub>SW</sub>	$V_{EN} = 0V, V_{SW} = 0V$	-	-	10	μA
Current Limit (*1)	loc	DC current, V <sub>OUT</sub> =1.8V	3.1	4.0	-	А
Oscillator Frequency	f <sub>sw</sub>	V <sub>OUT</sub> = 5V@1A Load	560	700	880	kHz
Soft-start time <sup>(*1)</sup>	tss	V <sub>FB</sub> = 0V to 0.7V	-	1.3	-	ms
Minimum ON-Time (*1)	ton		-	150	-	ns
Minimum OFF-Time (*1)	toff		-	250	360	ns
Under Voltage Lockout Threshold	VUVLOR	VIN Rising, VOUT = 1V	-	4.1	4.4	V
Under Voltage Lockout Threshold Hysteresis	Vuvloh	V <sub>OUT</sub> = 1V	-	250	-	mV
EN pin resistance to GND	R <sub>EN</sub>		0.7	1.2	1.8	MΩ
EN Up Threshold Voltage	$V_{\text{ENH}}$		1.8	-	18.0	V
EN Down Threshold Voltage	VENL		GND	-	0.4	V
Thermal Shutdown <sup>(*1)</sup>	T <sub>TSD</sub>		-	150	-	°C

Test Condition: Unless otherwise stated,  $V_{IN}$ =9V,  $V_{EN}$ =2V

(\*1) Design reference value.

### ■TYPICAL APPLICATION CIRCUIT



Figure.1 Typical Application Circuit

#### [Table1 : Recommended Component Selection]

Vout	R1	R2	CFB	L1	C1	C2	C3+C4 <sup>(*)</sup>	C5	C6
7.0\/	81.5kO	1040	Ontional	4 7.14	22µF/25V	0.1µF/25V	68µF	0.1µF	0.1µF
7.00	01.3K12	10K12	Optional	4.7µ11	Ceramic	Ceramic	Ceramic	Ceramic	Ceramic
E 0)/		1040	Ontional	2 2011	22µF/25V	0.1µF/25V	44~68µF	0.1µF	0.1µF
5.00	04.9K12	TUK12	Optional	а з.зµ⊓	Ceramic	Ceramic	Ceramic	Ceramic	Ceramic
2 2)/	22.240	1010	Ontional	3.3µH	22µF/25V	0.1µF/25V	44~68µF	0.1µF	0.1µF
3.3V	33.ZK12	10K12	Optional		Ceramic	Ceramic	Ceramic	Ceramic	Ceramic
1 0\/	12 6kO	1040	Ontional	2 2011	22µF/25V	0.1µF/25V	44~68µF	0.1µF	0.1µF
1.0V	13.0612	10K12	Optional	2.2µH	Ceramic	Ceramic	Ceramic	Ceramic	Ceramic
1.0V	2.0040	3.09kΩ 10kΩ	Optional	1 5.11	22µF/25V	0.1µF/25V	44~68µF	0.1µF	0.1µF
	3.09KD			i.oµ⊓	Ceramic	Ceramic	Ceramic	Ceramic	Ceramic

 $^{(*)}$ V<sub>OUT</sub>=0.8V ~ 5.0V : (C3+C4) = 44µF to 68µF (Ceramic)  $V_{OUT}=5.1V \sim 7.0V$  : (C3+C4) = 68µF (Ceramic)

#### [Table2 : Parts example for evaluation]

Name	Number	Value	Parts number	Vender
		1.5µH	CLF6045NIT-1R5N	TDK
11	(*1)	2.2µH	CLF6045NIT-2R2N	ТDК
		3.3µH	CLF6045NIT-3R3N	ТDК
		4.7µH	CLF6045NIT-4R7N	ТDК
C1	1	22µF	TMK212BBJ226MG	TAIYO YUDEN
C2	1	0.1µF	TMK105BJ104KP	TAIYO YUDEN
C2 C4	(*2)	22µF	LMK212BBJ226MG	TAIYO YUDEN
03, 04		68µF	C3216X5R1A686M160AC	TDK
C5	1	0.1µF	UMK105B7104KV	TAIYO YUDEN
C6	1	0.1µF	UMK105B7104KV	TAIYO YUDEN

 $^{(*1)}$  Select according to  $V_{OUT}$  setting voltage.  $^{(*2)}$  C3+C4=44µF : Two 22µF used in parallel

### ■Function Description

The main control loop of XCA204A0K1MR are adaptive on-time pulse width modulation (PWM) controller. The control mechanism combines adaptive on-time control with an internal compensation circuit for pseudo-fixed frequency and low external component count configuration with both low ESR and ceramic output capacitors. It is stable even with virtually no ripple at the output.

At the beginning of each cycle, the high-side MOSFET is turned on. This MOSFET is turned off after internal one shot timer expires. This one shot duration is set proportional to the converter input voltage, VIN, and inversely proportional to the output voltage, V<sub>OUT</sub>, to maintain a pseudo-fixed frequency over the input voltage range, hence it is called adaptive on-time control. The one-shot timer is reset and the high-side MOSFET is turned on again when the feedback voltage falls below the reference voltage.

#### < Enable >

The XCA204A0K1MR EN pin provides digital control to turn on/turn off the regulator. When the voltage of EN exceeds the threshold voltage, the regulator starts the soft start function. If the EN pin voltage is below than the shutdown threshold voltage, the regulator will be disable and into the shutdown mode.

#### < Output Over Voltage Protection >

When the FB pin voltage exceeds 25% of the regulation voltage, the output over voltage protection function will turn the high side MOSFET off.

#### < Input Under Voltage Lockout (UVLO) >

When the XCA204A0K1MR power on, the internal circuits are held inactive until VIN exceeds the input UVLO threshold voltage. And the regulator will be disabled when VIN below the input UVLO threshold voltage. The hysteretic of the UVLO comparator is 250mV.

#### < Short Protection >

The XCA204A0K1MR provides short protection function to prevent the device damage from short condition. When the output short to ground, the oscillator frequency is reduced to prevent the inductor current increasing beyond the current limit. In the meantime, the current limit is also reduced to lower the short current. Once the short condition is removed, the frequency and current limit will return to normal.

#### < Thermal Shutdown >

The XCA204A0K1MR incorporates an over temperature protection circuit to protect itself from overheating. When the junction temperature exceeds the thermal shutdown threshold temperature, the regulator will be shutdown.

### ■Application Information

#### < Output Voltage Setting >

The external resistor divider is used to set the output voltage. XCA204A0K1MR feedback resistors are unconcerned of compensation and provide an easy way to program output voltage. Table 1 shows a list of resistor selection for common output voltages:

$$V_{OUT} = 0.765 V x (1 + R1/R2)$$

< Selecting the Inductor >

A 4.7 $\mu$ H inductor with a DC current rating of at least 30% percent higher than the maximum load current is recommended for most applications. For highest efficiency, the inductor's DC resistance should be less than 50m $\Omega$ . For most designs, the required inductance value can be derived from the following equation:

$$\Delta I = 0.3 \times I_{L(MAX)}$$
$$L \ge \frac{VOUT}{fsw \cdot \Delta I} \cdot \left(1 - \frac{VOUT}{VIN}\right)$$

Where  $\Delta I$  is the inductor ripple current.

Choose the inductor ripple current to be 30% of the maximum load current. The maximum inductor peak current is calculated from:

$$I_{L(MAX)} = I_{LOAD(MAX)} + \frac{\Delta I}{2}$$

But when a coil with poor DC superimposition(DC bias) characteristics is used, the inductor peak current may increase due to the derating of the inductor during soft start. In this case, please use a coil with better DC superimposition (DC bias) characteristics.

Under light load conditions below 100mA, a larger inductance is recommended for improved efficiency.

#### < Selecting the Input Capacitor >

The input capacitor reduces the surge current drawn from the input supply and the switching noise from the device. The input capacitor impedance at the switching frequency should be less than the input source impedance to prevent high frequency switching current from passing through the input. Ceramic capacitors with X5R or X7R dielectrics are highly recommended because of their low ESR and small temperature coefficients. For most applications, a 22µF capacitor is sufficient.

#### < Selecting the Output Capacitor >

The output capacitor keeps the output voltage ripple small and one or two 22µF ceramic capacitor with X5R or X7R dielectrics is recommended for its low ESR characteristics.

#### < External Boost Diode Selection >

An external bootstrap diode is recommended if the input voltage is less than 5.5V, or duty cycle is high, or if there is a 5V system rail available. This diode helps improve the efficiency. Low cost diodes, such as B0520 are suitable for this application.



### ■Note on use

1. When the VIN voltage is lower than UVLO and the temperature is low, there is a possibility of step-down operation.

- 2. When the input voltage is low, the maximum load current may decrease.
- 3. Be especially careful of the capacitor characteristics and use X7R or X5R (EIA standard) ceramic capacitors. The capacitance decrease caused by the bias voltage may become remarkable depending on the external size of the capacitor.
- 4. Even within the operating temperature range, the thermal shutdown function may operate depending on the input voltage, output voltage and output current. An example of derating characteristics is shown below. Please evaluate IC well on your PCB.



#### < PCB Layout Recommendation >

The device's performance and stability is dramatically affected by PCB layout. It is recommended to follow these general guidelines show bellow:

- 1. Place the input capacitors, output capacitors as close to the device as possible. Trace to these capacitors should be as short and wide as possible to minimize parasitic inductance and resistance.
- 2. CIN must be closes to Pins VIN and GND. The loop area formed by CIN and VIN/GND pins must be minimized.
- 3. The GND trace between the output capacitor and the GND pin should be as wide as possible to minimize its race impedance.
- 4. Do not allow switching current to flow under the device.
- 5. Place feedback resistors close to the FB pin. A separate V<sub>OUT</sub> path should be connected to the upper feedback resistor
- 6. Keep the sensitive signal (FB) away from the switching signal (SW). The trace of the FB node should be as small as possible to avoid noise coupling.
- 7. Multi-layer PCB design is recommended.







Bottom Layer (4)





Dimensions : 40mm×45mm (4 layer) Copper thickness : 1oz Material : Glass Epoxy (FR-4) Thickness : 1.6mm

### ■TYPICAL PERFORMANCE CHARACTERISTICS

C1=22 $\mu$ F, C2=0.1 $\mu$ F, C3=C4=22 $\mu$ F, C5=C6=0.1 $\mu$ F, Ta = 25 $^{\circ}$ C, unless otherwise noted.



### ■PACKAGING INFORMATION

•SOT23-6 (unit:mm)





(PCB FOOTPRINT)





			Unit:mm
SYMBOLS	Min	Тур	Max
А	-	-	1.45
A1	0	-	0.15
A2	0.90	1.10	1.30
b	0.39	-	0.49
b1	0.38	0.4	0.45
С	0.12	-	0.19
c1	0.11	0.13	0.15
D	2.85	2.95	3.05
E	2.60	2.80	3.00
E1	1.55	1.65	1.75
е	0.85	0.95	1.05
e1	1.80	1.90	2.00
L	0.35	0.45	0.55
L1		0.59 REF	
L2		0.25 BSC	
R	0.05	-	-
R1	0.05	-	0.2
Θ	0°	-	8°
Θ1	8°	10°	12°
Θ2	8°	10°	12°
<b>a</b> aa		0.1	

#### NOTES:

- 1. JEDEC OUTLINE: N/A
- 2. DIMENSIONS "D" DOES NOT INCLUDE MOLD FLASH, PROTRUSIONS OR GATE BURRS.MOLD FLASH, PROTRUSIONS AND GATE BURRS SHALL NOT EXCEED .15mm (.006in) PER SIDE.
- 3. DIMENSIONS "E" DOES NOT INCLUDE INTER-LEAD FLASH, OR PROTRUSIONS. INTER-LEAD FLASH AND PROTRUSIONS SHALL NOT EXCEED .25mm (.010in) PER SIDE.

### ■MARKING RULE



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