

# XCL225/XCL226 Series

## 0.5A Inductor Built-in Step-down “micro DC/DC” Converter

ETR28014-004

### ■ GENERAL DESCRIPTION

The XCL225/XCL226 series are 18V operation synchronous step-down DC/DC converter ICs with a built-in high-side / low-side driver transistor. The XCL225/XCL226 series has operating voltage range of 3.0V~18.0V and it can support 500mA as an output current with high-efficiency. Compatible with Low ESR capacitors such as ceramic capacitors for the load capacitor ( $C_L$ ).

0.75V reference voltage source is incorporated in the IC, and the output voltage can be set to a value from 1.0V to 15.0V using external resistors ( $R_{FB1}$ ,  $R_{FB2}$ ).

1.2MHz can be selected for the switching frequency. In PWM/PFM automatic switchover control, IC can change the control method between PWM and PFM based on the output current requirement and as a result IC can achieve high efficiency over the full load range.

XCL225/XCL226 has a fixed internal soft start time which is 1.0ms (TYP.), additionally the time can be extended by using an external resistor and capacitor.

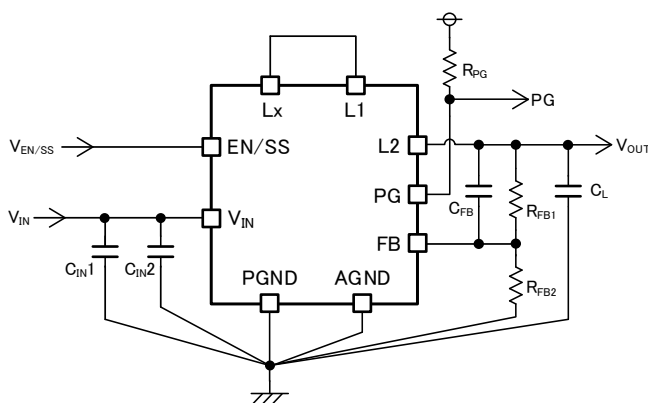
### ■ APPLICATIONS

- Hot water supply system
- Recorders, Camcorders
- Refrigerators, Air-conditioners
- Low Power Systems

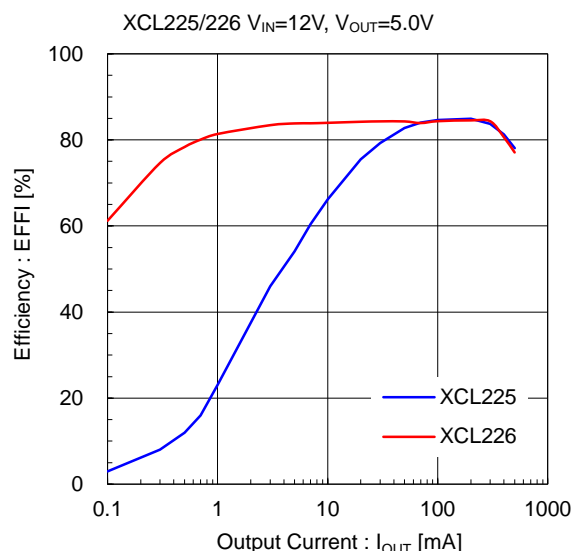
### ■ FEATURES

Input Voltage Range	: 3 ~ 18V (Absolute Max 20V)
FB Voltage	: 0.75V±1.5%
Oscillation Frequency	: 1.2MHz
Output Current	: 500mA
Control Methods	: PWM control PWM/PFM Automatic Efficiency 85%@12V→5V, 1mA
Soft-start Time	: Adjustable by RC
Protection Circuits	: Over Current Protection Integral Latch Method (XCL225A/XCL226A) Automatic Recovery (XCL225B/XCL226B) Thermal Shutdown
Package	: DFN3030-10B
Environmentally Friendly	: EU RoHS Compliant, Pb Free

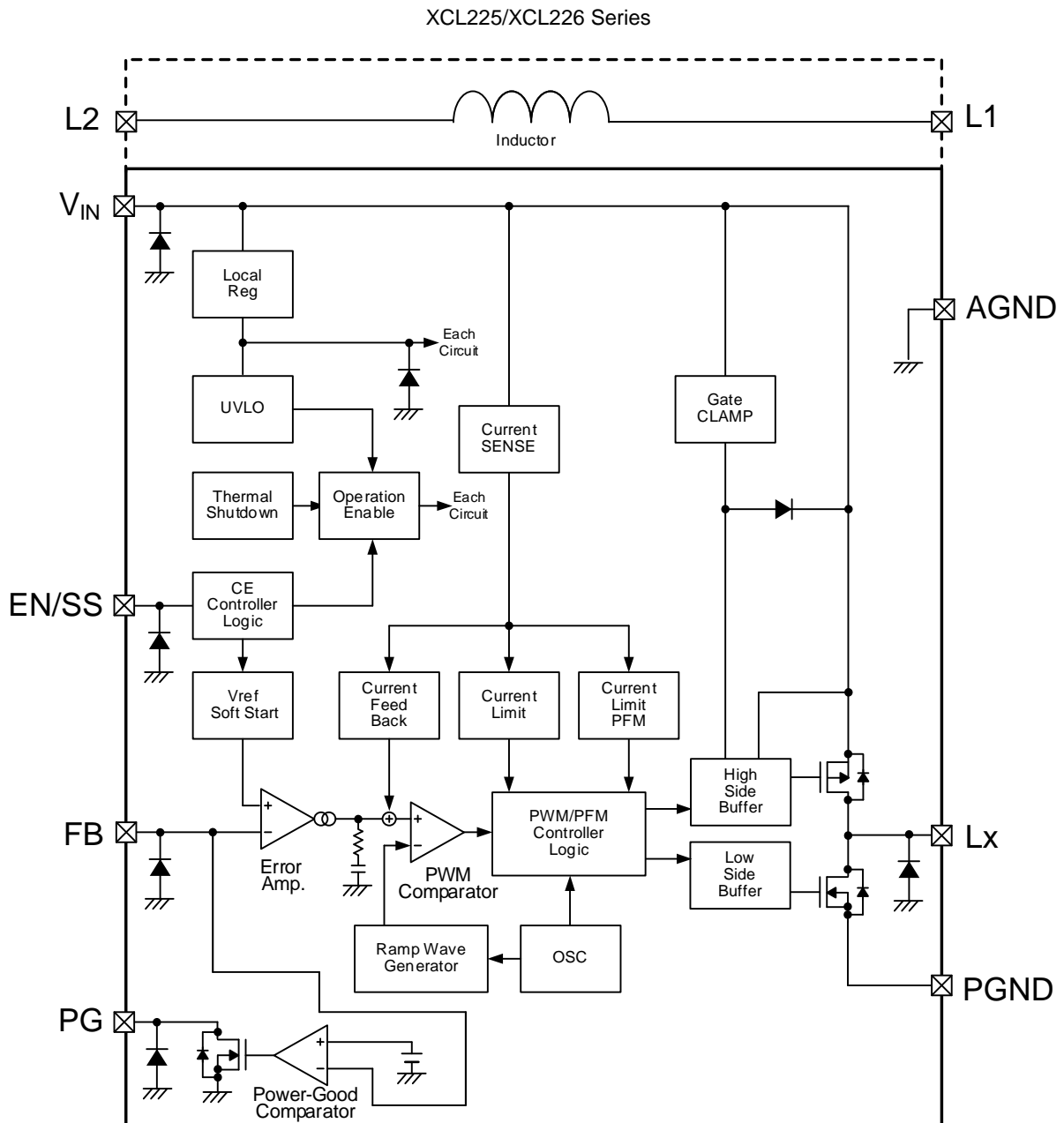
### ■ TYPICAL APPLICATION CIRCUIT



### ■ TYPICAL PERFORMANCE CHARACTERISTICS



## ■ BLOCK DIAGRAM



The XCL225 series is fixed to PWM control.

The XCL226 series is PWM / PFM automatic switching control.

Diodes inside the circuit are an ESD protection diodes and a parasitic diodes.

## ■ PRODUCT CLASSIFICATION

### ● Ordering Information

XCL225①②③④⑤⑥ PWM control

XCL226①②③④⑤⑥ PWM/PFM Automatic

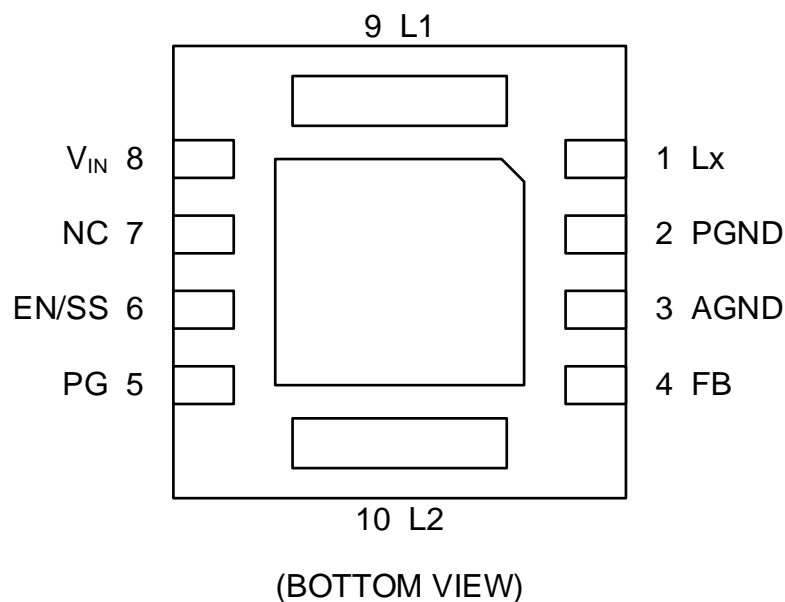
DESIGNATOR	ITEM	SYMBOL	DESCRIPTION
①	Type	A	Refer to Selection Guide
		B	
②③	FB Voltage	0K	ADJ=0.75V. Output voltage can be adjusted in 1V to 15V
④	Oscillation Frequency	1	1.2MHz
⑤⑥	Packages (Order Unit)	H2	DFN3030-10B(3000pcs/Reel)

### ● Selection Guide

FUNCTION	A TYPE	B TYPE
Chip Enable	Yes	Yes
UVLO	Yes	Yes
Thermal Shutdown	Yes	Yes
Soft Start	Yes	Yes
Power-Good	Yes	Yes
Current Limiter(Automatic Recovery)	-	Yes
Current Limiter(Latch Protection) <sup>(*)</sup>	Yes	-

<sup>(\*)</sup> The over-current protection latch is an integral latch type.

## ■ PIN CONFIGURATION



\* The dissipation pad for the DFN3030-10B package should be solder-plated in recommended mount pattern and metal masking so as to enhance mounting strength and heat release. If the pad needs to be connected to other pins, it should be connected to the GND (No. 2 and No.3) pin.

## PIN ASSIGNMENT

PIN NUMBER	PIN NAME	FUNCTION
1	Lx	Switching Output
2	PGND	Power Ground
3	AGND	Analog Ground
4	FB	Output Voltage Sense
5	PG	Power-good Output
6	EN/SS	Enable Soft-start
7	NC	No Connection
8	V <sub>IN</sub>	Power Input
9	L1	Inductor Electrodes
10	L2	Inductor Electrodes

\* This 7-pin NC terminal is not connected to an IC chip.

## FUNCTION CHART

PIN NAME	SIGNAL	STATUS
EN/SS	L	Stand-by
	H	Active
	OPEN	Undefined State <sup>(*)</sup>

<sup>(\*)</sup> Please do not leave the EN/SS pin open. Each should have a certain voltage

PIN NAME	CONDITIONS	SIGNAL
PG	V <sub>FB</sub> > V <sub>PGDET</sub>	H (High impedance)
	V <sub>FB</sub> ≤ V <sub>PGDET</sub>	L (Low impedance)
	Thermal Shutdown	L (Low impedance)
	UVLO (V <sub>IN</sub> < V <sub>UVLO1</sub> )	Undefined State
	EN/SS=L	Stand-by

## ABSOLUTE MAXIMUM RATINGS

T<sub>a</sub>=25°C

PARAMETER	SYMBOL	RATINGS	UNITS
V <sub>IN</sub> Pin Voltage	V <sub>IN</sub>	-0.3 ~ +20	V
EN/SS Pin Voltage	V <sub>EN/SS</sub>	-0.3 ~ +20	V
FB Pin Voltage	V <sub>FB</sub>	-0.3 ~ +6.2	V
PG Pin Voltage	V <sub>PG</sub>	-0.3 ~ +6.2	V
PG Pin Current	I <sub>PG</sub>	8	mA
Lx Pin Voltage	V <sub>Lx</sub>	-0.3 ~ V <sub>IN</sub> +0.3 or +20 <sup>(*)</sup>	V
Lx Pin Current	I <sub>Lx</sub>	1800	mA
Power Dissipation	P <sub>d</sub>	1950(JESD51-7 board) <sup>(**)</sup>	mW
Operating Ambient Temperature	T <sub>opr</sub>	-40 ~ +105	°C
Storage Temperature	T <sub>stg</sub>	-55 ~ +125	°C

\* All voltages are described based on the GND (AGND,PGND) pin.

<sup>(\*)</sup> The maximum value should be either V<sub>IN</sub>+0.3 or 20 in the lowest.

<sup>(\*\*)</sup> The power dissipation figure shown is PCB mounted and is for reference only.

Please refer to PACKAGING INFORMATION for the mounting condition.

## ELECTRICAL CHARACTERISTICS

XCL225/XCL226 Series

Ta=25°C

PARAMETER	SYMBOL	CONDITIONS	MIN.	TYP.	MAX.	UNIT	CIRCUIT
FB Voltage	V <sub>FB</sub>	V <sub>FB</sub> =0.739V→0.761V, V <sub>FB</sub> Voltage when Lx pin oscillates	0.739	0.750	0.761	V	②
Output Voltage Setting Range <sup>(*)</sup>	V <sub>OUTSET</sub>	-	1	-	15	V	-
Operating Voltage Range	V <sub>IN</sub>	-	3	-	18	V	-
UVLO Detect Voltage	V <sub>UVLO1</sub>	V <sub>IN</sub> :2.8V→2.6V, V <sub>FB</sub> =0.675V, V <sub>IN</sub> Voltage when Lx pin voltage changes from "H" level to "L" level	2.60	2.70	2.80	V	②
UVLO Release Voltage	V <sub>UVLO2</sub>	V <sub>IN</sub> :2.7V→2.9V, V <sub>FB</sub> =0.675V, V <sub>IN</sub> Voltage when Lx pin voltage changes from "L" level to "H" level	2.70	2.80	2.90	V	②
Quiescent Current(XCL226)	I <sub>q</sub>	V <sub>FB</sub> =0.825V	-	12.5	17.5	μA	④
Stand-by Current	I <sub>STB</sub>	V <sub>EN/SS</sub> =0V	-	1.65	2.5	μA	⑤
Oscillation Frequency	f <sub>OSC</sub>	Connected to external components, I <sub>OUT</sub> =100mA	1098	1200	1302	kHz	①
Minimum Duty Cycle	D <sub>MIN</sub>	V <sub>FB</sub> =0.825V	-	-	0	%	②
Maximum Duty Cycle	D <sub>MAX</sub>	V <sub>FB</sub> =0.675V	100	-	-	%	②
Lx SW "H" On Resistance	R <sub>LXH</sub>	V <sub>FB</sub> =0.675V, I <sub>LX</sub> =200mA	-	0.99	1.14	Ω	②
Lx SW "L" On Resistance	R <sub>LXL</sub>	V <sub>FB</sub> =0.825V, I <sub>LX</sub> =200mA	-	0.73 <sup>(*)</sup>	-	Ω	②
PFM Switch Current(XCL226)	I <sub>PFM</sub>	Connected to external components, I <sub>OUT</sub> =1mA	-	420	-	mA	①
High side Current Limit <sup>(*)</sup>	I <sub>LIMH</sub>	V <sub>FB</sub> =0.675V	920	1100	-	mA	②
Latch Time	t <sub>LAT</sub>	Type A only, Connected to external components, V <sub>FB</sub> =0V	0.5	1.0	1.7	ms	⑥
Internal Soft-Start Time	t <sub>SS1</sub>	V <sub>EN/SS</sub> =0V→12V, V <sub>FB</sub> =0.675V, Time until Lx pin oscillates	0.5	1.0	1.7	ms	②
External Soft-Start Time	t <sub>SS2</sub>	V <sub>EN/SS</sub> =0V→12V, V <sub>FB</sub> =0.675V, R <sub>SS</sub> =430KΩ, C <sub>SS</sub> =0.47μF Time until Lx pin oscillates	17	26	35	ms	③
PG Detect Voltage	V <sub>PGDET</sub>	V <sub>FB</sub> =0.712V→0.638V, R <sub>PG</sub> :100kΩ pull-up to 5V, V <sub>FB</sub> Voltage when PG pin voltage changes from "H" level to "L" level	0.638	0.675	0.712	V	②
PG Output Voltage	V <sub>PG</sub>	V <sub>FB</sub> =0.6V, I <sub>PG</sub> =1mA	AGND	-	0.3	V	②
Efficiency <sup>(*)</sup>	EFFI	Connected to external components, V <sub>OUT</sub> =5V, I <sub>OUT</sub> =1mA	-	82 <sup>(*)</sup>	-	%	①
FB Voltage Temperature Characteristics	ΔV <sub>FB</sub> / (ΔT <sub>opr</sub> ·V <sub>FB</sub> )	-40°C ≤ T <sub>opr</sub> ≤ 105°C	-	±100	-	ppm /°C	②
FB 'H' Current	I <sub>FBH</sub>	V <sub>IN</sub> =V <sub>EN/SS</sub> =18V, V <sub>FB</sub> =3.0V	-0.1	-	0.1	μA	④
FB 'L' Current	I <sub>FBL</sub>	V <sub>IN</sub> =V <sub>EN/SS</sub> =18V, V <sub>FB</sub> =0V	-0.1	-	0.1	μA	④
EN/SS 'H' Current	I <sub>EN/SSH</sub>	V <sub>IN</sub> =V <sub>EN/SS</sub> =18V, V <sub>FB</sub> =0.825V	-	0.1	0.3	μA	④
EN/SS 'L' Current	I <sub>EN/SSL</sub>	V <sub>IN</sub> =18V, V <sub>EN/SS</sub> =0V, V <sub>FB</sub> =0.825V	-0.1	-	0.1	μA	④
EN/SS 'H' Voltage	V <sub>EN/SSH</sub>	V <sub>EN/SS</sub> =0.3V→2.5V, V <sub>FB</sub> =0.71V, V <sub>EN/SS</sub> Voltage when Lx pin voltage changes from "L" level to "H"	2.5	-	18.0	V	②
EN/SS 'L' Voltage	V <sub>EN/SSL</sub>	V <sub>EN/SS</sub> =2.5V→0.3V, V <sub>FB</sub> =0.71V, V <sub>EN/SS</sub> Voltage when Lx pin voltage changes from "H" level to "L"	AGND	-	0.3	V	②
Thermal Shutdown Temperature	T <sub>TSD</sub>	Junction Temperature	-	150	-	°C	-
Hysteresis Width	T <sub>HYS</sub>	Junction Temperature	-	25	-	°C	-
Inductance	L	Test Freq.=1MHz	-	4.3	-	μH	-
Inductor Rated Current	I <sub>DC</sub>	ΔT=+40°C	-	1200	-	mA	-

Test Condition: Unless otherwise stated, V<sub>IN</sub>=12V, V<sub>EN/SS</sub>=12V

(\*) : Please use within the range of V<sub>OUT</sub>/V<sub>IN</sub> ≥ 0.14

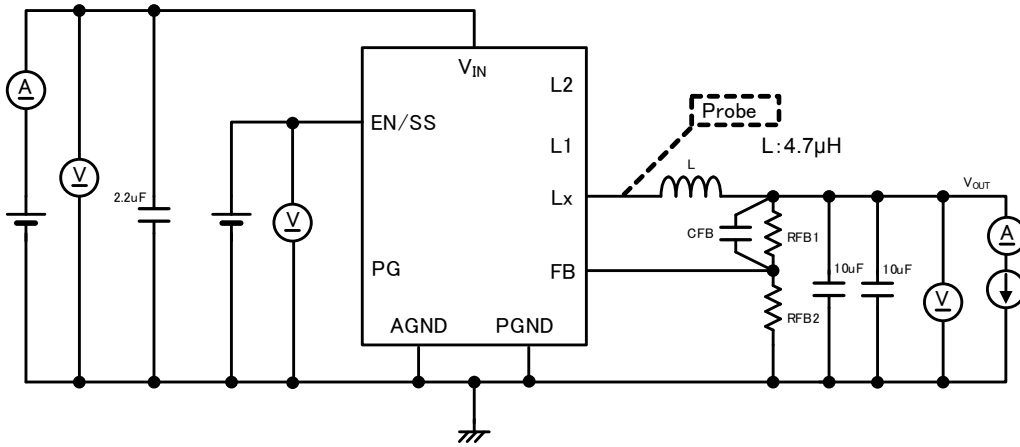
(\*) : Design reference value. This parameter is provided only for reference.

(\*) : Current limit denotes the level of detection at peak of coil current.

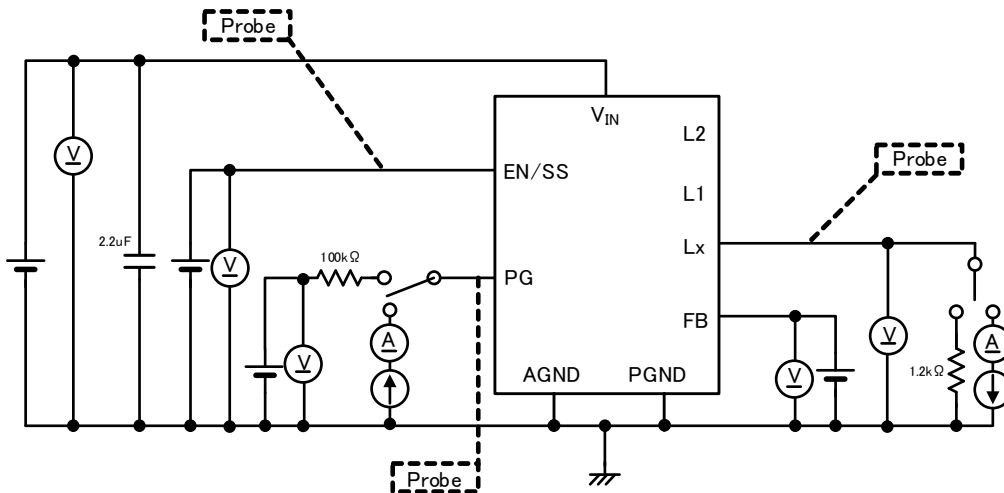
(\*) : EFFI = {(output voltage) x (output current)} / {(input voltage) x (input current)} x 100

## TEST CIRCUITS

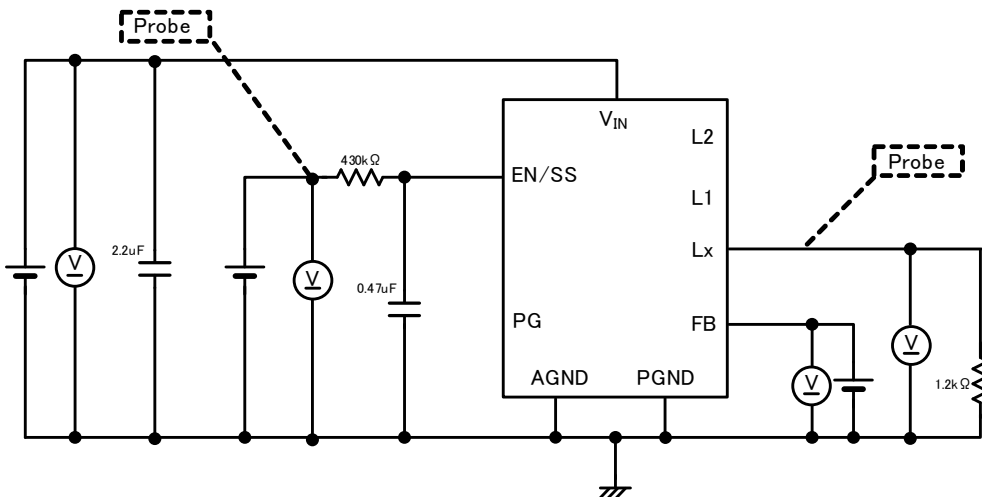
Circuit No. ①



Circuit No. ②

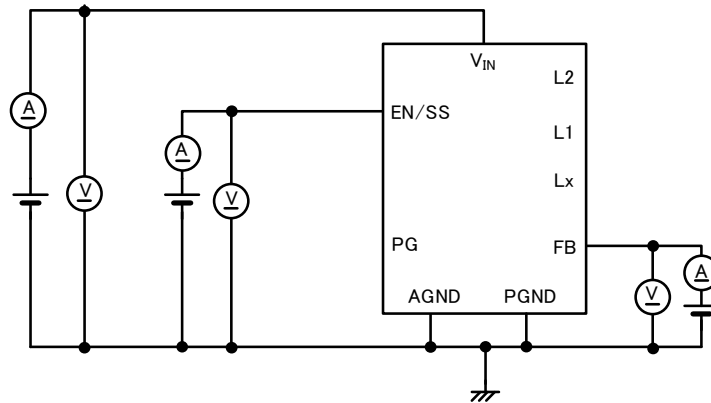


Circuit No. ③

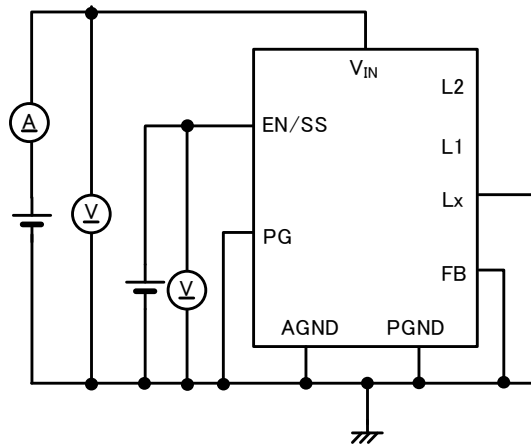


## TEST CIRCUITS

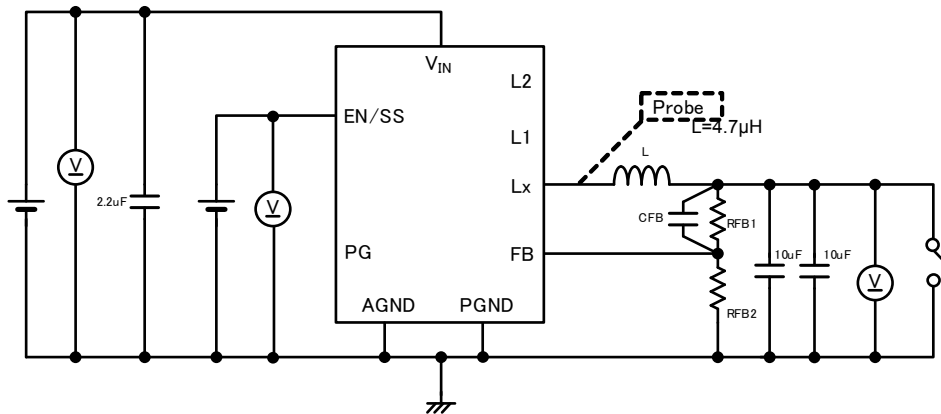
Circuit No.④



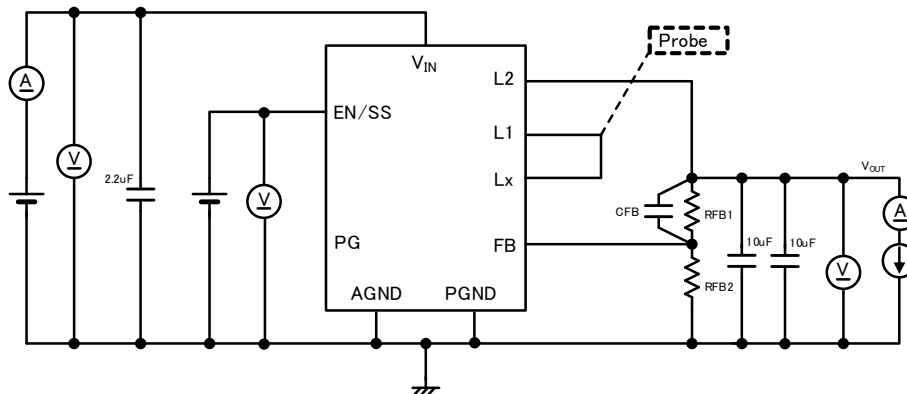
Circuit No.⑤



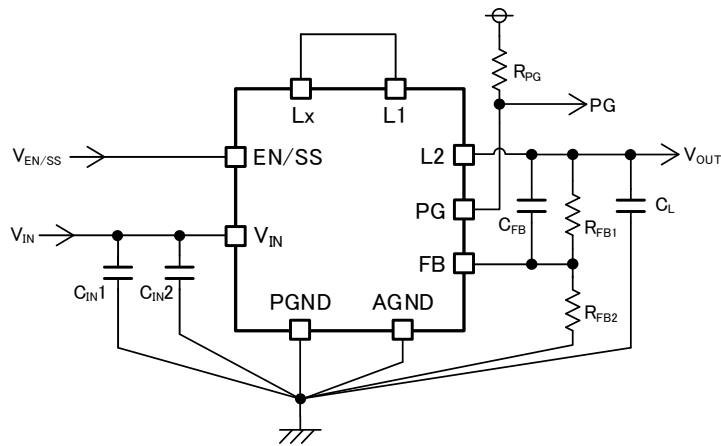
Circuit No.⑥



Circuit No.⑦



## TYPICAL APPLICATION CIRCUIT



\*Coil is dedicated to this product. Please do not use it for purposes other than this product.

### 【Typical example】

	VALUE	PRODUCT NUMBER	Notes
C <sub>IN1</sub>	25V/4.7μF	TMK107BBJ475KA (Taiyo Yuden)	X5R(-55 ~ +85°C)
		C1608X5R1E475K(TDK)	
		TMK212AB7475KG(Taiyo Yuden)	X7R(-55 ~ +125°C)
		C2012X7R1E475K(TDK)	
C <sub>IN2</sub>	25V/0.1μF	C1005X7R1E104K(TDK)	X5R(-55 ~ +85°C)
	50V/0.1μF	UMK105B7104KV(Taiyo Yuden)	X7R(-55 ~ +125°C)
C <sub>L</sub>	10V/10μF	LMK107BBJ106MALT(Taiyo Yuden)	X5R(-55 ~ +85°C)
		C1608X5R1A106M(TDK)	
		LMK212AB7106KG(Taiyo Yuden)	X7R(-55 ~ +125°C)
		C2012X7R1A106K(TDK)	
	25V/10μF	TMK316AB7106KL(Taiyo Yuden)	X5R(-55 ~ +85°C)
		C3216X7R1E106K(TDK)	X7R(-55 ~ +125°C)
	10V/22μF	LMK107BBJ226MA (Taiyo Yuden)	X5R(-55 ~ +85°C)
		LMK316AB7226ML (Taiyo Yuden)	X7R(-55 ~ +125°C)
	25V/22μF	TMK316BBJ226ML(Taiyo Yuden)	X5R(-55 ~ +85°C)
		TMK325B7226MM(Taiyo Yuden)	X7R(-55 ~ +125°C)

\*When under the condition of the voltage difference between input voltage and output voltage is low, please use 125°C product, which has small capacity drop.

### 【Typical example】

V <sub>OUT</sub>	R <sub>FB1</sub>	R <sub>FB2</sub>	C <sub>FB</sub> (C <sub>L</sub> =20μF)	C <sub>FB</sub> (C <sub>L</sub> =10μF)
1.2V	180kΩ	300kΩ	56pF	39pF
1.5V	220kΩ	220kΩ	47pF	33pF
1.8V	180kΩ	130kΩ	56pF	39pF
2.5V	560kΩ	240kΩ	15pF	12pF
3.0V	390kΩ	130kΩ	27pF	18pF
3.3V	510kΩ	150kΩ	22pF	15pF
5.0V	680kΩ	120kΩ	18pF	10pF
7.5V	270kΩ	30kΩ	39pF	27pF
10.0V	160kΩ	13kΩ	68pF	47pF
12.0V	360kΩ	24kΩ	33pF	22pF

•About R<sub>PG</sub>

The PG pin is an Nch open drain output, therefore a pull-up resistance (approx. 100kΩ) must be connected to the PG pin. When not using power good function connect the PG pin to GND or open it for use.



## ■ TYPICAL APPLICATION CIRCUIT(Continued)

### <Output voltage setting>

The output voltage can be set by adding an external dividing resistor.

The output voltage is determined by the equation below based on the values of  $R_{FB1}$  and  $R_{FB2}$ .

$$V_{OUT} = V_{FB} \times (R_{FB1} + R_{FB2}) / R_{FB2}$$

With  $R_{FB1} + R_{FB2} \leq 1M\Omega$

### < $C_{FB}$ setting>

Adjust the value of the phase compensation speed-up capacitor  $C_{FB}$  using the equation below.

$$C_{FB} = \frac{1}{2\pi \times f_{zfb} \times R_{FB1}}$$

A target value for  $f_{zfb}$  of optimum is below.

$$f_{zfb} = \frac{1}{2\pi \sqrt{C_L \times L}}$$

#### 【Setting Example】

When  $C_L$  is 10 $\mu$ F,  $f_{zfb}$ =24kHz

When  $C_L$  is 20 $\mu$ F,  $f_{zfb}$ =17kHz

#### 【Setting Example】

To set output voltage to 5V with  $f_{osc}$ =1.2MHz,  $C_L$ =20 $\mu$ F,  $L$ =4.3 $\mu$ H

When  $R_{FB1}$ =680k $\Omega$ ,  $R_{FB2}$ =120k $\Omega$ ,  $V_{OUT}$ =0.75V $\times$  (680k $\Omega$ +120k $\Omega$ ) / 120k $\Omega$ =5.0V

And  $f_{zfb}$  is set to a target of 17kHz using the above equation,

$C_{FB}$ =1/(2 $\times$   $\pi$   $\times$  17kHz $\times$  680k $\Omega$ )=13.8pF

\* The setting range for the output voltage is 1.0V to 15.0V.

The condition  $V_{OUT}/V_{IN} \geq 0.14$  must be satisfied.

### <Soft-start Time Setting>

The soft-start time can be adjusted by adding a capacitor and a resistor to the EN/SS pin.

Soft-start time ( $t_{SS2}$ ) is approximated by the equation below according to values of  $V_{EN/SS}$ ,  $R_{SS}$ , and  $C_{SS}$ .

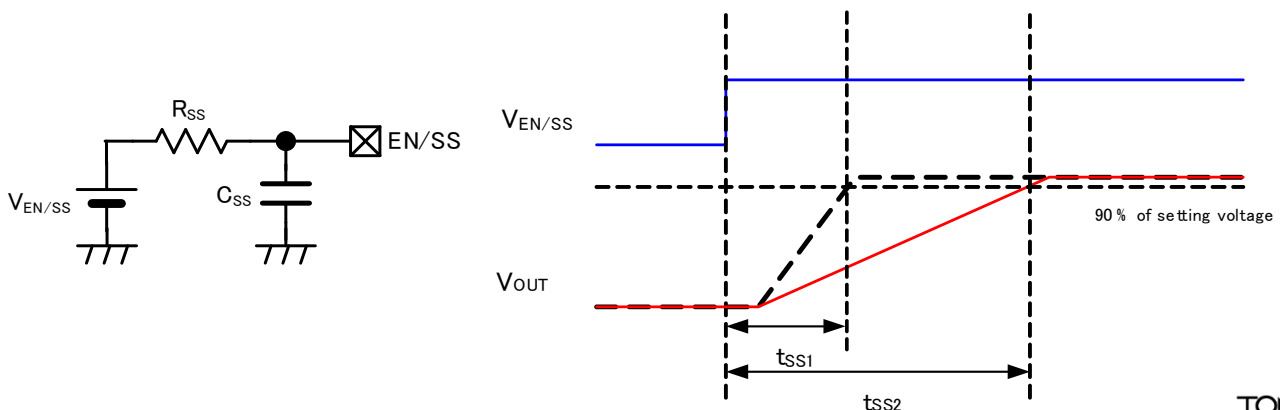
$$t_{SS2} = C_{SS} \times R_{SS} \times (-\ln((V_{EN/SS} - 1.45) / V_{EN/SS}))$$

#### 【Setting Example】

When  $C_{SS}$ =0.47 $\mu$ F,  $R_{SS}$ =430k $\Omega$  and  $V_{EN/SS}$ =12V,  $t_{SS2}$ =0.47 $\times 10^{-6}$   $\times$  430  $\times 10^3$   $\times$  (-ln((12-1.45)/12)) =26ms (Approx.)

\* The soft-start time is the time from the start of  $V_{EN/SS}$  until the output voltage reaches 90% of the set voltage.

If the EN/SS pin voltage rises steeply without connecting  $C_{SS}$  and  $R_{SS}$  ( $R_{SS}$ =0 $\Omega$ ), Output rises with taking the soft-start time of  $t_{SS1}$ =1.0ms (TYP.) which is fixed internally.

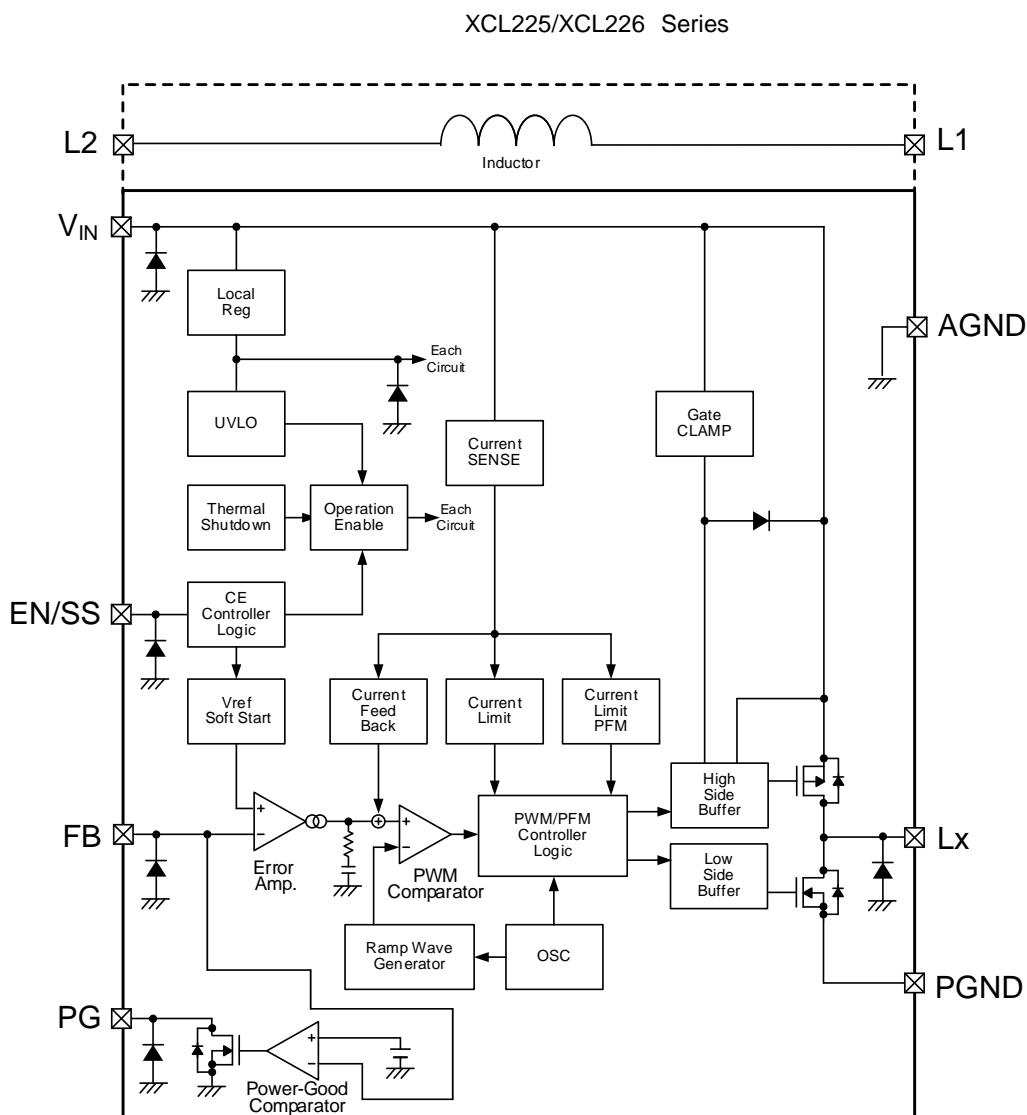


## OPERATIONAL EXPLANATION (Continued)

The XCL225/XCL226 series consists internally of a reference voltage supply with soft-start function, error amp, PWM comparator, ramp wave circuit, oscillator (OSC) circuit, phase compensation (Current feedback) circuit, current limiting circuit, current limit PFM circuit, High-side driver Tr., Low-side driver Tr., buffer drive circuit, internal power supply (Local Reg) circuit, under-voltage lockout (UVLO) circuit, gate clamp (CLAMP) circuit, thermal shutdown (TSD) circuit, power good comparator, control block and other elements.

The voltage feedback from the FB pin is compared to the internal reference voltage by the error amp, the output from the error amp is phase compensated, and the signal is input to the PWM comparator to determine the ON time of switching during PWM operation. The output signal from the error amp is compared to the ramp wave by the PWM comparator, and the output is sent to the buffer drive circuit and output from the Lx pin as the duty width of switching. This operation is performed continuously to stabilize the output voltage.

The driver transistor current is monitored at each switching by the phase compensation (Current feedback) circuit, and the output signal from the error amp is modulated as a multi-feedback signal. This allows a stable feedback system to be obtained even when a low ESR capacitor such as a ceramic capacitor is used, and this stabilizes the output voltage.



\*Diodes inside the circuit are an ESD protection diodes and a parasitic.

## ■ OPERATIONAL EXPLANATION (Continued)

### <Reference voltage source>

The reference voltage source provides the reference voltage to ensure stable output voltage of the DC/DC converter.

### <Oscillator circuit>

The ramp wave circuit determines switching frequency. The frequency is fixed internally 1.2MHz. Clock pulses generated in this circuit are used to produce ramp waveforms needed for PWM operation.

### <Error amplifier>

The error amplifier is designed to monitor output voltage. The amplifier compares the reference voltage with the feedback voltage divided by the internal voltage divider,  $R_{FB1}$  and  $R_{FB2}$ . When a voltage is lower than the reference voltage, then the voltage is fed back, the output voltage of the error amplifier increases. The error amplifier output is fixed internally to deliver an optimized signal to the mixer.

### <Current limiting>

The current limiting circuit of the XCL225/XCL226 series monitors the current that flows through the High-side driver transistor and Low-side driver transistor, and when over-current is detected, the current limiting function activates.

#### ① High-side driver Tr. current limiting

The current in the High-side driver Tr. is detected to equivalently monitor the peak value of the coil current. The High-side driver Tr. current limiting function forcibly turns off the High-side driver Tr. when the peak value of the coil current reaches the High-side driver current limit value  $I_{LIMH}$ . When the over-current state is released, normal operation resumes.

#### ② Low-side driver Tr. current limiting

The current in the Low-side driver Tr. is detected to equivalently monitor the bottom value of the coil current. The Low-side driver Tr. current limiting function prohibits the High-side driver Tr. from turning on in an over-current state where the bottom value of the coil current is higher than the Low-side driver Tr. current limit value  $I_{LIML}$  (TYP. 0.9A). Control to lower the switching frequency  $f_{OSC}$  is also performed. When the over-current state is released, normal operation resumes.

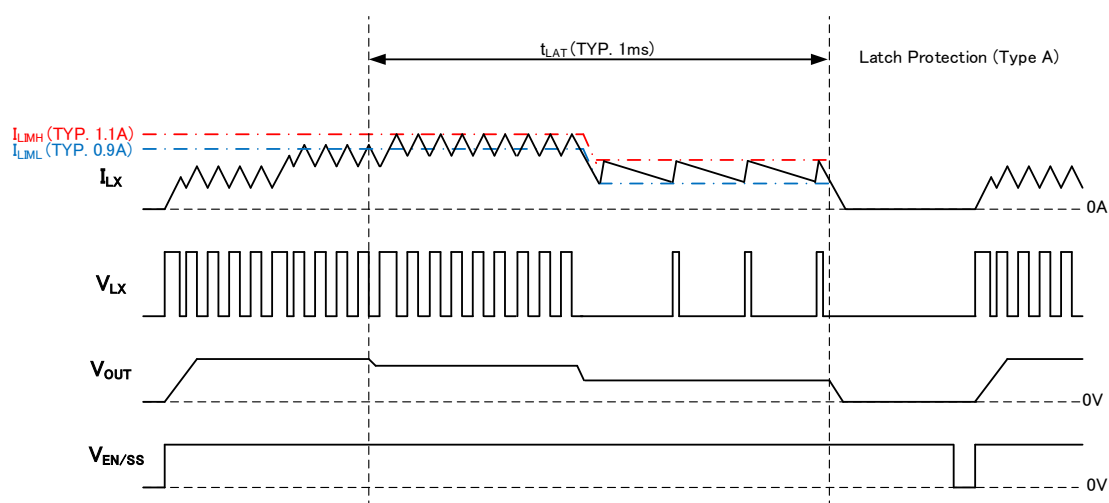
#### ③ Over-current latch (Type A)

Type A turns off the High-side and Low-side driver Tr. when state ① or ② continues for  $t_{LAT}$  (TYP. 1.0ms). The Lx pin is latch-stopped at the GND level (0V).

The latch-stopped state only stops the pulse output from the Lx pin; the internal circuitry of the IC continues to operate. To restart after latch-stopping, L level and then H level must be input into the EN/SS pin, or  $V_{IN}$  pin re-input must be performed (after lowering the voltage below the UVLO detection voltage) to resume operation by soft start.

The over-current latch function may occasionally be released from the current limit detection state by the effects of ambient noise, and it may also happen that the latch time becomes longer or latching does not take place due to board conditions. For this reason, place the input capacitor as close as possible to the IC.

Type B is an automatic recovery type that performs the operation of ① or ② until the over-current state is released.



< Current limiting timing chart >

## ■ OPERATIONAL EXPLANATION (Continued)

### <Soft-start function>

The reference voltage applied to the error amplifier is restricted by the start-up voltage of the EN/SS pin. This ensures that the error amplifier operates with its two inputs in balance, thereby preventing ON-time signal from becoming longer than necessary. Therefore, start-up time of the EN/SS pin becomes the set-time of soft-start. The soft-start time can be adjusted by adding a capacitor and a resistor to the EN/SS pin.

If the EN/SS pin voltage rises steeply without connecting  $C_{SS}$  and  $R_{SS}$  ( $R_{SS}=0\Omega$ ), Output rises with taking the soft-start time of  $t_{SS1}=1.0\text{ms}$  (TYP.) which is fixed internally.

The soft-start function operates when the voltage at the EN/SS pin is between 0.3V to 2.5V. If the voltage at the EN/SS pin does not start from 0V but from a middle level voltage when the power is switched on, the soft-start function will become ineffective and the possibilities of large inrush currents and ripple voltages occurring will be increased.

### <Thermal shutdown>

The thermal shutdown (TSD) as an over current limit is built in the XCL225/XCL226 series.

When the junction temperature reaches the detection temperature, the driver transistor is forcibly turned off. When the junction temperature falls to the release temperature while in the output stop state, restart takes place by soft-start.

### <UVLO>

When the  $V_{IN}$  pin voltage falls below  $V_{UVLO1}$  (TYP. 2.7V), the driver transistor is forcibly turned off to prevent false pulse output due to instable operation of the internal circuits. When the  $V_{IN}$  pin voltage rises above  $V_{UVLO2}$  (TYP. 2.8V), the UVLO function is released, the soft-start function activates, and output start operation begins. Stopping by UVLO is not shutdown; only pulse output is stopped and the internal circuits continue to operate.

### <Power good>

The output state can be monitored using the power good function.

The PG pin is an Nch open drain output, therefore a pull-up resistance (approx. 100k $\Omega$ ) must be connected to the PG pin. When not using power good function connect the PG pin to GND or open it for use.

CONDITION		SIGNAL
EN/SS = H	$V_{FB} > V_{PGDET}$	H (High impedance)
	$V_{FB} \leq V_{PGDET}$	L (Low impedance)
	Thermal Shutdown	L (Low impedance)
	UVLO ( $V_{IN} < V_{UVLO1}$ )	Undefined State
EN/SS = L	Stand-by	L (Low impedance)

## ■ NOTE ON USE(Continued)

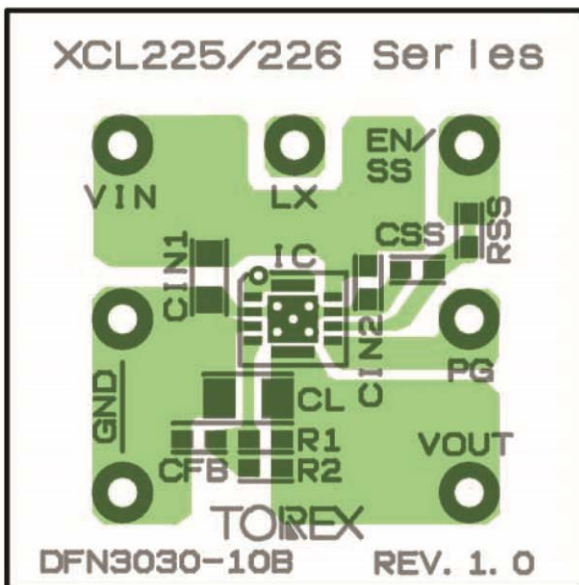
- 1) For the phenomenon of temporal and transitional voltage decrease or voltage increase, the IC may be damaged or deteriorated if IC is used beyond the absolute MAX. specifications.
- 2) Make sure that the absolute maximum ratings of the external components and of this IC are not exceeded.
- 3) The DC/DC converter characteristics depend greatly on the externally connected components as well as on the characteristics of this IC, so refer to the specifications and standard circuit examples of each component when carefully considering which components to select. Be especially careful of the capacitor characteristics and use B characteristics (JIS standard) or X7R, X5R (EIA standard) ceramic capacitors.  
The capacitance decrease caused by the bias voltage may become remarkable depending on the external size of the capacitor.
- 4) If there is a large dropout voltage, then there might be pulse-skip during light loads even with PWM control.
- 5) The DC/DC converter of this IC uses a current-limiting circuit to monitor the coil peak current. If the potential dropout voltage is large or the load current is large, the peak current will increase, which makes it easier for current limitation to be applied which in turn could cause the operation to become unstable.
- 6) If there is a large dropout voltage, a circuit delay could create the ramp-up of coil current with staircase waveform exceeding the current limit.
- 7) The ripple voltage could be increased when switching from discontinuous conduction mode to Continuous conduction mode. Please evaluate IC well on customer's PCB.
- 8) The operation of becomes unstable below the minimum operating voltage.
- 9) If the voltage at the EN/SS Pin does not start from 0V but it is at the midpoint potential when the power is switched on, the soft start function may not work properly and it may cause the larger inrush current and bigger ripple voltages.
- 10) The effects of ambient noise and the state of the circuit board may cause release from the current limiting state, and The latch time may lengthen or latch operation may not take place. Please evaluate IC well on customer's PCB.

## NOTE ON USE

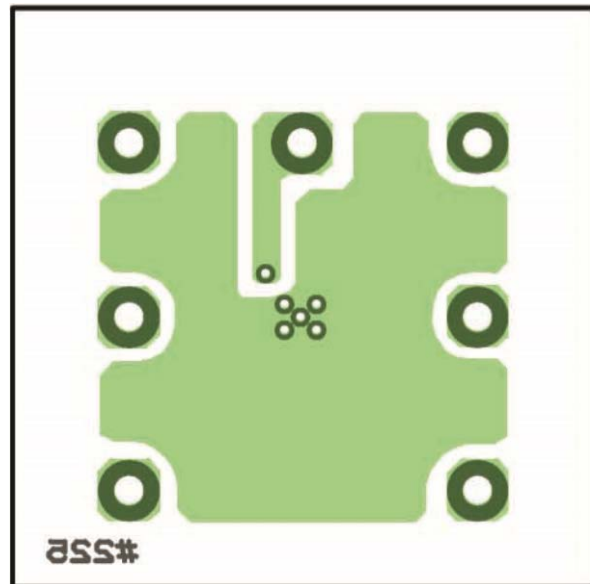
### 11) Instructions of pattern layouts

- (1) The operation may become unstable due to noise and/or phase lag from the output current when the wire impedance is high, please place the input capacitor( $C_{IN}$ ) and the output capacitor ( $C_L$ ) as close to the IC as possible.
- (2) In order to stabilize  $V_{IN}$  voltage level, we recommend that a by-pass capacitor ( $C_{IN}$ ) be connected as close as possible to the  $V_{IN}$  and PGND pins.
- (3) Please mount each external component as close to the IC as possible.
- (4) Wire external components as close to the IC as possible and use thick, short connecting traces to reduce the circuit impedance.
- (5) Make sure that the GND traces are as thick as possible, as variations in ground potential caused by high ground currents at the time of switching may result in instability of the IC.
- (6) Please note that internal driver transistors bring on heat because of the load current and ON resistance of High-side driver transistor, Low-side driver transistor.

< Reference Pattern Layout >



< Top view >



< Bottom side top view >

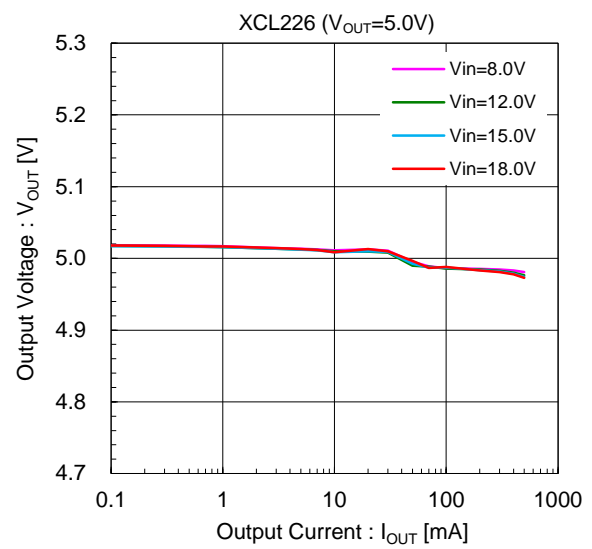
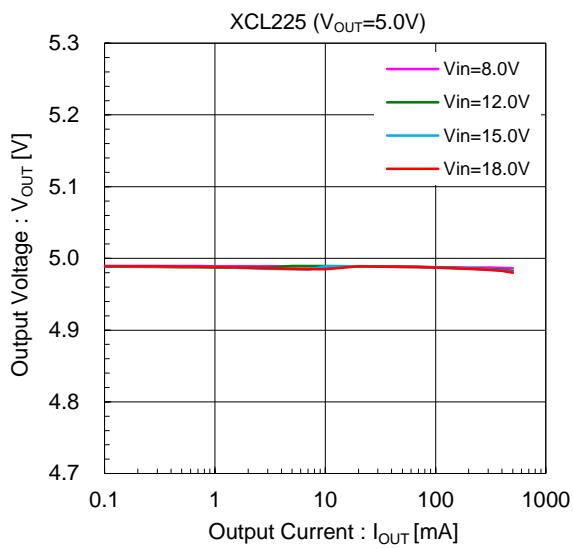
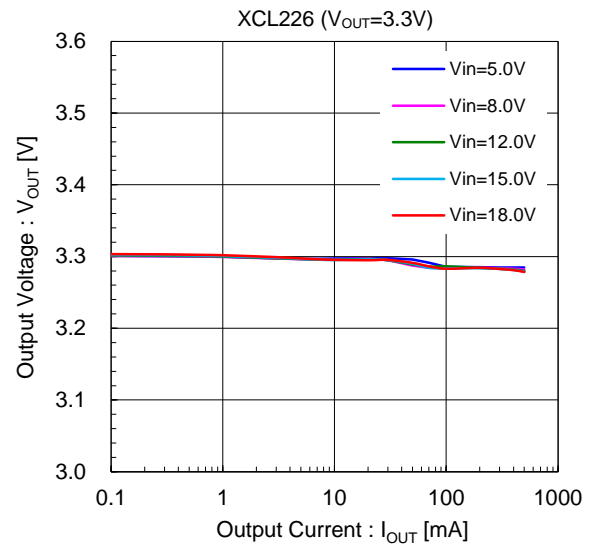
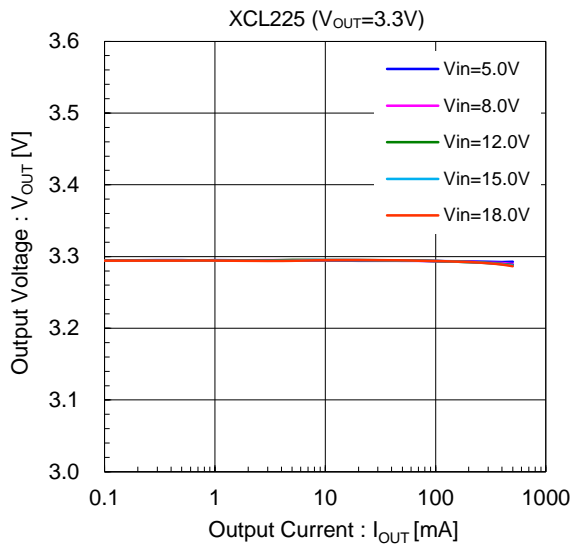
- 12) Torex places an importance on improving our products and their reliability. We request that users incorporate fail-safe designs and post-aging protection treatment when using Torex products in their systems.

## About the appearance (coil part)

- (1) Coils are compliant with general surface mount type chip coil (inductor) specifications and may have scratches, flux contamination and the like.
- (2) This product has a coil which is soldered on top of the package. There is no problem with normal board mounting reflow. If the excessive shock is provided during reflow, there is a possibility to cause the coil misaligned or have the coil fell off. Please be careful not to provide excessive shock to the PCB during the board mounting reflow.

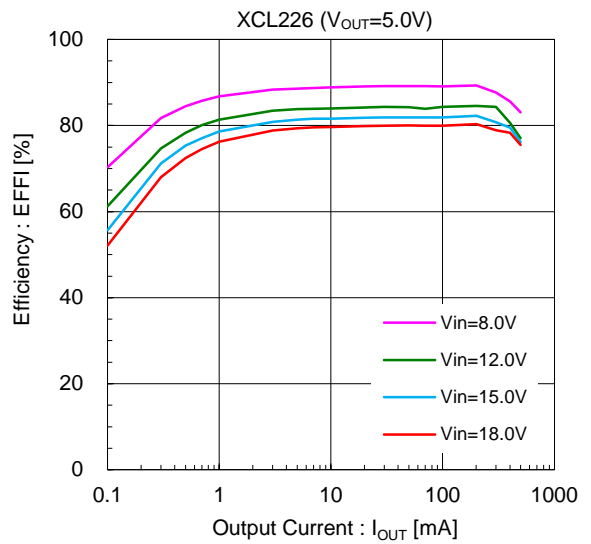
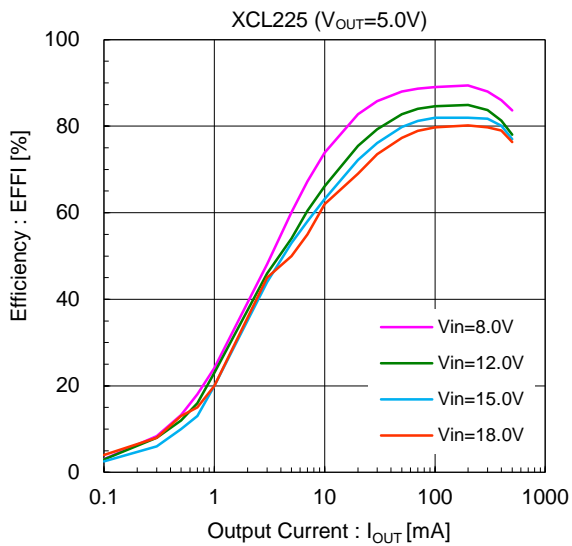
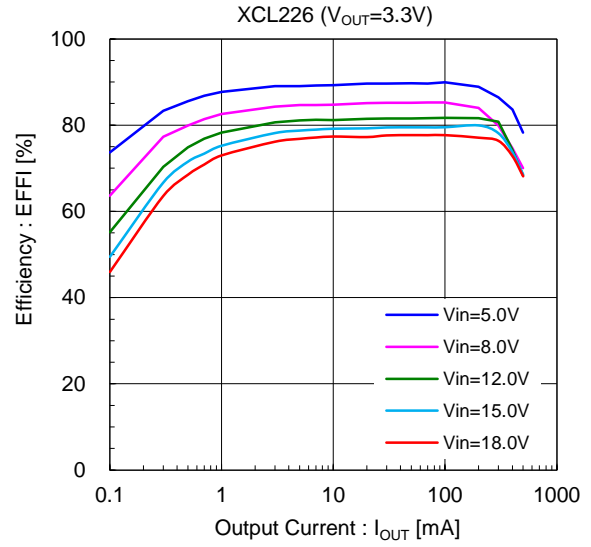
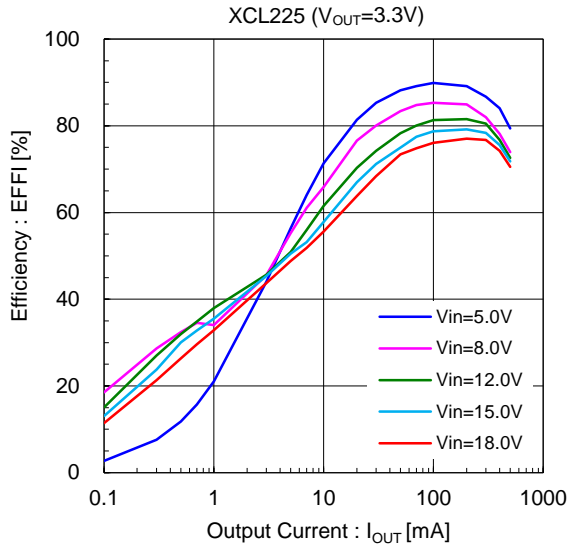
## TYPICAL PERFORMANCE CHARACTERISTICS

### (1) Output Voltage vs. Output Current



## TYPICAL PERFORMANCE CHARACTERISTICS

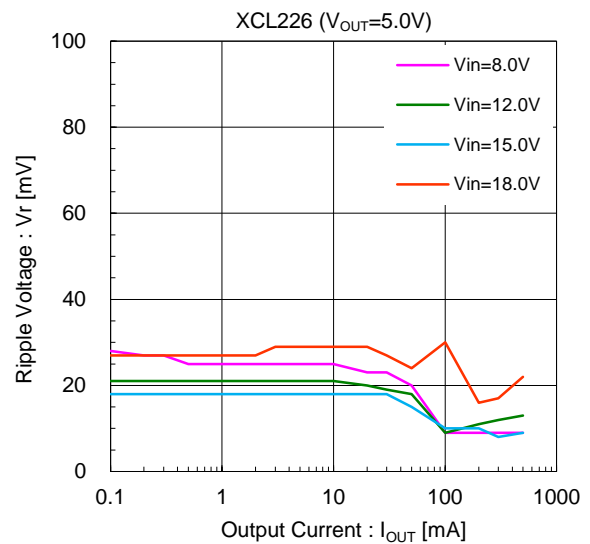
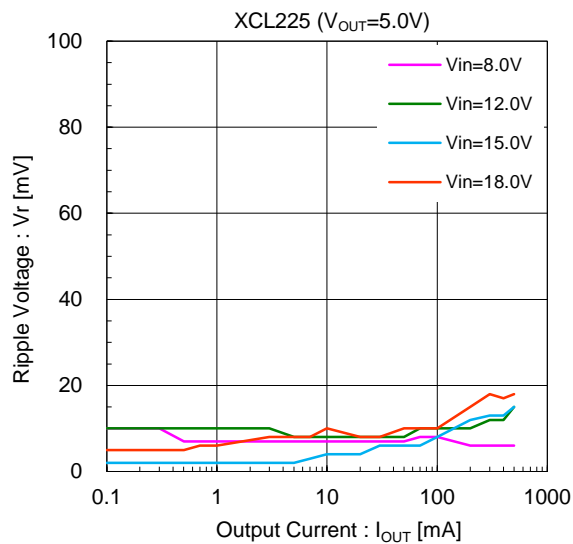
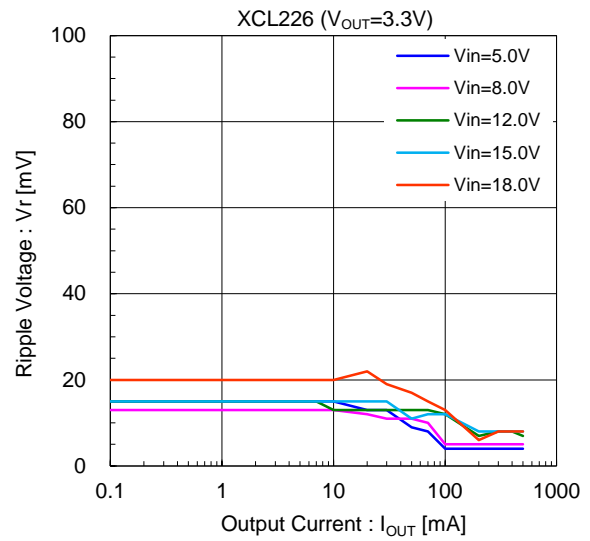
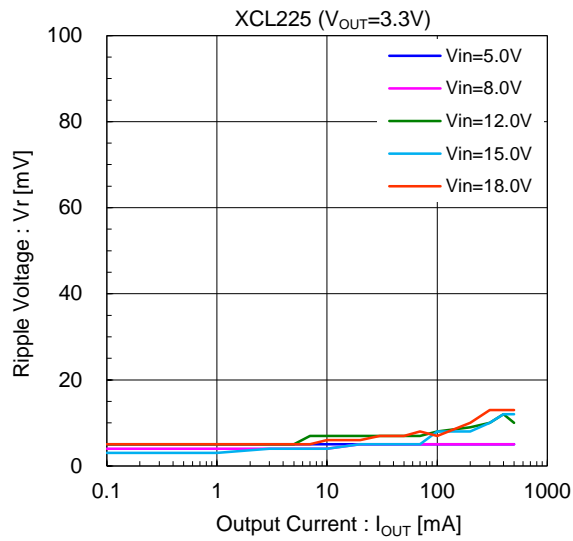
### (2) Efficiency vs. Output Current





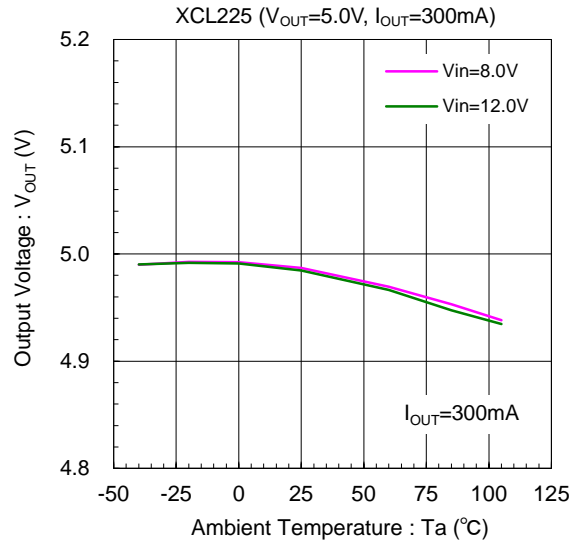
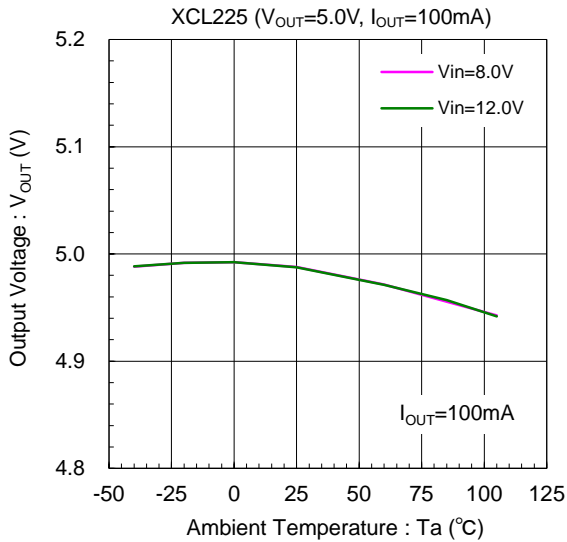
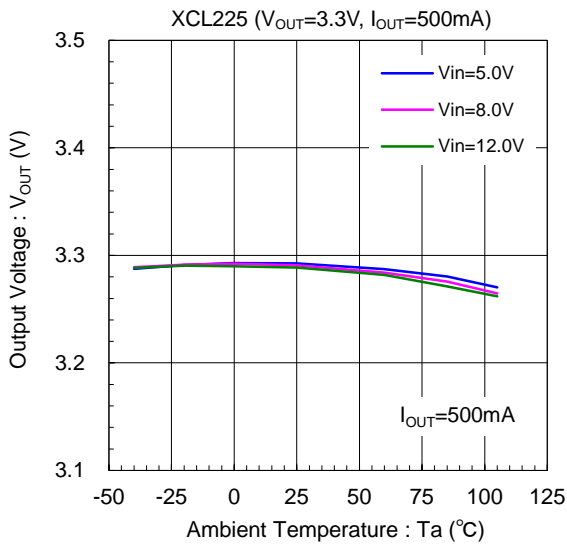
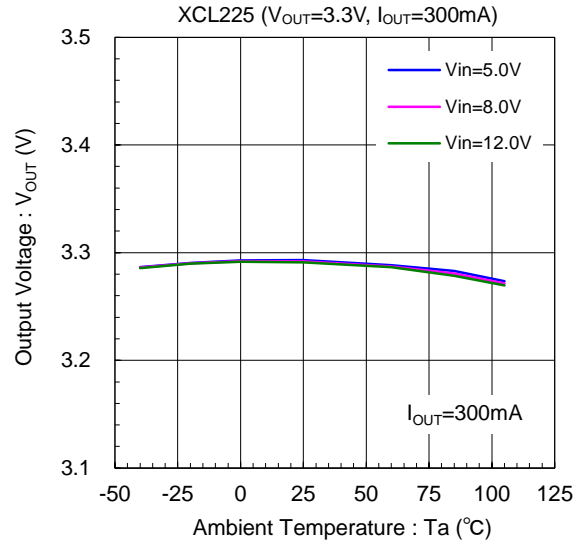
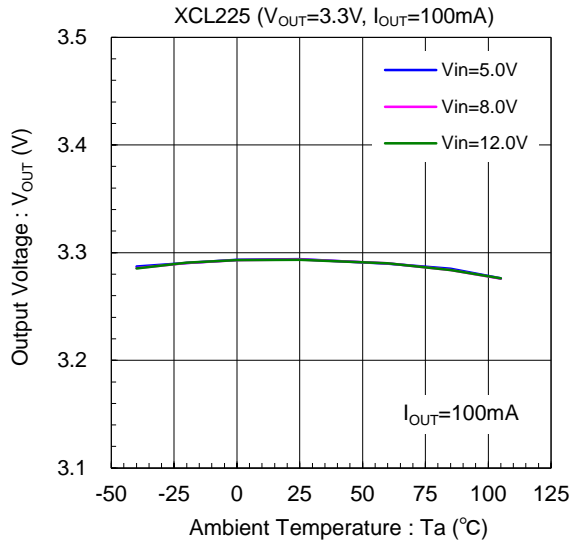
## TYPICAL PERFORMANCE CHARACTERISTICS

### (3) Ripple Voltage vs. Output Current



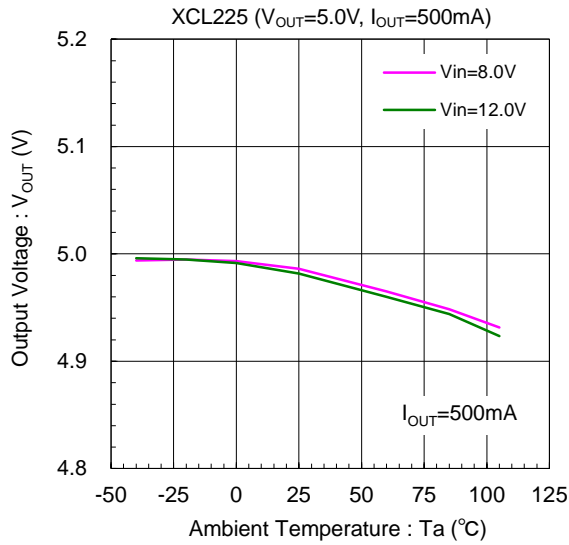
## TYPICAL PERFORMANCE CHARACTERISTICS

### (4) Output Voltage vs. Ambient Temperature

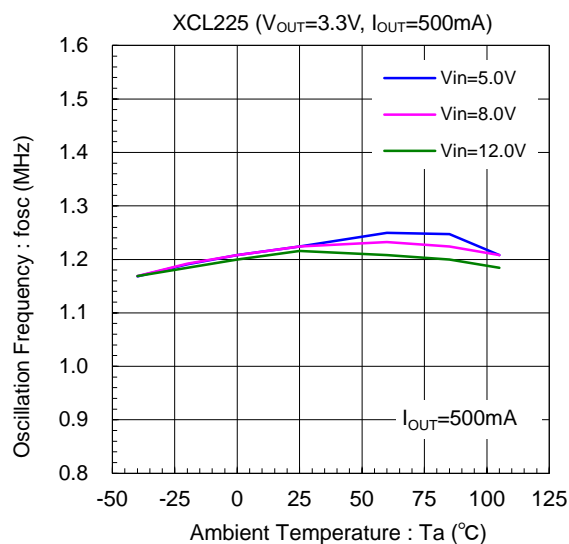
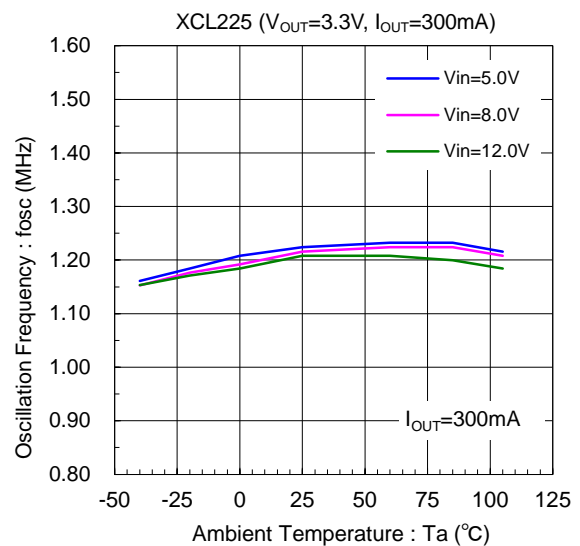
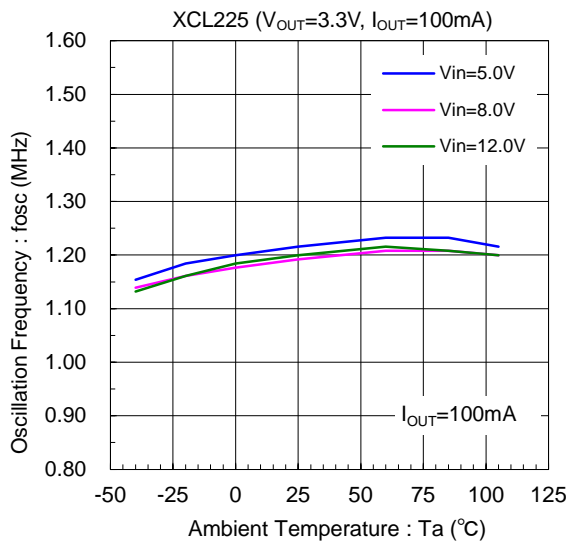


## TYPICAL PERFORMANCE CHARACTERISTICS

### (4) Output Voltage vs. Ambient Temperature

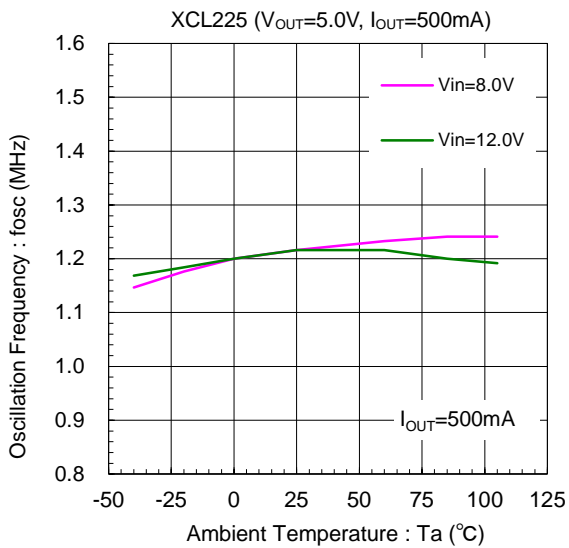
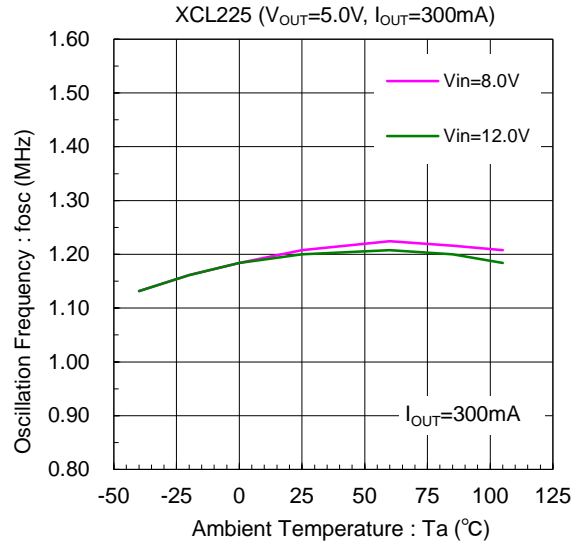
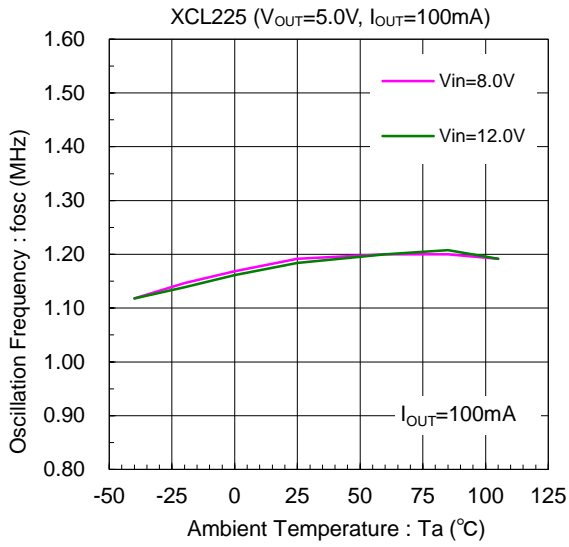


### (5) Oscillation Frequency vs. Ambient Temperature



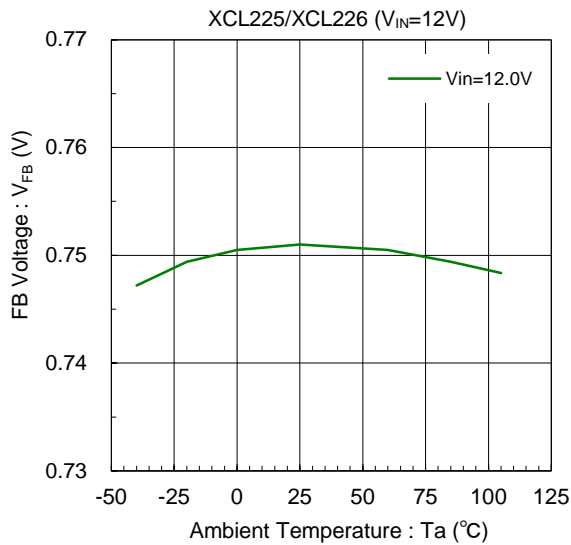
## TYPICAL PERFORMANCE CHARACTERISTICS

### (5) Oscillation Frequency vs. Ambient Temperature

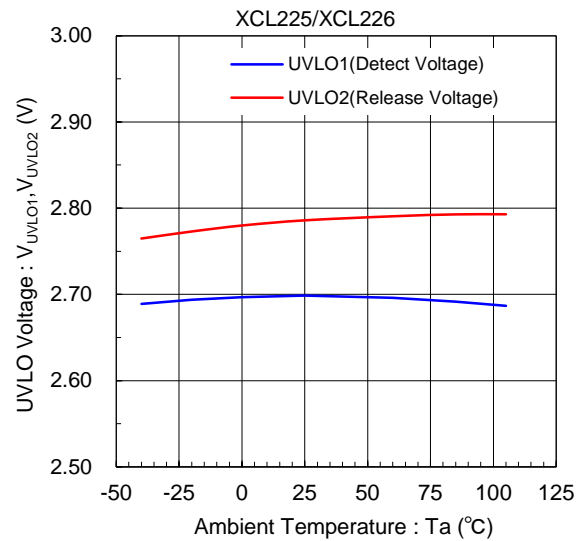


## TYPICAL PERFORMANCE CHARACTERISTICS

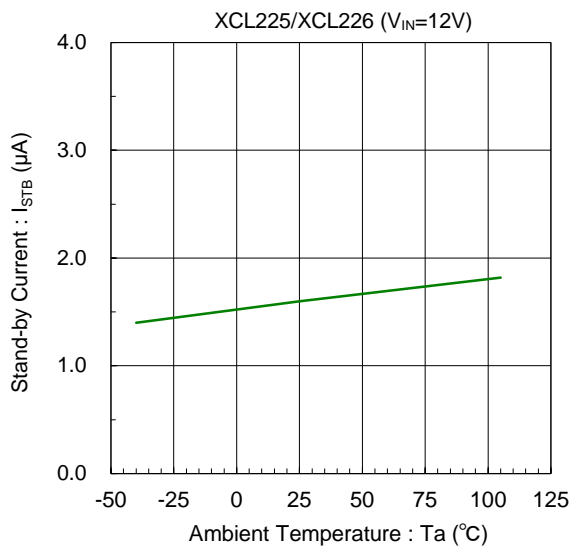
(6) FB Voltage vs. Ambient Temperature



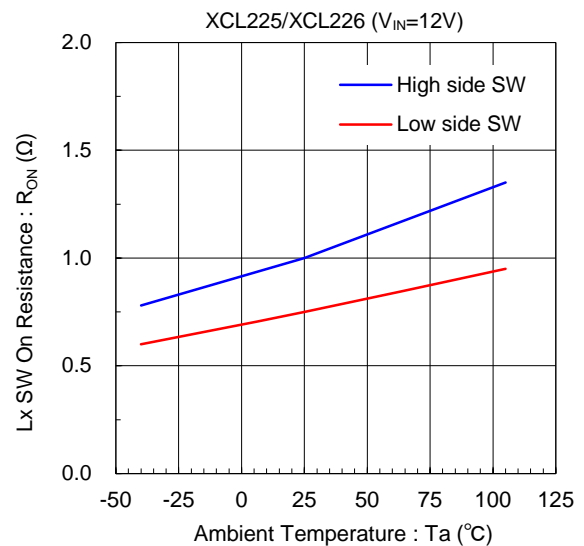
(7) UVLO Voltage vs. Ambient Temperature



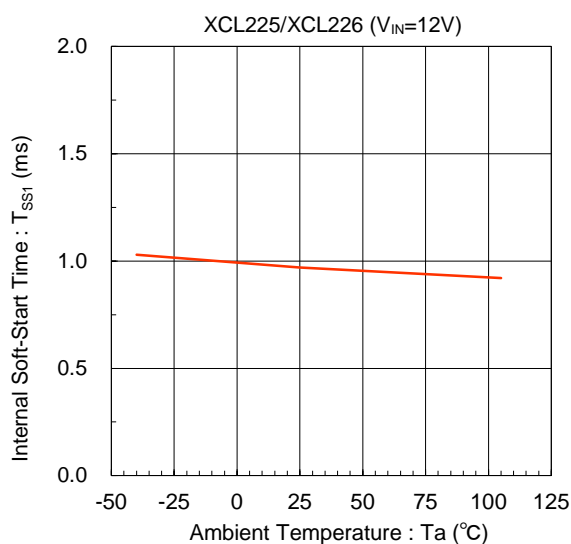
(8) Stand-by Current vs. Ambient Temperature



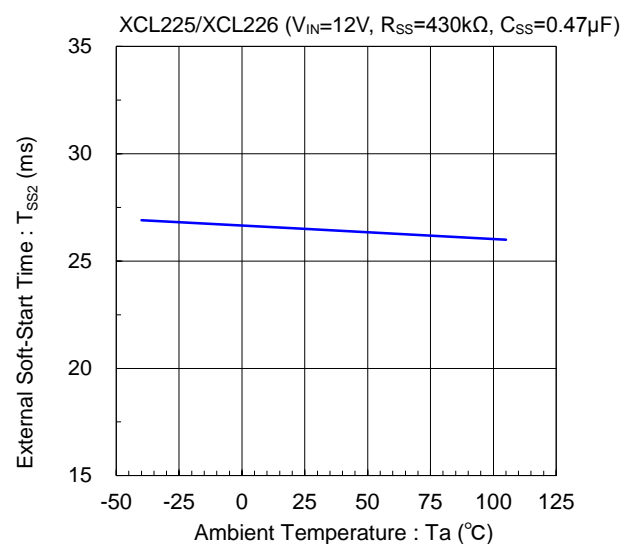
(9) Lx SW On Resistance vs. Ambient Temperature



(10) Internal Soft-Start Time vs. Ambient Temperature

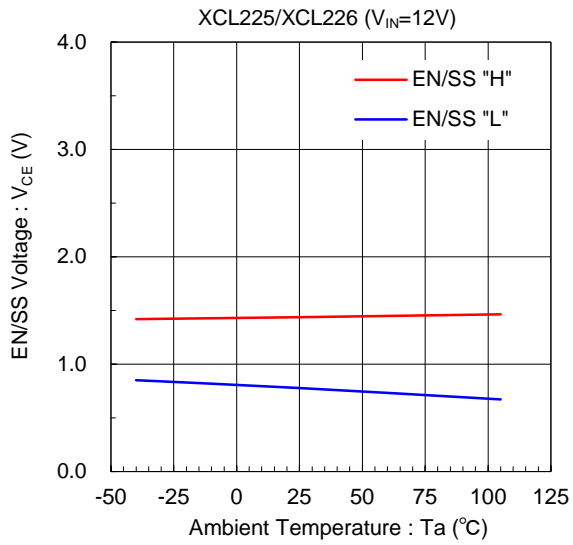


(11) External Soft-Start Time vs. Ambient Temperature

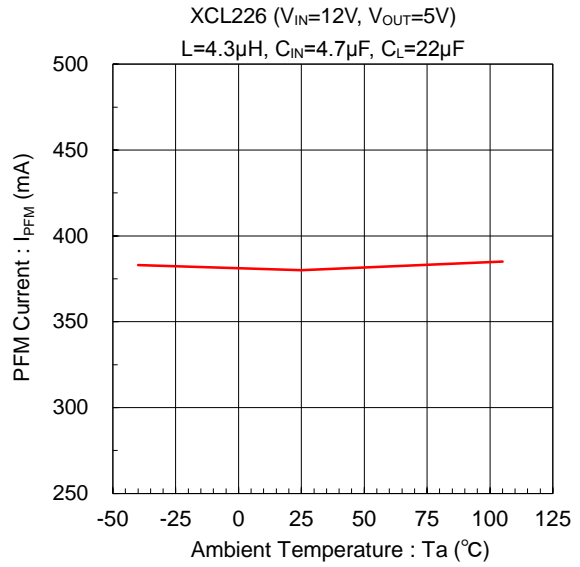


## TYPICAL PERFORMANCE CHARACTERISTICS

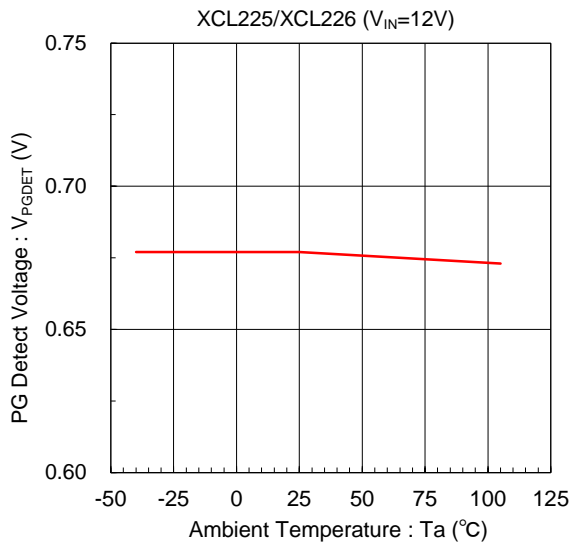
(12) EN/SS Voltage vs. Ambient Temperature



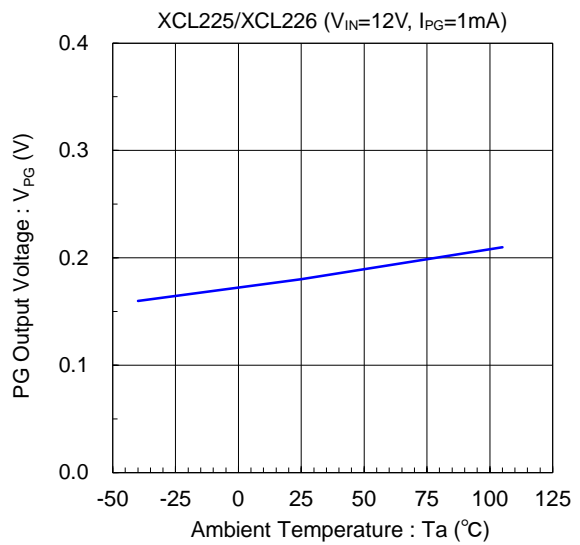
(13) PFM Current vs. Ambient Temperature



(14) PG Detect Voltage vs. Ambient Temperature



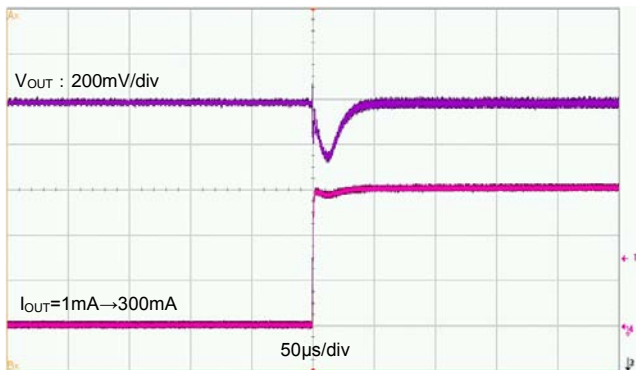
(15) PG Output Voltage vs. Ambient Temperature



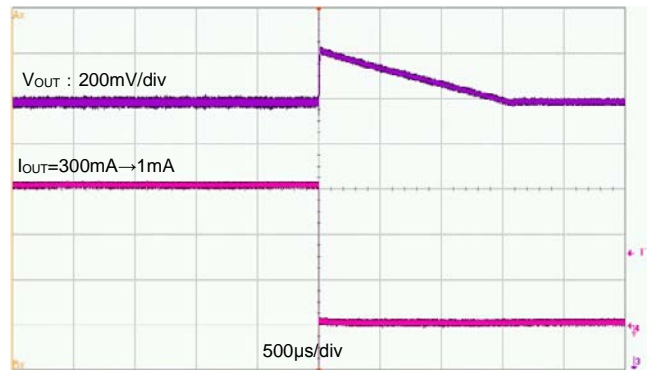
## TYPICAL PERFORMANCE CHARACTERISTICS

### (16) Load Transient Response

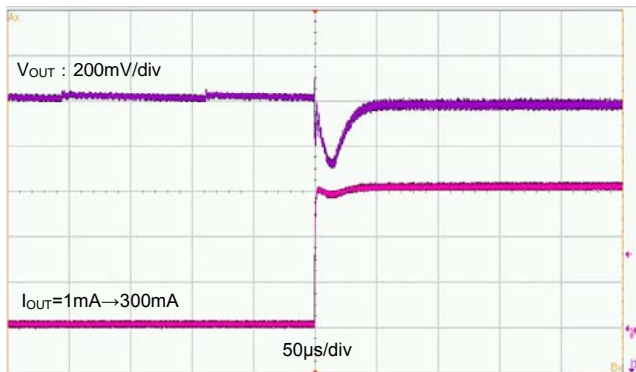
XCL225 ( $V_{IN}=12V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=1mA \rightarrow 300mA$ )



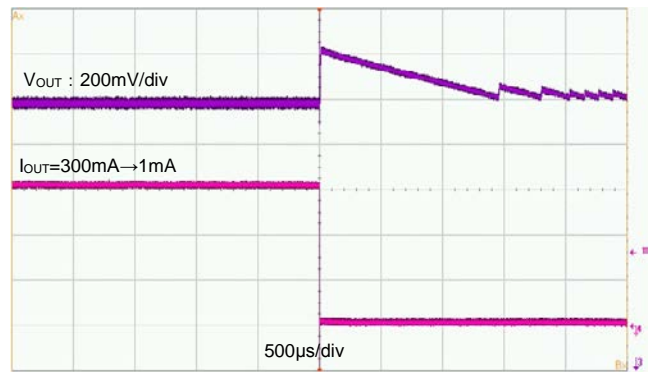
XCL225 ( $V_{IN}=12V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=300mA \rightarrow 1mA$ )



XCL226 ( $V_{IN}=12V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=1mA \rightarrow 300mA$ )



XCL226 ( $V_{IN}=12V$ ,  $V_{OUT}=5V$ ,  $I_{OUT}=300mA \rightarrow 1mA$ )



## ■ PACKAGING INFORMATION

For the latest package information go to, [www.torexsemi.com/technical-support/packages](http://www.torexsemi.com/technical-support/packages)

PACKAGE	OUTLINE / LAND PATTERN	THERMAL CHARACTERISTICS	
DFN3030-10B	<a href="#">DFN3030-10B PKG</a>	JESD51-7 Board	<a href="#">DFN3030-10B Power Dissipation</a>



## MARKING RULE

### DFN3030-10B

①,②,③ represents products series

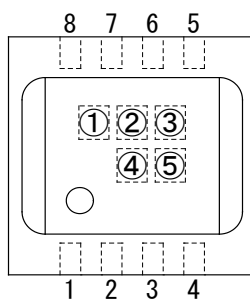
SYMBOL			製品	Type	PRODUCT
①	②	③			
1	1	B	XCL225	A	XCL225A0K1H2
1	1	E	XCL225	B	XCL225B0K1H2
1	1	K	XCL226	A	XCL226A0K1H2
1	1	N	XCL226	B	XCL226B0K1H2

④,⑤ represents production lot number

01~09, 0A~0Z, 11~9Z, A1~A9, AA~AZ, B1~ZZ in order. (G, I, J, O, Q, W excluded)

\* No character inversion used.

< DFN3030-10B >



1. The product and product specifications contained herein are subject to change without notice to improve performance characteristics. Consult us, or our representatives before use, to confirm that the information in this datasheet is up to date.
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5. Although we make continuous efforts to improve the quality and reliability of our products; nevertheless Semiconductors are likely to fail with a certain probability. So in order to prevent personal injury and/or property damage resulting from such failure, customers are required to incorporate adequate safety measures in their designs, such as system fail safes, redundancy and fire prevention features.
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7. Please use the product listed in this datasheet within the specified ranges.
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