TOSHIBA

32 Bit RISC Microcontroller TX00 Series

TMPM061FWFG

TOSHIBA ELECTRONIC DEVICES & STORAGE CORPORATION

TOSHIBA

Arm, Cortex and Thumb are registered trademarks of Arm Limited (or its subsidiaries) in the US and/or elsewhere. All rights reserved.

arm

Introduction: Notes on the description of SFR (Special Function Register) under this specification

An SFR (Special Function Register) is a control register for periperal circuits (IP).

The SFR addressses of IPs are described in the chapter on memory map, and the details of SFR are given in the chapter of each IP.

Definition of SFR used in this specification is in accordance with the following rules.

- a. SFR table of each IP as an example
	- ・ SFR tables in each chapter of IP provides register names, addresses and brief descriptions.
	- ・ All registers have a 32-bit unique address and the addresses of the registers are defined as follows, with some exceptions: "Base address + (Unique) address"

Note: **SAMCR register address is 32 bits wide from the address 0x0000_0004 (Base Address(0x00000000) + unique address (0x0004)).**

Note: **The register shown above is an example for explanation purpose and not for demonstration purpose. This register does not exist in this microcontroller.**

b. SFR(register)

- Each register basically consists of a 32-bit register (some exceptions).
- The description of each register provides bits, bit symbols, types, initial values after reset and functions.

1.2.2 SAMCR(Control register)

TOSHIBA

Note: **The Type is divided into three as shown below.**

c. Data descriptopn

Meanings of symbols used in the SFR description are as shown below.

- ・ x:channel numbers/ports
- ・ n,m:bit numbers

d. Register descriptoption

Registers are described as shown below.

• Register name <Bit Symbol>

Exmaple: SAMCR<MODE>="000" or SAMCR<MODE[2:0]>="000"

<MODE[2:0]> indicates bit 2 to bit 0 in bit symbol mode (3bit width).

・ Register name [Bit]

Example: SAMCR[9:7]="000"

It indicates bit 9 to bit 7 of the register SAMCR (32 bit width).

General precautions on the use of Toshiba MCUs

This Page explains general precautions on the use of Toshiba MCUs.

Note that if there is a difference between the general precautions and the description in the body of the document, the description in the body of document has higher priority.

1. The MCUs' operation at power-on

At power-on, internal state of the MCUs is unstable. Therefore, state of the pins is undefined until reset operation is completed.

When a reset is performed by an external reset pin, pins of the MCUs that use the reset pin are undefined until reset operation by the external pin is completed.

Also, when a reset is performed by the internal power-on reset, pins of the MCUs that use the internal power-on reset are undefined until power supply voltage reaches the voltage at which poweron reset is valid.

2. Unused pins

Unused input/output ports of the MCUs are prohibited to use. The pins are high-impedance.

Generally, if MCUs operate while the high-impedance pins left open, electrostatic damage or latchup may occur in the internal LSI due to induced voltage influenced from external noise.

Toshiba recommend that each unused pin should be connected to the power supply pins or GND pins via resistors.

3. Clock oscillation stability

A reset state must be released after the clock oscillation becomes stable. If the clock is changed to another clock while the program is in progress, wait until the clock is stable.

Revision History

Table of Contents

Introduction: Notes on the description of SFR (Special Function Register) under this specification

TMPM061FWFG

2. Product Information

3. Processor Core

4. Memory Map

5. Reset Operation

6. Clock/Mode control

7. Exceptions

8. Input / Output port

9. 16-bit Timer / Event Counters (TMRB)

10. 16-Bit Timer A (TMR16A)

11. Serial Channel (SIO/UART)

12. Serial Bus Interface (I2C/SIO)

13. 10-bit Analog/Digital Converter (ADC)

14. 24-bit ΔΣ Analog/Digital Converter (DSADC)

15. **Temperature Sensor (TEMP)**

16. Real Time Clock (RTC)

16.3.3.4 RTCDAYR (Day of the week column register (PAGE0/1)
16.3.3.5 RTCDAYR (Day of the week column register (PAGE0/1) RTCDAYR (Day of the week column register($PAGE0/1$))

17. [LCD Driver](#page-312-0)

18. Low Voltage Detection Circuit (LVD)

19. Watchdog Timer (WDT)

20. **Flash Memory Operation**

21. [Debug Interface](#page-398-0)

22. [Electrical Characteristics](#page-402-0)

23. Port Section Equivalent Circuit Schematic

24. Package Dimensions

CMOS 32-Bit Microcontroller

TMPM061FWFG

The TMPM061FWFG is a 32-bit RISC microprocessor series with an Arm®Cortex®-M0 microprocessor core. Features of the TMPM061FWFG are as follows:

1.1 Features

- 1. Arm Cortex-M0 microprocessor core
	- a. Improved code efficiency has been realized through the use of Thumb® −2 instruction.
	- b. Both high performance and low power consumption have been achieved.
		- [High performance]
		- \cdot A 32-bit multiplication (32×32=32 bit) can be executed with one clock.
		- [Low power consumption]
		- ・ Optimized design using a low power consumption library
		- ・ Standby function that stops the operation of the micro controller core
	- c. High-speed interrupt response suitable for real-time control
		- ・ An interruptible long instruction.
		- ・ Stack push automatically handled by hardware.
- 2. Endian: Little endian
- 3. On Chip program memory and data memory
	- On chip Flash ROM: 128 Kbyte
	- On chip RAM: 8 Kbyte
- 4. Power calculation engine (PCE) : DSP function for power calculation
- 5. 16-bit timer (TMRB): 2 channels
	- 16-bit interval timer mode
	- ・ 16-bit event counter mode
	- ・ 16-bit PPG output
	- ・ Input capture function
	- ・ Synchronous mode
- 6. 16-bit timer (TMR16A): 7 channels
- 7. General-purpose serial interface (SIO/UART): 4 channels

Either UART mode or synchronous mode can be selected

- * Port selection is possible (1 channel)
- * IR carrier pulse output is possible (3 channels)
- 8. Serial bus interface (I2C/SIO): 1channels

Either I2C bus mode or synchronous mode can be selected.

- 9. 10-bit AD converter (ADC): 1 unit
	- ・ Fixed channel/scan mode
	- ・ Single/repeat mode
	- ・ AD monitoring 2ch
	- ・ Conversion speed

16.2μsec (AVDD = 2.7 to 3.6V)

32.4μsec (AVDD = 1.8 to 3.6V)

10. 24-bit ΔΣAD converter (ADC): 3 units

- ・ Sampling frequency: 3 KHz, 6 KHz
- Input voltage range: -0.375 to $+0.375$ V
- Programmable gain amp: $\times 1$, $\times 2$, $\times 4$, $\times 8$ or $\times 16$ can be selected.
- ・ Conversion mode: Single or repeat
- ・ Adjustment of conversion start
- ・ Synchronous start of multiple units
- 11. Temperature sensor (TEMP)
- 12. Real time clock (RTC): 1 channel
	- ・ Clock (hour, minute and second)
	- ・ Calendar (month, week, date and leap year)
	- ・ Clock adjustment (by software)
- 13. LCD driver/controller (LCDD)
	- LCD direct drive is possible (40 seg \times 4 com)
	- \cdot 1/4, 1/3, 1/2 duties or static drive are selectable
	- ・ Bleeder resistance incorporated (external bleeder resistance is also usable)
- 14. Voltage detection circuit (LVD)
- 15. Watchdog timer (WDT): 1 channel

Watchdog timer (WDT) generates a reset or a non-maskable interrupt (NMI).

- 16. Interrupt source: The order of priority can be set to 4 levels.
	- ・ Internal: 28 factors
	- ・ External: 4 factors
- 17. Input/output ports (PORT): 64 pins

Three 5V-Tolerant inputs are prepared

18. Standby mode

Standby modes: IDLE, SLOW, SLEEP, STOP

19. Clock generator (CG)

Clock gear function: The high-speed clock can be divided into 1/1, 1/2, 1/4, 1/8 or 1/16.

- 20. Maximum operating frequency: 16 MHz
- 21. Operating voltage range: 1.8 to 3.6 V
	- 2.9 to 3.6 V: Use of 24-bit $\Delta \Sigma$ AD converter is possible

2.2 to 3.6 V: Use of LCD driver/controller is possible

2.7 to 3.6 V: Write/erase of Flash ROM is possible

22. Temperature range

- ・ -40 to 85 degrees (except during Flash writing/erasing)
- ・ 0 to 70 degrees (during Flash writing/erasing)

23. Package

LQFP100 (14mm \times 14mm, 0.5mm pitch)

1.2 Block Diagram

Figure 1-1 Block Diagram

1.3 Pin Layout (Top view)

Figure 1-2 shows the pin layout of TMPM061FWFG.

Figure 1-2 Pin Layout

1.4 Pin names and Functions

The input/output pin names and functions of the TMPM061FWFG are as follows:

Table 1-1 Pin Names and Functions Sorted by Pin (2/6)

Table 1-1 Pin Names and Functions Sorted by Pin (3/6)

Input/

Table 1-1 Pin Names and Functions Sorted by Pin (4/6)

Table 1-1 Pin Names and Functions Sorted by Pin (6/6)

Note 1: AVDD3 and VREH must be connected to power supply even if 10bit AD converter is not used.

Note 2: AVSS must be connected to GND even if the 10bit AD converter is not used.

Note 3: The same voltage must be supplied to DVDD3, AVDD3, RVDD3, DSRVDD3, SRVDD, VLC.

Note 4: When 24bit ΔΣAD converter is used, provide pin treatments as follows:

- Do not connect VREFINx to a reference voltage.

- Connect AGNDREFx to DVSS level.
- Connect a 1μF capacitor to between VREFINx and AGNDREFx.
- Note 5: When 24bit ΔΣAD converter is not used, below settings are required.
	- Connect AGNDREFx to DVSS level.

Note 6: When a temperature sensor is also not used, a reference voltage circuit requires below settings.

- Connect DSRVDD3 and SRVDD to DVDD3.
- Connect DSRVSS to DVSS.
- Note 7: Only when input is enabled, these pins tolerate 5V inputs.

Note that these pins cannot be pulled up over the power supply voltage when using as open-drain output.

1.5 Pin Numbers and Power Supply Pins

Table 1-2 Pin Numbers and Power Supplies

2. Product Information

This chapter describes the product-specific information about peripherals. Use this chapter in conjunction with Chapter of Peripherals.

2.1 16-bit Timer/Event Counter (TMRB)

TMPM061FWFG contains 2 channels of TMRB. The following are the product-specific functions:

1. Timer flip-flop output used as a capture trigger

An output of 16-bit timer A(TMR16A) can be used as a capture trigger for TMRB.

- T16A3OUT \rightarrow TMRB0
- \cdot T16A6OUT \rightarrow TMRB1
- 2. Start trigger in the timer synchronous mode

A synchronous start of multiple channels is possible.

• TMRB0 \rightarrow Start TMRB0 and TMRB1 simultaneously.

Table 2-1 describes the difference of TMRB.

Table 2-1 Difference of TMRB according to channels (n: channel number)

2.2 16-bit Timer A (TMR16A)

TMPM061FWFG contains 7 channels of TMR16A. Channel 4 does not provide a match interrupt signal and output signal to the rectangular wave pin.

[Table 2-2](#page-33-0) shows a list of rectangular wave output pins of TMR16A.

2.3 Serial Channel (SIO/UART)

2.3 Serial Channel (SIO/UART)

TMPM061FWFG contains 4 channels of SIO. The following are the product-specific functions:

1. Timer output used as a transfer clock

In the UART mode, a timer output can be used as a transfer clock.

- T16A0OUT \rightarrow SIO0, SIO1
- \cdot T16A1OUT \rightarrow SIO2, SIO3
- 2. Timer output used as carrier pulses

A timer flip-flop output can be used for transmission using carrier pulses. However, SIO3 does not have this function.

 \cdot T16A4OUT \rightarrow SIO0, SIO1, SIO2

Table 2-3 describes the difference of SIO.

SIO3 can choose a port to use. Either port must be valid. SIO3 does not have a function of IROUTn (carrier pulse transmission with data).

TOSHIBA

2.4 Analog/Digital Converter (ADC)

2.4.1 Non-Usable Functions

In the TMPM061FWFG, the following ADC functions cannot be used. Do not set the related registers.

2.4.2 Conversion Channel

In the TMPM061FWFG, 4 channels from 0 to 3 are used as input channels of the AD converter. Analog signals input to each channel are as follows.

Conversion channels are specified with ADMOD0<SCAN>, ADMOD1<ADSCN> and <ADCH>. Table 2-5 describes available settings.

2.5 ΔΣ Analog/Digital Converter (DSADC)

TMPM061FWFG contains 3 units of DSADC.

In the synchronous start function of DSADC, the following table is an assignment of a master unit and slave unit.

Table 2-6 Master/slave assignment

Master	Slave
Unit 0	Unit 1 Unit 2
3. Processor Core

The TMPM061FWFG series has a high-performance 32-bit processor core (the Arm Cortex-M0 processor core). For information on the operations of this processor core, please refer to the "Cortex-M0 Technical Reference Manual" issued by Arm Limited. This chapter describes the functions unique to the TMPM061FWFG series that are not explained in that document.

3.1 Information on the processor core

The following table shows the revision of the processor core in the TMPM061FWFG.

Refer to the detailed information about the CPU core and architecture, refer to the Arm manual "Cortex-M series processors" in the following URL:

http://infocenter.arm.com/help/index.jsp

3.2 Configurable Options

The Cortex-M0 core has optional blocks. The following tables shows the configurable options in the TMPM061FWFG.

Note: The fast multiplier provides a 32-bit × 32-bit multiply that yields the least-significant 32-bits.

3.3 Exceptions/ Interruptions

Exceptions and interruptions are described in the following section.

3.3.1 Number of Interrupt Inputs

The number of interrupt inputs can optionally be defined in the Cortex-M0 core.

TMPM061FWFG has 32 interrupt inputs.

3.3.2 SysTick

TMPM061FWFG has a SysTick timer which can generate SysTick exception.

For the detail of SysTick exception, refer to the section of "SysTick" in the exception and the register of SysTick in the NVIC register

3.3.3 SYSRESETREQ

The Cortex-M0 core outputs SYSRESETREQ signal when <SYSRESETREQ> bit of Application Interrupt and Reset Control Register are set.

TMPM061FWFG provides the same operation when SYSRESETREQ signal are output.

Note:The reset operation by <SYSRESETREQ> can not used while in SLOW mode.

3.3.4 LOCKUP

When irreparable exception generates, the Cortex-M0 core outputs LOCKUP signal to show a serious error included in software.

TMPM061FWFG does not use this signal. To return from LOCKUP status, it is necessary to use non-maskable interruput (NMI) or reset.

3.4 Events

The Cortex-M0 core has event output signals and event input signals. An event output signal is output by SEV instruction execution. If an event is input, the core returns from low-power consumption mode caused by WFE instruction.

TMPM061FWFG does not use event output signals and event input signals. Please do not use SEV instruction and WFE instruction.

3.5 Power Management

The Cortex-M0 core provides power management system which uses SLEEPING signal and SLEEPDEEP signal. SLEEPDEEP signals are output when <SLEEPDEEP> bit of System Control Register is set.

These signals are output in the following circumstances:

-Wait-For-Interrupt (WFI) instruction execution

-Wait-For-Event (WFE) instruction execution

-the timing when interrupt-service-routine (ISR) exit in case that <SLEEPONEXIT> bit of System Control Register is set.

TMPM061FWFG does not use SLEEPDEEP signal so that <SLEEPDEEP> bit must not be set. And also event signal is not used so that please do not use WFE instruction.

For detail of power management, refer to the Chapter "Clock/Mode control."

- 3. Processor Core
- 3.5 Power Management

4. Memory Map

4.1 Memory map

The memory maps for theTMPM061FWFG are based on the Arm Cortex-M0 processor core memory map.

The internal ROM is mapped to the code of the Cortex-M0 core memory, the internal RAM is mapped to the SRAM region and the special function register (SFR) is mapped to the peripheral region respectively.

The special function register (SFR) indicates I/O ports and control registers for the peripheral function. TMPM061FWFG has bit-band feature equivalent to Cortex-M3 and the SRAM and SFR regions of TMPM061FWFG are all included in the bit-band region.

The CPU register region is the processor core's internal register region.

For more information on each region, see the "Cortex-M0 Technical Reference Manual".

Note that access to regions indicated as "Fault" causes a hard fault. Do not access the vendor-specific region and the reserved region.

[Figure 4-1](#page-41-0) shows the memory map of the TMPM061FWFG.

Figure 4-1 Memory Map

4.2 Bus Structure

TMPM061FWFG contains the Cortex-M0 core and PCE acting as a bus master.

Slaves are the built-in ROM, built-in RAM0/1, built-in BOOT ROM, bridge0/1 from AHB-Lite bus to APB bus and bridge 0/1 from AHB-Lite bus to IO bus. Bridges numbered 0 are allocated to SFR0 area and bridges numbered 1 are allocated to SFR1 area. Peripheral functions are connected either the APB bus or IO bus via bridges.

An access to slaves from the core or PCE is executed simultaneously as long as the access is not to the same slave.

Note:PCE cannot access to the built-in BOOT ROM.

Figure 4-2 shows a bus structure.

Figure 4-2 Bus structure

4.3 Address lists of peripheral functions

Base addresses of the peripheral functions are shown below. Be careful of which area of SFR0 or SFR1 the peripheral function is allocated.

Do not access to addresses in the SFR area except control registers. For details of control registers, refer to Chapter of each peripheral functions.

5. Reset Operation

The following are sources of reset operation.

- \cdot RESET pin (RESET)
- Watch-dog timer (WDT)
- Application interrupt by CPU and a signal from the reset register bit <SYSRESETREQ>

To recognize a source of reset, check CGRSTFG in the clock generator register described in chapter of "Exception".

A reset by WDT is refer to the chapter on the "Watch-dog timer".

A reset by <SYSRESETREQ> is referred to "Cortex-M0 Technical Reference Manual".

Note 1: Once reset operation is done, internal RAM data is not assured. Note 2: In the SLOW mode, do not use a reset by <SYSRESETREQ>.

5.1 Cold Reset

When turning-on power, RESET pin must be kept "Low".

When turning-on power, it is necessary to take a stable time of built-in regulator into consideration. In the TMPM061FWFG, the internal regulator requires at least approximately 1ms to be stable. At cold reset, RESET pin must be kept "Low" for a duration of time sufficiently long enough for the internal regulator to be stable. Approximately 1.6ms after RESET pin becomes "High", internal reset will be released.

Figure 5-1 Cold Reset Operation Sequence

Note: The above sequence is applied as well when restoring power.

5.2 Warm Reset

To do reset TMPM061FWFG, the following conditions are required; power supply voltage is in the operational range; $\overline{\text{RESET}}$ pin is kept "Low" at least for 12 internal high-speed clocks. Approximately 1.6ms after $\overline{\text{RESET}}$ pin becomes "High", internal reset will be released.

In case of WDT reset or <SYSRESETREQ> reset, internal reset will be released approximately 30 internal highspeed clocks after reset.

5.3 After reset

All of the control register of the internal core and the peripheral function control register (SFR) are initialized by reset.

When reset is released, TMPM061FWFG starts operation by a clock of internal high-speed oscillator. External clock should be set if necessary.

6. Clock/Mode control

6.1 Features

The clock/mode control block enables to select clock gear, prescaler clock and warm-up of the oscillator. There is also the low power consumption mode which can reduce power consumption by mode transitions. This chapter describes how to control clock operating modes and mode transitions.

The clock/mode control block has the following functions:

- ・ Controls the system clock
- ・ Controls the prescaler clock
- ・ Controls the warm-up timer

In addition to NORMAL mode, the TMPM061FWFG can operate in variety of low power modes to reduce power consumption according to its usage conditions.

6.2 Registers

6.2.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

6.2.2 CGSYSCR (System control register)

Note:You cannot select fc/16 when the SysTick timer is being used.

6.2.3 CGOSCCR (Oscillation control register)

Note 1: Refer to Section ["6.3.4 Warm-up function"](#page-54-0) about the Warm-up setup.

Note 2: the external high-speed clock input $(f_{\sf EHCLKIN})$ cannot be used as warm-up clock.

Note 3: When using internal high-speed oscillator (IHOSC), do not use it as system clock which high accuracy assurance is required.

6.2.4 CGSTBYCR (Standby control register)

6.2 Registers

6.2.5 CGEHCLKSEL (External high-speed clock select register)

6.2.6 CGCKSEL (System clock selection register)

6.3 Clock control

6.3.1 Clock Type

Each clock is defined as follows:

The high-speed clock fc and the prescaler clock φT0 are dividable as follows.

6.3.2 Initial Values after Reset

Reset operation initializes the clock configuration as follows.

Reset operation causes all the clock configurations to be the same as f_{IHOSC}.

 $fc = f_{HOSC}$ $fsys = fc (=f_{IHOSC})$ $fperiph = fc (=f_{IHOSC})$ φ T0 = fperiph (= f_{HOSC})

6.3.3 Clock system Diagram

Figure 6-1 shows the clock system diagram.

The input clocks to selector shown with an arrow are set as default after reset.

Figure 6-1 Clock Block Diagram

6.3.4 Warm-up function

The warm-up function secures the stability time for the oscillator with the warm-up timer. When using stable external clock, warm-up function is not necessary.

The warm-up function is also used when returning from STOP/SLEEP mode. In this case, an interrupt for returning from the low power consumption mode triggers the automatic timer count. After the specified time is reached, the system clock is output and the CPU starts operation.

Note:Transition to the low power consumption mode while the warm-up timer is operating is prohibited.

How to configure the warm-up function.

1. Specify the count up clock

Specify the count up clock for the warm-up counter in the CGOSCCR <WUPSEL2> <WU-PSEL1> bit.

Note: the external high-speed clock input (f_{EHCLKN}) cannot be used as a warm-up clock. To use the f_{EHCLKN} as a system clock, specify the f_{lOSC}(internal high-speed oscillation) as a warm-up clock. In this case, you cannot stop the f_{losc}

2. Specify the warm-up counter value

CGOSCCR<WUPT[11:0]><WUPTL[1:0]> uses <WUPT> and <WUPTL> for counting with a low-speed clock and <WUPT> for a high-speed clock. Set "00" to <WUPTL> for using a highspeed clock.

Setting values can be calculated using the formula shown below and round off the lower four bits.

Note: Setting warm-up count value to CGOSCCR<WUPT><WUPTL>, wait until this value is reflected, then transit to standby mode by executing a command "WFI".

Note: The warm-up timer operates according to the oscillation clock, and it may contain errors if there is any fluctuation in the oscillation frequency. Therefore, the warm-up time should be taken as approximate time.

<example 1> When using high-speed oscillator 8MHz, and set warm-up time 5ms.

Round lower 4 bit off, set 0x9C4 to CGOSCCR<WUPT[11:0]>.

3. confirm the start and completion of warm-up

The CGOSCCR<WUEON><WUEF> is used to confirm the start and completion of warm-up through software (instruction). When CGOSCCR<WUEON> is set to "1", the warm-up start a count up. The completion of warm-up can be confirmed with CGOSCCR<WUEF>.

The example of warm-up function setup.

Table 6-1 <example> from STOP mode to NORMAL mode transition (internal high-speed oscillator is selected)

$CGOSCCR = "0x9C4"$: Specify the warm-up time				
CGOSCCR <wupt> read</wupt>	: Confirm warm-up time reflecting Repeat until the read data is "0x9C4".				
$CGOSCCR < XEN2 > = "1"$: high-speed oscillator (fosc) enable				
$CGOSCCR$: Start the warm-up timer				
CGOSCCR <wuef>read</wuef>	: Wait until the state becomes "0" (warm-up is finished)				

6.3.5 System clock

One of the following clocks can be used as a source clock of the system clock: internal high-speed oscillation clock, external high-speed oscillation clock (connected to an oscillator or a clock input) and external lowspeed clock (connected to an oscillator or a clock input).

Internal high-speed oscillation should not be used if high accuracy assurance is required.

The system clock can be divided by CGSYSCR<GEAR> when using a high-speed oscillator. Although the settings can be changed while operating, the actual switching takes place after a slight delay.

Operating frequency examples configured by clock gear settings are shown in Table 6-2.

Table 6-2 System clock frequency

fosc frequency (MHz)	Clock gear (CG)							
	1/1	1/2	1/4	1/8	1/16			
8	8		2					
10	10	5	2.5	1.25				
16	16	8		\mathcal{P}				

↑ Initial value after reset

Note:Do not use 1/16 when using SysTick.

TOSHIBA

6.3.5.1 System clock switching

The Figure 6-2 shows how to switch the system clock.

Figure 6-2 System clock switching

The internal high-speed oscillation clock is set as the system clock after releasing the reset signal. A high-speed clock can be selected from the internal high-speed oscillation clock, the external high-speed oscillation clock and external high-speed clock. A low-speed clock can be selected from the external lowspeed oscillation clock and the external low-speed clock. A clock cannot be switched between the external high-speed oscillation clock and external high-speed clock, and also between the external low-speed oscillation clock and external low-speed clock.

6.3.5.2 Precautions for transiting to SLOW, STOP and SLEEP mode

To transit to SLOW mode, STOP mode or SLEEP mode, switch the system clock to the internal highspeed oscillation clock by setting CGOSCCR<OSCSEL> first. While using the external high-speed oscillation clock or the external high-speed clock, mode cannot be switched to STOP or SLEEP.

6.3.5.3 Clock setting

The system clock can be selected by setting the CGOSCCR and CGEHCLKSEL. After selecting the clock, set the clock gear by setting the CGSYSCR if required.

The clock setup sequence is shown as below.

If a low-speed clock is set as the system clock, set the low-speed clock in NORMAL mode first and then shift to the SLOW mode.

TOSHIBA

2.High-speed clock setting sequence (Switch from external to internal high-speed clock)

6.3.6 Prescaler Clock Control

Peripheral IO (TMRB,SIO) has a prescaler for dividing a clock. As the clock φT0 to be input to each prescaler, the "fperiph" clock specified in the CGSYSCR<FPSEL0> can be divided according to the setting in the CGSYSCR<PRCK[2:0]>. After the controller is reset, fperiph/1 is selected as φT0.

When the clock fs is being used for the system clock fsys, fs can be used for the prescaler clock by setting "1" to CGSYSCR<FPSEL1>.

Note:To use the clock gear, ensure that you make the time setting such that prescaler output φTn from each peripheral function is slower than fsys (φTn < fsys). Do not switch the clock gear while the timer counter or other peripheral function is operating.

6.3.7 System Clock Pin Output Function

The TMPM061FWFG enables to output the system clock from a pin. The SCOUT pin can output the system clock fsys and fsys/2, and the prescaler input clock for peripheral I/O φT0.

Note:The phase difference (AC timing) between the system clock output by the SCOUT and the internal clock is not guaranteed.

By setting the port J registers, the PJCR<PD4C> and PJFR2<PJ4F2> to "1", the PJ4 pin becomes the SCOUT output pin. The output clock is selected by setting the CGSYSCR<SCOSEL[1:0]>.

[Table 6-3](#page-60-0) shows the pin status in each mode when the SCOUT pin is set to the SCOUT output.

Table 6-3 SCOUT Output Status in Each Mode

Note 1: The phase difference (AC timing) between the system clock output by the SCOUT and the internal clock is not guaranteed.

Note 2: If fsys is selected for SCOUT, output waveforms may undergo distortion when the clock gear is switched.

6.4 Modes and Mode Transitions

6.4.1 Mode Transitions

The NORMAL mode and the SLOW mode use the high-speed and low-speed clocks for the system clock respectively.

The IDLE, SLEEP and STOP modes can be used as the low power consumption mode that enables to reduce power consumption by halting processor core operation.

When the low-speed clock is not used, the SLOW and SLEEP modes cannot be used.

Figure 6-3 shows a mode transition diagram.

For a detail of sleep-on-exit, refer to "Cortex-M0 Technical Reference Manual."

Figure 6-3 Mode Transition Diagram

Note:Operation mode cannot be shifted to STOP or SLEEP mode when using the external high-speed oscillation clock (f_{EHOSC}) and the external high-speed clock ($f_{EHCLKIN}$). Switch the clock to the internal highspeed oscillation clock (f_{HOSC}) first.

6.5 Operation mode

Two operation modes, NORMAL and SLOW, are available. The features of each mode are described below.

6.5.1 NORMAL mode

This mode is to operate the CPU core and the peripheral hardware by using the high-speed clock.

It is shifted to the NORMAL mode after reset. The low-speed clock can also be used.

6.5.2 SLOW mode

This mode is to operate the CPU core and the peripheral hardware by using the low-speed clock with highspeed clock stopped. The SLOW mode reduces power consumption compared to the NORMAL mode.

Operable peripheral functions are limited in SLOW mode. The following peripheral functions are operable: I/O port (PORT), Power Calculation Engine (PCE), Timer (TMRB and TMR16A), watchdog timer (WDT), real-time clock (RTC), LCD driver, Low Voltage Detect (LVD) and temperature sensor (TEMP).

Note 1: In SLOW mode, stop all functions other than operable peripheral functions before shifting to SLOW mode. Note 2: In the slow mode, be sure not to perform reset using the Application Interrupt and Reset Control Register<SYSRESETREG> of the Cortex-M0 NVIC register.

6.6 Low Power Consumption Modes

The TMPM061FWFG has three low power consumption modes: IDLE, SLEEP and STOP. To shift to the low power consumption mode, specify the mode in the system control register CGSTBYCR<STBY[2:0]> and execute the WFI (Wait For Interrupt) instruction. In this case, execute reset or generate the interrupt to release the mode. Releasing by the interrupt requires settings in advance. See the chapter "Exceptions" for details.

- Note 1: The TMPM061FWFG does not offer any event for releasing the low power consumption mode. Transition to the low power consumption mode by executing the WFE (Wait For Event) instruction is prohibited.
- Note 2: The TMPM061FWFG does not support the low power consumption mode configured with the SLEEPDEEP bit in the Cortex-M0 core. Setting the <SLEEPDEEP> bit of the system control register is prohibited.
- Note 3: Transition to the low power consumption mode while the warm-up timer is operating is prohibited.
- Note 4: The operation mode cannot be shifted to STOP mode or SLEEP mode when using the external high-speed oscillation clock (f_{EHOSC}) and the external high-speed clock ($f_{EHOLKIN}$). Switch the clock to the internal highspeed oscillation clock (f_{IHOSC}) first.
- Note 5: To use the LCD driver in SLEEP mode, operation mode needs to be shifted from SLOW mode to SLEEP mode.

The features of IDLE, SLEEP, STOP mode are described as follows.

6.6.1 IDLE mode

This is a mode that the clock of the processor core stops. Each peripheral function has one bit in its control register for enabling or disabling operation in the IDLE mode. When the IDLE mode is entered, peripheral functions for which operation in the IDLE mode is disabled stop operation and hold the state at that time.

The following peripheral functions can be enabled or disabled in the IDLE mode. For setting details, see the chapter on each peripheral function.

- ・ 16-bit timer/event counter (TMRB)
- 16-bit timer A (TMR16A)
- ・ Serial channel (SIO/UART)
- ・ Serial bus interface (I2C/SIO)
- ・ Analog Digital converter (ADC)
- ・ Watch dog timer (WDT)

6.6.2 SLEEP mode

This is a mode that a high-speed clock stops. A real time clock (RTC) counter that operates with lowspeed clock signals keep operation.

By releasing the SLEEP mode, the device returns to the preceding mode of the SLEEP mode and starts operation.

6.6.3 STOP mode

All the clocks including an internal oscillator stop in STOP mode.

By releasing the STOP mode, the device returns to the preceding mode of the STOP mode and starts operation. Set 200μs or more as the warming up time after releasing the STOP mode.

The STOP mode enables to select the pin status by setting the CGSTBYCR<DRVE>. [Table 6-4](#page-64-0) shows the pin status in the STOP mode.

Table 6-4 Pin States in the STOP mode

ο : Valid input or output.

× : Invalid input or output.

Note:x: port number / m: corresponding bit

6.6.4 Low power Consumption Mode Setting

The low power consumption mode is specified by the setting of the standby control register CGSTBYCR<STBY[2:0]>.

Table 6-5 shows the mode setting in the <STBY[2:0]>.

Note:Do not set any value other than those shown above in <STBY[2:0]>.

6.6.5 Operational Status in Each Mode

Table 6-6 show the operational status in each mode.

Table 6-6 Operational Status in Each Mode

	NORMAL			IDLE			SLOW		SLEEP		
Block	Internal high- speed os- cillator use (f _{IH} OSC)	External high- speed os- cillator use (f _{EHOSC})	External high- speed clock use (f _{EHCLKIN})	Internal high- speed os- cillator use (f _{IH} osc)	External high- speed os- cillator use (f _{EH} _{OSC})	External high- speed clock use (f _{EHCLKIN})	External low-speed oscillator use (f_{ELOSC})	External low-speed clock use (f_{ELCLKIN})	External low-speed oscillator use (f_{ELOSC})	External low-speed clock use (f _{ELCLKIN})	STOP
Internal high- speed oscillator	\circ	Δ	\circ (Note 1)	\circ	Δ	o (Note 1)	Δ (Note 2)				
External high- speed oscillator	Δ	o	Δ	Δ	o	Δ	Δ (Note 2)				
External low- speed oscillator	Δ	Δ	Δ	Δ	Δ	Δ	\circ	Δ	$\mathsf{o}\,$	Δ	
CG	\circ		\circ			\circ		\circ			
Processor core	o		$\overline{}$		o (Note 6)		$\qquad \qquad -$		$\overline{}$		
Power Calcula- tion Engine	o		o		o						
IO port		o			o		\circ		$\mathsf{o}\,$		o(Note 3)
SIO/UART	o		Δ		#(Note 5)		\equiv		$\overline{}$		
I2C/SIO	o		Δ		#(Note 5)		$\overline{}$		$\overline{}$		
TMRB	o		Δ		$\mathsf{o}\xspace$		$\overline{}$		$\qquad \qquad -$		
TMR16A	o		Δ		$\mathsf{o}\xspace$		$\qquad \qquad -$		$\qquad \qquad -$		
WDT	o		$\Delta(Note 7)$		\circ		\equiv				
ADC	o		Δ		#(Note 5)		$\qquad \qquad -$		$\qquad \qquad -$		
ΔΣΑDC	\circ		\circ		#(Note 5)		$\overline{}$		$\qquad \qquad -$		
RTC	o		\circ		$\mathsf{o}\xspace$		\circ		$\overline{}$		
LCD driver	o		$\mathsf{o}\xspace$		$\mathsf{o}\xspace$		o(Note 4)		$\overline{}$		
Temperature sensor	o		o		o		o				
LVD o			o		o		o		o		

ο : Operation is available when in the target mode.

− : The clock to module stops automatically when transiting to the target mode.

: To Stop the function is required when transiting to the target mode.

Δ : Enables to select disabling module operation by software when in the target mode.

Note 1: When an external high-speed clock is used, an internal high-speed clock cannot be stopped.

- Note 2: The high-speed oscillator (EHOSC and IHOSC) does not automatically stop when the operation mode shifted from NOR-MAL mode to SLOW mode. Set the CGOSCCR<XEN1> or <XEN2> to stop the oscillation. Before the transition from SLOW mode to NORMAL mode is made, set CGOSCCR<XEN1> or <XEN2> to start oscillation, since the highspeed oscillators (EHOSC, IHOSC) do not automatically start oscillation.
- Note 3: This setting is depending on the CGSTBYCR<DRVE> setting.
- Note 4: To use the LCD driver in SLEEP mode, the operation mode needs to be shifted from SLOW mode to SLEEP mode.

Note 5: Stop ADC, ΔΣADC, SIO/UART and I2C/SIO in SLOW mode.

Note 6: SysTick timer cannot be used in the SLOW mode.

Note 7: In IDLE mode, the watchdog timer cannot be cleared by the processor core.

6.6.6 Releasing the Low Power Consumption Mode

The low power consumption mode can be released by an interrupt request, Non-Maskable Interrupt (NMI) or reset. The release source that can be used is determined by the low power consumption mode selected.

Details are shown in Table 6-7.

ο : Starts the interrupt handling after the mode is released. (The reset initializes the LSI)

× : Unavailable

Note 1: When releasing from low power consumption mode by interrupting level mode, hold the level until the interrupt handling starts. If the level is changed before that, the correct interrupt handling cannnot be started.

Note 2: For shifting to the low power consumption mode, set the CPU to prohibit all the interrupts other than the release source. If not, releasing may be executed by an unspecified interrupt.

・ Release by interrupt request

To release the low power consumption mode by an interrupt, the CPU must be set in advance to detect the interrupt. In addition to the setting in the CPU, the clock generator must be set to detect the interrupt to be used to release the STOP mode.

Release by SysTick interrupt

SysTick interrupt can only be used in the IDLE mode.

Release by Non-Maskable Interrupt (NMI)

INTWDT can only be used in the IDLE mode.

Release by reset

Any low power consumption mode can be released by reset from the RESET pin. After that, the mode switches to the NORMAL mode and all the registers are initialized as is the case with normal reset.

Refer to "Interrupts" for details.

6.6.7 Warm-up

Warm-up may be required for stable oscillation of internal oscillator in the mode transition.

In the mode transition from STOP to the NORMAL/SLOW or from SLEEP to NORMAL, the warm-up counter is activated automatically. And then the system clock output is started after the elapse of warm-up time. It is necessary to set a warm-up time in the CGOSCCR<WUPSEL2><WUPSEL1> and to set the warm-up time in the CGOSCCR<WUPT><WUPTL> before executing the instruction to enter the STOP/SLEEP mode.

Note: The external high-speed clock input (fEHCLKIN) cannot be used as a warm-up clock.

In the transition from NORMAL to SLOW/SLEEP, the warm-up is required so that the internal oscillator to stabilize if the low-speed oscillator is disabled. Enable the low-speed oscillator and then activate the warmup by software.

In the transition from SLOW to NORMAL when the high-speed oscillator is disabled, enable the highspeed oscillator and then activate the warm-up.

Table 6-8 shows whether the warm-up setting of each mode transition is required or not.

Table 6-8 Warm-up setting in mode transition

Note 1: If the low-speed oscillator is disabled, enable the low-speed oscillator and then activate the warm-up by software. Note 2: If the high-speed oscillator is disabled, enable the high-speed oscillator and then activate the warm-up by software. Note 3: Set 200μs or more as the warming up time after releasing the STOP mode.

6.6.8 Clock Operations in Mode Transition

The clock operations in mode transition are described as follows.

6.6.8.1 Transition of operation modes: NORMAL \rightarrow STOP \rightarrow NORMAL

When returning to the NORMAL mode from the STOP mode, the warm-up is activated automatically. It is necessary to set the warm-up time to CGOSCCR<WUPT> and select clock-source that is same as CGOSCCR<OSCSEL> to <WUPSEL2><WUPSEL1> before entering the STOP mode.

Note:The operation mode cannot be shifted to STOP mode or SLEEP mode when using the external highspeed oscillation clock (f_{EHOSC}) and the external high-speed clock (f_{EHCLKIN}). Switch the clock to the internal high-speed oscillation clock (f_{HOSC}) first.

Returning to the NORMAL mode by reset does not induce the automatic warm-up. Keep the reset signal asserted until the oscillator operation becomes stable.

6.6.8.2 Transition of operation modes: NORMAL \rightarrow SLEEP \rightarrow NORMAL

When returning to the NORMAL mode from the SLEEP mode, the warm-up is activated automatically. It is necessary to set the warm-up time to CGOSCCR<WUPT> and select clock-source that is same as CGOSCCR<OSCSEL> to <WUPSEL2><WUPSEL1> before entering the SLEEP mode.

Note:The operation mode cannot be shifted to STOP mode or SLEEP mode when using the external highspeed oscillation clock (f_{FHOSC}) and the external high-speed clock ($f_{FHCI KIN}$). Switch the clock to the internal high-speed oscillation clock (f_{IHOSC}) first.

Returning to the NORMAL mode by reset does not induce the automatic warm-up. Keep the reset signal asserted until the oscillator operation becomes stable.

6.6.8.3 Transition of operation modes: $SLOW \rightarrow STOP \rightarrow SLOW$

The warm-up is activated automatically. It is necessary to set the warm-up time before entering the STOP mode.

6.6.8.4 Transition of operation modes: $SLOW \rightarrow SLEEP \rightarrow SLOW$

TOSHIBA

The low-speed clock continues oscillation in the SLEEP mode. There is no need to make a warm-up setting.

- 6. Clock/Mode control
- 6.6 Low Power Consumption Modes
7. Exceptions

This chapter describes features, types and handling of exceptions.

Exceptions have close relation to the CPU core. Refer to "Cortex-M0 Technical Reference Manual" if needed.

7.1 Overview

An exception causes the CPU to stop the currently executing process and handle another process.

There are two types of exceptions: those that are generated when some error condition occurs or when an instruction to generate an exception is executed; and those that are generated by hardware, such as an interrupt request signal from an external pin or peripheral function.

All exceptions are handled by the Nested Vectored Interrupt Controller (NVIC) in the CPU according to the respective priority levels. When an exception occurs, the CPU stores the current state to the stack and branches to the corresponding interrupt service routine (ISR). Upon completion of the ISR, the information stored to the stack is automatically restored.

7.1.1 Exception Types

The following types of exceptions exist in the Cortex-M0.

For detailed descriptions on each exception, refer to "Cortex-M0 Technical Reference Manual".

- ・ Reset
- Non-Maskable Interrupt (NMI)
- ・ Hard Fault
- ・ SVCall (Supervisor Call)
- **PendSV**
- **SysTick**
- ・ External Interrupt

7.1.2 Handling Flowchart

```
The following shows how an exception/interrupt is handled. In the following descriptions,
indicates hardware handling. Indicates software handling.
```
Each step is described later in this chapter.

7.1.2.1 Exception Request and Detection

(1) Exception occurrence

Exception sources include instruction execution by the CPU, memory accesses, and interrupt requests from external interrupt pins or peripheral functions.

An exception occurs when the CPU executes an instruction that causes an exception or when an error condition occurs during instruction execution.

An exception also occurs by an instruction fetch from the Execute Never (XN) region or an access violation to the Fault region.

An interrupt request is generated from an external interrupt pin or peripheral function.For interrupts that are used for releasing a standby mode, relevant settings must be made in the clock generator.For details, refer to ["7.5 Interrupts"](#page-78-0).

OSHIBA

(2) Exception detection

If multiple exceptions occur simultaneously, the CPU takes the exception with the highest priority.

Table 7-1 shows the priority of exceptions. "Configurable" means that you can assign a priority level to that exception.

No.	Exception type	Priority	Description		
1	Reset	-3 (highest)	Reset pin, WDT or SYSRETREQ		
2	Non-Maskable Interrupt	-2	NMI pin or WDT		
3	Hard Fault	-1	Fault that cannot activate because a higher-priority fault is being han- dled or it is disabled		
$4 - 10$	Reserved	$\overline{}$			
11	SVCall	Configurable	System service call with SVC instruction		
$12 - 13$	Reserved	$\overline{}$			
14	PendSV	Configurable	Pendable system service request		
15	SysTick	Configurable	Notification from system timer		
$16 -$	External Interrupt	Configurable	External interrupt pin or peripheral function (Note 2)		

Table 7-1 Exception Types and Priority

Note:**External interrupts have different sources and numbers in each product. For details, see ["7.5.1.5 List of Interrupt Sources"](#page-80-0).**

(3) Priority setting

The external interrupt priority is set to the interrupt priority register and other exceptions are set to <PRI_n> bit in the system handler priority register.

The configuration of $\leq PRI_n$ is two bit, so the priority can be configured in the range from 0 to 3. The highest priority is "0". If multiple elements with the same priority exist, the smaller the number, the higher the priority becomes.

7.1.2.2 Exception Handling and Branch to the Interrupt Service Routine (Pre-emption)

When an exception occurs, the CPU suspends the currently executing process and branches to the interrupt service routine. This is called "pre-emption".

(1) Stacking

When the CPU detects an exception, it pushes the contents of the following eight registers to the stack in the following order:

- ・ Program Counter (PC)
- ・ Program Status Register (xPSR)
- \cdot r0 r3
- \cdot r12
- ・ Link Register (LR)

The SP is decremented by eight words by the completion of the stack push.The following shows the state of the stack after the register contents have been pushed.

(2) Fetching an ISR

The CPU enables instruction to fetch the interrupt processing with data store to the register.

Prepare a vector table containing the top addresses of ISRs for each exception.

The vector table should also contain the initial value of the main stack.

(3) Late-arriving

If the CPU detects a higher priority exception before executing the ISR for a previous exception, the CPU handles the higher priority exception first. This is called "late-arriving".

A late-arriving exception causes the CPU to fetch a new vector address for branching to the corresponding ISR, but the CPU does not newly push the register contents to the stack.

(4) Vector table

The vector table is configured as shown below.

You must always set the first four words (stack top address, reset ISR address, NMI ISR address, and Hard Fault ISR address).Set ISR addresses for other exceptions if necessary.

7.1.2.3 Executing an ISR

An ISR performs necessary processing for the corresponding exception. ISRs must be prepared by the user.

An ISR may need to include code for clearing the interrupt request so that the same interrupt will not occur again upon return to normal program execution.

For details about interrupt handling, see ["7.5 Interrupts".](#page-78-0)

If a higher priority exception occurs during ISR execution for the current exception, the CPU abandons the currently executing ISR and services the newly detected exception.

7.1.2.4 Exception exit

(1) Execution after returning from an ISR

When returning from an ISR, the CPU takes one of the following actions:

・ Tail-chaining

If a pending exception exists and there are no stacked exceptions or the pending exception has higher priority than all stacked exceptions, the CPU returns to the ISR of the pending exception.

In this case, the CPU skips the pop of eight registers and push of eight registers when exiting one ISR and entering another. This is called "tail-chaining".

Returning to the last stacked ISR

If there are no pending exceptions or if the highest priority stacked exception is of higher priority than the highest priority pending exception, the CPU returns to the last stacked ISR.

Returning to the previous program

If there are no pending or stacked exceptions, the CPU returns to the previous program.

(2) Exception exit sequence

When returning from an ISR, the CPU performs the following operations:

Pop eight registers

Pops the eight registers (PC, xPSR, r0 to r3, r12 and LR) from the stack and adjust the SP.

Load current active interrupt number

Loads the current active interrupt number from the stacked xPSR. The CPU uses this to track which interrupt to return to.

Select SP

If returning to an exception (Handler Mode), SP is SP_main. If returning to Thread Mode, SP can be SP_main or SP_process.

7.2 Reset Exceptions

Reset exceptions are generated from the following three sources.

Use the Reset Flag (CGRSTFLG) Register of the Clock Generator to identify the source of a reset.

・ External reset pin

A reset exception occurs when an external reset pin changes from "Low" to "High".

・ Reset exception by WDT

The watchdog timer (WDT) has a reset generating feature. For details, see the chapter on the WDT.

・ Reset exception by SYSRESETREQ

A reset can be generated by setting the SYSRESETREQ bit in the NVIC's Application Interrupt and Reset Control Register.

7.3 Non-Maskable Interrupts (NMI)

The watchdog timer (WDT) has a non-maskable interrupt generating feature. For details, see the chapter on the WDT.

7.4 SysTick

SysTick provides interrupt features using the CPU's system timer.

When you set a value in the SysTick Reload Value Register and enable the SysTick features in the SysTick Control and Status Register, the counter loads with the value set in the Reload Value Register and begins counting down. When the counter reaches "0", a SysTick exception occurs. You may be pending exceptions and use a flag to know when the timer reaches "0".

Note:In this product, the systick timer counts based on fosc which is selected by the bits <OSCSEL> of the register CGOSCCR.

7.5 Interrupts

This chapter describes routes, sources and required settings of interrupts.

The CPU is notified of interrupt requests by the interrupt signal from each interrupt source.

It sets priority on interrupts and handles an interrupt request with the highest priority.

Interrupt requests for clearing a standby mode are notified to the CPU via the clock generator. Therefore, appropriate settings must be made in the clock generator.

7.5.1 Interrupt Sources

7.5.1.1 Interrupt Route

Figure 7-1 shows an interrupt request route.

The interrupts issued by the peripheral function that is not used to release standby are directly input to the CPU (route1).

The peripheral function interrupts used to release standby (route 2) and interrupts from the external interrupt pin (route 3) are input to the clock generator and are input to the CPU through the logic for releasing standby (route 4 and 5).

If interrupts from the external interrupt pins are not used to release standby, they are directly input to the CPU, not through the logic for standby release (route 6).

7.5.1.2 Generation

An interrupt request is generated from an external pin or peripheral function assigned as an interrupt source or by setting the NVIC's Interrupt Set-Pending Register.

・ From external pin

Set the port control register so that the external pin can perform as an interrupt function pin.

From peripheral function

Set the peripheral function to make it possible to output interrupt requests.

See the chapter of each peripheral function for details.

By setting Interrupt Set-Pending Register (forced pending)

An interrupt request can be generated by setting the relevant bit of the Interrupt Set-Pending Register.

7.5.1.3 Transmission

An interrupt signal from an external pin or peripheral function is directly sent to the CPU unless it is used to exit a standby mode.

Interrupt requests from interrupt sources that can be used for clearing a standby mode are transmitted to the CPU via the clock generator. For these interrupt sources, appropriate settings must be made in the clock generator in advance. External interrupt sources not used for exiting a standby mode can be used without setting the clock generator.

7.5.1.4 Precautions when using external interrupt pins

If you use external interrupts, be aware the followings not to generate unexpected interrupts.

If input disabled (PxIE<PxmIE>="0"), inputs from external interrupt pins are "High". Also, if external interrupts are not used as a trigger to release standby (route 6 of ["Figure 7-1 Interrupt Route"\)](#page-78-0), input signals from the external interrupt pins are directly sent to the CPU. Since the CPU recognizes "High" input as an interrupt, interrupts occur if corresponding interrupts are enabled by the CPU as inputs are being disabled.

To use the external interrupt without setting it as a standby trigger, set the interrupt pin input as "Low" and enable it. Then, enable interrupts on the CPU.

7.5.1.5 List of Interrupt Sources

Table 7-2 shows the list of interrupt sources.

Table 7-2 List of Interrupt Sources

7.5.1.6 Active level

The active level indicates which change in signal of an interrupt source triggers an interrupt. The CPU recognizes interrupt signals in "High" level as interrupt. Interrupt signals directly sent from peripheral functions to the CPU are configured to output "High" to indicate an interrupt request.

Active level is set to the clock generator for interrupts which can be a trigger to release standby. Interrupt requests from peripheral functions are set as rising-edge or falling-edge triggered. Interrupt requests from interrupt pins can be set as level-sensitive ("High" or "Low") or edge-triggered (rising or falling).

If an interrupt source is used for clearing a standby mode, setting the relevant clock generator register is also required. Enable the CGIMCGx<INTxEN> bit and specify the active level in the CGIMCGx EMCGx bits. You must set the active level for interrupt requests from each peripheral function as shown in [Table 7-2](#page-80-0).

An interrupt request detected by the clock generator is notified to the CPU with a signal in "High" level.

OSHIBA

7.5.2 Interrupt Handling

7.5.2.1 Flowchart

The following shows how an interrupt is handled.

7.5.2.2 Preparation

When preparing for an interrupt, you need to pay attention to the order of configuration to avoid any unexpected interrupt on the way.

Initiating an interrupt or changing its configuration must be implemented in the following order basically. Disable the interrupt by the CPU. Configure from the farthest route from the CPU. Then enable the interrupt by the CPU.

To configure the clock generator, you must follow the order indicated here not to cause any unexpected interrupt. First, configure the precondition. Secondly, clear the data related to the interrupt in the clock generator and then enable the interrupt.

The following sections are listed in the order of interrupt handling and describe how to configure them.

- 1. Disabling interrupt by CPU
- 2. CPU registers setting
- 3. Preconfiguration (1) (Interrupt from external pin)
- 4. Preconfiguration (2) (Interrupt from peripheral function)
- 5. Preconfiguration (3) (Interrupt Set-Pending Register)
- 6. Configuring the clock generator
- 7. Enabling interrupt by CPU

(1) Disabling interrupt by CPU

To make the CPU for not accepting any interrupt, write "1" to the corresponding bit of the PRI-MASK Register. All interrupts and exceptions other than non-maskable interrupts and hard faults can be masked.

Use "MSR" instruction to set this register.

(2) CPU registers setting

You can assign a priority level by writing to <PRI n > field in an Interrupt Priority Register of the NVIC register.

Each interrupt source is provided with two bits for assigning a priority level from 0 to 3. Priority level 0 is the highest priority level. If multiple sources have the same priority, the smallest-numbered interrupt source has the highest priority.

Note:"n" indicates the corresponding exceptions/interrupts.

(3) Preconfiguration (1) (Interrupt from external pin)

Set "1" to the port function register of the corresponding pin. Setting PxFRn[m] allows the pin to be used as the function pin. Setting PxIE[m] allows the pin to be used as the input port.

Note:x: port number / m: corresponding bit / n: function register number

In modes other than STOP mode, setting PxIE to enable input enables the corresponding interrupt input regardless of the PxFR setting. Be careful not to enable interrupts that are not used. Also, be aware of the description of ["7.5.1.4 Precautions when using external interrupt](#page-79-0) [pins".](#page-79-0)

(4) Preconfiguration (2) (Interrupt from peripheral function)

The setting varies depending on the peripheral function to be used. See the chapter of each peripheral function for details.

(5) Preconfiguration (3) (Interrupt Set-Pending Register)

To generate an interrupt by using the Interrupt Set-Pending Register, set "1" to the corresponding bit of this register.

Note:**m: corresponding bit**

(6) Configuring the clock generator

For an interrupt source to be used for exiting a standby mode, you need to set the active level and enable interrupts in the CGIMCG register of the clock generator. The CGIMCG register is capable of configuring each source.

Before enabling an interrupt, clear the corresponding interrupt request already held. This can avoid unexpected interrupt.To clear corresponding interrupt request, write a value corresponding to the interrupt to be used to the CGICRCG register.See ["7.6.3.3 CGICRCG\(CG Interrupt Request](#page-102-0) [Clear Register\)"](#page-102-0) for each value.

Interrupt requests from external pins can be used without setting the clock generator if they are not used for exiting a standby mode. However, an "High" pulse or "High"-level signal must be input so that the CPU can detect it as an interrupt request. Also, be aware of the description o[f"7.5.1.4 Precautions when using external interrupt pins"](#page-79-0).

Note:**n: register number / m: number assigned to interrupt source**

(7) Enabling interrupt by CPU

Enable the interrupt by the CPU as shown below.

Clear the suspended interrupt in the Interrupt Clear-Pending Register. Enable the intended interrupt with the Interrupt Set-Enable Register. Each bit of the register is assigned to a single interrupt source.

Writing "1" to the corresponding bit of the Interrupt Clear-Pending Register clears the suspended interrupt. Writing "1" to the corresponding bit of the Interrupt Set-Enable Register enables the intended interrupt.

To generate interrupts in the Interrupt Set-Pending Register setting, factors to trigger interrupts are lost if pending interrupts are cleared. Thus, this operation is not necessary.

At the end, PRIMASK register is zero cleared.

Note:m : corresponding bit

7.5.2.3 Detection by Clock Generator

If an interrupt source is used for exiting a standby mode, an interrupt request is detected according to the active level specified in the clock generator, and is notified to the CPU.

An edge-triggered interrupt request, once detected, is held in the clock generator. A level-sensitive interrupt request must be held at the active level until it is detected, otherwise the interrupt request will cease to exist when the signal level changes from active to inactive.

When the clock generator detects an interrupt request, it keeps sending the interrupt signal in "High" level to the CPU until the interrupt request is cleared in the CG Interrupt Request Clear (CGICRCG) Register. If a standby mode is exited without clearing the interrupt request, the same interrupt will be detected again when normal operation is resumed. Be sure to clear each interrupt request in the ISR.

7.5.2.4 Detection by CPU

The CPU detects an interrupt request with the highest priority.

7.5.2.5 CPU processing

On detecting an interrupt, the CPU pushes the contents of PC, PSR, r0-r3, r12 and LR to the stack then enter the ISR.

7.5.2.6 Interrupt Service Routine (ISR)

An ISR requires specific programming according to the application to be used. This section describes what is recommended at the service routine programming and how the source is cleared.

(1) Pushing during ISR

An ISR normally pushes register contents to the stack and handles an interrupt as required. The Cor-tex-M0 core automatically pushes the contents of PC, PSR, r0-r3, r12 and LR to the stack. No extra programming is required for them. Push the contents of other registers if needed.

Interrupt requests with higher priority and exceptions such as NMI are accepted even when an ISR is being executed. We recommend you to push the contents of general-purpose registers that might be rewritten.

(2) Clearing an interrupt source

If an interrupt source is used for clearing a standby mode, each interrupt request must be cleared with the CG Interrupt Request Clear (CGICRCG) Register.

If an interrupt source is set as level-sensitive, an interrupt request continues to exist until it is cleared at its source. Therefore, the interrupt source must be cleared. Clearing the interrupt source automatically clears the interrupt request signal from the clock generator.

If an interrupt is set as edge-sensitive, clear an interrupt request by setting the corresponding value in the CGICRCG register. When an active edge occurs again, a new interrupt request will be detected.

7.6 Exception/Interrupt-Related Registers

The CPU's NVIC registers and clock generator registers described in this chapter are shown below with their respective addresses.

7.6.1 Register List

7.6.2 NVIC Registers

7.6.2.1 SysTick Control and Status Register

Note:In this product, the fosc (which is selected by the bits <OSCSEL> of the register CGOSCCR) is used as the external reference clock.

7.6.2.3 SysTick Current Value Register

7.6.2.4 SysTick Calibration Value Register

Note:TMPM061FWFG does not prepare the calibration value.

7.6.2.5 Interrupt Set-Enable Register

Note:For descriptions of interrupts and interrupt numbers, see Section ["7.5.1.5 List of Interrupt Sources".](#page-80-0)

7.6.2.6 Interrupt Clear-Enable Register

Note:For descriptions of interrupts and interrupt numbers, see Section ["7.5.1.5 List of Interrupt Sources".](#page-80-0)

7.6.2.7 Interrupt Set-Pending Register

Note:For descriptions of interrupts and interrupt numbers, see Section ["7.5.1.5 List of Interrupt Sources".](#page-80-0)

7.6.2.8 Interrupt Clear-Pending Register

Note: For descriptions of interrupts and interrupt numbers, see Section ["7.5.1.5 List of Interrupt Sources".](#page-80-0)

7.6.2.9 Interrupt Priority Register

The following shows the addresses of the Interrupt Priority Registers corresponding to interrupt numbers.

Cortex-M0 core uses two bits for assigning a priority.

The following shows the fields of the Interrupt Priority Registers for interrupt numbers 0 to 3. The Interrupt Priority Registers for all other interrupt numbers have the identical fields. Unused bits return "0" when read, and writing to unused bits has no effect.

7.6.2.10 Application Interrupt and Reset Control Register

Note 1: This product can be used as the little-endian memory format only.

Note 2: When SYSRESETREQ is output, reset is performed on this product. <SYSRESETREQ> is cleared by reset.

7.6.2.11 System Handler Priority Register

The following shows the addresses of the System Handler Priority Registers corresponding to each exception.

Cortex-M0 core uses two bits for assigning a priority.

The following shows the fields of the System Handler Priority Registers for Memory Management, Bus Fault and Usage Fault. Unused bits return "0" when read, and writing to unused bits has no effect.

7.6.2.12 System Handler Control and State Register

Note:You must clear or set the active bits with extreme caution because clearing and setting these bits does not repair stack contents.

7.6.3 Clock generator registers

7.6.3.1 CGIMCGA(CG Interrupt Mode Control Register A)

Note 1: <EMSTx> is effective only when <EMCGx[2:0]> is set to "100" for both rising and falling edge. The active level used for the reset of standby can be checked by referring <EMSTx>. If interrupts are cleared with the CGICRCG register, <EMSTx> is also cleared.

Note 2: Please specify the bit for the edge first and then specify the bit for the <INTxEN>. Setting them simultaneously is prohibited.

7.6.3.2 CGIMCGB(CG Interrupt Mode Control Register B)

Note 1: <EMSTx> is effective only when <EMCGx[2:0]> is set to "100" for both rising and falling edge. The active lev**el** used for the reset of standby can be checked by referring <EMSTx>. If interrupts are cleared with the CGICRCG register, <EMSTx> is also cleared.

Note 2: Please specify the bit for the edge first and then specify the bit for the <INTxEN>. Setting them simultaneously is prohibited.

7.6.3.3 CGICRCG(CG Interrupt Request Clear Register)

7.6.3.4 CGRSTFLG (Reset Flag Register)

Note 1: This flag indicates a reset generated by the SYSRESETREQ bit of the Application Interrupt and Reset Control Register of the CPU's NVIC.

Note 2: This register is not cleared automatically. Write "0" to clear the register.

8. Input / Output port

8.1 Port Function

8.1.1 Function List

The ports are also used as input / output pins for built-in peripheral functions.

Table 8-1 show list of port function.

Port	Pin name	Input / Out- put	Schmitt Input	Noise filter	Function pin				
Port A									
	PA ₀	I/O	o	$\overline{}$	SEG0				
	PA ₁	1/O	o	$\overline{}$	SEG1				
	PA ₂	I/O	o	-	SEG ₂				
	PA3	I/O	o	$\overline{}$	SEG3				
	PA4	I/O	o	$\overline{}$	SEG4				
	PA ₅	1/O	o	$\qquad \qquad -$	SEG5				
	PA6	I/O	o	-	SEG6				
	PA7	I/O	o	$\overline{}$	SEG7				
Port B									
	PB ₀	I/O	o	$\overline{}$	SEG8				
	PB ₁	I/O	o	$\overline{}$	SEG9				
	PB ₂	I/O	o	$\qquad \qquad -$	SEG10				
	PB ₃	I/O	$\mathsf{o}\,$	-	SEG11				
	PB4	I/O	o	$\overline{}$	SEG12				
	PB ₅	I/O	$\mathsf{o}\,$	$\qquad \qquad -$	SEG13				
	PB ₆	I/O	o	-	SEG14				
	PB7	I/O	o	$\qquad \qquad -$	SEG15				
Port C									
	PC ₀	I/O	$\mathsf{o}\,$	$\overline{}$	SEG16				
	PC ₁	I/O	o	$\overline{}$	SEG17				
	PC ₂	I/O	o	$\qquad \qquad -$	SEG18				
	PC ₃	I/O	o	$\qquad \qquad -$	SEG19				
	PC4	I/O	o	$\qquad \qquad -$	SEG20				
	PC ₅	I/O	o	$\qquad \qquad -$	SEG21				
	PC6	I/O	o	$\overline{}$	SEG22				
	PC7	I/O	o	-	SEG23				
Port D									

Table 8-1 List of Port Function (Port A to Port K)

Table 8-1 List of Port Function (Port A to Port K)

Port K

Port	Pin name	Input / Out- put	Schmitt Input	Noise filter	Function pin
	PK0	I/O	о	о	INT ₃
	PK1	I/O	о	$\overline{}$	TB1OUT

Table 8-1 List of Port Function (Port A to Port K)

Note:The noise elimination width of the noise filter is approximately 30 ns under typical condition.

8.2 Port Register General Description

When the port registers are used, the following registers must be set.

All registers are 32-bits. The configurations are different depend on the number of port bits and assignation of the function.

In the following examination, the 8 bits port are described. About the configuration of the port and the initial value in the each port, please refer to each port's section.

Note:"x" means the name of ports and "n" means the function number in the following description.

8.2.1 PxDATA : Port x data register

This register reads / writes port data.

8.2.2 PxCR : Port x output control register

This register controls output.

To enable / disable input with PxIE register.

8.2.3 PxFRn : Port x function register n

This register sets the function.

The assigned function can be enabled by setting "1". This register exists for the each function assigned to the port. In case of having some function, only one function can be enabled.

8.2.4 PxOD : Port x open-drain control register

This register controls programmable open-drain outputs.

Programmable open-drain outputs are set with PxOD. When output data is "1", output buffer is disabled and becomes a pseudo-open-drain output.

8.2.5 PxPUP : Port x pull-up control register

This register controls programmable pull-ups.

8.2.6 PxPDN : Port x pull-down control register

This register controls programmable pull-downs.

8.2.7 PxIE : Port x input control register

This register controls inputs.

8.3 Register List

8.4 Function details

This chapter describes the configuration of register, initial value and the function assigned by a function register. The bit 31 to 8 of the register and the bit hatched is read as "0". Writing to them does not influence.

8.4.1 Port A

Note:When port A is used as LCD segment outputs, set "1" to PAFR1 and clear other registers to "0".

8.4.2 Port B

Note:When port B is used as LCD segment outputs, set "1" to PBFR1 and clear other registers to "0".

8.4.3 Port C

Note:When port C is used as LCD segment outputs, set "1" to PCFR1 and clear other registers to "0".

8.4.4 Port D

Note:When port D is used as LCD segment outputs, set "1" to PDFR1 and clear other registers to "0".

8.4.5 Port E

Note 1: When port E is used as LCD segment outputs, set "1" to PEFR1 and clear other registers to "0". Note 2: Write as "0".

Note 3: Change setting if necessary.

8.4.6 Port F

When this port is used as analog input, the value of all register set to the initial value.

Note:In modes other than STOP mode, interrupt input is enabled regardless of the PxFRn register setting if input is enabled in PxIE. Make sure to disable unused interrupts when programming the device.

8.4 Function details

8.4.7 Port G

Note:In modes other than STOP mode, interrupt input is enabled regardless of the PxFRn register setting if input is enabled in PxIE. Make sure to disable unused interrupts when programming the device.

8.4.8 Port H

Note:PH3 is shared with DBGEN function. It is enabled to be input and pulled-up while RESET pin is low.

8.4.9 Port I

Note 1: PI0 is enabled to be input and pulled-up while RESET pin is low.

Note 2: Only when input is enabled, PI5 and PI6 tolerate 5V inputs.

Note that these pins cannot be pulled up over the power supply voltage when using as open-drain output.

8.4.10 Port J

Note 1: PJ0 is shared with BOOT function. It is enabled to be input and pulled-up while RESET pin is low.

Note 2: PJ5 is used as only input port.

Note 3: In modes other than STOP mode, interrupt input is enabled regardless of the PxFRn register setting if input is enabled in PxIE. Make sure to disable unused interrupts when programming the device.

Note 4: Only when input is enabled, PJ2 tolerates 5V input.

Note that these pins cannot be pulled up over the power supply voltage when using as open-drain output.

8.4.11 Port K

When bleeder resistor is connected with this port, the value of all register set to the initial value.

Note:In modes other than STOP mode, interrupt input is enabled regardless of the PxFRn register setting if input is enabled in PxIE. Make sure to disable unused interrupts when programming the device.

8.5 Block Diagrams of Ports

8.5.1 Port Type

The ports are classified as shown below. Please refer to the following pages for the block diagrams of each port type. A dotted box in the figure indicates the part of the equivalent circuit described in the "Block diagrams of ports".

Port types used in each pin are described in Chapter [8.6.](#page-129-0)

int: Interrupt input

−: No exist/Disabled

ο: Exist/Enabled

R: Forced disable during reset

EnR: Forced enable during reset

8.5.2 Type FT1

Figure 8-1 Port Type FT1

8.5.3 Type FT2

Figure 8-2 Port Type FT2

8.5.4 Type FT3

Figure 8-3 Port Type FT3

8.5.5 Type FT4

Figure 8-4 Port Type FT4

8.5.6 Type FT5

Figure 8-5 Port Type FT5

8.5.7 Type FT6

Figure 8-6 Port Type FT6

8.5.8 Type FT7

Figure 8-7 Port Type FT7

8.5.9 Type FT8

Figure 8-8 Port Type FT8

8.5.10 Type FT9

Figure 8-9 Port Type FT9

8.5.11 Type FT10

Figure 8-10 Port Type FT10

8.6 Appendix (List of Port Setting)

The following tables show port register settings in each pin.

The numbers "0" and "1" described below register names indicate a setting value and "x" means optional.

8.6.1 The Setting of I/O Port

When I/O port is used as for input port or output port, set its registers as follows:

8.6.2 The Setting of Input Dedicated Port

When an input dedicated port is used, set its registers as follows:

8.6.3 The Setting of Output Dedicated Port

When an output dedicated port is used, set its registers as follows:

8.6.4 The setting of peripheral's I/O port

This section describes the settings in case that ports are used for peripheral functions.

In almost all pins, the initial port state of port registers is all "0" and is inhibited to input/output after reset. Some pins are specified as a certain function after reset. In this case, the symbol "ο" is described in the "After reset" Column of the following tables.

The column of PxFRn indicates the function register that is required to set.

In the shaded areas of bits of the following tables indicate that "0" is read and write has no meaning.

8.6.4.1 Port A Setting

8.6.4.2 Port B Setting

8.6.4.3 Port C Setting

8.6.4.4 Port D Setting

8.6.4.5 Port E Setting

8.6.4.6 Port F Setting

8.6.4.7 Port G Setting

8.6.4.8 Port H Setting

8.6.4.9 Port I Setting

8.6.4.10 Port J Setting

8.6.4.11 Port K Setting

9. 16-bit Timer / Event Counters (TMRB)

9.1 Outline

TMRB has the operation modes shown as below.

- ・ Interval timer mode
- ・ Event counter mode
- ・ Programmable pulse generation (PPG) mode
- ・ Programmable pulse generation (PPG) external trigger mode

The use of the capture function allows TMRB to perform the following measurements.

- ・ Frequency measurement
- ・ Pulse width measurement

In the following explanation, "x" indicates a channel number.

9.2 Block Diagaram

TMRB consists of a 16-bit up-counter, two 16-bit timer register (Double-buffered), two 16-bit capture registers, two comparators, a capture input control, a timer flip-flop and its associated control circuit. Timer operation modes and the timer flip-flop are controlled by a register.

9.3 Registers

9.3.1 Register list

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

Note:During timer operation, timer control register, timer mode register and timer flip-flop control register should not be modified. After stopping timer operation, they should be modified.

9.3.2 TBxEN (Enable register)

9.3.3 TBxRUN (RUN register)

9.3.4 TBxCR (Control register)

9.3.5 TBxMOD (Mode register)

Note:Do not make any changes of TBxMOD register while the TMRBx is running.

9.3.6 TBxFFCR (Flip-flop control register)

9.3.7 TBxST (Status register)

Note 1: Even if mask configuration by TBxIM register is valid, the status is set to TBxST register.

Note 2: When the interrupt mask configuration is disabled by the corresponding bit of TBxIM register, the interrupt is issued to the CPU.

Note 3: To clear the flag, TBxST register should be read.

9.3.8 TBxIM (Interrupt mask register)

Note:Even if mask configuration by TBxIM register is valid, the status is set to TBxST register.

9.3.9 TBxUC (Up counter capture register)

9.3.10 TBxRG0 (Timer register 0)

9.3.11 TBxRG1 (Timer register 1)

9.3.12 TBxCP0 (Capture register 0)

9.3.13 TBxCP1 (Capture register 1)

9.4 Description of Operation

9.4.1 Prescaler

There is prescaler to generate the source clock for up-counter.

The prescaler input φT0 is fperiph/1, fperiph/2, fperiph/4, fperiph/8, fperiph/16 or fperiph/32 selected CGSYSCR<PRCK[2:0] in the CG circuit. The peripheral clock is either fgear, a clock selected by CGSYSCR<FPSEL> in the CG circuit, or fc, which is a clock before it is divided by the clock gear.

The operation or the stoppage of a prescaler is set with TBxRUN<TBPRUN> where writing "1" status counting and writing "0" clears and stops counting.

9.4.2 Up-counter (UC)

UC is a 16-bit binary counter.

9.4.2.1 Source clock

UC's source clock is specified by TBxMOD<TBCLK[2:0]>.

It can be selected from the prescaler output clock - φ T1, φ T4, φ T16, φ T32, φ T64, φ T128 and φ T256 or the external clock of the TBxIN pin.

9.4.2.2 Counter start / stop

To start the counter, there are a software start, external trigger start and synchronous start.

1. Software start

If <TBRUN> is set to "1", the counter will start. If "0" is set to the <TBRUN>, the counter will stop and the up-counter will be cleared at the same time.

2. External trigger start

In the external trigger mode, the counter will be started by external signals.

If TBxCR<CSSEL> is set to "1", the external trigger start mode is set. At this time, if <TBRUN> is set to "1", the condition of the counter will be trigger wait. The counter will start on the rising/falling edge of TBxIN.

TBxCR<TRGSEL> bit specifies the switching external trigger edges.

<TRGSEL>="0": Rising edge of TBxIN is selected.

<TRGSEL>="1": Falling edge of TBxIN is selected.

If <TBRUN> is set to "0", the counter will stop and the up-counter will be cleared at the same time.

3. Synchronous start

In the timer synchronous mode, synchronous start timers can be possible. If timer synchronous mode is used in the PPG output mode, motor drive application can be achieved.

Depending on products, the combination of master channels and slave channels have already been determined. For the combination of master channels and slave channels of this product, refer to Chapter Product Information.

TBxCR<TBSYNC> bit specifies the switching of synchronous mode. If <TBSYNC> bit of a slave channel is set to "1", the counter will start/stop synchronously with the software or external trigger start of a master channel. TBxRUN<TBPRUN, TBRUN> bit of a slave channel is not required to set. <TBSYNC> bit of a master channel must be set to "0".

Note that if the external trigger counter mode and timer synchronous mode are both set, the timer synchronous mode gains a higher priority.

9.4.2.3 Timing to clear UC

1. When a match with TBxRG1 is detected

By setting TBxMOD<TBCLE> = "1", UC is cleared if when the comparator detects a match between UC and TBxRG1.

2. When UC stops

UC stops and is cleared if $TBxRUN < TBRUN > = "0"$.

9.4.2.4 UC overflow

If UC overflow occurs, the INTTBx overflow interrupt is generated.

9.4.3 Timer registers (TBxRG0, TBxRG1)

TBxRG0 and TBxRG1 are registers for setting value to compare with up-counter values and two registers are built into each channel. If the comparator detects a match between a value set in timer register and in an up-counter, comparator outputs the match detection signal.

TBxRG0 and TBxRG1 are consisted of the double buffered configuration which are paired with register buffers. The double buffering is disabled in the initial state.

Controlling double buffering disable or enable is specified by $TBXCR \le TBWBF >$. if $\le TBWBF > 0$, the double buffering becomes disable, If <TBWBF> ="1", it becomes enable.

When the double buffering is enabled, data transfers from the register buffer to the timer register (TBxRG0/1) in the case that UC is matched with TBxRG1.

When UC is stopped even if double buffering is enabled, the double buffering operates as a single buffer, and data can be written to the TBxRG0 and TBxRG1 directly.

9.4.4 Capture control

This is a circuit that controls the timing of latching values from UC into the TBxCP0 and TBxCP1. The capture timing of UC is specified by TBxMOD<TBCPM[1:0]>.

Software can also capture the value of UC to capture registers. The value of UC are taken into the TBxCP0 each time "0" is written to TBxMOD<TBCP>.

9.4.5 Capture registers (TBxCP0, TBxCP1)

This register captures the value of UC.

9.4.6 Up-counter capture register (TBxUC)

If TBxUC register is read during the counter operation, the current value of up-counter will be captured and the value will be read. The value captured at the end is held while the counter is stopping.

9.4.7 Comparators (CP0, CP1)

This circuit compares with UC and the value set to TBxRG0/1 and detects match. If a match is detected, INTTBx is occurred.

9.4.8 Timer flip-flop (TBxFF0)

The timer flip-flop (TBxFF0) is reversed by a match signal from the comparator and a latch signal to the capture registers. It can be enabled or disabled to reverse by setting the TBxFFCR<TBC1T1, TBC0T1, TBC1T1, TBC1T0>.

The value of TBxFF0 becomes undefined after a reset. The flip-flop can be reversed by writing "00" to TBxFFCR<TBFF0C[1:0]>. It can be set to "1" by writing "01", and can be cleared o "0" by writing "10".

The value of TBxFF0 can be output to the timer output pin (TBxOUT). If the timer output is performed, the corresponding port settings should be programmed beforehand.

9.4.9 Capture interrupt (INTCAPx0, INTCAPx1)

INTCAPx0 and INTCAPx1 can be generated at the timing of latching value from UC into the TBxCP0 and TBxCP1.

9.5 Description of Operation for each mode

9.5.1 Interval timer mode

In the case of generating constant period interrupt, set the interval time to the timer register (TBxRG1) to generate the INTTBx interrupt.

Note:X; Don't care, *; optional value, -; Don't change

9.5.2 Event counter mode

It is possible to make TMRBx the event counter by using a source clock as an external clock (TBxIN pin input).

The UC counts up on the rising edge of TBxIN pin input. The value of UC can be captured by soft capture. It is possible to read the count value by reading it.

Note:X; Don't care, *; optional value, -; Don't change

9.5.3 Programmable pulse generation (PPG) output mode

Square wave with any frequency and any duty can be output. The output pulse can be either low-active or high-active.

TBxFF0 is reversed when UC matches the set value of TBxRG0 and TBxRG1. TBxFF0 can be output from TBxOUT pin.

Note that the set value of TBxRG0 and TBxRG1 must satisfy the following requirement.

Set value of TBxRG0 < Set value of TBxRG1

Figure 9-2 Example of Programmable pulse generation output

In this mode, by enabling the double buffering, The value of register buffer 0 and 1 are shifted into TBxRG0 and 1 when UC matches the value of TBxRG1.

It is possible to modify frequency and duty without timing of modifying TBxRG0 and TBxRG1.

Figure 9-3 Register Buffer Operation

The block diagram of this mode is shown below.

Figure 9-4 Block diagram of 16-bit PPG mode

Each register in the 16-bit PPG output mode should be programmed as listed below.

Note:X; Don't care, *; optional value, -; Don't change

9.5.4 Programmable pulse generation (PPG) external trigger output mode

A PPG wave with a short delay time can be output by using external trigger count start mode.

The example of an one-shot pulse output by external trigger count start mode is shown below.

To start count up by the rising edge of TBxIN, set TBxCR<CSSEL> to "1" and clear

TBxCR<TRGSEL> to "0" in stopping 16-bit up counter.

TBxRG0 is set the delay time (d) from an external trigger signal. TBxRG1 is set the value (d)+(p) which is added the delay time (d) and the width (p) of one-shot pulse.

To reverse TBxFF0 when UC matches TBxRG0 and TBxRG1, TBxFFCR<TBE1T1> and TBxFFCR<TBE1T1> are set to "1".

UC is readied to start UC by setting TBxRUN<TBPRUN> and TBxRUN<TBRUN> to "1".

UC starts by the rising edge of external trigger.

TBxFF0 is reversed when UC counts up to (d) and UC matches TBxRG0. TBxFF0 is "High" level.

TBxFF0 is reversed when UC counts up to $(d)+(p)$ and UC matches TBxRG1. TBxFF0 is "Low" level.

To fix the level of TBxFF0, clear TBxFFCR<TBE1T1> and TBxFFCR<TBE0T1> to "0" or stops UC by TBxRUN<TBPRUN><TBRUN> in INTTBx which is generated when UC matches TBxRG1.

Figure 9-5 One-shot pulse output with delay by external trigger start

The followings shows the setting in the case that 2 ms width one-shot pulse is output after 3 ms by triggering TBxIN input at the rising edge. In this example, the source clock is φT1.

TOSHIBA

Note:X; Don't care, *; optional value, -; Don't change

9.6 Applications using the capture function

The capture function can be used many applications.

The applications are shown below.

- 1. Frequency measurement
- 2. Pulse width measurement

9.6.1 Frequency measurement

The frequency of an external clock can be measured.

To measure frequency, TMRBm is used as 16-bit interval timer mode and TMRBn is used as 16-bit event counter mode.

To count UC of TMRBn freely by an external clock, set TMnMOD<TBCLK> to "000" and set TBnRUN<TBE1T1><TBE0T1> to "11".

To reverse TBmFF0 when UC of TMRBm matches TBmRG0 and TBmRG1, set TBmFFCR<TBE1T1> $<$ TBE0T1 $>$ to "11".

To capture UC to TBnCP0 at rising edge of TBmFF0 and UC to TBmCP1 at falling edge of TBmFF0, set TBxMOD<TBCPM> to "11".

Set TBmRG0 and TBmRG1 to time when UC counts an external clock and start TMRBm.

Rises-up TBmFF0 when UC of TMRBm matches TBmRG0 and captures the value of TMRBn's UC to TBnCP0. Falls-down TBmFF0 when UC of TMRBm matches TBmRG1 and captures the value of TMRBn's UC to TBnCP1.

A frequency is measured from (TBnCP1 - TBnCP0) ÷ (TBmRG1 - TBmRG0) in INTTBm.

For example, the difference between TBmRG1 and TBmRG0 is 0.5 s and the difference between TBnCP1 and TBnCP0 is 100, the frequency is 200 Hz (100 \div 0.5 s = 200Hz)

TBnCP1 - TBnCP0 may be less than zero depend on the changing timing of TBmFF0. Please correct the value if TBnCP1 - TBnCP0 is less than zero.

Figure 9-6 Frequency measurement

The following shows in the case that the measured pulse is input to TBxIN. In this example, the source clock is φT1.

TOSHIBA

Note:X; Don't care, *; optional value, -; Don't change

9.6.2 Pulse width measurement

"High" level width of the external pulse can be measured.

To capture UC to TBxCP0 at rising edge of TBxIN and UC to TBxCP1 at falling edge of TBxIN, set TBxMOD<TBCPM> to "10".

Enables INTCAPx1 interrupt.

Enables TMRBx operation.

Captures the vale of UC to TBxCP0 when the rising edge of the external pulse into TBxIN. Captures the value of UC to TBxCP1 when the falling edge of the external pulse into TBxIN and INTCAPx1 interrupt is occurred.

The "High" level width of the external pulse can be calculated by multiplying the difference between TBxCP0 and TBxCP1 by the clock cycle of a prescaler output clock.

For example, if the difference between TBxCP0 and TBxCP1 is 100 and the cycle of the prescaler output clock is 0.5 μs, the pulse width is 100×0.5 μs = 50 μs.

When the pulse width which is more than maximum count time of UC is measured, please correct the measured value.

The "Low" level width of an external pulse can also be measured.

In this case, enables INTCAPx0 interrupt. In twice process of INTCAPx0 interrupt, the difference between C2 generated the first time and C1 generated the second time in "Figure 9-7 Pulse width measurement" is multiplied by the cycle of the prescaler output clock.

Figure 9-7 Pulse width measurement

The following is shown that the "High" level width of the external pulse into TBxIN is measured. In this example, the source clock is φT1.

TOSHIBA

Note:X; Don't care, *; optional value, -; Don't change

9. 16-bit Timer / Event Counters (TMRB)

9.6 Applications using the capture function

TOSHIBA

10. 16-Bit Timer A (TMR16A)

10.1 Outline

TMR16A contains the following functions:

- ・ Match interrupt
- ・ Square waveform output
- ・ Read capture

In this chapter, "x" indicates a channel number.

10.2 Block Diagram

Figure 10-1 Block diagram of TMR16A

10.3 Registers

10.3.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

Note: When T16ARUN<RUN> is set to "1", do not modify T16AxEN, T16AxCR, T16AxRG and T16AxCP.

10.3.2 Details of Registers

10.3.2.1 T16AxEN (Enable Register)

10.3.2.2 T16AxRUN (RUN Register)

10.3.2.3 T16AxCR (Control Register)

10.3.2.4 T16AxRG (Timer Register)

Note:Do not set "0x0000".

10.3.2.5 T16AxCP (Capture Register)

10.4 Operation Description

10.4.1 Timer Operation

1. Preparation

Set "1" to T16AxCR<UCCR>.

Select a source clock with T16AxCR<CLK>. Write "0" to set fsys or write "1" to set ΦT0. Set a counter value to T16AxRG<RG>.

Modify T16AxCR and T16AxRG while the counter stops (T16AxRUN<RUN> is "0").

2. Counter operation

Before starting counter operation, set "0x0000" to T16AxCP<CP> to clear the counter.

To start count-up, set "1" to T16AxRUN<RUN>. If the counter value matches with a value of T16AxRG<RG>, it will be cleared to "0x0000" and continued to count-up.

3. Match detection interrupt generation

If a counter value matches with a value of T16AxRG, a match detection interrupt INTT16Ax will be output.

4. Stop

To stop counts, set "0" to T16AxRUN<RUN>. The counter value is held. Then clear the counter before counting is started by setting "1" to <RUN>.

Note:Modification of T16AxCR, T16AxRG and T16AxCP must be performed while the counter is stopping (T16AxRUN<RUN> is set to "0").

10.4.2 T16AxOUT Control

T16AxOUT is modified by register setting or by matching the counter with T16AxRG.

An initial state of T16AxOUT is "0".

1. Control by software

With T16AxCR<T16AFFC> setting, T16AxOUT can be specified; "1" is to set, "0" is to clear, and also the inverted setting is possible.

Modify T16AxCR while the counter stops (T16AxRUN<RUN> is "0").

2. Inverse due to matching counter

Write "1" to T16ACR<FFEN> to invert T16AxOUT. When T16AxRG matches with a counter value, T16AxOUT will invert. When the counter stops, a state of T16AxOUT will remain.

10.4.3 Read Capture

A current value of the counter can be captured by reading T16AxCP.

10.4.4 Automatic Stop

With the setting of T16AxEN, TMR16A automatically stops in the following conditions:

1. Transition to/from IDLE mode

With T16AEN<I2T16A> setting, TMR16A operation during IDLE mode can be specified. If "1" is set, TMR16A automatically stops count-up when the transition to the IDLE mode occurs. If TMR16A returns from IDLE mode, it will restart counting-up operation.

2. Debug halt

With T16AEN<HALT> setting, TMR16A operation during debug halt can be specified. If "0" is set, TMR16A automatically stops count-up when the transition to the debug halt mode occurs. If debug halt mode of the core is canceled, count-up will restart.

10.4 Operation Description

TOSHIBA

11. Serial Channel (SIO/UART)

11.1 Overview

Serial channel has the modes shown below.

- Synchronous communication mode (I/O interface mode)
- Asynchronous communication mode (UART mode)

Their features are given in the following.

- ・ Transfer Clock
	- Dividing by the prescaler, from the peripheral clock (φ T0) frequency into 1/2, 1/8, 1/32, 1/128.
	- Make it possible to divide from the prescaler output clock frequency into 1 to 16.
	- Make it possible to divide from the prescaler output clock frequency into N+m/16 (N=2 to 15, m=1 to 15). (only UART mode)
	- The usable system clock (only UART mode).
- ・ Double Buffer

The usable double buffer function.

- ・ I/O Interface Mode
	- Transfer Mode: the half duplex (transmit/receive), the full duplex
	- Clock: Output (fixed rising edge) /Input (selectable rising/falling edge)
	- Make it possible to specify the interval time of continuous transmission.
- ・ UART Mode
	- Data length: 7 bits, 8bits, 9bits
	- Add parity bit (to be against 9bits data length)
	- Serial links to use wake-up function
	- Handshaking function with $\overline{\text{CTSx}}$ pin
- ・ Output Signal using IR Carrier Pulses

In the following explanation, "x" represents channel number.

11.2 Configuration

Figure 11-1 shows Serial channel block diagram.

Figure 11-1 Serial Channel Block Diagram

11.3 Registers Description

11.3.1 Registers List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

Note:Do not modify any control register when data is being transmitted or received.

11.3 Registers Description

11.3.2 SCxEN (Enable Register)

11.3.3 SCxBUF (Buffer Register)

SCxBUF works as a transmit buffer for write operation and as a receive buffer for read operation.

11.3.4 SCxCR (Control Register)

Note:Any error flag (OERR, PERR, FERR) is cleared to "0" when read.

11.3.5 SCxMOD0 (Mode Control Register 0)

Note 1: Set <RXE> to "1" after setting each mode register (SCxMOD0, SCxMOD1 and SCxMOD2).

Note 2: Do not stop the receive operation (by setting SCxMOD0<RXE> to "0") when data is being received.

11.3.6 SCxMOD1 (Mode Control Register 1)

Note 1: Specify the all mode first and then enable the <TXE> bit.

Note 2: Do not stop the transmit operation (by setting <TXE> to "0")when data is being transmitted.

Note 1: While data transmission is in progress, any software reset operation must be executed twice in succession.

Note 2: A software reset requires 2 clocks-duration at the time between the end of recognition and the start of execution of software reset instruction.

11.3.8 SCxBRCR (Baud Rate Generator Control Register), SCxBRADD (Baud Rate Generator Control Register 2)

The division ratio of the baud rate generator can be specified in the registers shown below.

SCxBRCR

TOSHIBA

SCxBRADD

Table 11-1 lists the settings of baud rate generator division ratio.

Table 11-1 Setting division ratio

- Note 1: To use the "N + (16 K)/16" division function, be sure to set <BRADDE> to "1" after setting the K value to <BRK>. The "N + (16 - K)/16" division function can only be used in the UART mode.
- Note 2: As a division ratio, 1 ("0001") or 16 ("0000") can not be applied to N when using the "N + (16 K)/16" division function in the UART mode.
- Note 3: The division ratio "1" of the baud rate generator can be specified only when the double buffering is used in the I/O interface mode.
- Note 4: Specifying " $K = 0$ " is prohibited.

11.4 Operation in Each Mode

Table 11-2 shows the modes and data formats.

Table 11-2 Mode and Data format

Mode 0 is a synchronous communication and can be used to extend I/O. This mode transmits and receives data in synchronization with SCLK. SCLK can be used for both input and output.

The direction of data transfer can be selected from LSB first and MSB first. This mode is not allowed either to use parity bits or STOP bits.

The mode 1, mode 2 and mode 3 are asynchronous modes and the transfer direction is fixed to the LSB first.

Parity bits can be added in the mode 1 and mode 2. The mode 3 has a wakeup function in which the master controller can start up slave controllers via the serial link (multi-controller system).

STOP bit in transmission can be selected from 1 bit and 2 bits. The STOP bit length in reception is fixed to a one bit.

TOSHIBA

11.5 Data Format

11.5.1 Data Format List

Figure 11-2 shows data format.

11.5.2 Parity Control

The parity bit can be added with a transmitted data only in the 7- or 8-bit UART mode.

Setting "1" to SCxCR<PE> enables the parity.

The <EVEN> bit of SCxCR selects either even or odd parity.

11.5.2.1 Transmission

Upon data transmission, the parity control circuit automatically generates the parity with the data in the transmit buffer.

The parity bit will be stored in SCxBUF<TB7> in the 7-bit UART mode and SCxMOD<TB8> in the 8 bit UART mode.

The <PE> and <EVEN> settings must be completed before data is written to the transmit buffer.

11.5.2.2 Receiving Data

If the received data is moved from the receive shift register to the receive buffer, a parity is generated.

In the 7-bit UART mode, the generated parity is compared with the parity stored in SCxBUF<RB7>, in the 8-bit UART mode, it is compared with the one in SCxCR<RB8>.

If there is any difference, a parity error occurs and the SCxCR<PERR> is set to "1".

11.5.3 STOP Bit Length

The length of the STOP bit in the UART transmission mode can be selected from one bit or two bits by setting the SCxMOD2<SBLEN>. The length of the STOP bit data is determined as one-bit when it is received regardless of the setting of this bit.

11.6 Clock Control

The following figure shows the serial clock (SIOCLK) generation circuit. Before changing the serial clock setting, check if the setting satisfies AC electrical characteristics.

Figure 11-3 Serial clock generation circuit

11.6.1 Prescaler

There is a 7-bit prescaler to divide a prescaler input clock φT0 by 2, 8, 32 and 128.

Use the CGSYSCR register in the clock/mode control block to select the input clock φ T0 of the prescaler.

The prescaler becomes active only when the baud rate generator is selected as a transfer clock by $SCxMOD0 = "01".$

11.6.2 Serial Clock Generation Circuit

The serial clock circuit is a block to generate transmit and receive clocks (SIOCLK) and consists of the circuits in which clocks can be selected by the settings of the baud rates generator and modes.

11.6.2.1 Baud Rate Generator

The baud rate generator generates transmit and receive clocks to determine the serial channel transfer rate.

(1) Baud Rate Generator input clock

The input clock of the baud rate generator is selected from the prescaler outputs divided by 2, 8, 32 and 128.

This input clock is selected by setting the SCxBRCR<BRCK>.

(2) Baud Rate Generator output clock

The frequency division ratio of the output clock in the baud rate generator is set by SCxBRCR and SCxBRADD.

The following frequency divide ratios can be used; 1/N frequency division in the I/O interface mode, either $1/N$ or $N + (16-K)/16$ in the UART mode.

The table below shows the frequency division ratio which can be selected.

Note:1/N (N=1)frequency division ratio can be used only when a double buffer is enabled.

The input clock to the divider of baud rate generator is φTx , the baud rate in the case of 1/N and $N + (16-K)/16$ is shown below.

・ Divide by N

$$
Baud\ rate = \frac{\phi Tx}{N}
$$

• $N + (16-K)/16$ division

Bad that
$$
= \frac{\phi \, \text{Tx}}{N + \frac{(N - K)}{N}}
$$

11.6.2.2 Clock Selection Circuit

A clock can be selected by setting the modes and the register.

Modes can be specified by setting the SCxMOD0<SM>.

The input clock in I/O interface mode is selected by setting SCxCR. The clock in UART mode is selected by setting SCxMOD0<SC>.

(1) Transfer Clock in I/O interface mode

[Table 11-3](#page-188-0) shows clock selection in I/O interface mode.

To use SCLK input, the following conditions must be satisfied.

・ If double buffer is used

- SCLK cycle > 6/fsys

- ・ If double buffer is not used
	- SCLK cycle > 8/fsys

(2) Transfer clock in the UART mode

Table 11-4 shows the clock selection in the UART mode. In the UART mode, selected clock is divided by 16 in the receive counter or the transmit counter before use.

Mode SCxMOD0 <sm></sm>	Clock selection SCxMOD0 <sc></sc>
UART Mode	Timer output
	Baud rate generator
	fsys
	SCLKx input

Table 11-4 Clock Selection in UART Mode

To use SCLK input, the following conditions must be satisfied.

- SCLK cycle > 2/fsys

To enable the TMRB output, a timer output inverts when the value of the counter and that of TBxRG1 match. The SIOCLK clock frequency is "Setting value of TBxRG1 \times 2".

Baud rates can be obtained by using the following formula.

To enable the TMR16A output, a timer output inverts when the value of the counter and that of T16AxRG match. The SIOCLK clock frequency is "Setting value of T16AxRG \times 2".

Baud rates can be obtained by using the following formula.

Transfer rate =
$$
\frac{\text{Clock frequency selected by CGSYSCR>PRCK[1:0]>}{\text{(TBxRG1 × 2) × 16}}
$$

\nOne clock cycle is a period that the timer output is inverted twice.

OSHIBA

11.6.3 Transmit/Receive Buffer

11.6.3.1 Configuration

Figure 11-4 shows the configuration of transmit buffer, receive buffer.

Appropriate settings are required for using buffer. The configuration may be predefined depending on the mode.

Figure 11-4 The Configuration of Buffer

11.6.3.2 Transmit/Receive Buffer

Transmit buffer and receive buffer are double-buffered. The buffer configuration is specified by SCxMOD2<WBUF>.

In the case of a receive mode, if SCLK input is set in the I/O interface mode or the UART mode is selected, it's double buffered despite the <WBUF> settings. In other modes, it's according to the <WBUF> settings.

Table 11-5 shows correlation between modes and buffers.

11.7 Status Flag

The SCxMOD2 register has two types of flag. This bit is significant only when the double buffer is enabled.

<RBFLL> is a flag to show that the receive buffer is full. When one frame of data is received and the data is moved from the receive shift register to the receive buffers, this bit changes to "1" while reading this bit changes it to "0".

<TBEMP> shows that the transmit buffers are empty. When data in the transmit buffers is moved to the transmit shift register, this bit is set to "1" When data is set to the transmit buffers, the bit is cleared to "0".

11.8 Error Flag

Three error flags are provided in the SCxCR register. The meaning of the flags is changed depending on the modes. The table below shows the meanings in each mode.

These flags are cleared to "0" after reading the SCxCR register.

11.8.1 OERR Flag

In both UART and I/O interface modes, this bit is set to "1" when an error is generated by completing the reception of the next frame of receive data before the receive buffer has been read.

In the I/O interface with SCLK output mode, the SCLK output stops upon setting the flag.

Note:To switch the I/O interface SCLK output mode to other modes, read the SCxCR register and clear the over-run flag.

11.8.2 PERR Flag

This flag indicates a parity error in the UART mode and an under-run error or completion of transmit in the I/O interface mode.

In the UART mode, <PERR> is set to "1" when the parity generated from the received data is different from the parity received.

In the I/O interface mode, <PERR> is set to "1" under the following conditions when a double buffer is enabled.

In the SCLK input mode, <PERR> is set to "1" when the SCLK is input after completing data output of the transmit shift register with no data in the transmit buffer.

In the SCLK output mode, <PERR> is set to "1" after completing output of all data and the SCLK output stops.

Note: To switch the I/O interface SCLK output mode to other modes, read the SCxCR register and clear the under-run flag.

11.8.3 FERR Flag

A framing error is generated if the corresponding stop bit is determined to be "0" by sampling the bit at around the center. Regardless of the stop bit length settings in the SCxMOD2<SBLEN>register, the stop bit status is determined by only 1.

This bit is fixed to "0" in the I/O interface mode.

11.9 Receive

11.9.1 Receive Counter

The receive counter is a 4-bit binary counter and is up-counted by SIOCLK.

In the UART mode, sixteen SIOCLK clock pulses are used in receiving a single data bit and the data symbol is sampled at the seventh, eighth, and ninth pulses. From these three samples, majority logic is applied to decide the received data.

11.9.2 Receive Control Unit

11.9.2.1 I/O interface mode

In the SCLK output mode with SCxCR <IOC> set to "0", the RXDx pin is sampled on the rising edge of the shift clock outputted to the SCLKx pin.

In the SCLK input mode with SCxCR <IOC> set to "1", the serial receive data RXDx pin is sampled on the rising or falling edge of SCLKx pin input signal depending on the SCxCR <SCLKS> setting.

11.9.2.2 UART Mode

The receive control unit has a start bit detection circuit, which is used to initiate receive operation when a normal start bit is detected.

11.9.3 Receive Operation

11.9.3.1 Receive Buffer

The received data is stored by 1 bit in the receive shift register. When a complete set of bits has been stored, the interrupt INTRXx is generated.

When the double buffer is enabled, the data is moved to the receive buffer (SCxBUF) and the receive buffer full flag (SCxMOD2<RBFLL>) is set to "1". The receive buffer full flag is "0" cleared by reading the receive buffer. When the double buffer is disabled, the receive buffer full flag has no meaning.

11.9.3.2 I/O interface mode with SCLK output

In the I/O interface mode and SCLK output setting, SCLK output stops when all received data is stored in the receive buffer. So, in this mode, the over-run error flag has no meaning.

The timing of SCLK output stop and re-output depends on receive buffer.

(1) Case of single buffer

Stop SCLK output after receiving a data. In this mode, I/O interface can transfer each data with the transfer device by hand-shake.

When the data in a buffer is read, SCLK output is restarted.

(2) Case of double buffer

Stop SCLK output after receiving the data into a receive shift register and a receive buffer.

When the data is read, SCLK output is restarted.

11.9.3.3 Read Received Data

Read the received data from the receive buffer (SCxBUF).

The buffer full flag SCxMOD2<RBFLL> is cleared to "0" by this reading. In the case of the next data can be received in the receive shift register before reading a data from the receive buffer. The parity bit to be added in the 8-bit UART mode as well as the most significant bit in the 9-bit UART mode will be stored in SCxCR<RB8>.

11.9.3.4 Wake-up Function

In the 9-bit UART mode, the slave controller can be operated in the wake-up mode by setting the wakeup function SCxMOD0 <WU> to "1". In this case, the interrupt INTRXx will be generated only when SCxCR <RB8> is set to "1".

11.9.3.5 Over-run Error

The over-run error is occurred and set over-run flag without completing data read before receiving the next data. When over-run error is occurred, a content of receive buffer and SCxCR<RB8> is not lost, but a content of receive shift register is lost.

In the I/O interface mode with SCLK output setting, the clock output automatically stops, so this flag has no meaning.

Note:When the mode is changed from I/O interface SCLK output mode to the other mode, read SCxCR and clear over-run flag.

11.10 Transmission

11.10.1 Transmission Counter

The transmit counter is a 4-bit binary counter and is counted by SIOCLK as in the case of the receive counter. In UART mode, it generates a transmit clock (TXDCLK) on every 16th clock pulse.

11.10.2 Transmission Control

11.10.2.1 I/O Interface Mode

In the SCLK output mode with SCxCR<IOC> set to "0", each bit of data in the transmit buffer is outputted to the TXDx pin on the falling edge of the shift clock outputted from the SCLKx pin.

In the SCLK input mode with SCxCR<IOC> set to "1", each bit of data in the transmit buffer is outputted to the TXDx pin on the rising or falling edge of the SCLKx pin input signal according to the SCxCR<SCLKS> setting.

11.10.2.2 UART Mode

When the transmit data is written in the transmit buffer, data transmission is initiated on the rising edge of the next TXDCLK and the transmit shift clock signal is also generated.

11.10.3 Transmit Operation

11.10.3.1 Operation of Transmission Buffer

If double buffering is disabled, the CPU writes data only to transmit shift register and the transmit interrupt INTTXx is generated upon completion of data transmission.

If double buffering is enabled, data written to the transmit buffer is moved to the transmit shift register. The INTTXx interrupt is generated at the same time and the transmit buffer empty flag (SCxMOD2<TBEMP>) is set to "1". This flag indicates that the next transmit data can be written. When the next data is written to the transmit buffer, the <TBEMP> flag is cleared to "0".

OSHIBA

Figure 11-7 Operation of Transmission Buffer (Double-buffer is enabled)

11.10.3.2 I/O interface Mode/Transmission by SCLK Output

If SCLK is set to generate clock the I/O interface mode, the SCLK output automatically stops when all data transmission is completed and underrun error will not occur.

The timing of suspension and resume of SCLK output is different depending on the buffer and FIFO usage.

(1) Single Buffer

The SCLK output stops each time one frame of data is transferred. Handshaking for each data with the other side of communication can be enabled. The SCLK output resumes when the next data is written in the buffer.

(2) Double Buffer

The SCLK output stops upon completion of data transmission of the transmit shift register and the transmit buffer. The SCLK output resumes when the next data is written in the buffer.

11.10.3.3 Under-run error

In the I/O interface SCLK input mode and if no data is set in transmit buffer before the next frame clock input, which occurs upon completion of data transmission from transmit shift register, an under-run error occurs and SCxCR<PERR> is set to "1".

In the I/O interface mode with SCLK output setting, the clock output automatically stops, so this flag has no meaning.

Note:Before switching the I/O interface SCLK output mode to other modes, read the SCxCR register and clear the under-run flag.

11.11 Handshake function

The function of the handshake is to enable frame-by-frame data transmission by using the CTS (Clear to send) pin and to prevent over-run errors. This function can be enabled or disabled by SCxMOD0<CTSE>.

When the $\overline{\text{CTSx}}$ pin is set to "High" level, the current data transmission can be completed but the next data transmission is suspended until the $\overline{\text{CTSx}}$ pin returns to the "Low" level. However in this case, the INTTXx interrupt is generated in the normal timing, the next transmit data is written in the transmit buffer, and it waits until it is ready to transmit data.

- Note 1: If the CTS signal is set to "H" during transmission, the next data transmission is suspended after the current transmission is completed. (Point a in Figure 11-9)
- Note 2: Data transmission starts on the first falling edge of the TXDCLK clock after \overline{CTS} is set to "L". (Point b in Figure 11-9)

Although no RTS pin is provided, a handshake control function can easily implemented by assigning one bit of the port for the \overline{RTS} function. By setting the port to "High" level upon completion of data reception (in the receive interrupt routine), the transmit side can be requested to suspend data transmission.

Figure 11-9 CTS Signal timing

11.12 Output Signal using IR Carrier Pulses

IROUT pin outputs a signal using IR carrier pulses.

IROUT signal is generated by the timer output signals and TXD signals. Since the timer used to generate pulses varies depending on products, refer to Chapter Product Information to select timers.

The timer should be set to generate pulse outputs in a desired cycle within a settable range.

To output IROUT signal from a port, in the port where IROUT pin is assigned, specify the function setting of function register (PxFRn) to IROUT, then enable IROUT signal to be output by control register (PxCR).

SIO receives pulses from the timer and combines the pulses with a TXD signal to output a signal to IROUT pin. Timer outputs and TXD signals are asynchronous, so that pulse distortion occurs when TXD signals are changed.

Figure 11-10 Output signal using IR carrier pulse

11.13 Interrupt/Error Generation Timing

11.13.1 RX Interrupts

Figure 11-11 shows the data flow of receive operation and the route of read.

RX interrupts are generated at the time depends on the transfer mode and the buffer configurations, which are given as follows.

Note:Interrupts are not generated when an over-run error is occurred.

11.13.2 TX interrupts

Figure 11-12 shows the data flow of transmit operation and the route of read.

TX interrupts are generated at the time depends on the transfer mode and the buffer configurations, which are given as follows.

Note:If double buffer is enabled, a interrupt is also generated when the data is moved from the buffer to the shift register by writing to the buffer.

11.13.3 Error Generation

11.13.3.1 UART Mode

11.13.3.2 I/O Interface Mode

Note:Over-run error and Under-run error have no meaning in SCLK output mode.

11.14 Software Reset

Software reset is generated by writing SCxMOD2<SWRST[1:0]> as "10" followed by "01".

As a result, SCxMOD0<RXE>, SCxMOD1<TXE>, SCxMOD2<TBEMP><RBFLL><TXRUN>, SCxCR

<OERR><PERR><FERR> are initialized. And the receive circuit and the transmit circuit become initial state. Other states are maintained.

11.15 Operation in Each Mode

11.15.1 Mode 0 (I/O interface mode)

Mode 0 consists of two modes, the SCLK output mode to output synchronous clock and the SCLK input mode to accept synchronous clock from an external source.

11.15.1.1 Transmitting Data

(1) SCLK Output Mode

If the transmit double buffer is disabled $(SCxMOD2\le WBUF> = "0")$

Data is output from the TXDx pin and the clock is output from the SCLKx pin each time the CPU writes data to the transmit buffer. When all data is output, an interrupt (INTTXx) is generated.

• If the transmit double buffer is enabled $(SCxMOD2\le WBUF> = "1")$

Data is moved from the transmit buffer to the transmit shift register when the CPU writes data to the transmit buffer while data transmission is halted or when data transmission from the transmit buffer (shift register) is completed. Simultaneously, the transmit buffer empty flag SCxMOD2<TBEMP> is set to "1", and the INTTXx interrupt is generated.

When data is moved from the transmit buffer to the transmit shift register, if the transmit buffer has no data to be moved to the transmit shift register, INTTXx interrupt is not generated and the SCLK output stops.

TOSHIBA

Figure 11-13 Transmit Operation in the I/O Interface Mode (SCLK Output Mode)

(2) SCLK Input Mode

• If double buffering is disabled $(SCxMOD2\le WBUF> = "0")$

If the SCLK is input in the condition where data is written in the transmit buffer, 8-bit data is outputted from the TXDx pin. When all data is output, an interrupt INTTXx is generated. The next transmit data must be written before the timing point "A" as shown in [Fig](#page-204-0)[ure 11-14](#page-204-0).

If double buffer is enabled $(SCxMOD2\le WBUF> = "1")$

Data is moved from the transmit buffer to the transmit shift register when the CPU writes data to the transmit buffer before the SCLK input becomes active or when data transmission from the transmit shift register is completed. Simultaneously, the transmit buffer empty flag SCxMOD2<TBEMP> is set to "1", and the INTTXx interrupt is generated.

If the SCLK input becomes active while no data is in the transmit buffer, although the internal bit counter is started, an under-run error occurs and 8-bit dummy data (0xFF) is sent.

TOSHIBA

PERR (Functions to detect under-run errors)

<WBUF> = "1" (if double buffering is enabled and there is no data in buffer2)

Figure 11-14 Transmit Operation in the I/O Interface Mode (SCLK Input Mode)

11.15.1.2 Receive

(1) SCLK Output Mode

The SCLK output can be started by setting the receive enable bit SCxMOD0<RXE> to "1".

• If double buffer is disabled $(SCxMOD2\le WBUF> = "0")$

A clock pulse is outputted from the SCLKx pin and the next data is stored into the shift register each time the CPU reads received data. When all the 8 bits are received, the INTRXx interrupt is generated.

If double buffer is enabled $(SCxMOD2\le WBUF> = "1")$

Data stored in the shift register is moved to the receive buffer and the receive buffer can receive the next frame. A data is moved from the shift register to the receive buffer, the receive buffer full flag SCxMOD2<RBFLL> is set to "1" and the INTRXx is generated.

While data is in the receive buffer, if the data cannot be read from the receive buffer before completing reception of the next 8 bits, the INTRXx interrupt is not generated and the SCLK output stops. In this state, reading data from the receive buffer allows data in the shift register to move to the receive buffer and thus the INTRXx interrupt is generated and data reception resumes.

TOSHIBA

Figure 11-15 Receive Operation in the I/O Interface Mode (SCLK Output Mode)

(2) SCLK Input Mode

In the SCLK input mode, receiving double buffering is always enabled, the received frame can be moved to the receive buffer from the shift register, and the receive buffer can receive the next frame successively.

The INTRXx receive interrupt is generated each time received data is moved to the receive buffer.

If data cannot be read from buffer

Figure 11-16 Receive Operation in the I/O Interface Mode (SCLK Input Mode)

11.15.1.3 Transmit and Receive (Full-duplex)

- (1) SCLK Output Mode
	- If double buffers are disabled $(SCxMOD2\le WBUF> = "0")$

SCLK is outputted when the CPU writes data to the transmit buffer.

Subsequently, 8 bits of data are shifted into receive buffer and the INTRXx receive interrupt is generated. Concurrently, 8 bits of data written to the transmit buffer are outputted from the TXDx pin, the INTTXx transmit interrupt is generated when transmission of all data bits has been completed. Then, the SCLK output stops.

The next round of data transmission and reception starts when the data is read from the receive buffer and the next transmit data is written to the transmit buffer by the CPU. The order of reading the receive buffer and writing to the transmit buffer can be freely determined. Data transmission is resumed only when both conditions are satisfied.

If double buffers are enabled $(SCxMOD2\le WBUF> = "1")$

SCLK is outputted when the CPU writes data to the transmit buffer.

8 bits of data are shifted into the receive shift register, moved to the receive buffer, and the INTRXx interrupt is generated. While 8 bits of data is received, 8 bits of transmit data is outputted from the TXDx pin. When all data bits are sent out, the INTTXx interrupt is generated and the next data is moved from the transmit buffer to the transmit shift register.

If the transmit buffer has no data to be moved to the transmit buffer $(SCxMOD2\leq TBEMP> = "1")$ or when the receive buffer is full $(SCxMOD2\leq RBEULL> =$ "1"), the SCLK output is stopped. When both conditions, receive data is read and transmit data is written, are satisfied, the SCLK output is resumed and the next round of data transmission and reception is started.

Figure 11-17 Transmit/Receive Operation in the I/O Interface Mode (SCLK Output Mode)

(2) SCLK Input Mode

If double buffers are disabled. $(SCxMOD2\le WBUF> = "0")$

When receiving data, double buffer is always enabled regardless of the SCxMOD2 <WBUF> settings.

8-bit data written in the transmit buffer is outputted from the TXDx pin and 8 bit of data is shifted into the receive buffer when the SCLK input becomes active. The INTTXx interrupt is generated upon completion of data transmission. The INTRXx interrupt is generated when the data is moved from shift register to receive buffer after completion of data reception.

Note that transmit data must be written into the transmit buffer before the SCLK input for the next frame (data must be written before the point A in [Figure 11-18](#page-211-0)). Data must be read before completing reception of the next frame data.

If double buffers are enabled. $(SCxMOD2\le WBUF> = "1")$

The interrupt INTTXx is generated at the timing the transmit buffer data is moved to the transmit shift register after completing data transmission from the transmit shift register. At the same time, data received is shifted to the shift register, it is moved to the receive buffer, and the INTRXx interrupt is generated.

Note that transmit data must be written into the transmit buffer before the SCLK input for the next frame (data must be written before the point A in [Figure 11-18](#page-211-0)). Data must be read before completing reception of the next frame data.

Upon the SCLK input for the next frame, transmission from transmit shift register (in which data has been moved from transmit buffer) is started while receive data is shifted into receive shift register simultaneously.

If data in receive buffer has not been read when the last bit of the frame is received, an overrun error occurs. Similarly, if there is no data written to transmit buffer when SCLK for the next frame is input, an under-run error occurs and the dummy data (0xff) is output.

11.15 Operation in Each Mode

Figure 11-18 Transmit/Receive Operation in the I/O Interface Mode (SCLK Input Mode)

11.15.2 Mode 1 (7-bit UART mode)

The 7-bit UART mode can be selected by setting the serial mode control register $(ScxMOD\leq SM[1:0]$) to "01".

In this mode, parity bits can be added to the transmit data stream; the serial mode control register (SCxCR<PE>) controls the parity enable/disable setting.

When \leq PE $>$ is set to "1" (enable), either even or odd parity may be selected using the SCxCR \leq EVEN $>$ bit. The length of the stop bit can be specified using SCxMOD2<SBLEN>.

The following table shows the control register settings for transmitting in the following data format.

11.15.3 Mode 2 (8-bit UART mode)

The 8-bit UART mode can be selected by setting SCxMOD0<SM[1:0]> to "10". In this mode, parity bits can be added and parity enable/disable is controlled using $SCxCR < P E$ if $\langle PE \rangle = "1"$ (enabled), either even or odd parity can be selected using SCxCR<EVEN>.

The control register settings for receiving data in the following format are as follows:

11.15.4 Mode 3 (9-bit UART mode)

The 9-bit UART mode can be selected by setting SCxMOD0<SM[1:0]> to "11". In this mode, parity bits must be disabled (SCxCR<PE $>$ = "0").

The most significant bit (9th bit) is written to SCxMOD0<TB8> for transmitting data. The data is stored in SCxCR<RB8> for receiving data.

When writing or reading data to/from the buffers, the most significant bit must be written or read first before writing or reading to/from SCxBUF.

The stop bit length can be specified using SCxMOD2<SBLEN>.

11.15.4.1 Wakeup function

In the 9-bit UART mode, slave controllers can be operated in the wake-up mode by setting the wakeup function control bit SCxMOD0<WU> to "1".

In this case, the interrupt INTRXx will be generated only when SCxCR<RB8> is set to "1".

Note:The TXDx pin of the slave controller must be set to the open drain output mode using the PxOD register.

Figure 11-19 Serial Links to Use Wake-up Function

11.15.4.2 Protocol

- 1. Select the 9-bit UART mode for the master and slave controllers.
- 2. Set SCxMOD<WU> to "1" for the slave controllers to make them ready to receive data.
- 3. The master controller is to transmit a single frame of data that includes the slave controller select code (8 bits). In this, the most significant bit (bit 8) <TB8> must be set to "1".

- 4. Each slave controller receives the above data frame; if the code received matches with the controller's own select code, it clears the WU bit to "0".
- 5. The master controller transmits data to the designated slave controller (the controller of which SCxMOD<WU> bit is cleared to "0"). In this, the most significant bit (bit 8) <TB8> must be set to "0". **TMPM061FWFG**

Start $\sqrt{\frac{bit0}{1} \times \frac{2}{x}} \times \frac{3}{x} \times \frac{4}{x} \times \frac{5}{x} \times \frac{6}{x} \times \frac{7}{x} \times \frac{8}{x} \times \frac{100}{x} \times \frac{7}{x} \times \frac{100}{x}}$

Each slave controller receives the above data frame; if the code received matches with the c

$$
\xrightarrow{\text{start } \underbrace{\text{bit } 0} \bigtimes \underbrace{1 \bigtimes 2 \bigtimes 3 \bigtimes 4 \bigtimes 5 \bigtimes 6 \bigtimes 7 \bigtimes \text{bit } 8}_{\text{Data}}} \xrightarrow{\text{stop }} \bigcirc
$$

6. The slave controllers with the <WU> bit set to "1" ignore the receive data because the most significant bit (bit 8) <RB8> is set to "0" and thus no interrupt (INTRXx) is generated. Also, the slave controller with the <WU> bit set to "0" can transmit data to the master controller to inform that the data has been successfully received.

11.15 Operation in Each Mode
TOSHIBA

12. Serial Bus Interface (I2C/SIO)

12.1 Outline

Serial bus interface has two operation mode shown below.

- ・ I2C bus mode
- ・ Clock-synchronous 8-bit SIO mode

In the following examination, "x" indicates channel number.

12.2 Block Diagram

The block diagram is shown in Figure 12-1.

Figure 12-1 Block Diagram of Serial Bus Interface

12.3 I2C Bus Mode Data Format

Figure 12-2 shows the data formats used in the I2C bus mode.

P : Stop condition

Figure 12-2 I2C Bus Mode Data Formats

TOSHIBA

12.4 Registers

12.4.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

12.5 Control Registers in the I2C Bus Mode

The following registers control the serial bus interface in the I2C bus mode and provide its status information for monitoring.

12.5.1 SBIxCR0(Control register 0)

12.5.2 SBIxCR1(Control register 1)

TOSHIBA

Note 1: Clear <BC[2:0]> to "000" before switching the operation mode to the SIO mode.

Note 2: For details on the SCL line clock frequency, refer to ["12.6.2 Serial Clock".](#page-225-0)

- Note 3: After a reset, the <SCK[0]/SWRMON> bit is read as "1". However, if the SIO mode is selected at the SBIxCR2 register, the initial value of the <SCK[0]> bit is "0".
- Note 4: The initial value for selecting a frequency is <SCK[2:0]>=000 and is independent of the read initial value.

Note 5: When <BC[2:0]>="001" and <ACK>="0" in master mode, SCL line may be fixed to "L" by falling edge of SCL line after generation of STOP condition and the other devices can not use the bus. In the case of bus which is connected with several master devices, the number of bits per transfer should be set equal or more than 2 before generation of STOP condition.

12.5.3 SBIxCR2(Control register 2)

Note 1: Do not change the serial bus interface operating mode during the transfer.

Note 2: Changes to Port mode after conforming that SDAx/SCLx pin and SOx/SIx/SCKx pin are "High" level.

Note 3: Changes from port mode to I2C bus mode after conforming that SDAx/SCLx pin and SOx/SIx/SCKx pin are "High" level.

Note 4: SBIxCR2 is assigned at same address with SBIxSR. Thus, read-modify-write operation cannot be used.

12.5.4 SBIxSR (Status Register)

12.5.5 SBIxBR0(Serial bus interface baud rate register 0)

12.5.6 SBIxDBR (Serial bus interface data buffer register)

Note 1: The transmission data must be written in to the register from the MSB (bit 7). The received data is stored in the LSB.

Note 2: Since SBIxDBR has independent buffers for writing and reading, a written data cannot be read. Thus, read-modifywrite operation cannot be used.

12.5.7 SBIxI2CAR (I2C bus address register)

Note 1: Please set theSBIxI2CAR<ALS> to "0", except when you use a free data format. It operates as a free data format when setting it to "1". Selecting the master fixes to transmission. Selecting the slave fixes to reception.

Note 2: Do not set SBIxI2CAR to "0x00" in slave mode. (If SBIxI2CAR is set to "0x00", it's recognized that the slave address matches the START byte ("0x01") of the I2C standard received in slave mode.)

12.6 Control in the I2C bus mode

12.6.1 Setting operation mode

Set operation mode by SBIxCR2<SBIM[1:0]>. When SBI is used as I2C bus mode, set <SBIM[1:0]> to "10".

Note 1: Changes to Port mode after conforming that SDAx/SCLx pin and SOx/SIx/SCKx pin are "High" level.

Note 2: Changes from port mode to I2C bus mode after conforming that SDAx/SCLx pin and SOx/SIx/SCKx pin are "High" level.

12.6.2 Serial Clock

12.6.2.1 Clock source

SBIxCR1<SCK[2:0]> specifies the maximum frequency of the serial clock to be output from the SCLx pin in the master mode.

 $f\text{scI} = 1/(t_{\text{LOW}} + t_{\text{HIGH}})$ $=\frac{16}{2^{n}+72}$ fsys

Figure 12-3 Clock source

Note:The maximum speeds in the standard and high-speed modes are specified to 100kHz and 400kHz respectively following the communications standards. Notice that the internal SCL clock frequency is determined by the fsys used and the calculation formula shown above.

12.6.2.2 Clock Synchronization

The I2C bus is driven by using the wired-AND connection due to its pin structure. The first master that pulls its clock line to the "Low" level overrides other masters producing the "High" level on their clock lines. This must be detected and responded by the masters producing the "High" level.

Clock synchronization assures correct data transfer on a bus that has two or more master.

For example, the clock synchronization procedure for a bus with two masters is shown below.

Figure 12-4 Example of Clock Synchronization

At the point a, Master A pulls its internal SCL output to the "Low" level, bringing the SCL bus line to the "Low" level. Master B detects this transition, resets its "High" level period counter, and pulls its internal SCL output level to the "Low" level.

Master A completes counting of its "Low" level period at the point b, and brings its internal SCL output to the "High" level. However, Master B still keeps the SCL bus line at the "Low" level, and Master A stops counting of its "High" level period counting. After Master A detects that Master B brings its internal SCL output to the "High" level and brings the SCL bus line to the "High" level at the point c, it starts counting of its "High" level period.

After that Master finishes counting the "High" level period, the Master pulls the SCLx pin to "Low" and the SCL bus line becomes "Low".

This way, the clock on the bus is determined by the master with the shortest "High" level period and the master with the longest "Low" level period among those connected to the bus.

12.6.3 Setting the Acknowledgement Mode

Setting SBIxCR1<ACK> to "1" selects the acknowledge mode.

When operating as a master, the SBI adds one clock for acknowledgment signal.

In slave mode, the clock for acknowledgement signals is counted.

In transmitter mode, the SBI releases the SDAx pin during clock cycle of acknowledgement to receive acknowledgement signals from the receiver.

In receiver mode, the SBI pulls the SDAx pin to the "Low" level during the clock cycle of acknowledgement and generates acknowledgement signals. Also in slave mode, if a general-call is received, the SBI pulls the SDAx pin to the "Low" level during the clock cycle of acknowledgement and generates acknowledgement signals.

By setting <ACK> to "0", the non-acknowledgment mode is activated. When operating as a master, the SBI does not generate clock for acknowledgement signals. In slave mode, the clock for acknowledgement signals is not counted.

12.6.4 Setting the Number of Bits per Transfer

 $SBIXCR1 < BC[2:0]$ specifies the number of bits of the next data to be transmitted or received.

Under the start condition, $\langle BCI2:0|\rangle$ is set to "000", causing a slave address and the direction bit to be transferred in a packet of eight bits. At other times, $\langle BC[2:0] \rangle$ keeps a previously programmed value.

12.6.5 Slave Addressing and Address Recognition Mode

Setting "0" to SBIxI2CAR<ALS> and set a slave address in SBIxI2CAR<SA[6:0]> in the addressing format, and then the SBI recognizes a slave address transmitted by the master device and receives data in the addressing format.

If <ALS> is set to "1", the SBI does not recognize a slave address and receives data in the free data format. In the case of free data format, a slave address and a direction bit are not recognized; they are recognized as data immediately after generation of the start condition.

12.6.6 Configuring the SBI as a Master or a Slave

Setting SBIxCR2<MST> to "1" configures the SBI to operate as a master device.

Setting <MST> to "0" configures the SBI as a slave device.

<MST> is cleared to "0" by the hardware when SBI detects the stop condition on the bus or the arbitration lost.

12.6.7 Configuring the SBI as a Transmitter or a Receiver

Setting SBIxCR2<TRX> to "1" configures the SBI as a transmitter.

Setting <TRX> to "0" configures the SBI as a receiver.

<TRX> is cleared to "0" by the hardware when SBI detects the stop condition on the bus or the arbitration lost.

If SBI is used in free data format, <TRX> is not changed by the hardware.

If SBI is used in addressing format, <TRX> is set shown as below.

12.6.7.1 Master mode

As a master mode, if SBI receives acknowledgement from a slave device, <TRX> is set shown as below by a hardware.

If SBI does not acknowledgement, <TRX> retains the previous value.

- When the transmitted direction bit is "1", $\langle \text{TRX} \rangle$ is set to "0".
- When the transmitted direction bit is "0", $\langle \text{TRX} \rangle$ is set to "1".

12.6.7.2 Slave mode

As a slave mode, in case of addressing format, if below condition is satisfied, <TRX> is set depended on the direction bit which is sent by a master device.

- ・ When the received slave address is as same as the value set in SBIxI2CAR.
- ・ When SBI receives general-call

<TRX> is set shown as below.

- When the received direction bit is "1", $\langle \text{TRX} \rangle$ is set to "1".
- When the received direction bit is "0", $\langle \text{TRX} \rangle$ is set to "0".

12.6.8 Bus busy monitor

To conform the state of the bus, read SBIxSR<BB>.

OSHIBA

 \langle BB $>$ is set to "1" when SBI detects the start condition on the bus and is cleared to "0" when SBI detects the stop condition on the bus.

When \langle BB $>$ is "1", it is called as bus busy. When \langle BB $>$ is "0", it is called as bus free.

The master device can generate the start condition in only bus free. It should be conform that <BB> is "0".

When <BB> is "1", SBI generates the start condition, the start condition is not generated and the arbitration lost is occurred.

12.6.9 Interrupt Service Request and Release

When INTSBIx is generated, SBIxCR2<PIN> is cleared to "0" and SBI is in interrupt service request state.

SBI pulls SCLx pin to "Low" level.

<PIN> is set to "1" when data is written to or read from SBIxDBR. When the program writes "1" to <PIN>, it is set to "1". However, writing "0" does not clear <PIN> to "0".

If <PIN> is set "1", SCLx pin is released. It takes t_{LOW} from setting <PIN> to "1" to releasing SCLx pin.

Note:When arbitration lost is occurred in the master mode, <PIN> is not cleared to "0" if the received slave address does not match. But INTSBIx is occurred.

12.6.10 Arbitration Lost Detection Monitor

The I2C bus has the multi-master capability (there are two or more masters on a bus), and requires the bus arbitration procedure to ensure correct data transfer.

A master that attempts to generate the start condition while the bus is busy loses bus arbitration, with no start condition occurring on the SDA and SCL bus lines.The I2C-bus arbitration takes place on the SDA bus line.

The arbitration procedure for two masters on a bus is shown below.

Up until the point a, Master A and Master B output the same data. At the point a, Master A outputs the "Low" level and Master B outputs the "High" level.

Then Master A pulls the SDA bus line to the "Low" level because the line has the wired-AND connection. When the SCL line goes high at the point b, the slave device reads the SDA line data, i.e., data transmitted by Master A. At this time, data transmitted by Master B becomes invalid.

This condition of Master B is called "Arbitration Lost". Master B releases its SDAx pin, so that it does not affect the data transfer initiated by another master. If two or more masters have transmitted exactly the same first data word, the arbitration procedure continues with the second data word.

A master compares the SDA bus line level and the internal SDA output level at the rising of the SCL line. If there is a difference between these two values, Arbitration Lost occurs and SBIxSR<AL> is set to "1".

When <AL> is set to "1", SBIxSR<MST, TRX> are cleared to "0", causing the SBI to operate as a slave receiver.Therefore, the serial bus interface circuit stops the clock output during data transfer after <AL> is set to "1".

 $\langle A L \rangle$ is cleared to "0" when data is written to or read from SBIxDBR or data is written to SBIxCR2.

Figure 12-6 Example of Master B Lost Arbitration (D7A = D7B, D6A = D6B)

12.6.11 Slave Address Match Detection Monitor

When the SBI operates as a slave device in the addressing format (SBIxI2CAR<ALS>="0"), SBIxSR<AAS> is set to "1" on receiving the general-call or the slave address that matches the value specified at SBIxI2CAR.

When <ALS> is "1", <AAS> is set to "1" when the first data word has been received.

<AAS> is cleared to "0" when data is written to or read from SBIxDBR.

12.6.12 General-call Detection Monitor

When the SBI operates as a slave device, SBIxSR<ADO> is set to "1" when it receives the general-call address; i.e., the eight bits following the start condition are all zeros.

<ADO> is cleared to "0" when the start or stop condition is detected on the bus.

12.6.13 Last Received Bit Monitor

SBIxSR<LRB> is set to the SDA line value that was read at the rising of the SCL line.

In the acknowledgment mode, reading SBIxSR<LRB> immediately after generation of the INTSBIx interrupt request causes ACK signal to be read.

12.6.14 Data Buffer Register (SBIxDBR)

To read a data or to write a data, SBIxDBR is read or written from or to SBIxDBR.

When the SBI is in the master mode, after writing a slave address and a direction bit to this register in SBIxDBR, the start condition is generated, SBI transmits a slave address and a direction bit to slave device.

12.6.15 Baud Rate Register (SBIxBR0)

The SBIxBR0<I2SBI> register determines if the SBI operates or not when it enters the IDLE mode.

This register must be programmed before executing an instruction to switch to the standby mode.

12.6.16 Software Reset

If SBI locks up due to external noise, it can be initialized by using a software reset.

Writing "10" followed by "01" to SBIxCR2<SWRST[1:0]> generates a reset signal that initializes SBI. When writing SBIxCR2<SWRST[1:0]>, set "10" to SBIxCR2<SBIM[1:0]> for I2C bus mode. After a reset, all control registers and status flags are initialized to their reset values. When SBI is initialized, <SWRST> is automatically cleared to "00".

Note:A software reset causes the SBI operating mode to switch from the I2C mode to the port mode.

12.7 Data transfer Procedure in the I2C Bus mode.

12.7.1 Device Initialization

First, program SBIxCR1<ACK, SCK[2:0]>. Writing "000" to SBIxCR1<BC[2:0]> at the time.

Next, program SBIxI2CAR by specifying a slave address at <SA[6:0]> and an address recognition mode at <ALS>. (<ALS> must be cleared to "0" when using the addressing format).

To configure the Serial Bus Interface as a slave receiver, ensure that the serial bus interface pin is at "High" first. Then write "0" to SBIxCR2<MST, TRX, BB>, "1" to <PIN>, "10" to <SBIM[1:0]> and "0" to the bit 1 and 0.

Note:Initialization of the serial bus interface circuit must be completed within a period that any device does not generate start condition after all devices connected to the bus were initialized. If this rule is not followed, data may not be received correctly because other devices may start transfer before the initialization of the serial bus interface circuit is completed.

Specifies ACK and SCL clock. Specifies a slave address and an address recognition mode.

Note:X; Don't care

Configures the SBI as an I2C bus mode and a slave receiver.

12.7.2 Generating the Start Condition and a Slave Address

The following steps are required to generate the start condition and slave address.

First, ensure that the bus is bus free (SBIxSR<BB> = "0"), then SBIxCR1<ACK> is set to "1" to select the acknowledgment mode. Write to SBIxDBR a slave address and a direction bit to be transmitted.

When $\langle BB \rangle = "0"$, writing "1111" to SBIxCR2 $\langle MST, TRX, BB, PIN \rangle$ generates the start condition on the bus.

Following the start condition, the SBI generates nine clocks from the SCLx pin.

The SBI outputs the slave address and the direction bit specified at SBIxDBR with the first eight clocks, and releases the SDA line in the ninth clock to receive an acknowledgment signal from the slave device.

The INTSBIx interrupt request is generated on the falling of the ninth clock, and <PIN> is cleared to "0". The SBI holds the SCL line at the "Low" level while <PIN> is "0". <TRX> changes its value according to the transmitted direction bit at generation of the INTSBIx interrupt request, provided that an acknowledgment signal has been returned from the slave device.

Note:To output salve address, check with software that the bus is free before writing to SBIxDBR. If this rule is not followed, data being output on the bus may get ruined.

OSHIBA

Figure 12-7 Generation of the Start Condition and a Slave Address

12.7.3 Transferring a Data Word

At the end of a data word transfer, the INTSBIx interrupt is generated to test <MST> to determine whether the SBI is in the master or slave mode.

12.7.3.1 Master mode (<MST> = "1")

Test <TRX> to determine whether the SBI is configured as a transmitter or a receiver.

(1) Transmitter mode (<TRX> = "1")

Test <LRB>. If <LRB> is "1", that means the receiver requires no further data.

The master then generates the stop condition as described later to stop transmission.

If <LRB> is "0", that means the receiver requires further data.

If the next data to be transmitted has eight bits, the data is written into SBIxDBR. If the data has different length, $\langle BC[2:0] \rangle$ are programmed and the transmit data is written into SBIxDBR.

Writing the data makes $\langle PIN \rangle$ to "1", causing the SCL x pin to generate a serial clock for transferring a next data word, and the SDAx pin to transfer the data word.

After the transfer is completed, the INTSBIx interrupt request is generated, <PIN> is cleared to "0", and the SCLx pin is pulled to the "Low" level.

To transmit more data words, test <LRB> again and repeat the above procedure.

INTSBIx interrupt

Figure 12-8 <BC[2:0]>= "000",<ACK>= "1" (Transmitter Mode)

(2) Receiver mode (<TRX> = "0")

If the next data to be transmitted has eight bits, the received data is read from into SBIxDBR.

If the data has different length, $\langle BC|2:0\rangle$ are programmed and the received data is read from SBIxDBR to release the SCL line. (The data read immediately after transmission of a slave address is undefined.)

On reading the data, <PIN> is set to "1", and the serial clock is output to the SCLx pin to transfer the next data word. In the last bit, when the acknowledgment signal becomes the "Low" level, "0" is output to the SDAx pin.

After that, the INTSBIx interrupt request is generated, and $\langle PIN \rangle$ is cleared to "0", pulling the SCLx pin to the "Low" level. Each time the received data is read from SBIxDBR, one-word transfer clock and an acknowledgement signal are output.

Figure 12-9 <BC[2:0]>= "000",<ACK>= "1" (Receiver Mode)

To terminate the data transmission from the transmitter, <ACK> is cleared to "0" immediately before reading the data word second to last.

This disables generation of an acknowledgment clock for the last data word.

When the transfer is completed, an interrupt request is generated. In tis interrupt processing, <BC [2:0]> must be set to "001" and the data must be read so that a clock is generated for 1-bit transfer.

At this time, so the master is a receiver, the master holds the SDA line at the "High" level. The transmitter receives this "High" level as ACK signal, the master receiver can inform to the transmitter the end of transfer.

In this interrupt processing for terminating the reception of 1-bit data, the stop condition is generated to terminate the data transfer.

Example: When receiving N data word

Note:X; Don't care

12.7.3.2 Slave mode (<MST> = "0")

In the slave mode, SBI generates the INTSBIx interrupt request when SBI receives any slave address or general-call from master device, when SBI completes to transfers a data after SBI received its slave address or general-call.

Also, if the SBI detects Arbitration Lost in the master mode, it switches to the slave mode. When the completion of data word transfer in which Arbitration Lost is detected, the INTSBIx interrupt request is generated.

When INSBIx interrupt request, <PIN> is cleared to "0", and SCLx pin is pulled to the "Low" level.

When data is written to or read from SBIxDBR or when <PIN> is set to "1", SCLx pin is released after a period of t_{LOW} .

In the slave mode, the normal slave mode processing or the processing as a result of Arbitration Lost is carried out and it changes from master mode to slave mode.

SBIxSR<AL>, <TRX>, <AAS> and <ADO> are tested to determine the processing required.

["Table 12-1 Processing in Slave Mode"](#page-236-0)shows the slave mode states and required processing.

Example: When the received slave address matches the SBI's slave address and the direction bit is "1" in the slave receiver mode.

INTSBIx interrupt

TOSHIBA

Note:X; Don't care

Table 12-1 Processing in Slave Mode

$<$ TRX $>$	<al< th=""><th>$<$AAS></th><th>ADO</th><th>State</th><th>Processing</th></al<>	$<$ AAS>	ADO	State	Processing
1	1		Ω	Arbitration Lost is detected while the slave address was being transmitted and the SBI received a slave address with the direction bit "1" transmitted by an- other master.	Set the number of bits in a data word to <bc[2:0]> and write the transmit data into SBIxDBR.</bc[2:0]>
	Ω	1	Ω	In the slave receiver mode, the SBI received a slave address with the direction bit "1" transmitted by the master.	
		Ω	Ω	In the slave transmitter mode, the SBI has comple- ted a transmission of one data word.	Test LRB. If it has been set to "1", that means the re- ceiver does not require further data. Set <pin> to 1 and reset <trx> to 0 to release the bus. If <lrb> has been reset to "0", that means the receiver re- quires further data. Set the number of bits in the da- ta word to $\leq BC[2:0]$ and write the transmit data to the SBIxDBR.</lrb></trx></pin>
Ω	1	1	1/0	Arbitration Lost is detected while a slave address is being transmitted, and the SBI receives either a slave address with the direction bit "0" or a general- call address transmitted by another master.	Read the SBIxDBR (a dummy read) to set <pin> to 1, or write "1" to <pin>.</pin></pin>
		Ω	0	Arbitration Lost is detected while a slave address or a data word is being transmitted, and the trans- fer is terminated.	
	Ω	1	1/0	In the slave receiver mode, the SBI received either a slave address with the direction bit "0" or a general- call address transmitted by the master.	
		0	1/0	In the slave receiver mode, the SBI has completed a reception of a data word.	Set the number of bits in the data word to <bc [2:0] > and read the received data from SBIxDBR.</bc

12.7.4 Generating the Stop Condition

When SBIxSR<BB> is "1", writing "1" to SBIxCR2<MST, TRX, PIN> and "0" to <BB> causes SBI to start a sequence for generating the stop condition on the bus.

Do not alter the contents of <MST, TRX, BB, PIN> until the stop condition appears on the bus.

If another device is holding down the SCL bus line, SBI waits until the SCL line is released.

After that, the SDAx pin goes "High", causing the stop condition to be generated.

Figure 12-11 Generating the Stop Condition

12.7.5 Restart Procedure

Restart is used when a master device changes the data transfer direction without terminating the transfer to a slave device. The procedure of generating a restart in the master mode is described below.

First, write SBIxCR2<MST, TRX, BB> to "0" and write "1" to <PIN> to release the bus. At this time, the SDAx pin is held at the "High" level and the SCLx pin is released. Because no stop condition is generated on the bus, other devices recognize that the bus is busy.

Then, test SBIxSR<BB> and wait until it becomes "0" to ensure that the SCLx pin is released.

Next, test <LRB> and wait until it becomes "1" to ensure that no other device is pulling the SCL bus line to the "Low" level.

Once the bus is determined to be free by following the above procedures, follow the procedures described in ["12.7.2 Generating the Start Condition and a Slave Address"](#page-231-0) to generate the start condition.

To satisfy the setup time of restart, at least 4.7μs wait period (in the standard mode) must be created by the software after the bus is determined to be bus free.

Note 1: Do not write <MST> to "0" when it is "0". (Restart cannot be initiated.)

Note 2: When the master device is acting as a receiver, data transmission from the slave device which serves as a transmitter must be completed before generating a restart. To complete data transfer, slave device must receive a "High" level acknowledge signal. For this reason, <LBR> before generating a restart becomes "1", the rising edge of the SCL line is not detected even <LBR>= "1" is confirmed by following the restart procedure. To check the status of the SCL line, read the port.

Figure 12-12 Timing Chart of Generating a Restart

Precautions on Use of Multi-master

Prepare recovery process by software in case that communication is in lock state in multi-master mode.

Example of recovery process

- 1. Start timer for timeout detection synchronizing with starting communication.
- 2. If a serial interface interrupt (INTSBIx) does not occur within the specified time, a timeout occurs and the MCU determines that communication is locked up.
- 3. Do software reset on serial bus interface to release the condition that communication is locked up.
- 4. Adjust transmission timings. (Note)
- 5. Resend transmission data.
- Note: Adjust transmission timing between the MCUs to avoid overlapping the transmission timing.

12.8 Control register of SIO mode

The following registers control the serial bus interface in the clock-synchronous 8-bit SIO mode and provide its status information for monitoring.

12.8.1 SBIxCR0(control register 0)

12.8.2 SBIxCR1(Control register 1)

Note 1: After a reset, the <SCK[0]> bit is read as "1". However, if the SIO mode is selected at the SBIxCR2 register, the initial value is read as "0". In this document, the value written in the column "after reset" is the value after setting the SIO mode in the initial state. The descriptions of the SBIxCR2 register and the SBIxSR register are the same. Note 2: Set <SIOS> to "0" and <SIOINH> to "1" before programming the transfer mode and the serial clock.

12.8 Control register of SIO mode

12.8.3 SBIxDBR (Data buffer register)

Note:Since SBIxDBR has independent buffers for writing and reading, a written data cannot be read. Thus, readmodify-write operation cannot be used.

12.8.4 SBIxCR2(Control register 2)

Note 1: In this document, the value written in the column "after reset" is the value after setting the SIO mode in the initial state. Note 2: Make sure that modes are not changed during a communication session.

Note 3: Changes from port mode to I2C bus mode after conforming that SDAx/SCLx pin and SOx/SIx/SCKx pin are "High" level.

Note 4: SBIxCR2 is assigned at same address with SBIxSR. Thus, read-modify-write operation cannot be used.

12.8.5 SBIxSR (Status Register)

Note:In this document, the value written in the column "after reset" is the value after setting the SIO mode in the initial state.

12.8.6 SBIxBR0 (Baud rate register 0)

12.9 Control in SIO mode

12.9.1 Serial Clock

12.9.1.1 Clock source

Internal or external clocks can be selected by programming SBIxCR1<SCK[2:0]>.

(1) Internal clocks

In the internal clock mode, one of the seven frequencies can be selected as a serial clock, which is output to the outside through the SCKx pin.

At the beginning of a transfer, the SCKx pin output becomes the "High" level.

If the program cannot keep up with this serial clock rate in writing the transmit data or reading the received data, the SBI automatically enters a wait period. During this period, the serial clock is stopped automatically and the next shift operation is suspended until the processing is completed.

Figure 12-13 Automatic Wait

(2) External clock (<SCK[2:0]> = "111")

The SBI uses an external clock supplied from the outside to the SCKx pin as a serial clock.

For proper shift operations, the serial clock at the "High" and "Low" levels must have the pulse widths as shown below.

Figure 12-14 Maximum Transfer Frequency of External Clock Input

TOSHIBA

12.9.1.2 Shift Edge

Leading-edge shift is used in transmission. Trailing-edge shift is used in reception.

- Leading-edge shift

Data is shifted at the leading edge of the serial clock (or the falling edge of the SCKx pin input/output).

Trailing-edge shift

Data is shifted at the trailing edge of the serial clock (or the rising edge of the SCKx pin input/output).

Figure 12-15 Shift Edge

After writing the transmit data, writing "1" to SBIxCR1<SIOS> starts the transmission. The transmit data is moved from SBIxDBR to a shift register and output to the SOx pin, with the least-significant bit (LSB) first, in synchronization with the serial clock. Once the transmit data is transferred to the shift register, SBIxDBR becomes empty, and the INTSBIx (buffer-empty) interrupt is generated, requesting the next transmit data.

The transmit mode, the receive mode or the transmit/receive mode can be selected by programming

Set the control register to the transmit mode and write the transmit data to SBIxDBR.

In the internal clock mode, the serial clock will be stopped and automatically enter the wait state, if next data is not loaded after the 8-bit data has been fully transmitted. The wait state will be cleared when SBIxDBR is loaded with the next transmit data.

In the external clock mode, SBIxDBR must be loaded with data before the next data shift operation is started. Therefore, the data transfer rate varies depending on the maximum latency between when the interrupt request is generated and when SBIxDBR is loaded with data in the interrupt service program.

At the beginning of transmission, the same value as in the last bit of the previously transmitted data is output in a period from setting SBIxSR<SIOF> to "1" to the falling edge of SCK.

Transmission can be terminated by clearing <SIOS> to "0" or setting <SIOINH> to "1" in the IN-TSBIx interrupt service program. If <SIOS> is cleared, remaining data is output before transmission ends. The program checks SBIxSR<SIOF> to determine whether transmission has come to an end. \leq SIOF $>$ is cleared to "0" at the end of transmission. If \leq SIOINH $>$ is set to "1", the transmission is aborted immediately and <SIOF> is cleared to "0".

When in the external clock mode, <SIOS> must be cleared to "0" before next data shifting. If <SIOS> does not be cleared to "0" before next data shifting, SBI output dummy data and stopped.

12.9 Control in SIO mode

12.9.2 Transfer Modes

SBIxCR1<SIOM[1:0]>.

12.9.2.1 8-bit transmit mode

Example: Example of programming (external clock) to terminate transmission by <SIOS>

12.9.2.2 8-bit receive mode

Set the control register to the receive mode. Then writing "1" to SBIxCR1<SIOS> enables reception.Data is taken into the shift register from the SIx pin, with the least-significant bit (LSB) first, in synchronization with the serial clock. Once the shift register is loaded with the 8-bit data, it transfers the received data to SBIxDBR and the INTSBIx (buffer-full) interrupt request is generated to request reading the received data. The interrupt service program then reads the received data from SBIxDBR.

In the internal clock mode, the serial clock will be stopped and automatically be in the wait state until the received data is read from SBIxDBR.

In the external clock mode, shift operations are executed in synchronization with the external clock. The maximum data transfer rate varies, depending on the maximum latency between generating the interrupt request and reading the received data

Reception can be terminated by clearing <SIOS> to "0" or setting <SIOINH> to "1" in the INTSBIx interrupt service program. If <SIOS> is cleared, reception continues until all the bits of received data are written to SBIxDBR. The program checks SBIxSR<SIOF> to determine whether reception has come to an end. <SIOF> is cleared to "0" at the end of reception. After confirming the completion of the reception, last received data is read. If <SIOINH> is set to "1", the reception is aborted immediately and <SIOF> is cleared to "0". (The received data becomes invalid, and there is no need to read it out.)

Note:The contents of SBIxDBR will not be retained after the transfer mode is changed. The ongoing reception must be completed by clearing <SIOS> to "0" and the last received data must be read before the transfer mode is changed.

Reg. ← SBIxDBR Reads the received data.

Figure 12-17 Receive Mode (Example: Internal Clock)

12.9.2.3 8-bit transmit/receive mode

Set the control register to the transfer/receive mode. Then writing the transmit data to SBIxDBR and setting SBIxCR1<SIOS> to "1" enables transmission and reception.The transmit data is output through the SOx pin at the falling of the serial clock, and the received data is taken in through the SI pin at the rising of the serial clock, with the least-significant bit (LSB) first. Once the shift register is loaded with the 8 bit data, it transfers the received data to SBIxDBR and the INTSBIx interrupt request is generated.The interrupt service program reads the received data from the data buffer register and writes the next transmit data. Because SBIxDBR is shared between transmit and receive operations, the received data must be read before the next transmit data is written.

In the internal clock operation, the serial clock will be automatically in the wait state until the received data is read and the next transmit data is written.

In the external clock mode, shift operations are executed in synchronization with the external serial clock. Therefore, the received data must be read and the next transmit data must be written before the next shift operation is started.The maximum data transfer rate for the external clock operation varies depending on the maximum latency between when the interrupt request is generated and when the transmit data is written.

At the beginning of transmission, the same value as in the last bit of the previously transmitted data is output in a period from setting <SIOF> to "1" to the falling edge of SCKx.

Transmission and reception can be terminated by clearing <SIOS> to "0" or setting SBIxCR1<SIOINH> to "1" in the INTSBIx interrupt service program. If <SIOS> is cleared, transmission and reception continue until the received data is fully transferred to SBIxDBR. The program checks SBIxSR<SIOF> to determine whether transmission and reception have come to an end. <SIOF> is cleared to "0" at the end of transmission and reception.If <SIOINH> is set to "1", the transmission and reception is aborted immediately and <SIOF> is cleared to "0".

Note:The contents of SBIxDBR will not be retained after the transfer mode is changed. The ongoing transmission and reception must be completed by clearing <SIOS> to "0" and the last received data must be read before the transfer mode is changed.

12.9 Control in SIO mode

Figure 12-18 Transmit/Receive Mode (Example: Internal Clock)

12.9.2.4 Data retention time of the last bit at the end of transmission

Under the condition SBIxCR1<SIOS>= "0", the last bit of the transmitted data retains the data of SCK rising edge as shown below. Transmit mode and transmit/receive mode are the same.

Figure 12-19 Data retention time of the last bit at the end of transmission
13. 10-bit Analog/Digital Converter (ADC)

13.1 Outline

A 10-bit, sequential-conversion analog/digital converter (AD converter) is built into the TMPM061FWFG. For details, refer to "Product information" to confirm usable channels and settings.

13.2 Configuration

[Figure 13-1](#page-253-0) shows the block diagram of this AD converter.

Note:VREFL and AVSS are shared.

Figure 13-1 10-bit AD Converter Block Diagram

TOSHIBA

13.3 Registers

13.3.1 Register list

The control registers and addresses of the AD converter are as follows.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

13.3 Registers

13.3.2 ADCLK (Conversion Clock Setting Register)

Note 1: Do not change the setting of the AD conversion clock during AD conversion.

Note 2: The AD conversion clock ADCLK must be selected so that the relationship "ADCLK ≤ fsys" is satisfied .

A clock count is required to satisfy the condition that described below.

13.3.3 ADMOD0 (Mode Control Register 0)

Note:Before starting AD conversion, write "1" to the <VREFON> bit, wait for 3μs during which time the internal reference voltage should stabilize, and then write "1" to the ADMOD0<ADS>**.**

Selection of analog input channel

13.3.5 ADMOD2 (Mode Control Register 2)

Note:This bit is "0" cleared when it is read.

channel | AIN8 | AIN9 | AIN10 | AIN11 | AIN12 | AIN13 | AIN14 | AIN15

Conversion
channel

Selection of analog input channel

13.3.6 ADMOD3 (Mode Control Register 3)

13.3.7 ADMOD4 (Mode Control Register 4)

Note 1: For details, refer to "Product information" to confirm H/W source.

Note 2: The external trigger cannot be used for H/W activation of AD conversion when it is used for H/W activation of top priority AD conversion.

Note 3: A software reset initializes all the registers except for ADCLK<ADCLK>.

13.3.8 ADMOD5 (Mode Control Register 5)

13.3 Registers

13.3.9 ADREGn (Conversion Result Register n: n = 0 to 11)

Note:Access to this register must be a half word or a word access.

13.3.10 ADREGSP (AD Conversion Result Register SP)

Note:Access to this register must be a half word or a word access.

13.3.11 ADCMP0 (AD Conversion Result Comparison Register 0)

Note:To write values into this register, the AD monitor function 0 must be disabled (ADMOD3<ADBSV0>="0").

13.3.12 ADCMP1 (AD Conversion Result Comparison Register 1)

Note:**To write values into this register, the AD monitor function 1 must be disabled (AD-MOD5<ADBSV1>="0").**

13.4 Description of Operations

13.4.1 Analog Reference Voltage

The "High" level of the analog reference voltage shall be applied to the VRFEH pin, and the "Low" shall be applied to the VREFL pin.

To start AD conversion, make sure that you first write "1" to the <VREFON> bit, wait for 3 μs during which time the internal reference voltage should stabilize, and then write "1" to the ADMOD0<ADS> bit.

If you do not use ADC function, write "0" to the ADMOD1<DACON>. The consumption current of the analog circuit is reduced.

Note:VREFL and AVSS are shared by TMPM061FWFG.

13.4.2 AD Conversion Mode

Two types of AD conversion are supported: normal AD conversion and top-priority AD conversion.

For normal AD conversion, the following four operation modes are supported.

13.4.2.1 Normal AD conversion

For normal AD conversion, the following four operation modes are supported and the operation mode is selected with the ADMOD0 <REPEAT> <SCAN>.

- ・ Fixed channel single conversion mode
- ・ Channel scan single conversion mode
- ・ Fixed channel repeat conversion mode
- ・ Channel scan repeat conversion mode

For channel scan mode, the following three modes are supported and the operation mode is selected with the ADMOD1<ADSCN>.

- ・ 4-channel scan mode
- ・ 8-channel scan mode
- ・ 12-channel scan mode

(1) Fixed channel single conversion mode

If ADMOD0<REPEAT, SCAN> is set to "00", "AD conversion is performed in the fixed channel single conversion mode.

In this mode, AD conversion is performed once for one channel selected. After AD conversion is completed, ADMOD0<EOCFN> is set to "1", ADMOD0<ADBFN> is cleared to "0", and the AD conversion completion interrupt request (INTAD) is generated. <EOCFN> is cleared to "0" upon read.

(2) Channel scan single conversion mode

If ADMOD0 <REPEAT, SCAN> is set to "01," AD conversion is performed in the channel scan single conversion mode.

In this mode, AD conversion is performed once for each scan channel selected. After AD scan conversion is completed, ADMOD0<EOCFN> is set to "1", ADMOD0<ADBFN> is cleared to "0", and the conversion completion interrupt request (INTAD) is generated. <EOCFN> is cleared to "0" upon read.

(3) Fixed channel repeat conversion mode

If ADMOD0<REPEAT, SCAN> is set to "10", AD conversion is performed in fixed channel repeat conversation mode.

In this mode, AD conversion is performed repeatedly for one channel selected. After AD conversion is completed, ADMOD0<EOCFN> is set to "1". ADMOD0<ADBFN> is not cleared to "0". It remains at "1". The timing with which the conversion completion interrupt request (INTAD) is generated can be selected by setting ADMOD0<ITM> to an appropriate setting. <EOCFN> is set with the same timing as this interrupt INTAD is generated.

<EOCFN> is cleared to "0" upon read.

(4) Channel scan repeat conversion mode

If ADMOD0<REPEAT, SCAN> is set to "11", AD conversion is performed in the channel scan repeat conversion mode.

In this mode, AD conversion is performed repeatedly for a scan channel selected. Each time one AD scan conversion is completed, ADMOD0<EOCFN> is set to "1", and the conversion completion interrupt request (INTAD) is generated. ADMOD0<ADBFN> is not cleared to "0". It remains at "1". <EOCFN> is cleared to "0" upon read.

13.4.2.2 Top-priority AD conversion

By interrupting ongoing normal AD conversion, top-priority AD conversion can be performed.

The fixed-channel single conversion is automatically selected, irrespective of the ADMOD0 <RE-PEAT, SCAN> setting. When conditions to start operation are met, a conversion is performed just once for a channel designated by ADMOD2<HPADCH>. When conversion is completed, the top-priority AD conversion completion interrupt (INTADHP) is generated, and ADMOD2<EOCFHP> showing the completion of AD conversion is set to "1". <ADBFHP> returns to "0". EOCFHP flag is cleared to "0" upon read.

Top-priority AD conversion activated while top-priority AD conversion is under way is ignored.

13.4.3 AD Monitor Function

There are two channels of AD monitor function.

If ADMOD3<ADOBSV0> and ADMOD5<ADOBSV1> are set to "1", the AD monitor function is enabled. If the value of the conversion result register specified by ADMOD3 <ADREGS0> and ADMOD5 <ADREGS1> becomes larger or smaller ("Larger" or "Smaller" to be designated by ADMOD3 <ADOBIC0> and ADMOD5 <ADBIC1>) than the value of a comparison register, the AD monitor function interrupt (IN-TADM0, INTADM1) is generated. This comparison operation is performed each time a result is stored in a corresponding conversion result register.

If the conversion result register assigned to perform the AD monitor function is continuously used without reading the conversion result, the conversion result is overwritten. The conversion result storage flag $\langle ADRxRF\rangle$ and the overrun flag $\langle OVRx\rangle$ remain being set.

13.4.4 Selecting the Input Channel

How the input channel is selected is different depending on AD converter operation mode to be used.

- 1. Normal AD conversion mode
	- If the analog input channel is used in a fixed state (ADMOD0 $SCAN$ = "0")

One channel is selected from analog input pins by setting ADMOD1<ADCH> to an appropriate setting.

• If the analog input channel is used in a scan state $(ADMOD0 = "1")$

One scan mode is selected from the scan modes by setting ADMOD1 <ADCH> and <ADSCN> to an appropriate setting.

2. Top-priority AD conversion mode

One channel is selected from analog input pins by setting ADMOD2<HPADCH> to an appropriate setting.

13.4.5 AD Conversion Details

13.4.5.1 Starting AD Conversion

Normal AD conversion is activated by setting ADMOD0<ADS> to "1". Top-priority AD conversion is activated by setting ADMOD2<HPADCE> to "1".

Four operation modes are made available to normal AD conversion. In performing normal AD conversion, one of these operation modes must be selected by setting ADMOD0<REPEAT, SCAN> to an appropriate setting. For top-priority AD conversion, only one operation mode can be used: fixed channel single conversion mode.

Normal AD conversion can be activated using the H/W activation source selected by ADMOD4 <ADHS>, and top-priority AD conversion can be activated using the HW activation source selected by AD-MOD4 <HADHS>. If bits of <ADHS> and <HADHS> are "0", normal and top-priority AD conversions are activated in response to the input of a falling edge through the ADTRG pin. If these bits are "1", conversion is activated in response to match detection of timer.

To permit H/W activation, set ADMOD4 <ADHTG> to "1" for normal AD conversion and set AD-MOD4 <HADHTG> to "1" for top-priority AD conversion.

Software activation is still valid even after H/W activation has been permitted.

- Note 1: Some products don't provide the ADTRG pin.
- Note 2: When an external trigger is used for the HW activation source of a top-priority AD conversion, an external trigger cannot be set for activating normal AD conversion H/W.
- Note 3: For details, refer to "Product information" to confirm usable match detection of timer.

13.4.5.2 AD Conversion

When normal AD conversion starts, the AD conversion Busy flag (ADMOD0<ADBFN>) showing that AD conversion is under way is set to "1".

When top-priority AD conversion starts, the top-priority AD conversion Busy flag (ADMOD2 <ADBFHP>) showing that AD conversion is underway is set to "1".

At that time, the value of the Busy flag ADMOD0<ADBFN> for normal AD conversion before the start of top-priority AD conversions are retained. The value of the conversion completion flag AD-MOD0<EOCFN> for normal AD conversion before the start of top-priority AD conversion can also be retained.

Note:Normal AD conversion must not be activated when top-priority AD conversion is under way.

13.4.5.3 Top-priority AD conversion during normal AD conversion

If top-priority AD conversion has been activated during normal AD conversion, ongoing normal AD conversion is suspended, and restarts normal AD conversion after top-priority AD conversion is completed.

If ADMOD2<HPADCE> is set to "1" during normal AD conversion, ongoing normal AD conversion is suspended, and the top-priority AD conversion starts; specifically, AD conversion (fixed-channel single conversion) is executed for a channel designated by ADMOD2<HPADCH>. After the result of this toppriority AD conversion is stored in the storage register ADREGSP, normal AD conversion is resumed.

If H/W activation of top-priority AD conversion is authorized during normal AD conversion, ongoing AD conversion is discontinued when requirements for activation using a H/W activation resource are met, and top-priority AD conversion (fixed-channel single conversion) starts for a channel designated by ADMOD2<HPADCH>. After the result of this top-priority AD conversion is stored in the storage register ADREGSP, normal AD conversion is resumed.

For example, if channel repeat conversion is activated for channels AIN0 through AIN3 and if <HPADCE> is set to "1" during AIN2 conversion, AIN2 conversion is suspended, and conversion is performed for a channel designated by <HPADCH> (AIN11 in the case shown below). After the result of conversion is stored in ADREGSP, channel repeat conversion is resumed, starting from AIN2.

13.4.5.4 Stopping Repeat Conversion Mode

To stop the AD conversion operation in the repeat conversion mode (fixed-channel repeat conversion mode or channel scan conversion mode), write "0" to ADMOD0<REPEAT>. When ongoing AD conversion is completed, the repeat conversion mode terminates, and ADMOD0<ADBFN> is set to "0".

13.4.5.5 Reactivating normal AD conversion

To reactivate normal AD conversion while the conversion is underway, a software reset (ADMOD3 <ADRST>) must be performed before starting AD conversion. The H/W activation method must not be used to reactivate normal AD conversion.

'OSHIBA

13.4.5.6 Conversion completion

(1) Normal AD conversion completion

When normal AD conversion is completed, the AD conversion completion interrupt (INTAD) is generated. The result of AD conversion is stored in the storage register, and two registers change: the register ADMOD0 <EOCFN> which indicates the completion of AD conversion and the register ADMOD0 <ADBFN>.

For details, refer to [Table 13-2](#page-271-0) and [Table 13-3](#page-272-0) to confirm storage register corresponding to the conversion mode.

Interrupt requests, flag changes are as shown below.

・ Fixed-channel single conversion mode

After AD conversion completed, ADMOD0<EOCFN> is set to "1", AD-MOD0<ADBFN> is cleared to "0", and the interrupt request is generated.

・ Channel scan single conversion mode

After the channel scan conversion is completed, ADMOD0<EOCFN> is set to "1", AD-MOD0<ADBFN> is set to "0", and the interrupt request INTAD is generated.

・ Fixed-channel repeat conversion mode

ADMOD0<ADBFN> is not cleared to "0". It remains at "1". The timing with which the interrupt request INTAD is generated can be selected by setting ADMOD0<ITM> to an appropriate setting. ADMOD0<EOCFN> is set with the same timing as this interrupt INTAD is generated.

・ Channel scan repeat conversion mode

Each time one AD scan conversion is completed, ADMOD0 <EOCF> is set to "1" and interrupt request INTAD is generated. ADMOD0<ADBFN> is not cleared to "0". It remains at "1".

(2) Top-priority AD conversion completion

After the AD conversion is completed, the top-priority AD conversion completion interrupt (IN-TADHP) is generated, and ADMOD2<EOCFHP> which indicates the completion of top-priority AD conversion is set to "1".

AD conversion results are stored in the AD conversion result register SP.

(3) Data polling

To confirm the completion of AD conversion without using interrupts, data polling can be used. When AD conversion is completed, ADMOD0<EOCFN> is set to "1". To confirm the completion of AD conversion and to obtain the results, poll this bit.

AD conversion result storage register must be read by half word or word access. If \langle OVRx \rangle = "0" and $\langle ADRxRF \rangle =$ "1", a correct conversion result has been obtained.

13.4.5.7 Interrupt generation timings and AD conversion result storage register

Table 13-1 shows a relation in the following three items: AD conversion modes, interrupt generation timings and flag operations. Table 13-2 and [Table 13-3](#page-272-0) shows a relation between analog channel inputs and AD conversion result registers.

Note:ADMOD0<EOCFN> and ADMOD2<EOCFHP> are cleared upon read.

TOSHIBA

 Γ

Table 13-3 Result registers (Except for fixed-channel repet conversion mode)

T

Cautions

The result value of AD conversion may vary depending on the fluctuation of the supply voltage, or may be affected by noise. When using analog input pins and ports alternately, do not read and write ports during conversion because the conversion accuracy may be reduced. Also the conversion accuracy may be reduced if the output ports current fluctuate during AD conversion. Please take counteractive measures with the program such as averaging the AD conversion results.

13. 10-bit Analog/Digital Converter (ADC)

13.4 Description of Operations

TOSHIBA

14. 24-bit ΔΣ Analog/Digital Converter (DSADC)

A reference voltage circuit (BGR) used in the DSADC is shared with a temperature sensor and needs to set the control register (TEMPEN) of temperature sensor.

14.1 Features

DSADC has the following features:

・ Conversion start

Conversion started by software

- ・ Conversion modes
	- Single conversion
	- Repeat conversion
- ・ Status flags
	- Conversion result store flag
	- Overrun flag
	- Conversion end flag
	- Conversion flag
- ・ Conversion clock can be divided by below ratios.

fc/1, fc/2, fc/4, fc/8

- ・ Conversion end interrupt output
- ・ Conversion start correct function
- Synchronous start function for multiple units

When DSADC is used, provide pin treatments as follows:

- ・ Do not connect VREFINx to a reference voltage.
- ・ Connect AGNDREFx to DVSS level.
- ・ Connect a 1μF capacitor to between VREFINx and AGNDREFx.

When DSADC is not used, below settings are required.

・ Connect AGNDREFx to DVSS level.

When a temperature sensor is also not used, a reference voltage circuit requires below settings.

- ・ Connect DSRVDD3 and SRVDD to DVDD3.
- Connect DSRVSS to DVSS.

14.2 Block Diagram

14.3 Registers

14.3.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

14.3.2 Details of Registers

Note 1: Change <ADCLK[2:0]> register under the condition where DSADCR1<BIASEN>=<MODEN>=0 and AD conversion stops.

Note 2: When synchronous start function is used, select same conversion clock in all units.

Note:Valid only when DSADCR1<BIASEN>="1".

14.3.2.3 DSADCR1 (Control register 1)

14.3.2.4 DSADCR2 (Control register 2)

14.3.2.5 DSADCR3 (Control register 3)

14.3 Registers

Note:Set "1" to units only used for slaves. A unit used for a master is set to "0". For the combination of master channels and slave channels of this product, refer to Chapter Product Information.

14.3.2.6 DSADCR4 (Control register 4)

14.3.2.7 DSADADJ (Correction register)

14.3.2.8 DSADST (Conversion Status Register)

Note:This bit is cleared by reading the DSADST register.

Note:This bit is cleared by reading the DSADRES register.

14.4 Operation Description

14.4.1 Boot-up and Stop Procedures

This section explains the procedure how to start DSADC and how to stop DSADC for transiting to the low power consumption mode. The table below shows registers required to set.

Note:A reference voltage circuit is shared with a temperature sensor.

14.4.1.1 Boot-up

Figure 14-2 Boot-up procedure

Perform following procedure while supply voltage is stable.

1. Boot-up the reference voltage circuit

Set "1" to TEMPEN<EN0><EN1> and wait 1ms or more to be stable.

TEMPEN<EN1> must be enabled when TEMPEN<EN0> is enabled. Setting both <EN0> and <EN1> at the same time is capable.

2. Boot-up the bias circuit

Set "1" to DSADCR1<BIASEN> and wait 1ms or more to be stable.

Feed a conversion clock before DSADCR1<BIASEN> is set. For details of the conversion clock, refer to "1.4.2 Conversion Clock (ADCLK)".

3. Boot-up the modulator circuit

Set "1" to DSADCR1<MODEN>.

After ADCLK will have elapsed for 5 clocks, a conversion will be enabled.

Set the conversion mode (DSADCR3<REPEAT>) and gain setting (DSADCR4<DSGAIN>) before starting conversion.

14.4.1.2 Stop

Transfer to the low power consumption mode

Figure 14-3 Stop procedure

Perform following procedure while the conversion is finished. (DSADST<ADBF>="0")

1. Stopping the modulator circuit

Set "0" to DSADCR1<MODEN>

2. Stopping the bias circuit

Set "0" to DSADCR1<BIASEN> after ADCLK will have elapsed one clock or more since the modulator circuit is stopped.

3. Stopping the reference voltage generation circuit Set "0" to TEMPEN<EN0><EN1>

Note:The reference voltage circuit is shared with the temperature sensor. While the temperature sensor is operating, do not stop TEMPEN<EN0> ("0" setting means stop).

14.4.2 Conversion Clock (ADCLK)

A conversion clock fed to DSADC is shared with a sequential comparison type AD converter and will operate after reset. The clock is stopped with CGSYSCR<FCSTOP> and also can be divided by DSADCLK.

Change or stop the conversion clock under DSADC stops. (Both DSADCR1<BIASEN> and <MODEN> are set to "0".)

When synchronous start function is used, select same conversion clock in all units.

14.4.2.1 Conversion Time

A conversion time can be calculated by the formula below where the frequency of ADCLK is $f_{A D C L K}$.

Conversion time = $1 / f_{ADCLK} \times 2640 + Fixed$ delay time [s]

A fixed delay time is 673 to 675 clocks at the first conversion in the repeat conversion and single conversion. In the repeat conversion, a fixed delay time is 0 clock at the 2nd conversion or later.

For example, where fc/1 is selected at fc=16MHz, a conversion time will be 165μs after the second conversion or later.

14.4.2.2 Transition to Low Power Consumption Mode

When a transition to the SLOW mode takes place, stop DSADC along with the procedure ["14.4.1 Boot](#page-282-0)[up and Stop Procedures"](#page-282-0) then stop the clock fed to DSADC with CGSYSCR<FCSTOP>.

When a transition to the STOP or SLEEP mode takes place, stop DSADC along with the stopping procedure. The clock fed to DSADC is automatically stopped.

14.4.3 Conversion Mode

The DSADC provides two types of conversion mode such as the single mode and repeat mode. In the single mode, a conversion is performed once and in the repeat mode a conversion is sequentially performed. The mode is set with DSADCR3<REPEAT>.

14.4.4 Conversion Start

Set "1" to DSADCR2<ADS> to start conversion.

14.4.5 Conversion Status

A conversion status can be checked with DSADST.

During the conversion, DSADST<ADBF> becomes "1". After conversion, DSADST<EOCF> becomes "1". To clear <EOCF>, read DSADST.

In the repeat conversion, DSADST<ADBF> holds "1" during operation. When the repeat conversion is complete, DSADST<ADBF> is cleared to "0".

14.4.6 Conversion Stop

In the single mode, when the conversion is complete, DSADC will automatically stop.

In the repeat mode, in order to stop a conversion, set "0" to DSADCR3<REPEAT>. DSADC will suspend the current conversion then stop. At this time a conversion end interrupt will not occur.

Note:If repeat conversion is complete by setting "0" to <REPEAT>, do not modify other bits of DSADCR3.

14.4.7 Conversion End

When the conversion ends, a conversion end interrupt will occur. This conversion result is stored in DSA-DRES<ADR>, and DSADRES<ADRF> is set to "1".

If the conversion end interrupt is not used, poll DSADST<EOCF>. If <EOCF> is "1", the conversion is complete.

If \langle ADR $>$ is written to next result before reading the current value, \langle ADOVR $>$ is set to "1". To clear <ADRF> and <ADOVR>, read DSADRES.

14.4.8 Conversion Result

Because an input range of AINxP to AINxN is -0.375V to 0.375V, the maximum input amplitude is $\pm 0.75V$. At this time, conversion results are as follows:

Note:VREFINx = 2.75V

14.5 Synchronous Start Function

A master unit and slave unit can start conversion simultaneously. For details, refer to "Product information" to confirm an assignment of a master and slave in this product.

14.5.1 Boot-up

The following flowchart shows the setting procedure when the synchronous start function is used.

Figure 14-4 Synchronous start function procedure

1. Boot-up the reference voltage circuit

Set "1" to TEMPEN<EN0><EN1> and wait 1ms or more to be stable.

TEMPEN<EN1> must be enabled when TEMPEN<EN0> is enabled. Setting both <EN0> and <EN1> at the same time is capable.

2. Set the synchronous operation

When the synchronous start function is used, set the slave unit DSADCR3<ADSYNC> to "1" and set the master unit <ADSYNC> to "0".

3. Boot-up the bias circuit

Set "1" to DSADCR1<BIASEN> and wait 1ms or more to be stable. In the case of $\langle BIASEN \rangle$, set slave side first, then set the master side.

Feed a conversion clock before DSADCR<BIASEN> is set. For details of the conversion clock, refer to "1.4.2 Conversion Clock (ADCLK)".

4. Boot-up the modulation circuit

Set "1" to DSADCR1<MODEN>.

After ADCLK will have elapsed for 5 clocks, a conversion will be enabled.

Conversion mode (DSADCR3<REPEAT> and gain setting (DSADCR4<DSGAIN>) can be set in each unit. Set them in each unit before conversion starts.

14.5.2 Stop

In the single conversion mode, DSADC stops when conversion is complete in each unit.

In the repeat conversion mode, stop DSADC as follows:

・ Stop only slave side

Change the conversion mode of slave (set DSADCR3<REPEAT> to "0") or perform a software reset.

At this time, master side continues conversion.

・ Stop only master side

Cancel the synchronous operation by setting DSADCR3<ADSYNC> of the slave to "0", then stop the master side by changing the conversion mode or by performing a software reset.

At this time, slave side continues conversion.

Stop both master side and slave side

Stop slave side then master side by changing the conversion mode or by performing a software reset.

14.6 Conversion Start Correction Function

By using a conversion start correction function, the conversion start time can be delayed from the time when DSADCR2<ADS> is set to "1".

This function is enabled by setting "1" to DSADADJ<ADJ>. The delay time is set with DSADADJ<OFF-SET>. After delay time of <OFFSET>×1/fsys has elapsed, conversion will start synchronously with ADCLK.

In the synchronous operation, a desired delay time can be set to <OFFSET> in each unit. Conversion will start after the time defined in <OFFSET> has elapsed from the time when <ADS> of master side is set to "1".

Duration between setting "1" to <ADS> and starting conversion, do not modify a value of <OFFSET>.
15. Temperature Sensor (TEMP)

15.1 Outline

The MCU measures a relative temperature using a temperature sensor.

A temperature sensor outputs a voltage based on the reference voltage circuit (BGR) according to temperatures. The output voltage is input to Channel 2 in the analog/digital converter (ADC)Unit D in the $\Delta\Sigma$ analog/digital converter (DSADC). With AD conversion, a corresponding digital value to temperatures is obtained.

Note:The reference voltage circuit (BGR) is shared with a ΔΣ type analog/digital converter (DSADC).

A difference among the temperature sensor output voltages is linearity related to temperature changes. To obtain a relative temperature, collect data under several conditions.

Channel 3 of ADC is input a 1V from BGR. In variable power voltage system, power voltage can be relatively identified by the result where BGR voltage was performed AD conversion.

If the temperature sensor or DSADC is not used, the reference voltage circuit requires below settings.

- ・ Connect DSRVDD3 and SRVDD to DVDD3
- Connect DSRVSS to DGND

15.2 Block diagram

15.2 Block diagram

AGNDREFx

Figure 15-1 Block diagram of Temperature sensor

TOSHIBA

15.3 Registers

15.3.1 Register List

Then table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

15.3.2 Details of Register

15.3.2.1 TEMPEN (Enable register)

Note:TEMPEN<EN1> must be enabled when TEMPEN<EN0> is enabled. Setting both <EN0> and <EN1> at the same time is possible.

- Note 1: TEMPEN<EN1> must be enabled when TEMPEN<EN0> is enabled. Setting both <EN0> and <EN1> at the same time is possible.
- Note 2: Do not enable AMP when the reference voltage is applied to VREFINx using the ΔΣADC.

15.3.2.2 TEMPCR (Control Register)

15.4 Operation Description

1. Boot-up

Perform the following procedure while power supply voltage is stable.

1. Boot-up the reference voltage circuit

Set "1" to TEMPEN<EN0> and wait 1 ms or more to be stable.

2. Boot-up the temperature sensor Set "1" to TEMPCR<CR0>

Approximately after 10μs, an output voltage is enabled.

2. Stop

Perform the following procedure.

1. Stop the temperature sensor

Set "0" to TEMPCR<CR0>

2. Stop the reference voltage circuit

Set "0" to TEMPEN<EN0>

Note: The reference voltage circuit is shared with ΔΣADC. Do not set TEMPEN<EN0> to "0" (stop) during the ΔΣADC operation.

15.4 Operation Description

TMPM061FWFG

TOSHIBA

16. Real Time Clock (RTC)

16.1 Function

- 1. Clock (hour, minute and second)
- 2. Calendar (month, week, date and leap year)
- 3. Selectable 12 (am/ pm) and 24 hour display
- 4. Time adjustment + or -30 seconds (by software)
- 5. Alarm function (available only in the products that have ALARM pin.)
- 6. Alarm interrupt
- 7. Clock correction function
- 8. 1 Hz clock output

16.2 Block Diagram

Figure 16-1 Block Diagram

- Note 1: Western calendar year column:This product uses only the final two digits of the year. The year following 99 is 00 years. Please take into account the first two digits when handling years in the western calendar.
- Note 2: Leap year:A leap year is divisible by 4 excluding a year divisible by 100; the year divisible by 100 is not considered to be a leap year. Any year divisible by 400 is a leap year. This product is considered the year divisible by 4 to be a leap year and does not take into account the above exceptions. It needs adjustments for the exceptions.

16.3 Detailed Description Register

16.3.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

RTC has two functions, PAGE0 (clock) and PAGE1 (alarm), which share some parts of registers.

The PAGE can be selected by setting RTCPAGER<PAGE >.

Note:"0" is read by reading the address. Writing is disregarded.

16.3.2 Control Register

Reset operation initializes the following registers.

- ・ RTCPAGER<PAGE>, <ADJUST>, <INTENA>
- **RTCRESTR**
- ・ RTCPROTECT
- ・ RTCADJCTL
- ・ RTCADJDAT

Other clock-related registers are not initialized by reset operation.

Before using the RTC, set the time, month, day, day of the week, year and leap year in the relevant registers.

Caution is required in setting clock data, adjusting seconds or resetting the clock.

Refer to ["16.4.3 Entering the Low Power Consumption Mode"](#page-308-0) for more information.

Table 16-1 PAGE0 (clock function) register

Note:Reading RTCSECR, RTCMINR, RTCHOURR, RTCDAYR, RTCMONTHR, RTCYEARR of PAGE0 captures the current state.

Table 16-2 PAGE1 (alarm function) registers

Note 1: Reading RTCMINR, RTCHOURR, RTCDAYR, RTCMONTHR, RTCYEARR of PAGE1 captures the current state. Note 2: RTCSECR, RTCMINR, RTCHOURR, RTCDAYR, RTCDATER, RTCMONTHR, RTCYEARR of PAGE0 and RTCYEARR of PAGE1 (for leap year) must be read twice and compare the data captured.

16.3.3 Detailed Description of Control Register

16.3.3.1 RTCSECR (Second column register (for PAGE0 only))

Note:The setting other than listed above is prohibited.

16.3.3.2 RTCMINR (Minute column register (PAGE0/1))

Note:The setting other than listed above is prohibited.

TOSHIBA

16.3.3.3 RTCHOURR (Hour column register(PAGE0/1))

(1) 24-hour clock mode (RTCMONTHR<MO0>= "1")

Note:The setting other than listed above is prohibited.

(2) 12-hour clock mode (RTCMONTHR<MO0> = "0")

Note:The setting other than listed above is prohibited.

Note:The setting other than listed above is prohibited.

16.3.3.5 RTCDATER (Day column register (for PAGE0/1 only))

Note 1: The setting other than listed above is prohibited.

Note 2: Do not set for non-existent days (e.g. 30th Feb.).

16.3.3.6 RTCMONTHR (Month column register (for PAGE0 only))

Note:The setting other than listed above is prohibited.

16.3.3.7 RTCMONTHR (Selection of 24-hour clock or 12-hour clock (for PAGE1 only))

Note:Do not change the RTCMONTHR<MO0> while the RTC is in operation.

16.3.3.8 RTCYEARR (Year column register (for PAGE0 only))

Note:The setting other than listed above is prohibited.

16.3.3.9 RTCYEARR (Leap year register (for PAGE1 only))

TOSHIBA

16.3.3.10 RTCPAGER(PAGE register(PAGE0/1))

Note 1: A read-modify-write operation cannot be performed.

Note 2: To set interrupt enable bits to <ENATMR>, <ENAALM> and <INTENA>, you must follow the order specified here. Make sure not to set them at the same time (make sure that there is time lag between interrupt enable and clock/ alarm enable).To change the setting of <ENATMR> and <ENAALM>, <INTENA> must be disabled first.

Example: Clock setting/Alarm setting

16.3 Detailed Description Register

16.3.3.11 RTCRESTR (Reset register (for PAGE0/1))

Note:A read-modify-write operation cannot be performed.

The setting of <DIS1HZ>, <DIW2HZ>, <DIS4HZ> and <DIS16MHZ>, RTCPAGER<ENAALM> used for alarm, 1Hz, 2Hz, 4Hz, 8Hz and 16Hz interrupt is shown as below.

16.3.3.12 RTCPROTECT(Protect register)

16.3.3.13 RTCADJCTL (Correction Function Control Register)

16.3.3.14 RTCADJDAT (Correction Value Register)

16.4 Operational Description

The RTC incorporates a second counter that generates a 1Hz signal from a 32.768 kHz signal.

The second counter operation must be taken into account when using the RTC.

16.4.1 Reading clock data

1. Using 1Hz interrupt

The 1Hz interrupt is generated being synchronized with counting up of the second counter. Data can be read correctly if reading data after 1Hz interrupt occurred.

2. Using pair reading

There is a possibility that the clock data may be read incorrectly if the internal counter operates carry during reading. To ensure correct data reading, read the clock data twice as shown below. A pair of data read successively needs to match.

Figure 16-2 Flowchart of the clock data reading

16.4.2 Writing clock data

A carry during writing ruins correct data writing. The following procedure ensures the correct data writing.

1. Using 1 Hz interrupt

The 1Hz interrupt is generated by being synchronized with counting up of the second counter. If data is written in the time between 1Hz interrupt and subsequent one second count, it completes correctly.

2. Resetting counter

Write data after resetting the second counter.

The 1Hz-interrupt is generated one second after enabling the interrupt subsequent to counter reset.

The time must be set within one second after the interrupt.

Figure 16-3 Flowchart of the clock data writing

3. Disabling the clock

Writing "0" to RTCPAGER<ENATMR> disables clock operation including a carry. Stop the clock after the 1Hz-interrupt. The second counter keeps counting. Set the clock again and enable the clock within one second before next 1Hz-interrupt

Figure 16-4 Flowchart of the disabling clock

16.4.3 Entering the Low Power Consumption Mode

To enter SLEEP mode, in which the system clock stops, after changing clock data, adjusting seconds or resetting the clock, be sure to observe one of the following procedures

- 1. After changing the clock setting registers, setting the RTCPAGER<ADJUST> bit or setting the RTCRESTR<RSTTMR> bit, wait for one second for an interrupt to be generated.
- 2. After changing the clock setting registers, setting the RTCPAGER<ADJUST> bit or setting the RTCRESTR<RSTTMR> bit, read the corresponding clock register values, <ADJUST> or <RSTTMR> to make sure that the setting you have made is reflected.

16.5 Alarm function

By writing "1" to RTCPAGER<PAGE>, the alarm function of the PAGE1 registers is enabled. One of the following signals is output to the ALARM pin if the product provide the ALARM pin.

1. "Low" pulse (when the alarm register corresponds with the clock)

2. 1, 2, 4, 8 or 16Hz cycle "Low" pulse

In any cases shown above, the RTC outputs one cycle pulse of low-speed clock. It outputs the INTRTC interrupt request simultaneously.

The INTRTC interrupt signal is falling edge triggered. Specify the falling edge as the active state in the CG Interrupt Mode Control Register

16.5.1 Usage of alarm function

"Low" pulse is output to the ALARM pin when the values of the PAGE0 clock register and the PAGE1 alarm register correspond. The INTRTC interrupt is generated and the alarm is triggered.

The alarm settings

Initialize the alarm with alarm prohibited. Write "1" to RTCRESTR<RSTALM>.

It makes the alarm setting to be 00 minute, 00 hour, 01 day and Sunday.

Setting alarm for min., hour, date and day is done by writing data to the relevant PAGE1 register.

Enable the alarm with the RTCPAGER <ENAALM> bit. Enable the interrupt with the RTCPAGER <INTE-NA> bit.

The following is an example program for outputting an alarm from the $\overline{\text{ALARM}}$ pin at noon (12:00) on Monday 5th.

If some alarm registers are set to "1", RTC doesn't compare the term. For example, if RTCDATER is set to "11 \pm 1111" and RTCDAYR is set to "111", the alarm will be output at noon (12:00) every day.

The above alarm works in synchronization with the low-speed clock. When the CPU is operating at high frequency oscillation, a maximum of one clock delay at fs (about 30μs) may occur for the time register setting to become valid.

16.5.2 1, 2, 4, 8 or 16 Hz cycle "Low" pulse

The RTC outputs a "Low" pulse cycle to the \overline{ALARM} pin by setting RTCPAGER<INTENA>="1" after setting RTCPAGER<ENAALM>= "0" and RTCRESTR. It is required that one of RTCRESTR<DIS1HZ>, \langle DIS2HZ>, \langle DIS4HZ>, \langle DIS8HZ> or \langle DIS16HZ> is set to "0".

The RTC outputs one cycle pulse of low-speed clock which correspond to RTCRESTR setting. It generates an INTRTC interrupt simultaneously.

16.6 Clock Correction Function

The clock correction function can precisely adjust the deviation of the clock.

In the Figure 16-5, T1 indicates one second. One second is generated by counting fs (32768Hz) 32768 times. The clock correction function adjusts the number of counts of T2 that is an one second of the correction reference time (Tall). The correction reference time is selected either among 1, 10, 20, 30 seconds or 1 minute with RTCADJCTL<AJSEL>. A count value of T2 can be adjustable from 32768-255 to 32768+256 with RTCADJ-DAT<ADJDAT>.

Figure 16-5 Clock Correction

The correction function related register, RTCADJCTL and RTCADJDAT, can be disabled with the RTCPRO-TECT register. In the initial state, RTCPROTECT is "0xC1" and write enable. If RTCPROTECT is set to a value other than "0xC1", RTCADJCTL and RTCADJDAT will be write disable.

16.7 1Hz Clock Output Function

RTCOUT pin outputs 1Hz clock. This clock is adjusted to operate on a 50% duty ratio. If the clock correction function is used, a duty ratio may be varied due to the error corrections.

TOSHIBA

17. LCD Driver

The TMPM061FWFG has a driver and control circuit to directly drive a liquid crystal display (LCD) device. The pins to be connected to the LCD are as follows:

In addition, the VLC pin is provided as a drive power supply pin, and the LV1 and LV2 pins are provided as external bleeder resistance connection pins.

Note:When the static, 1/3 or 1/2 duties are selected, unused common output pins should be opened. (It outputs bias voltage)

The LCD driver can directly drive the following five types of LCD:

17.1 Configuration

Figure 17-1 LCD Driver

17.2 Registers

17.2.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

17.2.2 Details of Registers

17.2.2.1 LCDEN (Enable register)

Note:In SLOW mode, do not set SLF to "0000" to "0100" (i.e. frequencies based on fsys). If SLF is set to one of these frequencies, pulses of an unexpected frame frequency will be output from the common and segment output pins.

17.2.2.3 LCDCR2 (Control register 2)

Note:The LRSE and BRH settings are effective only when BRSEL is set to select internal bleeder resistance.

17.2.2.4 LCDBUF00 (Buffer register 00)

17.2.2.5 LCDBUF01 (Buffer register 01)

17.2.2.7 LCDBUF03 (Buffer register 03)

17.2.2.8 LCDBUF04 (Buffer register 04)

17.2.2.9 LCDBUF05 (Buffer register 05)

17.2.2.10 LCDBUF06 (Buffer register 06)

17.2.2.11 LCDBUF07 (Buffer register 07)

17.2.2.12 LCDBUF08 (Buffer register 08)

17.2.2.13 LCDBUF09 (Buffer register 09)

17.2.2.14 LCDBUF10 (Buffer register 10)

17.2.2.15 LCDBUF11 (Buffer register 11)

17.2.2.16 LCDBUF12 (Buffer register 12)

17.2.2.17 LCDBUF13 (Buffer register 13)

17.2.2.18 LCDBUF14 (Buffer register 14)

17.2.2.19 LCDBUF15 (Buffer register 15)

17.2.2.20 LCDBUF16 (Buffer register 16)

17.2.2.21 LCDBUF17 (Buffer register 17)

17.2.2.22 LCDBUF18 (Buffer register 18)

17.2.2.23 LCDBUF19 (Buffer register 19)

17.3 Functions

17.3.1 LCD Display Control

If LCDCR1<EDSP> is set to "1", the power switch of LCD driver is on and a VLC voltage will be applied to a LCD driver. As a result, the LCD driver display will be enabled. If <EDSP> is set to "0", the power switch of LCD driver is off and a VLC voltage is shutdown. This causes the driver to display blank.

The following tables describes the condition of the LCD connection pins when <EDSP> is "1".

1. Common output pin

Common output pin is only controlled with LCDCR1<EDSP>.

2. External bleeder resistor pin

External bleeder resistor pin is a dual-purpose pin with a general-purpose input/output port. When this pin is used for a external bleeder resistor pin, set "0" to LCDCR2<BRSEL>. A setting of port control register is not required.

3. Segment output pin

A segment output pin is a dual-purpose pin with a general-purpose input/output port. To function as a segment output, set "1" to the corresponding function register PxFRn. Settings of other port control registers are not required except PxFRn.

Note) In register names, "x" indicates a port number and "n" indicates a function register number.

17.3.2 Operation at Reset

When a reset occurs, LCDCR1<EDSP> is initialized to "0" and the power switch of the LCD driver is automatically turned off, shutting off VLC voltage. At this time, the common output pins are fixed to the "Low" level. The multiplexed pins (input/output port or segment output) are configured as port input pins (high impedance). Therefore, if external reset operation takes time, the LCD display may become blurred.

17.3.3 Operation in SLEEP/STOP mode

If a transition to SLEEP/STOP mode occurs while LCDCR1<EDSP> is "1", the following conditions will occur.

Note:To use a LCD in the SLEEP mode, perform a transition to SLEEP mode in the SLOW mode. If a transition from the NORMAL to SLEEP mode occurs, a display will be blank.

SLEEP mode

A common pin and segment pin remain the same state as before a transition occurs and a display is being used.

STOP mode

If a transition to the STOP mode occurs, LCDCR1 \leq EDSP> will automatically initialized to "0", a display will be blank. To redisplay the LCD after returning from the STOP mode, set "1" to LCDCR1<EDSP>.

17.3.4 Operation in SLOW mode

When the LCD is used in both NORMAL mode and SLOW mode, it is recommended that LCDCR1<SLF> be set to a frequency based on fs ("1000" or "1001"). (This will eliminate the need for changing the LCDCR1<SLF> setting each time the operating mode is switched between NORMAL mode and SLOW mode.)

If a frequency based on fsys is used in NORMAL mode, it is necessary to clear LCDCR1<EDSP> to "0" before switching to SLOW mode. Then, after entering SLOW mode, it is necessary to change LCDCR1<SLF> to a frequency based on fs and to set LCDCR1<EDSP> to "1". Likewise, in switching from SLOW mode to NORMAL mode, it is necessary to clear LCDCR1<EDSP> to "0" before switching to NORMAL mode. Then, after entering NORMAL mode, it is necessary to change LCDCR1<SLF> to a frequency based on fsys and to set LCDCR1<EDSP> to "1".

17.3.5 Fail-safe

Be careful about the following because LCD cannot display.

- 1. When LCDCR1<EDSP>="1", setting LCDEN<LCDE> to "0" blanks the LCD display. The display can be re-enabled by setting LCDEN<LCDE> to "1".
- 2. When LCDCR1<SLF> is set to "0000" to "0110", the high-speed clock must be activated (OSCCR $\langle XEN1 \rangle = "1"$ or OSCCR $\langle XEN \rangle = "1"$) and allowed to achieve stable oscillation before LCDCR1 <EDSP> can be set to "1". If LCDCR1<EDSP> is set to "1" while the high-speed clock is stopped, the LCD display cannot be enabled. (Although LCDCR1<EDSP> changes to "1", the LCD display remains blank.)
- 3. When LCDCR1<SLF> is set to "1000" or "1001", the low-speed clock must be activated (OSCCR $\langle XTEN \rangle = "1"$) and allowed to achieve stable oscillation before LCDCR1 $\langle EDSP \rangle$ can be set to "1". If LCDCR1<EDSP> is set to "1" while the low-speed clock is stopped, the LCD display cannot be enabled. (Although LCDCR1<EDSP> changes to "1", the LCD display remains blank.)

17.3.6 LCD Drive Methods (LCDCR1<DUTY>)

The LCD drive method can be selected from the following five types by the setting of LCDCR1<DUTY>.

Note 2: V_{LCD3} = LCD drive voltage (= V_{LC} – V_{SS})

Figure 17-2 LCD Drive Waveforms (Potential Difference between COM and SEG Pins)

17. LCD Driver 17.3 Functions

17.3.7 Frame Frequency (LCDCR1<SLF>)

The frame frequency (f_F) is determined based on the drive method and base frequency, as shown in Table 17-1. The base frequency is selected by LCDCR1<SLF>.

		Frame frequency [Hz]					
SLF	Base frequency [Hz]	1/4 Duty	1/3 Duty $(4/3)$ x fsys $/2^{18}$ 81 $(4/3)$ x fsys $/2^{17}$ 163 81 $(4/3)$ x fsys $/2^{16}$ 163 81 $(4/3)$ x fsys $/2^{15}$ 163 81 $(4/3)$ x fsys $/2^{14}$ 163	1/2 Duty	Static		
	fsys $/2^{18}$	fsys / 2 ¹⁸		$(4/2)$ x fsys $/2^{18}$	fsys $/2^{18}$		
0000	$(fsys = 16 MHz)$	61		122	61		
	fsys $/ 2^{17}$	fsys $/ 2^{17}$		$(4/2)$ x fsys $/2^{17}$	fsys $/ 2^{17}$		
0001	$(fsvs = 16 MHz)$	122		244	122		
	$(fsys = 8 MHz)$	61		122	61		
0010	fsys $/2^{16}$	fsys $/2^{16}$		$(4/2)$ x fsys $/2^{16}$	fsys / 2^{16}		
	$(fsys = 8 MHz)$	122		244	122		
	$(fsys = 4 MHz)$	61		122	61		
0011	fsys $/2^{15}$	fsys $/2^{15}$		$(4/2)$ x fsys $/2^{15}$	fsys $/2^{15}$		
	$(fsys = 4 MHz)$	122		244	122		
	$(fsys = 2 MHz)$	61		122	61		
	fsvs $/ 2^{14}$	fsys / 2 ¹⁴		$(4/2)$ x fsys $/2^{14}$	fsys $/ 2^{14}$		
0100	$(fsys = 2 MHz)$	122		244	122		
	$(fsys = 1 MHz)$	61	81	122	61		
1000	fs $/2^9$	fs $/2^9$	$(4/3)$ x fs $/2^9$	$(4/2)$ x fs $/2^9$	fs $/2^9$		
	$(fs = 32.768 kHz)$	64	85	128	64		
	fs $/2^8$	fs $/2^8$	$(4/3)$ x fs $/2^8$	$(4/2)$ x fs $/2^8$	fs $/2^8$		
1001	$(fs = 32.768 kHz)$	128	171	256	128		

Table 17-1 Frame Frequency Settings

Note:fsys = Gear clock frequency [Hz], fs = Low-frequency clock frequency [Hz]

17.3.8 Internal/External Bleeder Resistance Switching Control

The LCD bias voltage is generated by bleeder resistance. Either external or internal bleeder resistance can be used.

To use internal bleeder resistance, set LCDCR2<BRSEL> to "1". In this case, the multiplexed pins (input/output port or external bleeder resistance connection) can be used as input/output ports.

To use external bleeder resistance, set LCDCR2<BRSEL> to "0" and connect external resistance to the external bleeder resistance connection pins (LV1, LV2). In this case, the multiplexed pins (input/output port, external bleeder resistance connection, or segment output) can only be used as external bleeder resistance connection pins.

See [Figure 17-4](#page-335-0) for how to connect bleeder resistance.

17.3.9 Low Internal Bleeder Resistance Connection Time Selection (LCDCR2<LRSE>)

Internal bleeder resistance is comprised of two parts: high resistance and low resistance. The high and low resistance parts are connected in parallel for each bias voltage. The low bleeder resistance is provided with an analog switch, and the time to turn on the low bleeder resistance can be adjusted by LCDCR2<LRSE>. While the analog switch is turned on, the low resistance is connected in parallel to the high resistance. This reduces the total amount of resistance, allowing the drive capability of the LCD driver to be increased.

Typically, the longer the period of connecting the low resistance, the higher the drive capability of the LCD panel, but the higher the power consumption. Conversely, the shorter the period of connecting the low resistance, the lower the drive capability, but the lower the power consumption. Insufficient drive capability will cause adverse effects on the LCD display, such as blurring. Choose the optimum drive capability for the LCD panel to be used.

[Table 17-2](#page-333-0) shows the connection time (percentage) of the low bleeder resistance per frame and the estimated amount of current that flows through the entire bleeder resistance in each case.

[Figure 17-3](#page-333-0) shows the bleeder resistance control timing for the 1/4 duty and 1/3 bias LCD.

See ["17.3.10 High Internal Bleeder Resistance Selection \(LCDCR2<BRH>\)"f](#page-334-0)or the setting of LCDCR2<BRH>.

Note:The bleeder resistance current values shown above are estimated values. The actual current values may vary depending on the amount of LCD load and manufacturing variations in resistance values.

Low Bleeder Resistance Connection Time

- RLt : Period during which low resistance is connected (high resistance and low resistance are connected in parallel)
- RHt : Period during which low resistance is not connected (only high resistance is connected)

Figure 17-3 Bleeder Resistance Selection by LCDCR2<LRSE> (1/4 Duty, 1/3 Bias)

17.3.10 High Internal Bleeder Resistance Selection (LCDCR2<BRH>)

The value of high internal bleeder resistance can be selected from two levels (500 kΩ (Typ.) or 200 kΩ (Typ.)) by the setting of LCDCR2<BRH>. Typically, the lower the resistance value, the higher the drive capability of the LCD panel, but the higher the power consumption. Conversely, the higher the resistance value, the lower the drive capability, but the lower the pz ower consumption.

The resistance value of the low resistance is fixed to 20 k Ω (Typ.). Since the low resistance is connected in parallel to the high resistance via an analog switch, the total amount of resistance can be adjusted by the setting of LCDCR2<LRSE> as shown in Table 17-3.

For example, setting LCDCR2<BRH> to "1" selects a synthesized resistance of 19.23 kΩ (Typ.) when the low resistance is connected, and a high resistance of 500 kΩ (Typ.) when the low resistance is not connected.

LCDCR2 <brh></brh>	When low resistance is not connected	When low resistance is connected		
	500 k Ω (Typ.)	19.23 k Ω (Typ.)		
	200 k Ω (Typ.)	18.18 kΩ (Typ.)		

Table 17-3 Bleeder Resistance Values

17.3.11 LCD Display Operation

The LCD drive voltage V_{LCD} is given by the potential difference between the VLC and VSS pins (V_{LC} − V_{SS}). The LCD lights up when the potential difference between segment and common outputs is \pm VL_{CD}. At other times, the LCD is turned off.

Power supply connections should be made to satisfy the condition $V_{CC} \leq V_{DD}$. Connection examples are shown in [Figure 17-4.](#page-335-0)

When internal bleeder resistance is used (LCDCR2<BRSEL>="1")

- Note 1: When the CPU operating voltage is the same as the LCD drive voltage, the VLC pin should be connected to the VDD pin.
- Note 2: At reset, the common output pins become low. However, the multiplexed pins (input/output port or segment output) are configured as port input pins (high impedance). Therefore, if the multiplexed pins (input/output port or segment output) are used as segment output pins and an external reset signal is input for a prolonged period of time, the LCD display may be adversely affected, such as blurring. The multiplexed pins (input/output port or external bleeder resistance connection) are configured as external bleeder resistance connection pins.

17.3.12 Display Data Setting

Display data is stored in the Buffer register 00 to 19.

The display data stored in the display data area is automatically read out and sent to the LCD driver by hardware. The LCD driver generates the segment and common signals according to the display data and drive method. Therefore, display patterns can be changed by simply overwriting the contents of the display data area[.Ta](#page-336-0)[ble 17-5](#page-336-0) shows the correspondence between the display data area and the SEG and COM pins.

The LCD lights up when display data is "1" and is turned off when display data is "0".

At reset, the data in the Buffer registers are initialized to "0".

Since the number of pixels that can be driven varies with the LCD drive method, the number of bits used for storing display data also varies accordingly. Therefore, the bits not used for storing display data and the data memory locations corresponding to addresses not connected to the LCD can be used for storing general user data (see Table 17-4).

Drive method	Bits 7/3	Bits 6/2	Bits 5/1	Bits 4/0
$1/4$ duty	COM ₃	COM ₂	COM ₁	COM ₀
$1/3$ duty ٠		COM ₂	COM ₁	COM ₀
$1/2$ duty ۰		۰	COM ₁	COM ₀
Static	۰	۰	-	COM ₀

Table 17-4 Bits To Be Used for Storing Display Data

Note:" - " denotes bits not used for storing display data.

Table 17-5 LCD Display Data Area

17.4 Examples of How To Control the LCD Driver

17.4.1 Initialization

Figure 17-5 is a flowchart showing the initialization process of the LCD driver.

Example: When the LCD driver is to be operated with the following conditions:

- Drive method: 1/4 duty, 1/3 bias
- LCD frame frequency: fsys/2¹⁸ [Hz]
- Connection time of low bleeder resistance (internal): 2¹⁵/fsys
- High bleeder resistance (internal): 200 kΩ

Figure 17-5 Flowchart for the LCD Driver Initialization

17.4.2 Display Data Setting

Display data is normally prepared as fixed data in the program memory (ROM) and transferred to the display data area by instructions.

Example1: The following shows an example of how to set display data for displaying a number corresponding to BCD data stored at address 0x90 in the data memory by using the 1/4 duty and 1/3 bias LCD. [Figure 17-6](#page-338-0) shows an example of how the COM and SEG pins are connected to the LCD, and [Table 17-6](#page-338-0) shows how display data is set for this example.

Figure 17-6 Example of COM and SEG Pin Connections (1/4 Duty)

Table 17-6 Example of Display Data (1/4 Duty)

Example2: The following shows an example of how to set display data for displaying a number as explained in example 1 by using the 1/2 duty LCD. [Figure 17-7](#page-339-0) shows an example of how the SEG and COM pins are connected to the LCD, and [Table 17-7](#page-339-0) shows how display data is set for this example.

Figure 17-7 Example of COM and SEG Pin Connections

Note:An asterisk (*) denotes "don't care".

TOSHIBA

17.4.3 Drive Output Examples

Figure 17-8 1/4 Duty (1/3 Bias) Drive

Figure 17-9 1/3 Duty (1/3 Bias) Drive

TOSHIBA

Figure 17-10 1/3 Duty (1/2 Bias) Drive

17. LCD Driver 17.4 Examples of How To Control the LCD Driver

Figure 17-11 1/2 Duty (1/2 Bias) Drive

TOSHIBA

Figure 17-12 Static Drive

- 17. LCD Driver
- 17.4 Examples of How To Control the LCD Driver

TOSHIBA

18. Low Voltage Detection Circuit (LVD)

Low voltage detection circuit generates an interrupt signal by detecting a decreasing/increasing voltage.

Supply voltage is indicated as RVDD3.

Note:Due to the fluctuation of supply voltage, the power-on reset circuit may not operate properly. Users should give due consideration based on the electrical characteristic in the device designing.

18.1 Structure

The low voltage detection circuit consists of a reference voltage generation circuit, comparators and control registers.

Supply voltage is divided by a ladder resistor and input to the voltage selection circuit. In the voltage selection circuit, a voltage is chosen according to the detected voltage then compared with the reference voltage in the comparator. If the supply voltage is upper/lower than the detected voltage, an interrupt signal occurs.

Figure 18-1 Block diagram of LVD (described only LVD interrupt circuit)

18.2 Registers

18.2.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

18.2.2 LVDICR (LVD-INTcontrol register)

Note:LVDICR is initialized by reset with RESET pin.

Note:LVDSR is initialized by reset with RESET pin.

18.3 Operation Description

18.3.1 Detection Voltage Selection and Enabling/Disabling the Operation

The LVDICR register sets the following; choosing the detection voltage, setting the operation to enable/disable, choosing output conditions and setting the output to enable/disable. The LVDICR register is initialized with the reset by RESET pin.

The LVDICR<LVDLVL2[2:0]> bit chooses the detection voltage. If LVDICR<LVDEN2> is set to "1", low voltage detection operation is enabled.

Note:While supply voltage is lower than the detection voltage, if low voltage detection operation is enabled, INTLVD will generate at this timing.

18.3.2 Lower Voltage Detection

If supply voltage is lower than the detection voltage level, INTLVD generates. When the LVDICR<IN-TSEL> is set to "1", if supply voltage is upper than the detection voltage, INTLVD generates.

After lower voltage detection, to detect INTLVD is required a certain time. If this period is shorter than expected, INTLVD may not generate.

If supply voltage is lower than 1.8V, MCU operation is not guaranteed. In this case, supply voltage must be decreased to 0V and then power-on.

Figure 18-2 low voltage detection Timing

TOSHIBA

19. Watchdog Timer (WDT)

The watchdog timer (WDT) is for detecting malfunctions (runaway) of the CPU caused by noises or other disturbances and remedying them to return the CPU to normal operation.

If the watchdog timer detects a runaway, it generates a INTWDT interrupt or reset.

Note:INTWDT interrupt is a factor of the non-maskable interrupts (NMI).

Also, the watchdog timer notifies of the detecting malfunction to the external peripheral devices from the watchdog timer pin (WDTOUT) by outputting "Low".

Note:TMPM061FWFG does not have the watchdog timer out pin (WDTOUT).

19.1 Configuration

Figure 19-1 shows the block diagram of the watchdog timer.

Figure 19-1 Block Diagram of the watchdog Timer

19.2 Register

19.2.1 Register List

The table below shows control registers and their addresses.

For detail of the base address, refer to "Address lists of peripheral functions" of "Memory Map" chapter.

19.2.2 WDMOD (Watchdog Timer Mode Register)

Note:INTWDT interrupt is a factor of the non-mask interrupt.

19.2.3 WDCR (Watchdog Timer Control Register)

19.3 Description of Operation

19.3.1 Basic Operation

The watchdog timer is consists of the binary counter that works using the system clock (fsys) as an input.

Detecting time can be selected between 2^{15} , 2^{17} , 2^{19} , 2^{21} , 2^{23} and 2^{25} by the WDMOD<WDTP[2:0]>.

The detecting time as specified is elapsed, the watchdog timer interrupt (INTWDT) is generated, and the watchdog timer out pin (WDTOUT) outputs "Low".

To detect malfunctions (runaways) of the CPU caused by noise or other disturbances, the binary counter of the watchdog timer should be cleared by software instruction before INTWDT interrupt is generated. If the binary counter is not cleared, the non-maskable interrupt is generated by INTWDT. Thus CPU detect malfunction (runaway), malfunction countermeasure program is performed to return to the normal operation.

Additionally, it is possible to resolve the problem of a malfunction (runaway) of the CPU by connecting the watchdog timer out pin to reset pins of peripheral devices.

Note:TMPM061FWFG does not have a watchdog timer out pin (WDTOUT).

19.3.2 Operation Mode and Status

The watchdog timer begins operation immediately after a reset is released. If not using the watchdog timer, it should be disabled.

The watchdog timer can not be used at the high-speed frequency clock is stopped. Before transition to below operation modes, the watchdog timer should be disabled.

In IDLE mode, its operation depend on WDMOD<I2WDT> setting.

- STOP mode
- SLEEP mode
- SLOW mode

Also, the binary counter is automatically stopped during debug mode.

19.3.3 Operation when malfunction (runaway) is detected.

19.3.3.1 INTWDT interrupt generation

[Figure 19-2](#page-354-0) shows the case that INTWDT interrupt is generated (WDMOD<RESCR>="0").

When an overflow of the binary counter occurs, INTWDT interrupt is generated. It is a factor of nonmaskable interrupt (NMI). Thus CPU detects non-maskable interrupt and perform the countermeasure program.

When INTWDT interrupt is generated, simultaneously the watchdog timer out ($\overline{\text{WDTOUT}}$) output "Low".

WDTOUT becomes "High" by the watchdog timer clearing that is writing clear code 0x4E to the WDCR.

Note:TMPM061FWFG does not have a watchdog timer out pin (WDTOUT).

19.3.3.2 Internal Reset Generation

Figure 19-3 shows the internal reset generation (WDMOD<RESCR>="1").

MCU is reset by the overflow of the binary counter. In this case, reset status continues for 32 states.

Figure 19-3 Internal reset generation

19.4 Control of the watchdog timer

19.4.1 Disable control

By writing the disable code (0xB1) to WDCR after setting WDMOD<WDTE> to "0", the watchdog timer can be disabled and the binary counter can be cleared.

19.4.2 Enable control

Set WDMOD<WDTE> to "1".

19.4.3 Watchdog timer clearing control

Writing the clear code (0x4E) to WDCR clears the binary counter and it restarts counting.

19.4.4 Detection time of watchdog timer

Set WDMOD<WDTP[2:0]> depend on the detection time.

For example, in the case that $2^{21}/f_{SVS}$ is used, set "011" to WDMOD<WDTP[2:0]>.

- 19. Watchdog Timer (WDT)
- 19.4 Control of the watchdog timer

20. Flash Memory Operation

This section describes the hardware configuration and operation of Flash memory. In this section, "1-word" means 32 bits.

20.1 Features

20.1.1 Memory Size and Configuration

Table 20-1 and Figure 20-1 show a built-in memory size and configuration of TMPM061FWFG.

Table 20-1 Memory size and configuration

	Block configuration					Write time		Erase time		
Memory size	128 KB	64 KB	32 KB	16 KB	# of words # of pages per page	page	Total area	Block erase	Chip erase	
128 KB	$\overline{}$	-	4	-	32	1024	.25ms	1.28 sec	0.1 sec	$0.2~\rm{sec}$

Note:The above values are theoretical values not including data transfer time. The write time per chip depends on the write method used by a user.

Figure 20-1 Block configuration

Flash memory configuration units ares described as "block" and "page".

・ Page

One page is 32 words. Same address [31:7] is used in a page. First address of the group is [6:0] $= 0$ and the last address of the group is $[6:0] = 0x7F$.

Block

One block is 32KB and flash memory is consists of four blocks.

Write operation is performed per page. The write time per page is 1.25ms. (Typ.)

Erase is performed per block (auto block erase command use) or performed on entire flash memory (use of auto chip erase command). Erase time varies on commands. If auto block command is used, the erase time will be 0.1 sec per block (Typ.). If the auto chip erase command is used to erase entire area, the time will be 0.2 sec (Typ.).

In addition, the protect function can be used per block. For detail of the protect function, refer to ["20.1.5 Pro](#page-359-0)[tect/Security Function"](#page-359-0).

20.1.2 Function

Flash memory built-in this device is generally compliant with the JEDEC standards except for some specific functions. Therefore, if a user is currently using a flash memory as an external memory, it is easy to implement the functions into this device. Furthermore, to provide easy write or erase operation, this product contains a dedicated circuit to perform write or chip erase automatically.

20.1.3 Operation Mode

20.1.3.1 Mode Description

This device provides the single chip mode and single boot mode. The single chip mode contains the normal mode and user boot mode. Figure 20-2 shows the mode transition.

Figure 20-2 Mode transition

(1) Single chip mode

The single chip mode is a mode where the device can boot-up from Flash memory after reset. The mode contains two sub-modes in below.

・ Normal mode

The mode where user application program is executed.

・ User boot mode

The mode where flash memory is re-programmed on the user's set.

Users can switch the normal mode to user boot mode freely. For example, a user can set if PA0 of port A is "1", the mode is the normal mode. If PA0 of port A is "0", the mode is the user boot mode. The user must prepare a routine program in the application program to determine the switching.

(2) Single boot mode

The mode where flash memory can boot-up from the built-in BOOT ROM (Mask ROM) after reset.

The BOOT ROM contains the algorithm that can rewrite Flash memory via serial port of this device on the user's set. With connecting the serial port to external host, data transfer is performed in above-mentioned protocol and re-programed Flash memory.

(3) On-board programming mode

The user boot mode and single boot mode are the modes where flash memory can be re-programmable on the user's set. These two modes are called "on-board programming mode".

20.1.3.2 Mode Determination

Either the single chip or single boot operation mode can be selected by the level of the \overline{BOOT} pin when reset is released.

	Pin				
Operation mode	RESET	BOOT			
Single chip mode	$0 \rightarrow 1$				
Single boot mode					

Table 20-2 Operation mode setting

20.1.4 Memory Map

Figure 20-3 shows a comparison of the memory map in the single chip mode and single boot mode. In the single boot mode, built-in Flash memory is mapped to 0x3F80 0000 and subsequent addresses, and the builtin BOOT ROM is mapped to $0x0000$ 0000 through $0x0000$ OFFF.

Flash memory and RAM addresses are shown below.

Figure 20-3 Comparison of memory map

20.1.5 Protect/Security Function

This device has the protect and security functions for Flash memory.

1. Protect function

The write/erase operation can be inhibited per block.

2. Security function

The read operation from a flash writer can be inhibited. Usage restrictions on debug functions
20.1.5.1 Protect Function

This function inhibits the write/erase operation per block.

To enable the protect function, a protect bit corresponding to a block is set to "1" using the protect bit program command. If a protect bit is set to "0" using the protect bit erase command, a block protect can be canceled. The protect bit can be monitored with FCFLCS<BLPRO[3:0]>.

A program of protect bit can be programmed by 1-bit unit and can be erased by 4-bit unit. For detail of programming/erasing of protect bits, refer to ["20.2.4 Command Description".](#page-364-0)

20.1.5.2 Security Function

Table 20-3 shows operations when the security function is enabled.

Table 20-3 Operations when the security function is enabled.

The security function is enabled under the following conditions;

- 1. FCSECBIT<SECBIT> is set to "1".
- 2. All protect bits (FCFLCS<BLPRO>) are set to "1".

FCSECBIT<SECBIT> is set to "1" by the pin reset. Rewriting of FCSECBIT <SECBIT> is described in below.

Note:Use a 32-bit transfer instruction when the following writing operations, item1 and 2.

- 1. Write the specified code (0xa74a9d23) to FCSECBIT
- 2. Write data within 16 clocks after the operation of item 1.

20.1.6 Register

20.1.6.1 Register List

Base Address = 0x41FF_F000

20.1.6.2 FCFLCS (Flash control register)

Note 1: Make sure that flash memory is ready before commands are issued.

Note 2: A value will correspond to the protection status.

Note 3: Access the Flash memory after changing <RDY/BSY> "0" into "1" and passing for 200μs or more.

20.1.6.3 FCSECBIT (Security bit register)

Note:This register is initialized by pin reset.

20.2 Detail of Flash Memory

In on-board programming, the CPU executes commands for reprogramming or erasing Flash memory. This reprogramming/erase control program should be prepared by the user beforehand. Since Flash memory content cannot be read while Flash memory is being written or erased, it is necessary to run the reprogram/erase control program on the built-in RAM. Do not generate interrupt/fault to avoid abnormal program termination.

20.2.1 Function

Flash memory is generally compliant with the JEDEC standards except for some specific functions. However; a method of address designation of operation command is different from standard commands.

If write/erase operation is executed, commands are input to flash memory using 32-bit (1-word) store instruction command. After command input, write or erase operation is automatically executed in inside.

Table 20-4 Flash memory function

Main function	Description
Automatic page program	Writes data automatically.
Automatic chip erase	Erases the entire area of Flash memory automatically.
Automatic block erase	Erases a selected block automatically.
Write/erase protect	The write or erase operation can be individually inhibited for each block.

20.2.2 Operation Mode of Flash Memory

Flash memory provides mainly two types of operation modes;

- The mode to read memory data (Read mode)
- The mode to erase or rewrite memory data automatically (Automatic operation mode)

After power-on, after rest or after automatic operation mode is finished normally, Flash memory becomes read mode. Instruction stored in Flash memory or data read is executed in the read mode.

If commands is input during the read mode, the operation mode becomes the automatic operation. If the command process is normally finished, the operation mode returns to the read mode except the ID-Read command. During the automatic operation, data read and instruction execution stored in Flash memory cannot be performed.

20.2.3 How to Execute Command

The command execution is performed by writing command sequences to Flash memory with a store instruction. Flash memory executes each automatic operation command according to the combination of input addresses and data. For detail of the command execution, refer to ["20.2.4 Command Description".](#page-364-0)

An execution of store instruction to the Flash memory is called "bus write cycle". Each command consists of some bus write cycles. In Flash memory, when address and data of bus write cycle are performed in the specified order, the automatic command operation is performed. When the cycle is performed in non-specified order, Flash memory stops command execution and returns to the read mode.

If you cancel the command during the command sequence or input a different command sequence, execute the read command or read/reset command. Then Flash memory stops command execution and returns to the read mode. The read command and read/reset command are called "software reset".

When write command sequence ends, the automatic operation starts and FCFLCS<RDY/BSY> is set to "0". When the automatic operation normally ends, FCFLCS<RDY/BSY> = "1" is set and Flash memory returns to the read mode. Access the Flash memory after changing <RDY/BSY> "0" into "1" and passing for 200μs or more.

New command sequences are not accepted during the automatic operation.

Notes on the command execution;

- 1. To recognize command, command sequencer need to be in the read mode before command starting. Confirm FCFLCS<RDY/BSY> = 1 is set prior to the first bus write cycle of each command. Consecutively, it is recommended that the read command is executed.
- 2. Execute each command sequence from outside of Flash memory.
- 3. Execute sequentially each bus write cycle by data transfer instruction in one-word (32-bit).
- 4. Do not access Flash memory during the each command sequence. Do not generate any interrupt or fault.
- 5. Upon issuing a command, if any address or data is incorrectly written, make sure to return to the read mode by using software reset.

20.2.4 Command Description

This section explains each command content. For detail of specific command sequences, refer to ["20.2.5](#page-368-0) [Command Sequence"](#page-368-0).

20.2.4.1 Automatic Page Program

(1) Operation Description

The automatic page program writes data per page. When the program writes data to multiple pages, a page command need to be executed in page by page. Writing across pages is not possible.

Writing to Flash memory means that data cell of "1" becomes data of "0". It is not possible to become data cell of "1" from data of "0". To become data cell of "1" from "0", the erase operation is required.

The automatic page program is allowed only once to each page already erased. Either data cell of "1" or "0" cannot be written data twice or more. If rewriting to a page that has already been written once, the automatic page program is needed to be set again after the automatic block erase or automatic chip erase command is executed.

Note 1: Page program execution to the same page twice or more without erasing operation may damage the device.

Note 2: Writing to the protected block is not possible.

(2) How to Set

The 1st to 3rd bus write cycles indicate the automatic page program command.

In the 4th bus write cycle, the first address and data of the page are written. On and after 5th bus cycle, one page data will be written sequentially. Data is written in one-word unit (32-bit).

If a part of the page is written, set "0xFFFFFFFF" as data, which means not required to write, for entire one page.

No automatic verify operation is performed internally in the device. So, be sure to read the data programmed to confirm that it has been correctly written.

If the automatic page program is abnormally terminated, that page has been failed to write. It is recommended not to use the device or not to use the block including the failed address.

20.2.4.2 Automatic Chip Erase

(1) Operation Description

The automatic chip erase is executed to the memory cell of all addresses. If protected blocks are contained, these blocks will not be erased. If all blocks are protected, the automatic chip erase operation will not performed and will return to the read mode after a command sequence is input.

(2) How to Set

The 1st to 6th bus write cycles indicate the automatic chip erase command. After the command sequence is input, the automatic chip erase operation starts.

No automatic verify operation is performed internally in the device. So, be sure to read the data to confirm that it has been correctly erased.

20.2.4.3 Automatic Block Erase

(1) Operation Description

The automatic erase command performs erase operation to the specified block. If the specified block is protected, erase operation is not executed.

(2) How to Set

The 1st to 5th bus write cycles indicate the automatic block erase command. In the 6th bus write cycle, the block to be erased is specified. After the command sequence is input, the automatic block erase operation starts.

No automatic verify operation is performed internally in the device. So, be sure to read the data to confirm that it has been correctly erased.

20.2.4.4 Automatic Protect Bit Program

(1) Operation Description

The automatic protect bit program writes "1" to a protect bit at a time. To set "0" to a protect bit, use the automatic protect bit erase command.

For detail of the protect function, refer to ["20.1.5 Protect/Security Function"](#page-359-0).

(2) How to Set

The 1st to 6th bus write cycles indicate the automatic protect bit program command. In the 7th bus write cycle, the protect bit to be written is specified. After the command sequence is input, the automatic protect bit program starts. Check whether write operation is normally terminated with FCFLCS<BLPRO>.

OSHIBA

20.2.4.5 Auto Protect Bit Erase

(1) Operation Description

The automatic protect bit erase command operation depends on the security status. For detail of security status, refer to ["20.1.5 Protect/Security Function"](#page-359-0).

・ Non-security status

Clear the specified protect bit to "0". Protect bit erase is performed in 4-bit unit.

Security status

Erase all protect bits after all addresses of Flash memory are erased.

(2) How to Set

The 1st to 6th bus write cycles indicate the automatic protect bit erase command. In the 7th bus write cycle, the protect bit to be erased is specified. After the command sequence is input, the automatic protect bit erase operation starts.

In the non-security status, specified protect bit is erased. Check whether erase operation is normally terminated with FCFLCS<BLPRO>.

In the security status, all addresses and all protect of Flash memory bits are erased. Confirm if data and protect bits are erased normally. If necessary, execute the automatic protect bit erase, automatic chip erase or automatic block erase.

All cases are the same as other commands, FCFLCS<RDY/BSY> becomes "0" during the automatic protect bit erase command operation. After the operation is complete, FCFLCS<RDY/BSY> becomes "1" and Flash memory will return to the read mode.

20.2.4.6 ID-Read

(1) Operation Description

The ID-Read command can read information including Flash memory type and three types of codes such as a maker code, device code and macro code.

(2) How to Set

The 1st to 3rd bus write cycles indicate the ID-Read command. In the 4th bus write cycle, the code to be read is specified. After the 4th bus write cycle, read operation in the arbitrary flash area acquires codes.

The ID-Read can be executed successively. The 4th bus write cycle and reading ID value can be executed repeatedly.

The ID-Read command does not automatically return to the read mode. To return to the read mode, execute the read command or read/reset command.

20.2.4.7 Read Command and Read/reset Command (Software Reset)

(1) Operation Description

A command to return Flash memory to the read mode.

When the ID-Read command is executed, macro stops at the current status without automatically return to the read mode. To return to the read mode from this situation, use the read command or read/ reset command. It is also used to cancel the command when commands are input to the middle.

(2) How to Set

The 1st bus cycle indicates the read command. The 1st to 3rd bus write cycles indicate the read/reset command. After either command sequence is executed, Flash memory returns to the read mode.

20.2.5 Command Sequence

20.2.5.1 Command Sequence List

Table 20-5 shows addresses and data of bus write cycle in each command.

All command cycles except the 5th bus cycle of ID-Read command are bus write cycles. A bus write cycle is performed by 32-bit (1-word) data transfer instruction. (Following table shows only lower 8 bits of data.)

For detail of addresses, refer to [Table 20-6](#page-369-0). Use below values to "command" described in a column of Addr[15:9] in the [Table 20-6](#page-369-0).

Note 1) Always set to "0" to the address bit [1:0].

Note 2) Set below values to the address bit [19] according to Flash memory size. Memory size is 1MB or less : Always set to "0" Memory size is over 1MB : If bus write to 1MB area or less, the bit is set to "0". If bus write to over 1MB area, the bit is set to "1".

Table 20-5 Command Sequence

Command	1st bus cycle	2nd bus cycle	3rd bus cycle	4th bus cycle	5th bus cycle	6th bus cycle	7th bus cycle
	Addr.	Addr.	Addr.	Addr.	Addr.	Addr.	Addr.
	Data	Data	Data	Data	Data	Data	Data
Read	0xXX		$\overline{}$	-			-
	0xF0		$\overline{}$				
Read/reset	0xX55X	0xXAAX	0xX55X				-
	0xAA	0x55	0xF0				
ID-Read	0xX55X	0xXAAX	0xX55X	IA	0xXX		
	0xAA	0x55	0x90	0x00	ID		
Automatic page program	0xX55X	0xXAAX	0xX55X	PA	PA	PA	PA
	0xAA	0x55	0xA0	P _D ₀	PD ₁	PD ₂	PD ₃
Automatic chip erase	0xX55X	0xXAAX	0xX55X	0xX55X	0xXAAX	0xX55X	$\overline{}$
	0xAA	0x55	0x80	0xAA	0x55	0x10	$\qquad \qquad -$
Automatic block erase	0xX55X	0xXAAX	0xX55X	0xX55X	0xXAAX	BA	$\qquad \qquad -$
	0xAA	0x55	0x80	0xAA	0x55	0x30	-
Automatic protect bit pro- gram	0xX55X	0xXAAX	0xX55X	0xX55X	0xXAAX	0xX55X	PBA
	0xAA	0x55	0x9A	0xAA	0x55	0x9A	0x9A
Automatic protect bit erase	0xX55X	0xXAAX	0xX55X	0xX55X	0xXAAX	0xX55X	0xXX
	0xAA	0x55	0x6A	0xAA	0x55	0x6A	0x6A

Supplementary explanation

- ・ IA: IDAddress
- ・ ID: ID data
- PA: Program page address
- ・ PD: Program data (32-bit data)

After the 4th bus cycle, input data in the order of the addresses per page

・ BA: Block address (see [Table 20-7](#page-369-0))

・ PBA: Protect bit address (see [Table 20-8](#page-370-0))

20.2.5.2 Address Bit Configuration in the Bus Cycle

Table 20-6 is used in conjunction with ["Table 20-5 Command Sequence"](#page-368-0).

Set the address setting according to the normal bus write cycle address configuration from the first bus cycle.

Table 20-6 Address bit configuration in the bus write cycle

20.2.5.3 Block Address(BA)

Table 20-7 shows block addresses. Specify any address included in the block to be erased in the 6th bus write cycle of the automatic block erase command.

Table 20-7 Block address

20.2.5.4 How to Specify Protect Bit (PBA)

The protect bit is specified in 1-bit unit in programming and in 4-bit unit in erasing.

Table 20-8 shows a protect bit selection table of the automatic protect bit program. The column of address example indicates an address described in upper side is used in the use boot mode and the lower side is used in the single boot mode.

Four protect bits are erased by the automatic protect bit erase command in all.

	Address of 7th bus write cycle					
Protect bit Block		Address Address [14:9] [8]		Address $^{[7]}$	Address example [31:0]	
Block ₀	<blpro[0]></blpro[0]>		$\mathbf 0$	Ω	0x0000 0000 0x3F80 0000	
Block1	<blpro[1]></blpro[1]>		$\mathbf 0$		0x0000 0080 0x3F80 0080	
Block ₂	<blpro[2]></blpro[2]>	Fix to "0"		Ω	0x0000 0100 0x3F80 0100	
Block3	<blpro[3]></blpro[3]>				0x0000 0180 0x3F80 0180	

Table 20-8 Protect bit program address

20.2.5.5 ID-Read Code (IA, ID)

Table 20-9 shows how to specify a code and the content using ID-Read command.

The column of address example indicates an address described in the upper side is used in the use boot mode and the lower side is used in the single boot mode

Code	ID[7:0]	IA[13:12]	Address Example [31:0]
Manufacture code	0x98	0 _V 00	0x0000 0000 0x3F80 0000
Device code	0x5A	0y01	0x0000 1000 0x3F80 1000
	Reserved	0v10	
Macro code	0x33	0y11	0x0000 3000 0x3F80 3000

Table 20-9 ID-Read Command codes and contents

20.2.5.6 Example of Command Sequence

(1) use boot mode

(2) Data single boot mode

TOSHIBA

20.2.6 Flowchart

20.2.6.1 Automatic Program

Automatic Page Programming Command Sequence (Address/ Command)

Figure 20-4 Flowchart of automatic program

20.2.6.2 Automatic Erase

Figure 20-5 Flowchart of automatic erase

20.3 How to Reprogram Flash using Single Boot Mode

The single boot mode utilizes a program contained in built-in BOOT ROM for reprogrammig Flash memory. In this mode, BOOT ROM is mapped to the area containg interrupt vector tables and Flash memory is mapped to another address area other than BOOT ROM area.

In the boot mode, Flash memory is reprogrammed using serial command/data transfer. With connecting serial channel (SIO/UART) of this device to the external host, a reprogramming program is copied from the external host to the built-in RAM. A reprogramming routine in the RAM is executed to reprogram Flash memory. For details of communication with host, follow the protocol described later.

Even in the single boot mode, do not generate interrupt/fault to avoid abnormal program termination.

To secure the contents of Flash memory in the single chip mode (normal operation mode), once re-programming is complete, it is recommended to protect relevant flash blocks against accidental erasure during subsequent single chip operations.

20.3.1 Mode Setting

In order to execute the on-board programming, this device is booted-up in the single boot mode. Below setting is for the single boot mode setting.

```
\overline{BOOT} = 0\overline{\text{RESET}} = 0 \rightarrow 1
```
While **BOOT** pin is set to the above in advance, set **RESET** pin to "0". Then release **RESET** pin, the device will boot-up in the single boot mode.

20.3.2 Interface Specification

This section describes SIO/UART communication format in the single boot mode. The serial operation supports both UART (asynchronous communication) and I/O interface modes. In order to execute the on-board programming, set the communication format of the programming controller as well.

UART communication

Communication channel: channel 0

Serial transfer mode: UART (asynchronous), half-duplex, LSB first

Data length: 8-bit

Parity bit: None

STOP bit: 1-bit

Baud rate: Arbitrary baud rate

I/O interface mode

Communication channel: channel 0

Serial transfer mode: I/O interface, full-duplex, LSB first

Synchronous signal (SCLK0): Input mode, rising edge setting

Handshaking signal: PH3 (output mode)

Baud rate: Arbitrary baud rate

The boot program operates the clock/mode control block setting as an initial condition. For detail of the initial setting of the clock, refer to "Clock/Mode control".

As explained in the ["20.3.5.1 Serial Operation Mode Determination"](#page-376-0), a baud rate is determined by the 16 bit timer (TMRB). When determining the baud rate, communication is executed by 1/16 of a desired baud rate. Therefore, the communication baud rate must be within the measurable range. The timer count clock operates at ΦT1 (fc/2).

A handshaking pin of I/O interface mode outputs "Low" waiting in receive state and outputs "High" in transmission state. Check the handshaking pin before communications and must follow the communication protocol.

Table 20-10 shows the pins used in the boot program. Other than these pins are not used by the boot program.

Table 20-10 Pin connection

ο:used ×:unused

20.3.3 Restrictions on Internal Memories

Note that the single boot mode places restrictions on the built-in RAM and built-in flash memory as shown in Table 20-11.

Note:If a password is erased data (0xFF), it is difficult to protect data secure due to an easy-to-guess password. Even if the single boot mode is not used, it is recommended to set a unique value as a password.

20.3.4 Operation Command

The boot program provides the following operation commands.

Table 20-12 Operation command data

Operation command da- ta	Operation mode
0x10	RAM transfer
0x40	Flash memory chip erase and protect bit erase

20.3.4.1 RAM Transfer

The RAM transfer is to store data from the controller to the built-in RAM . When the transfer is complete normally, a user program starts. User program can use the memory address of $0x2000$ 0400 or later except 0x2000 0000 to 0x2000 03FF for the boot program. CPU will start execution from RAM store start address. The start address must be even address.

This RAM transfer function enables user-specific on-board programming control. In order to execute the on-board programming by a user program, use Flash memory command sequence explained in [20.2.5.](#page-368-0)

20.3.4.2 Flash Memory Chip Erase and Protect Bit Erase

Flash memory chip erase and protect bit erase commands erase the entire blocks of Flash memory and write/erase protects of all blocks regardless of write/erase protect or security status.

Note:Even if a command is completed normally, ACK may be returned as the abnormal termination. In this case, check the erasure again.

20.3.5 Common Operation regardless of Command

This section describes common operation under the boot program execution.

20.3.5.1 Serial Operation Mode Determination

When the controller communicates via UART, set the 1st byte to 0x86 at the desired baud rate. When the controller communicate via I/O interface mode, set the 1st byte to 0x30 at 1/16 of the desired baud rate. Figure 20-6 shows waveforms in each case.

Figure 20-6 Serial operation mode determination data

[Figure](#page-378-0) 20-7 shows a flowchart of boot program. Using 16-bit timer (TMRB) with the time of tAB, tAC and tAD, the 1st byte of serial operation mode determination data (0x86, 0x30) after reset is provided. In [Figure 20-7](#page-378-0), the CPU monitors level of the receive pin, and obtains a timer value at the moment when the receive pin's level is changed. Consequently, the timer values of tAB, tAC and tAD have a margin of error. In addition, note that if the transfer goes at a high baud rate, the CPU may not be able to determine the level of receive pin. In particular, I/O Interface tends to generate this problem since its baud rate is generally much higher than those of UART. To avoid this, the controller should send data at 1/16 of the desired baud rate in the I/O interface mode.

The flowchart in [Figure 20-8](#page-379-0) shows the serial operation mode is determined that the time length of the receive pin is long or short. If the length is $tAB \leq tCD$, the serial operation mode is determined as UART mode. The time of tAD is used whether the automatic baud rate setting is enable or not. If the length is tAB > tCD, the serial operation mode is determined as I/O Interface mode. Note that timer values of

tAB, tAC and tAD have a margin of error. If the baud rate is high and operation frequency is low, each timer value becomes small. This may generates unexpected determination occurs. (To prevent this problem, re-set UART within the programming routine.)

For example, the serial operation mode may be determined to be I/O Interface mode when the intended mode is UART mode. To avoid such a situation, when UART mode is utilized, the controller should allow for a time-out period where the time is expected to receive an echo-back (0x86) from the target board. The controller should give up the communication if it fails to get that echo-back within the allowed time. When I/O Interface mode is utilized, once the first serial byte has been transmitted, the controller should send the SCLK clock after a certain idle time to get an acknowledge response. If the received acknowledge response is not 0x30, the controller should give up further communications.

When the intended mode is I/O interface mode, it is not necessary that the first byte is 0x30 as long as tAB >tCD as shown above. 0x91, 0xA1 or 0xB1 can be sent as the first byte code to determine the falling edges of Point A and Point C and the rising edges of Point B and Point D. If tAB>tCD is established and SIO is selected by the resolution of the operation mode determination, the second byte code is 0x30 even though the transmitted code on the first byte is not 0x30 (The first byte code to determine I/O interface mode is described as 0x30).

Figure 20-7 Serial operation mode receive flowchart

20.3.5.2 Acknowledge Response Data

The boot program represents processing states in specific codes and sends them to the controller. Table 20-13 to [Table 20-16](#page-380-0) show the values of acknowledge responses to each receive data.

In Table 20-14 to [Table 20-16,](#page-380-0) the upper four bits of the acknowledge response are equal to those of the operation command data. The 3rd bit indicates a receive error. The 0th bit indicates an invalid operation command error, a checksum error or a password error. The 1st bit and 2nd bit are always "0". Receive error checking is not performed in I/O Interface mode.

Table 20-13 ACK response to the serial operation determination data

Note:When the serial operation is determined as UART, if the baud rate setting is determined as unacceptable, the boot program aborts without sending back any response.

Note:The upper 4 bits of the ACK response data are the same as those of the previous command data.

Table 20-15 ACK response to the CHECK SUM data

Note:The upper 4 bits of the ACK response data are the same as those of the operation command data.

Table 20-16 ACK response to Flash memory chip erase and protect bit erase operation

20.3.5.3 Password Determination

The boot program use the below area to determine whether a password is required or use as a password.

The RAM Transfer command performs a password verification regardless of necessity judging data. Flash memory chip erase or protect bit erase command performs a password verification only when necessity judging is determined as "required".

If a password is set to 0xFF (erased data), it is difficult to protect data securely due to an easy-to-guess password. Even if Single Boot mode is not used, it is recommended to set a unique value as a password.

(1) Password verification using RAM transfer command

If all these address locations contain the same bytes of data other than 0xFF, this condition is determined as a password area error as shown in [Figure 20-9.](#page-381-0) In this case, the boot program returns an error acknowledge (0x11) in response to the 17th byte of checksum value regardless of the password verification.

The boot program verifies 5th byte to 16th byte of receive data (password data). A password error occurs if all 12 bytes do not match. If the password error is determined, an ACK response data to the 17th of CHECK SUM data is a password error.

The password verification is performed even if the security function is enabled.

Figure 20-9 Password area check flowchart

(2) Password verification to Flash memory chip erase and protect bit erase command

When a password is enable in the erase password necessity determination area as shown in Figure 20-10 and the passwords are identical data, a password area error occurs. If a password area error is determined, an ACK response to the 17th byte of CHECK SUM sends 0x41 regardless of the password verification.

The boot program verifies 5th byte to 16th byte of receive data (password data). A password error occurs if all 12 bytes do not match. If the password error is determined, an ACK response data to the 17th of CHECK SUM data is a password error.

The password verification is performed even if the security function is enabled.

20.3.5.4 CHECK SUM Calculation

The checksum is calculated by 8-bit addition to transmit data, dropping the carries, and taking the two's complement of the total sum. The controller must perform the same checksum operation in transmitting checksum bytes.

Example of CHECK SUM

To calculate the checksum for a series of 0xE5 and 0xF6, perform 8-bit addition.

 $0xE5 + 0xF6 = 0x1DB$

Take the two's complement of the sum to the lower 8-bit, and that is a checksum value. So the boot program sends 0x25 to the controller.

 $0 - 0xDB = 0x25$

20.3.6 Transfer Format at RAM Transfer

This section shows a RAM transfer command format. Transfer directions in the table are indicated as follows: Transfer direction (C→T): Controller to TMPM061FWFG Transfer direction (C←T): TMPM061FWFG to Controller

20.3.7 Transfer Format of Flash memory Chip Erase and Protect Bit Erase

This section shows a transfer format of Flash memory chip erase and protect bit erase commands. Transfer directions in the table are indicated as follows:

Transfer direction (C→T): Controller to TMPM061FWFG

Transfer direction (C←T): TMPM061FWFG to Controller

TOSHIBA

20.3.8 Boot Program Whole Flowchart

This section shows a boot program whole flowchart.

20.3.9 Reprogramming Procedure of Flash using reprogramming algorithm in the onchip BOOT ROM

This section describes the reprogramming procedure of the flash using reprogramming algorithm in the onchip boot ROM.

20.3.9.1 Step-1

The condition of Flash memory does not care whether a user program made of former versions has been written or erased. Since a programming routine and programming data are transferred via the SIO (SIO0), the SIO0 must be connected to an external host. A programming routine (a) is prepared on the host.

20.3.9.2 Step-2

Release the reset by pin condition setting in the boot mode and boot-up the BOOT ROM. According to the procedure of boot mode, transfer the programming routine (a) via SIO0 from the source (host). A password verification with the the password in the user application program is perfomed. (If Flash memory is erased, an erase data (0xFF) is dealt with a password.)

TOSHIBA

20.3.9.3 Step-3

If the password verification is complete, the boot program transfer a programming routine (a) from the host into the on-chip RAM. The programming routine must be stored in the range from 0x2000_0400 to the end address of RAM.

20.3.9.4 Step-4

The boot program jumps to the programming routine (a) in the on-chip RAM to erase the flash block containing old application program codes. The Block Erase or Chip Erase command is be used.

20.3.9.5 Step-5

The boot program executes the programming routine (a) to download new application program codes from the host and programs it into the erased flash area. When the programming is complete, the writing or erase protection of that flash area in the user's program must be set.

In the example below, new program code comes from the same host via the same SIO0 channel as for the programming routine. However, once the programming routine has begun to execute, it is free to change the transfer path and the source of the transfer. Create a hardware board and programming routine to suit your particular needs.

20.3.9.6 Step-6

When programming of Flash memory is complete, power off the board and disconnect the cable leading from the host to the target board. Turn on the power again so that the device re-boots in the singlechip (Normal) mode to execute the new program.

20.4 Programming in the User Boot Mode

A user Boot mode is to use flash memory programming routine defined by users. It is used when the data transfer buses for flash memory program code on the user application is different from the serial I/O. It operates in the single chip mode; therefore, a switch from normal mode in which user application is activated in the use boot mode to the user boot mode for programming flash is required. Specifically, add a mode judgment routine to the reset service routine in the user application program.

The condition to switch the modes needs to be set according to the user's system setup condition. Also, a flash memory programming routine that the user uniquely makes up needs to be set in the new application. This routine is used for programming after being switched to the user boot mode. The data in built-in Flash memory cannot be read out during erase/reprogramming mode. Thus, reprogramming routine must be take place while it is stored in the area outside of Flash memory area. Once re-programming is complete, it is recommended to protect relevant flash blocks from accidental reprogramming. Be sure not to generate interrupt/fault to avoid abnormal termination during the user boot mode.

Taking examples from two cases such as the method that reprogramming routine stored in Flash memory (1-A) and transferred from the external device (1-B), the following section explains the procedure. For a detail of the program/erase to Flash memory, refer to ["20.2 Detail of Flash Memory".](#page-363-0)

20.4.1 (1-A) Procedure that a Programming Routine Stored in Flash memory

20.4.1.1 Step-1

A user determines the conditions (e.g., pin status) to enter the user boot mode and the I/O bus to be used to transfer data. Then suitable circuit design and program are created. Before installing the device on a printed circuit board, write the following three program routines into an arbitrary flash block using programming equipment such as a flash writer.

(a) Mode determination routine: A program to determine to switch to user boot mode or not (b) Flash programming routine: A program to download new program from the host controller and re-program Flash memory

(c) Copy routine: A program to copy the data described in (a) to the built-in RAM or external memory device

20.4.1.2 Step-2

This section explains the case that a programming routine storied in the reset routine. First, the reset routine determines to enter the user boot mode. If mode switching conditions are met, the device enters the user boot mode to reprogram data.

20.4.1.3 Step-3

Once the device enters the user boot mode, execute the copy routine (C) to download the flash programming routine (b) from the host controller to the built-in RAM .

20.4.1.4 Step-4

Jump to the the reprogramming routine in the built-in RAM to release the write/erase protection for the old application program, and to erase a flash in block unit.

20.4.1.5 Step-5

Continue to execute the flash programming routine to download new program data from the host controller and program it into the erased flash block. When the programming is complete, the write/erase protection of that flash block in the user program area must be set.

20.4.1.6 Step-6

Set RESET to "0". Upon reset, Flash memory is set to the normal mode. After reset, the CPU will start along with the new application program.

20.4.2 (1-B) Procedure that a Programming Routine is transferred from External Host

20.4.2.1 Step-1

A user determines the conditions (e.g., pin status) to enter the user boot mode and the I/O bus to be used to transfer data. Then suitable circuit design and program are created. Before installing the device on a printed circuit board, write the following two program routines into an arbitrary flash block using programming equipment such as a flash writer.

In addition, prepare a reprogramming routine shown below must be stored on the host controller.

(c) Reprogramming routine: A program to reprogram data

20.4.2.2 Step-2

This section explains the case that a programming routine storied in the reset routine. First, the reset routine determines to enter the user boot mode. If mode switching conditions are met, the device enters the user boot mode to reprogram data.

20.4.2.3 Step-3

Once the device enters the user boot mode, execute the transfer routine (b) to download the programming routine (c) from the host controller to the built-in RAM .

20.4.2.4 Step-4

Jump to the the reprogramming routine in the built-in RAM to release the write/erase protection for the old application program, and to erase a flash in block unit.

20.4.2.5 Step-5

Continue to execute the flash programming routine (c) to download new program data from the host controller and program it into the erased flash block. When the programming is complete, the write/erase protection of that flash block in the user program area must be set.

20.4.2.6 Step-6

Set RESET to "0". Upon reset, Flash memory is set to the normal mode. After reset, the CPU will start along with the new application program.

21. Debug Interface

21.1 Specification Overview

TMPM061FWFG contains the Serial Wire Debug Port (SW-DP) units as the interface with the debugging tools. SW-DP supports the Serial Wire Debug Port (SWCLK, SWDIO).

For details about SW-DP, refer to the Arm manual "Cortex-M0 Technical Reference Manual".

21.2 Pin Functions

The debug interface pins can also be used as general-purpose ports.

After reset, the debug interface pins are configured as debug port function pins. To use the debug interface pins as other function, change of setting is required.

Table 21-1 shows the settings of the debug interface pins after reset.

Debug Function	Value of Related port settings after reset						
	Function (PxFR)	Input (PxIE)	Output (PxCR)	Pull-up (PxPUP)	Pull-down (PxPDN)		
SWCLK							
SWDIO							

Table 21-1 Debug Interface Pins and Related Port Settings after Reset

21.3 Reset Vector Break

TMPM061FWFG is prohibited from transmission with debug tools while reset caused by RESET pin is effective. When setting a break by using reset vector, set the following procedure after reset; set break points from the debug tools, then set the application interrupt and the <SYSRESETREQ> bit of the reset control register to reset again.

21.4 Precautions on Use of Debug Interface Pin Used as General-purpose Port

If a debug interface pin is set to a general-purpose port by a user program after reset, then debug tools cannot control a MCU. In order to connect a debug tool to MCU again, a certain mechanism in which a general-purpose port can be changed to a debugging interface setting must be prepared by a user.

21.5 Debug Enable Pin

TMPM061FWFG prepares debug enable pins (DBGEN). These debug enable pins are used for reliable communication with a debug tool.

Debug Enable pins are valid to be input during reset. If "Low" signal is sampled on the rising edge of reset signal, a debug enable condition is set and its function is fixed to the interface. This condition remains until next pin reset.

A debug tool communicates with a processor core after reset. This is effective when a user program changes a port setting of debug interface shortly after reset.

Below register can monitor the debug conditions.

21.6 Peripheral Functions in Halt Mode

When Cortex-M0 core enters in the halt mode, the watchdog-timer (WDT) automatically stops. It is selectable that 16-bit Timer (TMRB and TMR16A) continue or stop counting. Other peripherals are continue operating.

21.7 About connection with debug tool

Concerning a connection with debug tools, refer to manufactures recommendations.

Debug interface pins contain a pull-up resistor and a pull-down resistor. When debug interface pins are connected with external pull-up or pull-down, please pay attention to input level.

- 21. Debug Interface
- 21.7 About connection with debug tool

TOSHIBA

22. Electrical Characteristics

22.1 Absolute Maximum Ratings

Note:Absolute maximum ratings are limiting values of operating and environmental conditions which should not be exceeded under the worst possible conditions. The equipment manufacturer should design so that no Absolute maximum rating value is exceeded with respect to current, voltage, power consumption, temperature, etc. Exposure to conditions beyond those listed above may cause permanent damage to the device or affect device reliability, which could increase potential risks of personal injury due to IC blowup and/or burning.

22.2 DC Electrical Characteristics (1/3)

Ta = −40 to 85 °C

Note 1: Ta = 25 °C, DVDD3 = RVDD3 = AVDD3 = DSRVDD = SRVDD = VLC = 3.3 V, unless otherwise noted. Note 2: The same voltage must be supplied to DVDD3, AVDD3, RVDD3, DSRVDD3, SRVDD and VLC.

22.3 DC Electrical Characteristics (2/3)

Parameter	Symbol	Condition	Min	Typ.	Max	Unit
Low-level output current	I_{OL}	Per pin	-	-	2	mA
	ΣI_{OL}	Total, all Port	-	-	35	mA
High-level output current	Iон	Per pin	-	-	-2	mA
	ΣI_{OH}	Total, all Port	-		-35	mA

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 1.8 V to 3.6 V, Ta = −40 to 85 °C

Note:The same voltage must be supplied to DVDD3, AVDD3, RVDD3, DSRVDD3, SRVDD and VLC.

22.4 DC Electrical Characteristics (3/3)

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 1.8 V ~ 3.6 V, Ta = -40 ~ 85 °C

Note 1: Ta = 25 °C, DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 3.3 V, unless otherwise noted.

Note 2: I_{DD} NORMAL: Measured with the dhrystone ver. 2.1 operated in FLASH.

All functions operates excluding reference voltage of the AD converter, ΔΣ AD converter, temperature sensor, bleeder resistance of LCD and I/O port.

Note 3: I_{DD} IDLE: Measured with all functions stopped. The currents flow through DVDD3, AVDD3, RVDD3, DSRVDD3, SRVDD and VLC are included.

22.5 10-bit ADC Electrical Characteristics

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC =1.8 V to 3.6 V AVSS = DVSS = 0V, Ta = -40 to 85 °C.

Note 1: 1LSB = (AVREFH − AVREFL)/1024 [V]

Note 2: This characteristics is shown in operating only ADC.

22.6 24-bit ΔΣADC Electrical Characteristics

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 2.9 V to 3.6 V

Note 1: This characteristics is shown in operating only ADC.

Note 2: A value when AINN=0V.

22.7 Temperature Sensor Characteristics

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 1.8 V ~ 3.6 V, Ta = -40 ~ 85 °C

Note 1: The reference voltage circuit is shared with a ΔΣ analog/digital converter.

Note 2: These are design assurance values of single temperature sensor, which were obtained from a straight-line approximation based on the measured values at 30 °C and 60 °C.

22.8 LCD Characteristics

22.9 AC Electrical Characteristics

22.9.1 AC measurement condition

The AC characteristics data of this chapter is measured under the following conditions unless otherwise noted

- Output levels: High = $0.8 \times$ DVDD3A, $0.8 \times$ DVDD3, Low = $0.2 \times$ DVDD3A, $0.2 \times$ DVDD3B
- ・ Input levels: Refer to low-level input voltage and high-level input voltage in "DC Electrical Characteristics".
- Load capacity: $CL = 30pF$

22.9.2 Serial Channel (SIO/UART)

22.9.2.1 I/O Interface mode

In the table below, the letter x represents the SIO operation clock cycle time which is identical to the fsys cycle time. It varies depending on the programming of the clock gear function.

(1) SCLK input mode

Note:SCLK rise or fall ; Measured relative to the programmed active edge of SCLK.

(2) SCLK output mode

$DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 2.7 to 3.6V$

DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 1.8 V

22.9.3 Serial Bus Interface (I2C/SIO)

22.9.3.1 I2C Mode

In the table below, the letter x represents the I2C operation clock cycle time which is identical to the fsys cycle time. It varies depending on the programming of the clock gear function.

n denotes the value of n programmed into the SCK (SCL output frequency select) field in the SBIxCR.

Note 1: SCL clock Low width (output) = $(2^{n-1} + 58)/x$

Note 2: SCL clock High width (output) = $(2^{n-1} + 14)/x$ On I2C-bus specification, Maximum Speed of Standard Mode is 100kHz, Fast mode is 400khz. Internal SCL Frequency setting should comply with Note1 & Note2 shown above.

Note 3: The output data hold time is equal to 4x of internal SCL.

Note 4: The Philips I2C-bus specification states that a device must internally provide a hold time of at least 300 ns for the SDA signal to bridge the undefined region of the falling edge of SCL. However, this SBI does not satisfy this requirement. Also, the output buffer for SCL does not incorporate slope control of the falling edges; therefore, the equipment manufacturer should design so that the input data hold time shown in the table is satisfied, including tr/tf of the SCL and SDA lines.

Note 5: Software -dependent

Note 6: The Philips I2C-bus specification instructs that if the power supply to a Fast-mode device is switched off, the SDA and SCL I/O pins must be floating so that they don't obstruct the bus lines. However, this SBI does not satisfy this requirement.

22.9.3.2 Clock-Synchronous 8-Bit SIO mode

In the table below, the letter x represents the I2C operation clock cycle time which is identical to the fsys cycle time. It varies depending on the programming of the clock gear function.

(1) SCK Input Mode (The electrical specifications below are for an SCK signal with a 50% duty cycle.)

(2) SCK Output Mode (The electrical specifications below are for an SCK signal with a 50% duty cycle.)

 $DVDD3 = AVDD3 = RVDD3 = DSRVDD3 = SRVDD = VLC = 2.7 to 3.6V$

	Symbol	Equation		16 MHz		
Parameter		Min	Max	Min	Max	Unit
SCK cycle (programmable)	t _{SCY}	16x (Note 1)		1000		
Output Data \leftarrow SCK rise	toss	$t_{SCY}/2 - 30$ (Note2)		470		
SCK rise \rightarrow Output Data hold	t _{OHS}	$t_{SCY}/2 - 30$		470		ns
Valid Data input \leftarrow SCK rise	t _{SRD}	$x + 45$		107.5		
SCK rise \rightarrow Input Data hold	^t HSR	0		0		

Note 1: SCK cycle after automatic wait becomes 14x.

Note 2: SO data output after automatic wait may be $t_{SCY}/2 - x - 20$.

22.9.4 Event Counter

In the table below, the letter x represents the TMRB operation clock cycle time which is identical to the fsys cycle time. It varies depending on the programming of the clock gear function.

22.9.5 Capture

In the table below, the letter x represents the TMRB operation clock cycle time which is identical to the fsys cycle time. It varies depending on the programming of the clock gear function.

22.9.6 External Interrupt

In the table below, the letter x represents the fsys cycle time.

1. Except STOP release interrupts

TOSHIBA

2. STOP release interrupts

22.9.7 SCOUT Pin AC Characteristic

Note:In the above table, the letter T represents the cycle time of the SCOUT output clock.

22.9.8 Debug Communication

22.9.9 Flash Characteristics

22.9.10 On chip oscillator

22.10 Recommended Oscillation Circuit

TOSHIBA

Figure 22-1 High-frequency oscillation connection

Figure 22-2 Low-frequency oscillation connection

Note:To obtain a stable oscillation, load capacity and the position of the oscillator must be configured properly. Since these factors are strongly affected by substratepatterns, please evaluate oscillation stability using the substrate you use.

22.10.1 Ceramic oscillator

This product has been evaluated by the ceramic oscillator by Murata Manufacturing Co., Ltd.

Please refer to the company's website for details.

22.10.2 Crystal oscillator

This product has been evaluated by the crystal oscillator by Seiko Instruments Inc..

Please refer to the company's website for details.

22.10 Recommended Oscillation Circuit

23. Port Section Equivalent Circuit Schematic

Basically, the gate symbols written are the same as those used for the standard CMOS logic IC [74HCXX] series.

The input protection resistance ranges from several tens of Ω to several hundreds of Ω . Damping resistors X2 and XT2 are shown with a typical value.

Note:Resistors without values in the figure show input protection resistors.

23.1 PORT pin

23.2 Analog pin

TOSHIBA

23.3 Control pin

23.4 Clock pin

24. Package Dimensions

Type: LQFP100-P-1414-0.50G

RESTRICTIONS ON PRODUCT USE

Toshiba Corporation and its subsidiaries and affiliates are collectively referred to as "TOSHIBA". Hardware, software and systems described in this document are collectively referred to as "Product".

- TOSHIBA reserves the right to make changes to the information in this document and related Product without notice.
- This document and any information herein may not be reproduced without prior written permission from TOSHIBA. Even with TOSHIBA's written permission, reproduction is permissible only if reproduction is without alteration/omission.
- Though TOSHIBA works continually to improve Product's quality and reliability, Product can malfunction or fail. Customers are responsible for complying with safety standards and for providing adequate designs and safeguards for their hardware, software and systems which minimize risk and avoid situations in which a malfunction or failure of Product could cause loss of human life, bodily injury or damage to property, including data loss or corruption. Before customers use the Product, create designs including the Product, or incorporate the Product into their own applications, customers must also refer to and comply with (a) the latest versions of all relevant TOSHIBA information, including without limitation, this document, the specifications, the data sheets and application notes for Product and the precautions and conditions set forth in the "TOSHIBA Semiconductor Reliability Handbook" and (b) the instructions for the application with which the Product will be used with or for. Customers are solely responsible for all aspects of their own product design or applications, including but not limited to (a) determining the appropriateness of the use of this Product in such design or applications; (b) evaluating and determining the applicability of any information contained in this document, or in charts, diagrams, programs, algorithms, sample application circuits, or any other referenced documents; and (c) validating all operating parameters for such designs and applications. **TOSHIBA ASSUMES NO LIABILITY FOR CUSTOMERS' PRODUCT DESIGN OR APPLICATIONS.**
- **PRODUCT IS NEITHER INTENDED NOR WARRANTED FOR USE IN EQUIPMENTS OR SYSTEMS THAT REQUIRE EXTRAORDINARILY HIGH LEVELS OF QUALITY AND/OR RELIABILITY, AND/OR A MALFUNCTION OR FAILURE OF WHICH MAY CAUSE LOSS OF HUMAN LIFE, BODILY INJURY, SERIOUS PROPERTY DAMAGE AND/OR SERIOUS PUBLIC IMPACT ("UNINTENDED USE").** Except for specific applications as expressly stated in this document, Unintended Use includes, without limitation, equipment used in nuclear facilities, equipment used in the aerospace industry, lifesaving and/or life supporting medical equipment, equipment used for automobiles, trains, ships and other transportation, traffic signaling equipment, equipment used to control combustions or explosions, safety devices, elevators and escalators, and devices related to power plant. IF YOU USE PRODUCT FOR UNINTENDED USE, TOSHIBA ASSUMES NO LIABILITY FOR PRODUCT. For details, please contact your TOSHIBA sales representative or contact us via our website.
- Do not disassemble, analyze, reverse-engineer, alter, modify, translate or copy Product, whether in whole or in part.
- Product shall not be used for or incorporated into any products or systems whose manufacture, use, or sale is prohibited under any applicable laws or regulations.
- The information contained herein is presented only as guidance for Product use. No responsibility is assumed by TOSHIBA for any infringement of patents or any other intellectual property rights of third parties that may result from the use of Product. No license to any intellectual property right is granted by this document, whether express or implied, by estoppel or otherwise.
- **ABSENT A WRITTEN SIGNED AGREEMENT, EXCEPT AS PROVIDED IN THE RELEVANT TERMS AND CONDITIONS OF SALE FOR PRODUCT, AND TO THE MAXIMUM EXTENT ALLOWABLE BY LAW, TOSHIBA (1) ASSUMES NO LIABILITY WHATSOEVER, INCLUDING WITHOUT LIMITATION, INDIRECT, CONSEQUENTIAL, SPECIAL, OR INCIDENTAL DAMAGES OR LOSS, INCLUDING WITHOUT LIMITATION, LOSS OF PROFITS, LOSS OF OPPORTUNITIES, BUSINESS INTERRUPTION AND LOSS OF DATA, AND (2) DISCLAIMS ANY AND ALL EXPRESS OR IMPLIED WARRANTIES AND CONDITIONS RELATED TO SALE, USE OF PRODUCT, OR INFORMATION, INCLUDING WARRANTIES OR CONDITIONS OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, ACCURACY OF INFORMATION, OR NONINFRINGEMENT.**
- Do not use or otherwise make available Product or related software or technology for any military purposes, including without limitation, for the design, development, use, stockpiling or manufacturing of nuclear, chemical, or biological weapons or missile technology products (mass destruction weapons). Product and related software and technology may be controlled under the applicable export laws and regulations including, without limitation, the Japanese Foreign Exchange and Foreign Trade Law and the U.S. Export Administration Regulations. Export and re-export of Product or related software or technology are strictly prohibited except in compliance with all applicable export laws and regulations.
- Please contact your TOSHIBA sales representative for details as to environmental matters such as the RoHS compatibility of Product. Please use Product in compliance with all applicable laws and regulations that regulate the inclusion or use of controlled substances. including without limitation, the EU RoHS Directive. **TOSHIBA ASSUMES NO LIABILITY FOR DAMAGES OR LOSSES OCCURRING AS A RESULT OF NONCOMPLIANCE WITH APPLICABLE LAWS AND REGULATIONS.**

TOSHIBA ELECTRONIC DEVICES & STORAGE CORPORATION

https://toshiba.semicon-storage.com/

单击下面可查看定价,库存,交付和生命周期等信息

[>>Toshiba\(东芝\)](https://www.oneyac.com/brand/1164.html)