

SGM8903 Capless 3Vrms Line Driver with Adjustable Gain

GENERAL DESCRIPTION

The SGM8903 is a 3Vrms pop/click-free stereo line driver designed to allow the removal of the output DC-blocking capacitors for reduced component count and cost. The device is ideal for single supply electronics where size and cost are critical design parameters.

The SGM8903 is capable of driving 3Vrms into a $2.5 k\Omega$ load with 5V supply voltage. The device has differential inputs and uses external gain setting resistors that supports a gain range of $\pm 1 \text{V/V}$ to $\pm 10 \text{V/V}$. The use of external gain resistors also allows the implementation of a 2nd-order low pass filter to compliment DAC's and SOC converters. The SGM8903 has build-in shutdown control for pop/click-free on/off control.

Using the SGM8903 in audio products can reduce component count compared to traditional methods of generating a 3Vrms output. The SGM8903 doesn't require a power supply greater than 5V to generate an 8.5V_{PP} output, nor does the device require a split rail power supply. The SGM8903 integrates a charge pump to generate a negative supply rail that provides a clean, pop/click-free ground-biased 3Vrms output.

The SGM8903 is available in Green TSSOP-14 package. It operates over an ambient temperature range of -40°C to +85°C.

FEATURES

- Capless Structure
 Eliminates Pop/Clicks
 Eliminates Output DC-Blocking Capacitors
 Provides Flat Frequency Response from DC to 20kHz
- Low Noise and THD
 Typical SNR = 114dB
 Typical V_N = 5.5µVrms
 Typical THD+N = 0.001% (f = 1kHz)
- 3Vrms Output Voltage into 2.5kΩ Load with 5V Supply Voltage
- Differential Input
- -40°C to +85°C Operating Temperature Range
- Available in the Green TSSOP-14 Package

APPLICATIONS

Set-Top Box LCD TV Blue-Ray DVD-Players Home Theater in a Box



PACKAGE/ORDERING INFORMATION

| MODEL | PACKAGE DESCRIPTION | SPECIFIED TEMPERATURE RANGE | ORDERING NUMBER | PACKAGE MARKING | PACKING OPTION |
|---------|------------------------|-----------------------------------|--------------------|---------------------------|---------------------|
| SGM8903 | TSSOP-14 | -40°C to +85°C | SGM8903YTS14G/TR | SGM8903 YTS14 XXXXX | Tape and Reel, 3000 |

NOTE: XXXXX = Date Code and Vendor Code.

Green (RoHS & HSF): SG Micro Corp defines "Green" to mean Pb-Free (RoHS compatible) and free of halogen substances. If you have additional comments or questions, please contact your SGMICRO representative directly.

ABSOLUTE MAXIMUM RATINGS

| Supply Voltage | 0.3V to 6V |
|--|------------------------------------|
| Input Voltage | V_{SS} - 0.3V to V_{DD} + 0.3V |
| Minimum Load Impedance (R _L) | 600Ω |
| EN to GND | 0.3V to V _{DD} + 0.3V |
| Junction Temperature | +150°C |
| Storage Temperature Range | 65°C to +150°C |
| Lead Temperature (Soldering, 10s) . | +260°C |
| ESD Susceptibility | |
| HBM | 6000V |
| MM | 300V |
| CDM | 1000V |

RECOMMENDED OPERATING CONDITIONS

| Supply Voltage Range | 3V to 5.5V |
|-----------------------------|----------------|
| Operating Temperature Range | -40°C to +85°C |

OVERSTRESS CAUTION

Stresses beyond those listed may cause permanent damage to the device. Functional operation of the device at these or any other conditions beyond those indicated in the operational section of the specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.

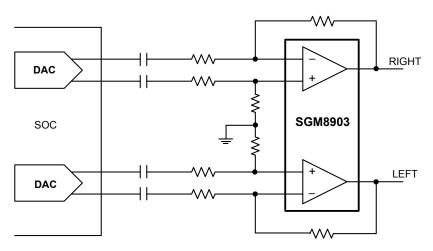
ESD SENSITIVITY CAUTION

This integrated circuit can be damaged by ESD if you don't pay attention to ESD protection. SGMICRO recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage. ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

DISCLAIMER

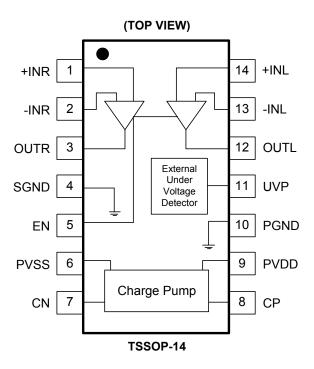
SG Micro Corp reserves the right to make any change in circuit design, specification or other related things if necessary without notice at any time.

TYPICAL OPERATION CIRCUIT





PIN CONFIGURATION



PIN DESCRIPTION

| PIN | NAME | FUNCTION |
|-----|------|---|
| 1 | +INR | Right Channel OPAMP Positive Input. |
| 2 | -INR | Right Channel OPAMP Negative Input. |
| 3 | OUTR | Right Channel OPAMP Output. |
| 4 | SGND | Signal Ground. |
| 5 | EN | Enable Input. Active high. |
| 6 | PVSS | Negative Supply Voltage Output. |
| 7 | CN | Charge Pump Flying Capacitor Negative Terminal. |
| 8 | CP | Charge Pump Flying Capacitor Positive Terminal. |
| 9 | PVDD | Positive Supply. |
| 10 | PGND | Power Ground. |
| 11 | UVP | Under-Voltage Protection Input. |
| 12 | OUTL | Left Channel OPAMP Output. |
| 13 | -INL | Left Channel OPAMP Negative Input. |
| 14 | +INL | Left Channel OPAMP Positive Input. |

ELECTRICAL CHARACTERISTICS

 $(T_A = +25^{\circ}C, \text{ unless otherwise noted.})$

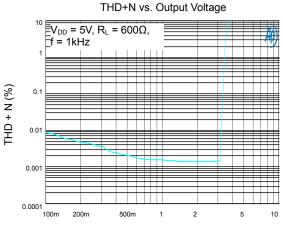
| PARAMETER | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|--|------------------------|-----------------------|-------|-------|
| ELECTRICAL CHARACTERISTICS | | _ | | | _ |
| DC Supply Voltage (V _{DD}) | | 3 | | 5.5 | V |
| Output Offset Voltage (Vos) | V _{DD} = 3V to 5V | | 1.2 | 5 | |
| Power Supply Rejection Ratio (PSRR) | V _{DD} = 3V to 5V | | 97 | | dB |
| High-Level Output Voltage (V _{OH}) | $V_{DD} = 3.3V, R_L = 2.5k\Omega$ | 3.18 | | | V |
| Low-Level Output Voltage (V _{OL}) | $V_{DD} = 3.3V, R_L = 2.5k\Omega$ | | | -3.05 | V |
| High-Level Input Current (EN) (I _{IH}) | $V_{DD} = 5V$, $V_1 = V_{DD}$ | | | 1 | μA |
| Low-Level Input Current (EN) (I _{IL}) | $V_{DD} = 5V$, $V_1 = 0V$ | | | 1 | μA |
| | V _{DD} = 3.3V, No load, EN = V _{DD} | | 10.5 | 14.5 | |
| Supply Current (I _{DD}) | V_{DD} = 5V, No load, EN = V_{DD} | | 11.3 | 15.5 | mA |
| | Shutdown mode, V _{DD} = 3V to 5V | | 0.13 | 0.18 | |
| OPERATING CHARACTERISTICS (V _{DD} = 3.3V, R | $_{L} = 2.5k\Omega$, $C_{PUMP} = C_{PVSS} = 1\mu F$, $C_{IN} = 10\mu F$, $R_{IN} = 1$ | 0kΩ, R _{FB} = | 20kΩ.) ⁽¹⁾ | I. | 1 |
| | THD = 1%, V _{DD} = 3.3V, f = 1kHz | 2.05 | | | |
| Output Voltage (Outputs in Phase) (V _O) | THD = 1%, V _{DD} = 5V, f = 1kHz | 3.05 | | | Vrms |
| | THD = 1%, V_{DD} = 5V, f = 1kHz, R_{L} = 100k Ω | 3.1 | | | |
| Total Harmonic Distortion Plus Noise (THD+N) | V _O = 2Vrms, f = 1kHz | | 0.001 | | % |
| Crosstalk | V _O = 2Vrms, f = 1kHz | | 113 | | dB |
| Output Current Limit (I _O) | V _{DD} = 3.3V | | 20 | | mA |
| Input Resistor Range (R _{IN}) | | | 10 | | kΩ |
| Feedback Resistor Range (R _{FB}) | | | 20 | | kΩ |
| Slew Rate | | | 10 | | V/µs |
| Maximum Capacitive Load | | | 220 | | pF |
| Noise Output Voltage (V _N) | A-weighted, BW = 20kHz | | 5.5 | | μVrms |
| Signal to Noise Ratio (SNR) | V_O = 3Vrms, THD+N = 0.1%, BW = 20kHz, A-weighted | | 114 | | dB |
| Unity Gain Bandwidth (G _{BW}) | | | 7.8 | | MHz |
| Open-Loop Voltage Gain (A _{VO}) | | | 120 | | dB |
| Charge Pump Frequency (F _{CP}) | | 330 | 450 | 560 | kHz |
| External Under-Voltage Detection (V _{UVP}) | | 1.05 | 1.13 | 1.25 | V |
| External Under-Voltage Detection Hysteresis Current (I _{Hys}) | | | 4.8 | | μΑ |
| EN PIN | | | | | |
| Input High Voltage (V _{INH}) | EN | 1.2 | | | V |
| Input Low Voltage (V _{INL}) | EN | | | 0.6 | V |

NOTE:

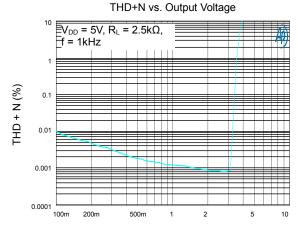
1. For C_{PUMP} , C_{PVSS} , C_{IN} , R_{IN} and etc, please refer to the TYPICAL APPLICATION CIRCUIT on page 7.

TYPICAL PERFORMANCE CHARACTERISTICS

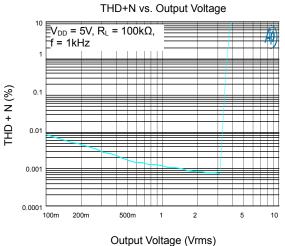
 $V_{DD} = 3.3V, T_A = +25^{\circ}C, R_L = 2.5k\Omega, C_{PUMP} = C_{PVSS} = 1\mu F, C_{IN} = 10\mu F, R_{IN} = 10k\Omega, R_{FB} = 20k\Omega, unless otherwise noted.$

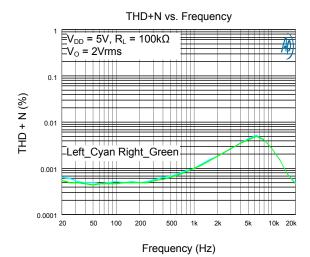






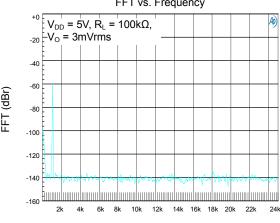
Output Voltage (Vrms)





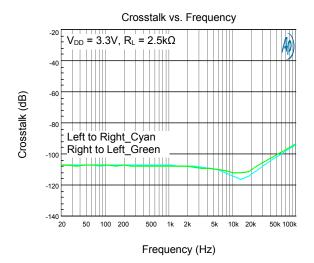
FFT vs. Frequency $V_{DD} = 5V, R_L = 100k\Omega,$ $V_0 = 3mVrms$ -60 -100

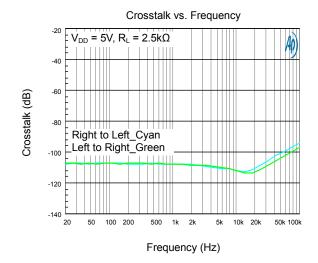
Frequency (Hz)

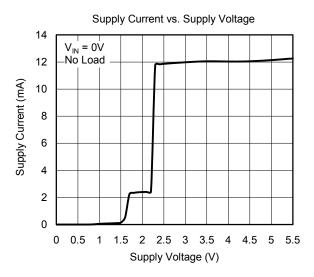


TYPICAL PERFORMANCE CHARACTERISTICS (continued)

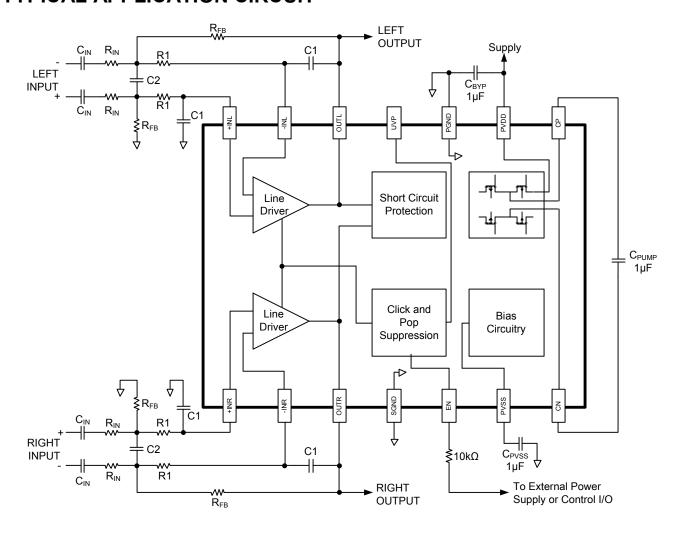
 $V_{DD} = 3.3V, T_A = +25^{\circ}C, R_L = 2.5k\Omega, C_{PUMP} = C_{PVSS} = 1\mu F, C_{IN} = 10\mu F, R_{IN} = 10k\Omega, R_{FB} = 20k\Omega, unless otherwise noted.$







TYPICAL APPLICATION CIRCUIT



NOTES:

- 1. In order to get good performance, it's important to select the right C_{PUMP} , C_{PVSS} and C_{BYP} in application. All tests are performed with circuit set up with X5R and X7R capacitors. Capacitors having high dissipative loss, such as Y5V capacitor, may cause performance degradation and unexpected system behavior.
- 2. A $10k\Omega$ resistor is recommended to be put between EN pin and external power supply or control I/O.

APPLICATION INFORMATION

Decoupling Capacitors

The SGM8903 is a capless line driver amplifier that requires adequate power supply decoupling to ensure that the noise and total harmonic distortion (THD) are low. A good low equivalent-series-resistance (ESR) ceramic capacitor, typically $1\mu F$, placed as close as possible to the device V_{DD} lead, works best. Placing this decoupling capacitor close to the SGM8903 is important for the performance of the amplifier. For filtering lower frequency noise signals, a $10\mu F$ or larger capacitor placed near the audio power amplifier would also help, but it is not required in most applications because of the high PSRR of this device.

Gain Setting Resistors Ranges

The gain setting resistors, R_{IN} and R_{FB} , must be chosen so that noise, stability and input capacitor size of the SGM8903 are kept within acceptable limits. Voltage gain is defined as R_{FB} divided by R_{IN} .

Selecting values that are too low demands a large input AC coupling capacitor, C_{IN} . Selecting values that are too high increases the noise of the amplifier. Table 1 lists the recommended resistor values for different gain settings.

Table 1. Recommended Resistor Values

| INPUT RESISTOR VALUE, R _{IN} | FEEDBACK RESISTOR VALUE, R _{FB} | DIFFERENTIAL INPUT GAIN | INVERTING INPUT GAIN | NON INVERTING INPUT GAIN |
|--|---|----------------------------|-------------------------|-----------------------------|
| 22kΩ | 22kΩ | 1.0V/V | -1.0V/V | 2.0V/V |
| 20kΩ | 30kΩ | 1.5V/V | -1.5V/V | 2.5V/V |
| 33kΩ | 68kΩ | 2.1V/V | -2.1V/V | 3.1V/V |
| 10kΩ | 100kΩ | 10.0V/V | -10.0V/V | 11.0V/V |

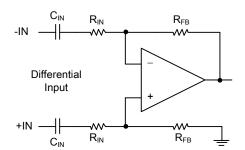


Figure 1. Differential Input

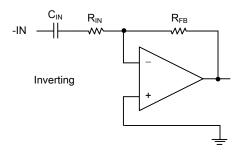


Figure 2. Inverting

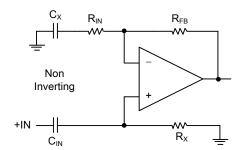


Figure 3. Non-Inverting

APPLICATION INFORMATION (continued)

Input-Blocking Capacitors

DC input-blocking capacitors are required to be added in series with the audio signal into the input pins of the SGM8903. These capacitors block the DC portion of the audio source and allow the SGM8903 inputs to be properly biased to provide maximum performance. The input-blocking capacitors also limit the DC-gain to one, limiting the DC-offset voltage at the output.

These capacitors form a high-pass filter with the input resistor, R_{IN} . The cutoff frequency is calculated using Equation 1. For this calculation, the capacitance used is the input-blocking capacitor and the resistance is the input resistor chosen from Table 1, then the frequency and/or capacitance can be determined when one of the two values are given.

$$fc_{I_N} = \frac{1}{2\pi R_{I_N} C_{I_N}} \text{ or } C_{I_N} = \frac{1}{2\pi f c_{I_N} R_{I_N}}$$
 (1)

Using the SGM8903 as 2nd-Order Filter

Several audio DACs used today require an external low-pass filter to remove out of band noise. This is possible with the SGM8903 as it can be used like a standard OPAMP.

Several filter topologies can be implemented both single-endedly and differentially. In Figure 4, a Multi Feedback (MFB), with differential input and single-ended input is shown.

An AC coupling capacitor to remove DC-content from the source is shown. It serves to block any DC-content from the source and lowers the DC-gain to one, helping reducing the output DC-offset to minimum.

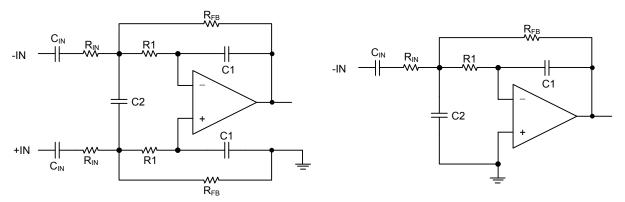


Figure 4. 2nd-Order Active Low Pass Filter

The resistor values should be kept low to obtain low noise, but should also be high enough to get a small size AC coupling cap. Using $5.6k\Omega$ for the resistors, C1 = 220pF, and C2 = 470pF, an SNR of 114dB can be achieved with a $10\mu\text{F}$ input AC coupling capacitor.

Pop-Free Power-Up

Pop-free power-up is ensured by keeping the \overline{SD} (EN) (shutdown pin) low during power supply ramp up and down. The EN pin should be kept low until the input AC coupling capacitors are fully charged before asserting the EN pin high. This way proper precharge of the AC coupling is performed, and pop-free power-up is achieved. Figure 5 illustrates the preferred sequence.

APPLICATION INFORMATION (continued)

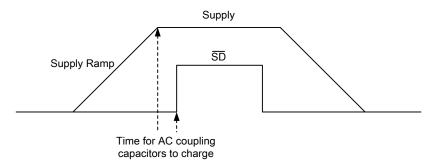


Figure 5. Power-Up Sequence

External Under-Voltage Detection

External under-voltage detection can be used to mute/shut down the SGM8903 before an input device can generate a pop.

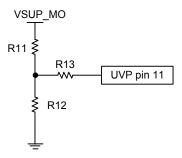
The threshold seen at the UVP pin is 1.13V. A hysteresis is introduced with a resistive divider, where thresholds for startup and shutdown are determined respectively as follows:

Startup Threshold: $V_{UDPR} = 1.13V \times (R11 + R12) / R12$

Shutdown Threshold: $V_{UDPF} = 1.13V \times (R11+R12) / R12 - 4.8\mu A \times (R13 + R11 || R12) \times (R11 + R12) / R12$

Hysteresis: 4.8µA × (R13 + R11 || R12) × (R11 + R12) / R12

The R13 is optional. If the R13 is not used, the UVP pin connects to the divider center tap directly.



Capacitive Load

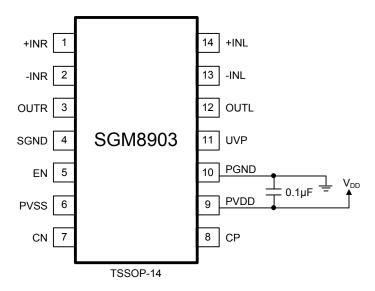
The SGM8903 has the ability to drive large capacitive load up to 220pF directly, and larger capacitive loads can be accepted by adding a series resistor of 47Ω or larger.

Gain-Setting Resistors

The gain setting resistors, R_{IN} and R_{FB} , must be placed close to the input pins to minimize the capacitive loading on these pins and to ensure maximum stability of the SGM8903.

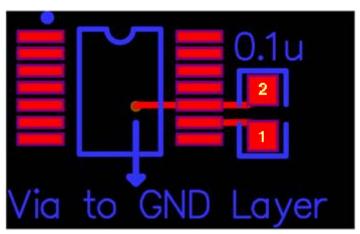
APPLICATION INFORMATION (continued)

PCB Layout Guide

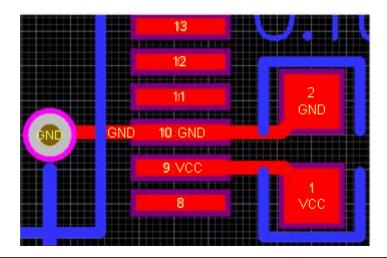


NOTE: $0.1\mu F$ decoupling capacitor must be close to PGND and PVDD pins; capacitor can be connected between PVDD and PGND pins directly and then connect PGND pin to GND layer.

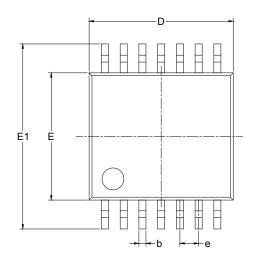
The reference PCB layout is shown in below:

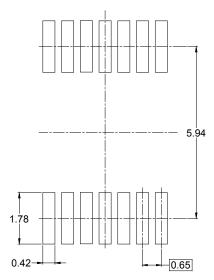


Zoomed in:

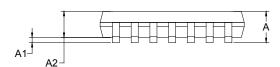


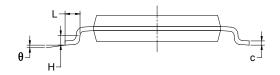
PACKAGE OUTLINE DIMENSIONS TSSOP-14





RECOMMENDED LAND PATTERN (Unit: mm)

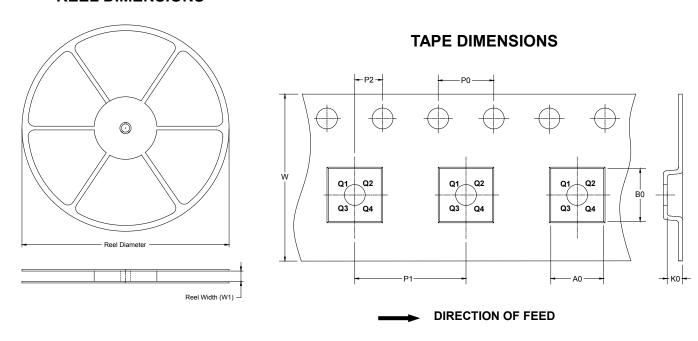




| Symbol | - | nsions imeters | Dimensions In Inches | | |
|--------|-------------|-------------------|-------------------------|-------|--|
| 1 | MIN | MAX | MIN | MAX | |
| А | | 1.200 | | 0.047 | |
| A1 | 0.050 | 0.150 | 0.002 | 0.006 | |
| A2 | 0.800 | 1.050 | 0.031 | 0.041 | |
| b | 0.190 | 0.300 | 0.007 | 0.012 | |
| С | 0.090 | 0.200 | 0.004 | 0.008 | |
| D | 4.860 | 5.100 | 0.191 | 0.201 | |
| Е | 4.300 | 4.500 | 0.169 | 0.177 | |
| E1 | 6.250 6.550 | | 0.246 | 0.258 | |
| е | 0.650 | 0.650 BSC | | BSC | |
| L | 0.500 | 0.700 | 0.02 | 0.028 | |
| Н | 0.25 | 0.25 TYP | | TYP | |
| θ | 1° | 7° | 1° | 7° | |

TAPE AND REEL INFORMATION

REEL DIMENSIONS

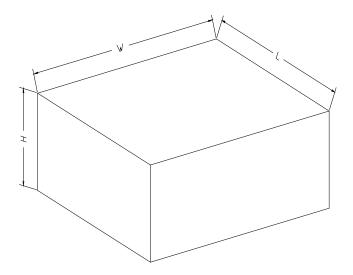


NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF TAPE AND REEL

| Package Type | Reel Diameter | Reel Width W1 (mm) | A0 (mm) | B0 (mm) | K0 (mm) | P0 (mm) | P1 (mm) | P2 (mm) | W (mm) | Pin1 Quadrant |
|--------------|------------------|--------------------------|------------|------------|------------|------------|------------|------------|-----------|------------------|
| TSSOP-14 | 13" | 12.4 | 6.95 | 5.60 | 1.20 | 4.0 | 8.0 | 2.0 | 12.0 | Q1 |

CARTON BOX DIMENSIONS



NOTE: The picture is only for reference. Please make the object as the standard.

KEY PARAMETER LIST OF CARTON BOX

| Reel Type | Length (mm) | Width (mm) | Height (mm) | Pizza/Carton | |
|-----------|----------------|---------------|----------------|--------------|--------|
| 13" | 386 | 280 | 370 | 5 | DD0002 |

单击下面可查看定价,库存,交付和生命周期等信息

>>SGMICRO(圣邦微电子)